#### Errata

Title & Document Type: 4276A LCZ Meter Operating & Service Manual

Manual Part Number: 04276-90000

**Revision Date:** July 1983

### **HP References in this Manual**

This manual may contain references to HP or Hewlett-Packard. Please note that Hewlett-Packard's former test and measurement, semiconductor products and chemical analysis businesses are now part of Agilent Technologies. We have made no changes to this manual copy. The HP XXXX referred to in this document is now the Agilent XXXX. For example, model number HP8648A is now model number Agilent 8648A.

# **About this Manual**

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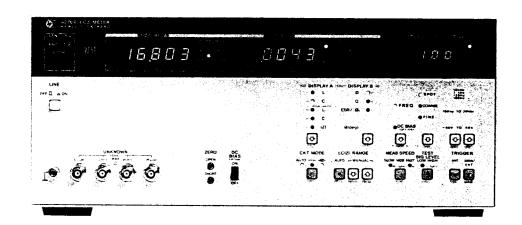
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# 4276A LCZ METER



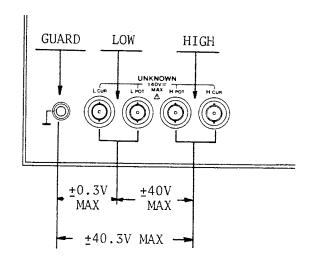


# CAUTIONS ON OPERATION

# USING EXTERNAL BIASING

When measuring any device that is biased from an external bias source, you must observe certain precautions to protect the 4276A's sensitive measurement circuitry.

1) DO NOT apply a dc bias voltage exceeding ±40V between the LOW and HIGH UNKNOWN terminals as shown in the figure below.



Interterminal Bias Limitations

2) DO NOT, under any circumstances, connect a charged capacitor directly to the 4276A's UNKNOWN terminals.

If either of these precautions is ignored, the instrument may be damaged. Symptoms of the damage that may result are listed below.

- (1) No test signal at the H cur terminal
- (2) Excessive display fluctuation during measurement
- (3) During SELF TEST, error codes E37 through E39 and E41 through E45 are displayed on DISPLAY A.

If your 4276A exhibits these symptoms, contact the nearest Hewlett-Packard Sales and Service Office.

#### Note

When making impedance measurements on an active circuit or a device biased from an external source, set the DC BIAS slide switch on the rear panel to EXT and connect nothing to the EXT INPUT/INT MONITOR connector.

#### SAFETY SUMMARY

The following general safety precautions must be observed during all phases of operation, service, and repair of this instrument. Failure to comply with these precautions or with specific warnings given elsewhere in this manual violates safety standards of design, manufacture, and intended use of the instrument. Hewlett-Packard Company assumes no liability for the customer's failure to comply with these requirements.

#### GROUND THE INSTRUMENT

To minimize shock hazard, the instrument chassis and cabinet must be connected to an electrical ground. The instrument is equipped with a three-conductor ac power cable. The power cable must either be plugged into an approved three-contact electrical outlet or used with a three-contact to two-contact adapter with the grounding wire (green) firmly connected to an electrical ground (safety ground) at the power outlet. The power jack and the mating plug of the power cable meet International Electrotechnical Commission (IEC) safety standards.

# DO NOT OPERATE IN AN EXPLOSIVE ATMOSPHERE

Do not operate the instrument in the presence of flammable gases or fumes. Operation of any electrical instrument in such an environment constitutes a definite safety hazard.

#### KEEP AWAY FROM LIVE CIRCUITS

Operating personnel must not remove instrument covers. Component replacement and internal adjustments must be made by qualified maintenance personnel. Do not replace components with power cable connected. Under certain conditions, dangerous voltages may exist even with the power cable removed. To avoid injuries, always disconnect power and discharge circuits before touching them.

# DO NOT SERVICE OR ADJUST ALONE

Do not attempt internal service or adjustment unless another person, capable of rendering first aid and resuscitation, is present.

# DO NOT SUBSTITUTE PARTS OR MODIFY INSTRUMENT

Because of the danger of introducing additional hazards, do not install substitute parts or perform any unauthorized modification to the instrument. Return the instrument to a Hewlett-Packard Sales and Service Office for service and repair to ensure that safety features are maintained.

# DANGEROUS PROCEDURE WARNINGS

Warnings, such as the example below, precede potentially dangerous procedures throughout this manual. Instructions contained in the warnings must be followed.

# WARNING

Dangerous voltages, capable of causing death, are present in this instrument. Use extreme caution when handling, testing, and adjusting.

#### Herstellerbescheinigung

Hiermit wird bescheinigt, daß das Gerät HP 4276A (LCZ Meter) in Übereinstimmung mit den Bestimmungen von Postverfügung 1046/84 funkentstört ist.

Der Deutschen Bundespost wurde das Inverkehrbringen dieses Gerätes angezeigt und die Berechtigung zur Überprüfung der Serie auf Einhaltung der Bestimmungen eingeräumt.

Anm: Werden Meß- und Testgeräte mit ungeschirmten Kabeln und/oder in offenen Meßaufbauten verwendet, so ist vom Betreiber sicherzustellen, daß die Funk-Entstörbestimmungen unter Betriebsbedingungen an seiner Grundstücksgrenze eingehalten werden.

#### Manufacturer's Declaration

This is to certify that this product, the HP 4276A LCZ Meter, meets the radio frequency interference requirements of directive 1046/84. The German Bundespost has been notified that this equipment was put into circulation and was granted the right to check the product type for compliance with these requirements.

Note: If test and measurement equipment is operated with unshielded cables and/or used for measurements on open setups, the user must insure that under these operating conditions, the radio frequency interference limits are met at the border of his premises.

#### SAFETY SYMBOLS

General Definitions of Safety Symbols Used On Equipment or In Manuals.



Instruction manual symbol: the product will be marked with this symbol when it is necessary for the user to refer to the instruction manual in order to protect against damage to the instrument.



Indicates dangerous voltage (terminals fed from the interior by voltage exceeding 1000 volts must be so marked).



Protective conductor terminal. For protection against electrical shock in case of a fault. Used with field wiring terminals to indicate the terminal which must be connected to ground before operating equipment.



Low-noise or noiseless, clean ground (earth) terminal. Used for a signal common, as well as providing protection against electrical shock in case of fault. A terminal marked with this symbol must be connected to ground in the manner described in the installation (operating) manual, and before operating the equipment.



Frame or chassis terminal. A connection to the frame (chassis) of the equipment which normally includes all exposed metal structures.



Alternating current (power line).

\_\_\_

Direct current (power line).

 $\overline{\phantom{a}}$ 

Alternating or direct current (power line).

WARNING

A WARNING denotes a hazard. It calls attention to a procedure, practice, condition or the like, which, if not correctly performed or adhered to, could result in injury or death to personnel.

CAUTION

The CAUTION sign denotes a hazard. It calls attention to an operating procedure, practice, condition or the like, which, if not correctly performed or adhered to, could result in damage to or destruction of part or all of the product.

Note

A Note denotes important information. It calls attention to a procedure, practice, condition or the like, which is essential to highlight.



# OPERATION AND SERVICE MANUAL

# MODEL 4276A LCZ METER

(Including Options 001 and 002)

# SERIAL NUMBERS

This manual applies directly to instruments with serial numbers prefixed 2227J.

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Model 4276A SECTION I

# SECTION I GENERAL INFORMATION

#### 1-1. INTRODUCTION

1-2. This operation and service manual contains the information required to install, operate, test, adjust, and service the Hewlett-Packard Model 4276A LCZ Meter. Figure 1-1 shows the instrument and its supplied accessories. This section covers specifications, instrument identification, description, options, accessories, and other basic information.

1-3. Listed on the title page of this manual is a microfiche part number. This number can be used to order  $4 \times 6$  inch microfilm transparencies of the manual. Each microfiche contains up to 60 photo-duplicates of the manual pages. The microfiche package also includes the latest manual changes supplement as well as all pertinent service notes. To order an additional manual, use the part number listed on the title page of this manual.

#### 1-4. DESCRIPTION

l-5. The HP Model 4276A LCZ Meter is a fully automatic, high performance test instrument designed to measure the inductance. capacitance, dissipation factor, quality factor, conductance, equivalent series resistance, impedance magnitude, and phase of electronic components and devices. Its built-in test signal source covers the frequency range of 100Hz to 20kHz and provides 801 spot frequencies. Test frequency resolution is 1Hz (maximum), and frequency accuracy is ±0.01% of the selected frequency. Frequently used spot frequencies--100Hz, 120Hz. lkHz. and 10kHz-can be quickly selected by the SPOT key. Test signal level is selectable at lVrms (HIGH) or 50mVrms (LOW). In exceptional measurement ranges, the test signal in HIGH mode is 2Vrms. The instrument's 5 terminal configuratin provides a basic measurement accuracy of 0.1% over a wide measurement range.

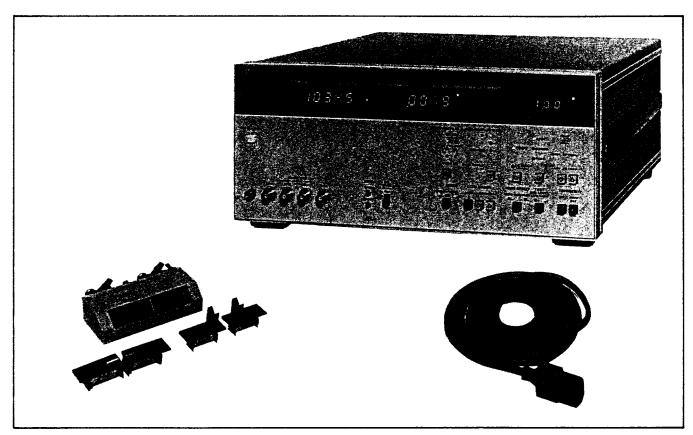


Figure 1-1. Model 4276A and Accessories.

SECTION I Model 4276A

The 4276A has three measurement speed modes: SLOW, MED, and FAST. When MED mode is selected, total time required for a C-D or L-Q measurement is approximately 100ms (at FAST mode measurement time is approximately 40 percent shorter than that of MED mode. Also, the HIGH SPEED C and HIGH L measurement functions reduce measurement time to approximately half that of a normal C-D or L-Q measurement. Shortest measurement time is approximately 25ms (HIGH SPEED C or L, FAST mode, at 20kHz). The 4276A is equipped with a (delta) measurement function to permit temperature dependency or de bias dependency measurements.

- 1-7. All instrument operations--measurement, front panel control settings, self test, continuous memory, etc.—are controlled by a Z80 microprocessor. The built-in self test function can be initiated at any time to verify correct operation of the instrument's basic capabilities. The 4276A is also equipped with a continuous memory function that is automatically activated when the instrument is turned off or experiences a power failure. All front panel control settings (except dc bias), zero offset data, and comparator limits (Option 002) are memorized and automatically recalled when the instrument is turned on again.
- 1-8. The Hewlett-Packard Interface Bus (HP-IB) is standard on the 4276A. All of the instrument's standard and optional functions (except power on/off and DC BIAS ON/OFF) can be remotely controlled from an HP-IB compatible controller. When set to TALK ONLY mode, the 4276A can send measurement data to an external device (a printer, for example) without a controller.
- 1-9. The 4276A can be equipped with two special options: Option 001 Internal DC Bias and Option 002 Comparator/Handler Interface. Refer to paragraph 1-21 for a brief description of these options.
- 1-10. A wide selection of accessories—test fixtures and test leads—is available. A description of furnished accessories is given in paragraph 1-30. For details on available accessories, refer to paragraph 1-32.

## 1-11. SPECIFICATIONS

1-12. Complete specifications of the Model 4276A are given in Table 1-1. These specifications are the performance standards or limits against which the instrument is tested. The test procedures for verifying the specifications are covered in Section IV,

Performance Tests. Table 1-2 lilsts supplemental performance characteristics. Supplemental performance characteristics are not specifications but are typical characteristics included as additional information for the operator. When the 4276A is shipped from the factory, it meets the specifications listed in Table 1-1.

#### 1-13. SAFETY CONSIDERATIONS

- 1-14. The Model 4276A has been designed to conform to the safety requirements of an IEC (International Electromechanical Committee) Safety Class I instrument and is shipped from the factory in a safe condition.
- 1-15. This operating and service manual contains information, cautions, and warnings which must be followed by the user to ensure safe operation and to maintain the instrument in a safe condition.

#### 1-16. INSTRUMENTS COVERED BY MANUAL

- l-17. Hewlett-Packard uses a two-section nine character serial number which is stamped on the serial number plate (Figure 1-2) attached to the instrument's rear-panel. The first four digits and the letter are the serial prefix and the last five digits are the suffix. The letter placed between the two sections identifies the country where the instrument was manufactured. The prefix is the same for all identical instruments; it changes only when a change is made to the instrument. The suffix, however, is assigned sequentially and is different for each instrument. The contents of this manual apply to instruments with the serial number prefix(es) listed under SERIAL NUMBERS on the title page.
- 1-18. An instrument manufactured after the printing of this manual may have a serial number prefix that is not listed on the title page. This unlisted serial number prefix indicates the instrument is different from the one described in this manual. The manual for this newer instrument may be accompanied by a yellow Manual Changes supplement or have a different manual part number. This supplement contains "change information" that explains how to adapt the manual to the newer instrument.

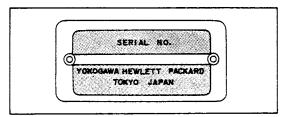


Figure 1-2. Serial Number Plate.

Model 4276A SECTION I

1-19. In addition to change information, the information for supplement may contain correcting errors (called Errata) in the manual. To keep this manual as current and accurate as possible, Hewlett-Packard recommends that you periodically request the latest Manual Changes supplement. The supplement for this manual is identified with this manual's print date and part number, both of which appear on the manual's title page. Complimentary copies of the supplement are available from Hewlett-Packard. If the serial prefix or number of an instrument is lower than that on the title page of this manual, see Section VII, Manual Changes.

1-20. For information concerning a serial number prefix that is not listed on the title page or in the Manual Change supplement, contact the nearest Hewlett-Packard office.

#### 1-21. OPTIONS

1-22. Options are modifications to the standard instrument that implement the user's special requirements for minor functional changes. The 4276A has six options:

Option 001: Internal DC Bias Supply

 $(0 - \pm 40 \text{V})$ 

Option 002: Comparator/Handler Interface

Option 907: Front Handle Kit

Option 908: Rack Flange Kit

Option 909: Rack Flange and Front Handle

Kit

Option 910: Extra Manual

# 1-23. OPTION 001

1-24. Option 001 equips the standard 4276A with a built-in dc voltage source for biasing the device under test. Output voltage is user-selectable from 0 to  $\pm 40 \text{V}$  with 10 mV (0V to  $\pm 9.99 \text{V}$  range) or 100 mV ( $\pm 10 \text{V}$  to  $\pm 40 \text{V}$  range) setting resolution, and can be keyed in directly from the front panel or remotely programmed via the HP-IB. Maximum display resolution is 3 digits.

#### 1-25. OPTION 002

1-26. Option 002 equips the standard 4276A with the 16064A Comparator/Handler Interface for go/no-go testing and automatic bin sorting. Up to nine sets of HIGH/LOW limits for one DISPLAY A function (L, C, or |Z|) and one set of HIGH/LOW limits for one DISPLAY B

function (D, Q, ESR, or G) can be manually keyed in from the 16064A, or entered from a remote controller via the HP-IB. Comparison results--HIGH, IN, or LOW—for each measurement parameter are indicated by LED lamps on the 16064A and by TTL level voltages (open collector) output from the handler interface connector.

HIGH: Measured value exceeds the HIGH limit.

IN: Measured value is within the HIGH and LOW limits, inclusive.

LOW: Measured value is lower than the LOW limit.

#### 1-27. OTHER OPTIONS

1-28. The following options provide the mechanical parts necessary for rack mounting and hand carrying:

Option 907: Front Handle Kit. Furnishes

carrying handles for both ends

of the front-panel.

Option 908: Rack Flange Kit. Furnishes

flanges for rack mounting.

Option 909: Rack Flange and Front Handle

Kit. Furnishes front handles (Opt. 907) and rack flanges

(Opt. 908).

Installation instructions for these options are given in Section II.

1-29. Option 910 adds an extra copy of the Operation and Service Manual.

# 1-30. ACCESSORIES SUPPLIED

1-31. The standard HP Model 4276A LCZ Meter, along with its furnished accessories, is shown in Figure 1-1. The furnished accessories are also listed below:

16047A Test Fixture

(Refer to Table 1-3 for a brief description)

Power Cable HP Part No. 8120-1378

Fuse HP Part No. 2110-0007

or 2110-0360

Table 1-1. Specifications (Sheet 1 of 14)

# Specifications

#### Parameters Measured:

C (capacitance), L (inductance), |Z| (impedance), D (dissipation factor), Q (quality factor), ESR (equivalent series resistance), G (conductance),  $\theta$  (phase angle), HIGH SPEED C, HIGH SPEED L,  $\Delta$  (deviation)

#### Parameter Combinations:

Circuit Mode	Parameter Combination
Series <del>•□•••</del>	C-D, C-Q, C-ESR, L-D, L-Q, L-ESR, $ z $ - $\theta$ , HIGH SPEED C, HIGH SPEED L
Parallel •₩•	C-D, C-Q, C-G, L-D, L-Q, L-G,  Z -0, HIGH SPEED C, HIGH SPEED L

Measurement Circuit Modes:

Auto, Series (-----), and Parallel (-----)

Measurement Speed Modes:

SLOW, MED, and FAST

### Displays:

Measurement Speed Mode	Display Digits	Maximum Display
SLOW	4 1/2	19999 counts
MED		
FAST	3 1/2	1999 counts

### Note

Number of display digits depends on the test frequency, the test signal level, and the measurement range.

Measurement Terminals:

5-terminal configuration with guard terminal

Ranging Modes:

Auto and Manual (UP/DOWN keys)

## Test Frequencies:

Test Frequency Range	Reduction
100Hz to 200Hz	1Hz
200Hz to 500Hz	2Hz
500Hz to 1kHz	5Hz
1kHz to 2kHz	10Hz
2kHz to 5kHz	20Hz
5kHz to 10kHz	50Hz
10kHz to 20kHz	100Hz

Frequency Control Modes:

SPOT (100Hz, 120Hz, 1kHz, 10kHz) COARSE (10 Freq. points/decade) FINE (Maximum resolution)

Frequency Accuracy: ±0.01%

Test Signal Level:

HIGH (1Vrms) or LOW (50mVrms)

Note

HIGH test signal level is 2Vrms on the ranges shown in Tables A and B.

Table A.

Capacitance	Test Frequency Range								
Range	100Hz	to	199Hz	200Hz	to	1.99kHz	2kHz	to	20kHz
10mF									
lmF				) :	2V ri	ns	l		
100µF							)		
10μF									
lμF									
100nF	1Vrms								
10nF	1								
lnF									
100pF									
10pF				`					

Note: C Measurement in Series CKT Mode

#### Table B.

Impedance	Test Frequency Range				
Range	100Hz to 20kHz				
10ΜΩ					
1ΜΩ					
100kΩ					
10kΩ	lVrms				
1kΩ					
100Ω					
10Ω					
1Ω					
100mΩ	2Vrms				

Note: | Z | Measurement

# Level Accuracy:

Test Signal Level	Test Frequency			
1030 013	1kHz	Other than 1kHz		
HIGH	±10%	=50%		
LOW	±20%			

Output Impedance: 100Ω±20%

# Deviation Measurement:

Calculates and displays the difference between stored reference values and measured values.

# ZERO Offset Adjustment:

Compensation for residual impedance and stray admittance of the test fixture connected to the UNKNOWN terminals is automatically done by the ZERO OPEN/SHORT buttons.

#### \* Compensation frequencies:

20kHz, 16kHz, 10kHz, 5kHz, 2kHz, 1kHz, 500Hz, 200Hz, and 100Hz. Compensation at other frequencies is automatically done by secondary interpolation.

# \* Maximum offset values:

C: Up to 20pF (OPEN)
G: Up to .2 μS (OPEN)
| Z |: Up to 2Ω (SHORT)

#### SELF TEST:

Checks the 4276A's basic operation when the instrument is turned on or when the SELF TEST key is pressed. If an abnormality is detected, an error code is displayed on DISPLAY A.

#### External DC Bias:

Up to ±40V dc can be applied to the DUT from an external voltage source connected to the EXT INPUT/INT MONITOR BNC connector on the rear-panel.

Output impedance is  $1020\Omega \pm 10\%$ .

#### Trigger:

Internal, External, Manual, or HP-IB remote control

# HP-IB (Hewlett-Packard Interface Bus):

Data output and remote control. Based on IEEE Std 488 and ANSI-MCl.1.

#### Interface Capabilities:

SH1, AH1, T5, L4, SR1, RL1, DC1, DT1, and El

#### Remote Control:

All front panel control settings (except power switch, and DC BIAS ON/OFF switch) and all 16064A Comparator/Handler Interface settings (option 002)

# Data Output:

Parameter measured, equivalent circuit mode, display status, measured values, and comparator output. Output format is ASCII format or Binary Packed format.

### Continuous Memory:

Memorizes all front panel control settings (except DC BIAS voltage setting), zero offset adjustment data, reference values, and comparator limits (option 002) when the instrument is turned off or experiences a power failure. Settings and data are recalled when the instrument is turned on.

Warm-up Time: Maximum 30minutes

## Ambient Temperature:

23 °C±5 °C (At 0 °C to 55 °C, error doubles)

Table 1-1. Specifications (Sheet 3 of 14)

# CAPACITANCE MEASUREMENT ACCURACY

# C-D Measurement Accuracy:

C Accuracy: ±[(% of reading) + (number of counts)], see Tables A-1 and A-2.

D Accuracy:  $\pm$ [(% of reading) + (D error) + (number of counts)], see Tables A-1 and A-2.

Note: Use Table A-1 when the test frequency is 100Hz, 120Hz, 1kHz, and 10kHz. Use Table A-2 for all other frequencies.

# C-Q Measurement Accuracy:

C Accuracy: ±[(C accuracy of C-D measurement)]

Q Accuracy:  $\pm$ [(D accuracy  $\pm$  measured D value x 100)% of Q reading + 1 count]

Note: Q is the reciprocal of D.

Note: Q accuracy is calculated from the measured D value. Refer to Figure 3-16.

Table A-1. C-D Accuracies (100Hz, 120Hz, 1kHz, 10kHz only)

			•					
Capacitance	Test Frequency Range							
Range	100Hz and 120Hz	1kHz	10kHz					
10mF	3% + 4 3% + .03 + 6							
, lmF	.75% + <u>2</u> .75% + .015 + <u>3</u>	2% + 4 2% + .02 + 6						
100µF	.45% + 2 .45% + .015 + 3	.5% + <u>2</u> .5% + .01 + <u>3</u>	3% + 2 3% + .03 + <u>3</u>					
10µF	3 8	$.3\% + \frac{2}{.01 + \frac{3}{.01}}$	1.5% + <u>2</u> 1.5% + .03 + <u>3</u>					
lμF	1.15% + 5		.9% + <u>2</u> .9% + .03 + <u>3</u>					
100nF	.15% + .0009A + <u>5</u>	.1% + <u>5</u>	.3% + <u>5</u> .3% + .0018A + <u>5</u>					
10nF	]   	.1% + .0006A + <u>5</u>	.3% + .0018A + 5					
lnF	.45% + <u>5</u> .45% + .0045A + <u>5</u>	a	.6% + 10					
100pF		.3% + <u>5</u>   .3% + .003A + <u>5</u>   .3% + .003A - <u>5</u>	.6% + <u>10</u> .6% + <u>.</u> 0036A + <u>10</u>					
10pF			1.2% + 4 11.2% + .0036A + 6					

Table 1-1. Specifications (Sheet 4 of 14)

#### Table A-2. C-D Accuracies Test Frequency Range Capacitance Range 101Hz to 199Hz\* 200Hz to 498Hz 500Hz to 995Hz 1.01kHz to 1.99kHz 2kHz to 4.98kHz 5kHz to 9.95kHz 10.1kHz to 20kHz 10mF $\frac{2\%}{2\%} + \frac{2}{.02} + 3$ lmF $\frac{15 + 2}{15 + .02 + 3}$ 3% + 2 3% + .03 + 3 5% + 2 5% + .05 + 3 100uF $2.5\% + \frac{2}{2.5\% + .05 + 3}$ .6% + 2 .6% + .02 + 3 10µF $.9\% + \frac{2}{.03}$ $.9\% + .03 + \frac{3}{.03}$ 1uF 100nF .3% + <u>5</u> .3% + .0018A + <u>5</u> .5% + 5 .5% + .003A + 5 .2% + 5 .2% + .0012A + 5 10nF

.6% + 5 .6% + .006A +

.4% + 2 .4% + .0012A +

\* Except 120Hz

Equations in Tables A-1 and A-2 represent

6% + <u>5</u> 6% - .006A -

100pF

10pF

C Accuracy
D Accuracy

α: Full-scale factor (= measured C value ÷ full-scale C value). For example, when the measured C value is 850pF on the 1000pF range, α is 0.85.

A: =  $[\alpha + (1/\alpha)]/2$ 

Note 1: Tables A-1 and A-2 are applicable under the following conditions:

- (1) Test Signal Level: HIGH
- (2) Measurement Speed: MED or SLOW
- (3) Sample's D Value: ≤ 0.1
- (4) Zero offset adjustment has been performed with the OPEN and SHORT terminations of the Model 16074A.

.6% + 10 .6% + .0036A + 10

.2% + 4 .2% + .0036A + 1% + 10 1% + .006A + 10

2% + <u>20</u> 2% + <u>.0</u>12A + <u>20</u>

 $2\% + \frac{4}{.006A}$ 

- Note 2: Error doubles when LOW test signal level (50mVrms) is used. LOW test signal level can be used only on the ranges enclosed in the bold line in Tables A-l and A-2.
- Note 3: When FAST measurement speed is used, accuracies double on the ranges outside the area enclosed in the bold line in Tables A-1 and A-2.

HIGH SPEED C Measurement Accuracy:

C Accuracy: ±[(C accuracy of C-D measurement) + (X% of reading)], see Table A-3.

Note: HIGH SPEED C accuracy is specified on the ranges enclosed in the dotted line in Tables A-1 and A-2.

Table A-3. Additional Error

Test Signal Level	Sample's D Value			
l cost organization	D≤.0004	.0004 <d≤.002< td=""><td>.002<d≤.1< td=""><td>D&gt;.1</td></d≤.1<></td></d≤.002<>	.002 <d≤.1< td=""><td>D&gt;.1</td></d≤.1<>	D>.1
HIGH	X = 0		X = 20D	Not specified.
LOW	$X = 0 \qquad X = 100D$		1	

Note: Table A-3 is applicable under the following condition:

(1) Zero offset adjustment has been performed with the OPEN and SHORT terminations of the Model 16074A.

Table 1-1. Specifications (Sheet 5 of 14)

# C-ESR/G Measurement Accuracy:

C Accuracy: ±[(C accuracy of C-D measurement)]

ESR Accuracy: ±[(% of reading) + (ESR error in ohms), see Tables A-4 and A-5.

G Accuracy:  $\pm$ [(% of reading) + (G error in siemens) + (number of counts)], see Tables A-4 and A-5.

Note: Use Table A-4 when the test frequency is 100Hz, 120Hz, lkHz, or 10kHz. Use Table A-5 for all other frequencies.

Note: ESR range and G range depend on the selected C range and test frequency. Refer to Table A-6.

Note: DISPLAY B function, when ESR/G is selected, depends on the CIRCUIT MODE.

Table A-4. C-ESR/G Accuracies (100Hz, 120Hz, 1kHz, 10kHz only)

<u> </u>		,	Test Frequency Rang	e	
ESR/G Range		100Hz and 120Hz	1kHz	10kHz	
ESR	10ΜΩ	See Note 1	See Note 1	See Note 1	
G	1μS	$.6\% + 6 \alpha nS + 5$	$.3\% + 3\alpha nS + 5$	$1.2\% + 12 \alpha nS + 4$	
ESR	1ΜΩ	See Note 1		See Note 1	
G	10μS	.2% + 40αnS + <u>5</u>	.1% + 20αnS + <u>5</u>	.6% + .12 <sub>αμ</sub> S + <u>10</u>	
ESR	100kΩ	See Note 1	See Note 1	See Note 1	
G	100µS	.2% + .4αμS + <u>5</u>	.1% + .2αμS + <u>5</u>	.6% + 1.2αμS + <u>10</u>	
ESR	10kΩ	See Note 1	See Note 1	See Note 1	
G	1mS	.2% + 4αμS + <u>5</u>	.1% + 2αμS + <u>5</u>	.3% + 6αμS + <u>5</u>	
ESR	1kΩ	See Note 1	See Note 1	See Note 1	
G	10mS	.2% + 40αμS + <u>5</u>	.1% + 20αμS + <u>5</u>	.3% + 60αμS + <u>5</u>	
ESR	100Ω	$.4\% + .4/\alpha\Omega + 5$	.2% + .2/αΩ + <u>5</u>	$.6\% + .6/\alpha\Omega + 5$	
G	100mS	See Note 2	See Note 2	See Note 2	
ESR	10Ω	$1\% + .1/\alpha\Omega + 2$	$.5\% + 50/\alpha m\Omega + 2$	1.5% + .15/αΩ + <u>2</u>	
G	1S	See Note 2	See Note 2	See Note 2	
ESR	1Ω	2% + 20/αmΩ + 2	1% + 10/αmΩ + <u>2</u>	$3\% + 30/\alpha m\Omega + 2$	
G	10S	See Note 2	See Note 2	See Note 2	

Table 1-1. Specifications (Sheet 6 of 14)

				Table A-5.	C-ESR/G Acc	uracies		
					Test Frequency Rai	nge		
ESR/G	Range	101Hz to 199Hz*	200Hz to 498Hz	500Hz to 995Hz	1.01kHz to 1.99kHz	2kHz to 4.98kHz	5kHz to 9.95kHz	10.1kHz to 20kH
ESR	10ΜΩ	See Note 1		See Note 1	See Note 1		See Note 1	See Note 1
G	1µS	.6% + 6anS + <u>5</u>		.4% + 4 anS + 2	.6% + 6anS + <u>5</u>		1.2% + 12anS + 4	2% + 20anS +
ESR	1MΩ			See Note 1			See Note 1	See Note 1
G	10µS		.2% + 40anS + <u>5</u>				.6% + .12αμS + <u>10</u>	2% + .4œuS + <u>2</u>
ESR	100kΩ	See Note 1				See Note 1	See Note 1	
G	100µS	.2% + .4auS + <u>5</u>				.6% + 1.2auS + 10	13 + 2αμS + <u>1</u>	
ESR	10kΩ		See Note 1					See Note 1
G	lmS		. 2% + 4αμS + <u>5</u>				.3% + 6auS + <u>5</u>	.5% + 10aus +
ESR	1kΩ			See Note 1	. #		1	See Note 1
G	10mS			.2% + 40αµS ·	<u> 5</u>		.3% + 60αμS + <u>5</u>	.5% + .1amS +
ESR	100Ω			.4% + .4/αΩ +	· <u>5</u>		.6% + .6/αΩ + <u>5</u>	
G	100mS			See Note 2			See Note 2	See Note 2
ESR	10Ω	1°s + .1/αΩ + <u>2</u>				$1.5\% + .15/\alpha\Omega + 2$ See Note 2	2.5% + .25/αΩ +	
G	15		See Note 2				See Note 2	See Note 2
ESR	1Ω		2° + 20/cmΩ + <u>2</u>				3% + 30/αmΩ + <u>2</u>	5% + 50/αmΩ +
G	105			See Note 2			See Note 2	See Note 2

<sup>\*</sup> Except 120Hz

Equations in Tables A-4 and A-5 represent:

ESR Accuracy
G Accuracy

- lpha: Full-scale factor (= measured C value  $\div$  full-scale C value). For example, when the measured C value is 850pF on the 1000pF range, lpha is 0.85.
- Note 1: ESR accuracy is ±[2 (C accuracy ÷ measured C x 100)% of ESR reading + (G accuracy ÷ measured G x 100)% of ESR reading + 1 count].
- Note 2: G accuracy is ±[2 (C accuracy ÷ measured C x 100)% of G reading + (ESR accuracy ÷ measured ESR x 100)% of G reading) + 1 count].
- Note 3: Tables A-4 and A-5 are applicable under the following conditions:
- (1) Test Signal Level: HIGH
- (2) Measurement Speed Mode: MED or SLOW
- (3) Sample's D Value:  $\leq 0.1$
- (4) Zero offset adjustment has been performed with the OPEN and SHORT terminations of the Model 16074A.

- Note 4: Error doubles when LOW test signal level (50mVrms) is used.

  LOW test signal level can be used only on the ranges enclosed in the bold line in Tables A-4 and A-5.
- Note 5: When FAST measurement speed is used, accuracies doule on the ranges outside the area enclosed in the bold line in Tables A-4 and A-5.

Table A-6. ESR/G Range Selection

Capacitance	Test Frequency Range						
Range	100Hz to 199Hz	200Hz to 498Hz 500Hz to 1.99kH	z 2kHz to 4,98kHz 5kHz to 20kHz				
10mF							
lmF		1Ω 10S					
100µF		10Ω 1S					
10 LF		100Ω 100mS					
luF		1kC 10mS					
100nF		10kΩ 1mS					
10nF		100kΩ 100us					
ln₽	10M? 1uS	IMΩ 10uS					
100pF		10MΩ 1US					
10pF			10MN 1uS				

Table 1-1. Specifications (Sheet 7 of 14)

# INDUCTANCE MEASUREMENT ACCURACY

## L-D Measurement Accuracy:

L Accuracy: ±[(% of reading) + (L error) + (number of counts)], see Tables B-1 and

B-2.

D Accuracy: ±[(% of reading) + (D error) + (number of counts)], see Tables B-1 and

B-2.

Note: Use Table B-1 when the test frequency is 100Hz, 120Hz, 1kHz, or

10kHz. Use Table B-2 for all other frequencies.

# L-Q Measurement Accuracy:

L Accuracy: ±[(Laccuracy of L-D measurement)]

Q Accuracy:  $\pm$ [(D accuracy  $\div$  measured D value x 100)% of Q reading + 1 count]

Note: Q value is the reciprocal of D.

Note: Q accuracy is calculated from the measured D value. Refer to Figure

3-15.

Table B-1. L-D Accuracies (100Hz, 120Hz, 1kHz, 10kHz only)

Inductance	,	Test Frequency Ran	ge
Range	100Hz and 120Hz	1kHz	10kHz
1kH			
100Н		$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	
10Н	$.4 (1 + \alpha)\% + \frac{2}{2}$ $.8\% + .02 + \frac{2}{2}$		$\begin{array}{c} 1.5 & (1 + \alpha)\% + 2 \\ 3\% + .03 + 2 \end{array}$
1H	     	$\begin{array}{c} .2 & (1 + \alpha)\% + 2 \\ .4\% + .01 + 2 \end{array}$	
100mH	.4% + <u>5</u> .6% + .006/α + <u>5</u>		$\begin{array}{c} .6 \ (1 + \alpha)\% + \underline{2} \\ 1.2\% + .03 + \underline{2} \end{array}$
10mH	.6% + <u>5</u> .8% + .008/α + <u>5</u>	.2% + <u>5</u> .3% + .003/α + <u>5</u>	
1mH		.3% + <u>5</u> .4% + .004/α + <u>5</u>	.6% + <u>5</u> .9% + .009/α + <u>5</u>
100μΗ			.9% + <u>5</u> 1.2% + .012/α + <u>5</u>

Table 1-1. Specifications (Sheet 8 of 14)

#### Table B-2. L-D Accuracies Test Frequency Range Inductance Range 10.1kHz to 20kHz 101Hz to 995Hz\* 1.01kHz to 4.98kHz | 5kHz to 9.95kHz 1kH $(1 + \alpha)$ % + $\frac{2}{2}$ 2% + .02 + $\frac{2}{2}$ $\frac{1.5(1 + \alpha)\% + 2}{3\% + .03 + 2}$ 100H 2.5 $(1 + \alpha)$ % + 2 10H 5% + .05 + 2 .4 (1 + α) % + <u>2</u> $.6 (1 + \alpha)\% + 2$ 1.2% + .03 + 2 .8% + .02 + <u>2</u> 1H $(1 + \alpha)$ % + $\frac{2}{2}$ 2% + .05 + $\frac{2}{2}$ 100mH .4% + <u>5</u> .6% + .006/α + <u>5</u> .6% + <u>5</u> .9% + .009/α + <u>5</u> 10mH .6% + <u>5</u> .8% + .008/α + <u>5</u> $1\% + \frac{5}{4}$ 1.5% + .015/\alpha + $.9\% + \frac{5}{1.2\% + 0.012/\alpha + 5}$ 1mH 100uH $2\% + .0\overline{2}/\alpha$

\* Except 120Hz

Equations in Tables B-1 and B-2 represent:

L Accuracy
D Accuracy

α: Full-scale factor (= measured L value ÷ full-scale L value). For example, when the measured L is 850nH on the 1000nH range, α is 0.85.

Note 1: Tables B-1 and B-2 are applicable under the following conditions:

(1) Test Signal Level: HIGH

- (2) Measurement Speed Mode: MED or SLOW
- (3) Sample's D Value: ≤ 0.1
- (4) Zero offset adjustment has been performed with the OPEN and SHORT terminations of the Model 16074A.

Note 2: When FAST measurement speed is used, accuracies double on the ranges outside the area enclosed in the bold line in Tables B-1 and B-2.

Note 3: LOW test signal level cannot be used in L measurement mode.

HIGH SPEED L Measurement Accuracy:

L Accuracy: ±[(L accuracy of L-D measurement) + (Y% of reading)], see Table B-3.

Note: HIGH SPEED L accuracy is specified on the ranges enclosed in the dotted line in Tables B-1 and B-2.

Table B-3. Additional Error

Test Signal Level		Sample	's D Value	
1000 015	D≤.0004	.0004 <d≤.002< td=""><td>.002<d≤.1< td=""><td>D&gt;.1</td></d≤.1<></td></d≤.002<>	.002 <d≤.1< td=""><td>D&gt;.1</td></d≤.1<>	D>.1
HIGH		Y = 0	Y = 20D	Not specified.

Note: Table B-3 is applicable under the following conditions:

 Zero offset adjustment has been performed with the OPEN and SHORT terminations of the Model 16074A.

Note: LOW test signal level cannot be used in L measurement mode.

Table 1-1. Specifications (Sheet 9 of 14)

# L-ESR/G Measurement Accuracy:

L Accuracy: ±[(Laccuracy of L-D measurement)]

ESR Accuracy: ±[(% of reading) + (ESR error in ohms) + (number of counts)], see Tables B-4 and B-5.

G Accuracy: ±[(% of reading) + (G error in siemens) + (number of counts)], see Tables B-4 and B-5.

Note: Use Table B-4 when the test frequency is 100Hz, 120Hz, 1kHz, or 10kHz. Use Table B-5 for all other frequencies.

Note: ESR range and G range depend on the selected L range and test frequency. Refer to Table B-6.

Note: DISPLAY B function, when ESR/G is selected, depends on the CIRCUIT MODE.

Table B-4. L-ESR/G Accuracies (100Hz, 120Hz, 1kHz, 10kHz only)

		Test Frequency Range			
ESR/G	Range	100Hz and 120Hz	1kHz	10kHz	
ESR	1ΜΩ	See Note 1	See Note 1	See Note 1	
G	10µS	$.4\% + 40/\alpha nS + 2$	$.2\% + 20/\alpha nS + 2$	1.2% + .12/αμS + <u>4</u>	
ESR	100kΩ		See Note 1		
G	100µՏ	.4% + .4/αμS + <u>5</u>	.2% + .2/αμS + <u>5</u>	1.2% + 1.2/αμS + <u>10</u>	
ESR	10kΩ	See Note 1	See Note 1 .2% + 2/αμS + <u>5</u>	See Note 1 .6% + 6/αμS + <u>5</u>	
G	1mS	.4% + 4/αμS + <u>5</u>			
ESR	1kΩ	See Note 1	i i	See Note 1 .6% + 60/αμS + <u>5</u>	
G	10mS	.4% + 40/αμS + <u>5</u>			
ESR	100Ω	$.2\% + .4\alpha\Omega + 5$	.1% + .2αΩ + 5	$.3\% + .6\alpha\Omega + 5$	
G	100mS	See Note 2	See Note 2	See Note 2	
ESR	10Ω	.6% + 60αmΩ + 10	$.3\% + 30 \text{cm}\Omega + 10$	.9% + 90απΩ + 10	
G	1S	See Note 2		See Note 2	

Table 1-1. Specifications (Sheet 10 of 14)

Table B-5. L-ESR/G Accuracies

50D (0			Test Frequency Range		
ESR/G	Range	101Hz to 4.98kHz*	5kHz to 9.95kHz	10.1kHz to 20kHz	
ESR	1ΜΩ	See Note 1	See Note 1	See Note 1	
G	10µS	$.4\% + 40/\alpha nS + 2$	1.2% + .12/αμS + <u>4</u>	4% + .4/αμS + <u>8</u>	
ESR	100kΩ	See Note 1	See Note 1	See Note 1	
G	100µS	.4% + .4/αμS + <u>5</u>	1.2% + 1.2/αμS + <u>10</u>	2% + 2/αμS + <u>10</u>	
ESR	10kΩ	See Note 1	See Note 1	See Note 1	
G	1mS	.4% + 4/αμS + <u>5</u>	.6% + 6/αμS + <u>5</u>	1% + 10/αuS + <u>5</u>	
ESR	1kΩ	See Note 1	See Note 1	See Note 1	
G	10mS	.4% + 40/αμS + <u>5</u>	.6% + 60/αμS + <u>5</u>	1% + 100/αμS + <u>5</u>	
ESR	100Ω	$.2\% + .4\alpha\Omega + 5$	.3% + .6αΩ + 5	.5% + αΩ + 5	
G	100mS	See Note 2	See Note 2	See Note 2	
ESR	10Ω	.6% + 60cmΩ + 10	.9% + 90amΩ + 10	$1.5\% + .15\alpha\Omega + 10$	
G	18	See Note 2	See Note 2	See Note 2	

<sup>\*</sup> Except 120Hz and 1kHz

Equations in Tables B-4 and B-5 represent:

ESR Accuracy
G Accuracy

 $\alpha$ : Full-scale factor (= measured L value  $\div$  full-scale L value). For example, when measured C value is 850nH on the 1000nH range,  $\alpha$  is 0.85.

Note 1: ESR accuracy is  $\pm$ [(L accuracy  $\div$  measured L x 100)% of ESR reading + (G accuracy  $\div$  measured G x 100)% of ESR reading + 1 count].

Note 2: G accuracy is ±[2 (L accuracy ÷ measured L x 100)% of G reading + (ESR accuracy ÷ measured ESR x 100)% of G reading + 1 count].

Note 3: Tables B-4 and B-5 are applicable under the following conditions:

- (1) Test Signal Level: HIGH
- (2) Measurement Speed Mode: MED or SLOW
- (3) Sample's D Value: ≤ 0.1
- (4) Zero offset adjustment has been performed with the OPEN and SHORT terminations of the Model 16074A.

Note 4: When FAST measurement speed is used, accuracies double on the ranges outside the area enclosed in the bold line in Tables B-4 and B-5.

Note 5: LOW test signal level cannot be used in L measurement mode.

Table B-6. ESR/G Range Selection

Inductance	Te	est Frequency Ran	ge
Range	100Hz to 995Hz	1kHz to 9.95kHz	10kHz to 20kHz
lkH			
100Н		1MΩ 10μS	
10H		100kΩ 100uS	
1H		10kΩ 1mS	
100mH		1kΩ 10mS	
10mH		100Ω 100mS	
1 mH		10Ω 1S	
100µН	-		

Table 1-1. Specifications (Sheet 11 of 14)

# IMPEDANCE MEASUREMENT ACCURACY

 $|Z| - \theta$  Measurement Accuracy:

| Z | Accuracy: ±[(% of reading) + (number of counts)], see Tables C-1 and C-2.

 $\theta$  Accuracy:  $\pm$ [( $\theta$  error in degrees) + (number of counts)], see Tables C-1 and C-2.

Note: Use Table C-1 when the test frequency is 100Hz, 120Hz, 1kHz, or 10kHz. Use Table C-2 for all other frequencies.

Table C-1.  $|Z| - \theta$  Accuracies (100 Hz, 120 Hz, 1kHz, 10kHz only)

Impedance	Te	st Frequency Ran	ıge
Range	100Hz and 120Hz	1kHz	10kHz
10ΜΩ	(1 + α) (1 + α)	)% + <u>2</u> )° + <u>2</u>	$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$
1ΜΩ	.5 (1 + .5 (1 +	$\alpha)\% + \frac{2}{2}$ $\alpha)° + \frac{2}{2}$	$(1 + \alpha)$ % + $\frac{2}{2}$ $(1 + \alpha)$ ° + $\frac{2}{2}$
100kΩ			
10kΩ	.2 (1 + .2 (1 +	$(\alpha)^{\frac{\alpha}{6}} + \frac{2}{2}$	$\begin{array}{c ccccccccccccccccccccccccccccccccccc$
1kΩ			
100Ω	.1% - .1/α'	+ <u>5</u>	.2% + <u>5</u> .2/α° + <u>5</u>
10Ω	.3% - .3/α'	+ <u>5</u> ° + <u>5</u>	.6% + <u>5</u> .6/α° + <u>5</u>
1Ω	.7% .5/α'	+ <u>2</u> • + <u>2</u>	$\frac{1.4\% + 2}{1/\alpha^{\circ} + 2}$
100mΩ	1% + 1/α°	<u>20</u> + <u>2</u>	$\frac{2\% + 20}{2/\alpha^{\circ} + 2}$

Table 1-1. Specifications (Sheet 12 of 14)

Table C-2.  $|Z| - \theta$  Accuracies

Impedance	Т	est Frequency Range	
Range	101Hz to 995Hz*	1.01kHz to 9.95kHz	10.1kHz to 20kHz
10ΜΩ	$(1 + \alpha)^{9} + \frac{2}{2}$ $(1 + \alpha)^{\circ} + \frac{2}{2}$	$\begin{array}{cccccccccccccccccccccccccccccccccccc$	$\begin{array}{cccccccccccccccccccccccccccccccccccc$
1ΜΩ	.5 $(1 + \alpha)$ % + 2 .5 $(1 + \alpha)$ ° + $\overline{2}$	$(1 + \alpha)\% + \frac{2}{2}$ $(1 + \alpha)^{\circ} + \frac{2}{2}$	2.5 $(1 + \alpha)$ % + 2 2.5 $(1 + \alpha)$ ° + $\frac{2}{2}$
100kΩ			
10kΩ	.2 $(1 + \alpha)$ % + 2 .2 $(1 + \alpha)$ ° + $\frac{2}{2}$	.4 $(1 + \alpha)$ % + $\frac{2}{2}$ .4 $(1 + \alpha)$ ° + $\frac{2}{2}$	$(1 + \alpha)\% + \frac{2}{2}$ $(1 + \alpha)^{\circ} + \frac{2}{2}$
1kΩ			
100Ω	.1% + <u>5</u> .1/α° + <u>5</u>	.2% + <u>5</u> .2/α° + <u>5</u>	.5% + <u>5</u> .5/α° + <u>5</u>
10Ω	$.3\% + \frac{5}{5}$ $.3/\alpha^{\circ} + \frac{5}{5}$	.6% + <u>5</u> .6/α° + <u>5</u>	1.5% + <u>5</u> 1.5/α° + <u>5</u>
1Ω	.7% + <u>2</u> .5/α° + <u>2</u>	$\frac{1.4\% + 2}{1/\alpha^{\circ} + 2}$	3.5% + <u>2</u> 2.5/α° + <u>2</u>
$100$ m $\Omega$	$\frac{1\% + 20}{1/\alpha^{\circ} + 2}$	$\frac{2\% + 20}{2/\alpha^{\circ} + 2}$	5% + <u>20</u> 5/α° + <u>2</u>

\* Except 120Hz

Equations in Tables C-1 and C-2 represent:

 $\alpha$ : Full-scale factor (= measured | Z | value ÷ full-scale | Z | value). For example, when measured | Z | value is  $850\Omega$  on the  $1000\Omega$  range,  $\alpha$  is 0.85.

Note l: Tables C-l and C-2 are available under the following conditions:

- (1) Test Signal Level: HIGH
- (2) Measurement Speed Mode: MED or SLOW
- (3) Sample's D Value: ≤ 0.1
- (4) Zero offset adjustment has been performed with the OPEN and SHORT terminations of the Model 16074A.

- Note 2: When FAST measurement speed is used, accuracies double on the ranges outside the area enclosed in thd bold line in Tables C-1 and C-2.
- Note 3: LOW test signal level cannot be used in  $\mid Z \mid$  measurement mode.

Table 1-1. Specifications (Sheet 13 of 14)

# **Options**

# Option 001:

Internal DC Bias. Equips the standard 4276A with a variable 0 to ±40V dc voltage source for biasing DUTs connected to the UNKNOWN terminals. Output voltage can be set from the front panel or via the HP-IB.

Bias Control Range and Accuracy:

Voltage Range	Step	Temperature	Accuracy
10.0 to 40.0V	100mV	23°C ± 5°C 0°C to 55°C	±(.5%+35mV) ±(1%+70mV)
.00 to 9.99V	10mV	23°C ± 5°C 0°C to 55°C	±(.3%+10mV) ±(1%+20mV)
-9.99 to01V	10mV	23°C ± 5°C 0°C to 55°C	±(1%+10mV) ±(2%+20mV)
-40.0 to -10.0V	100mV	23°C ± 5°C 0°C to 55°C	±(1%+35mV) ±(2%+70mV)

#### Note

DC bias voltage is specified 2 minutes after the voltage is set.

Output Impedance: 1020Ω±10%

Bias Voltage Monitor:

Bias voltage across the DUT can be monitored at the EXT INPUT/INT MONITOR BNC connector on the rear panel. INT MONITOR output impedance is approximately  $730\Omega$ .

Output Characteristics:

Range Resistor Value	Maximum Current
100Ω	1mA
lkΩ	0.5mA
10kΩ	50µA
100kΩ	5µA

# Note

Refer to Figure 3-16 for the range resistor value.

#### Note

Measurement accuracies are not guaranteed if output current exceeds the maximum current for each range resistor.

Option 002: C

COMPARATOR/HANDLER

INTERFACE

### Contents:

Model 16064A COMPARATOR/ HANDLER INTERFACE (Includes the 16064-66502 Interface board assembly and 1251-0084 36-pin male Amphenol connector)

# Comparator Function:

Compares measured values to 9 sets (Bins) of stored high/low limits. Displays LOW/IN/HIGH judgements and bin number.

Handler Interface Function: Outputs comparison results and handler control signals (TTL, open-collectors). Detects KEY LOCK and EXT TRIGGER signals sent from component handler.

Option 907: Front handle kit

(Part No. 5061-0090)

Option 908: Rack flange kit

(Part No. 5061-0078)

Option 909: Rack flange and handle kit

(Part No. 5061-0084)

Option 910: Extra manual

Model 4276A SECTION I

Table 1-1. Specifications (Sheet 14 of 14)

# Accessories Supplied

Test Fixture:

16047A Test Fixture. Includes three kinds of

contact inserts

Power Cord: HP Part No. 8120-1378

Fuse:

Part No. 2110-0007 (100V/120V) Part No. 2110-0360 (220V/240V)

Protective Fuse:

Part No. 2110-0011 (for de bias input)

Accessories Available

HP-IB Cable:

10833A (1m) 10833C (4m) 10833B (2m) 10833D (0.5m)

Test Fixtures and Test Leads: Refer to

Table 1-3.

# General Specifications

Operating Temperature: 0 °C to 55 °C

Relative Humidity: 95% at 40 °C

Storage Temperature: -40 °C to +70 °C

Power Requirements:

90V to 132V, 198V to 250V. 48Hz to 66Hz.

Power Consumption:

65 V Amax with any option

Dimensions:

425.5 (W) x 188 (H) x 430 (D) mm

Weight: Approximately 8.5kg

#### 1-32. ACCESSORIES AVAILABLE

1-33. In addition to the furnished 16047A Test Fixture, seven special purpose test fixtures and test leads are available. Each is intended for a particular measurement or DUT type, and all were designed with careful consideration to accuracy, reliability, ease of use, and compatibility with other HP instruments. A brief description of each available accessory is given in Table 1-3.

SECTION I Model 4276A

Table 1-2. Supplemental Performance Characteristics (Sheet 1 of 2)

# Supplemental Performance Characteristics

Measurement Accuracy:

Additional Error in lm/2m Test Lead Usage:

Add the following errors to the measurement accuracies listed in Table 1-1 when a 1m or 2m long test lead is used.

Measurement Function	Additional Error	
L, C,  Z	M% of reading	
D	Р	
θ	60P <b>°</b>	
G (C-G)	M% of reading + (G full-scale value) x $\alpha P$	
G (L-G)	M% of reading + (G full-scale value) x $P/\alpha$	

where,  $\alpha$  is the full scale factor, and M and P are represented as follows:

$$M = K (.16 + .1 \ell) \times (.3f^2 + f)/100$$
  
 $P = K (.16 + .1 \ell) \times (.09f^2 + 1.6f)/1000$ 

where,  $\ell$  is the cable length in meters, f is test frequency in kHz, and K is the range resistor factor. See Figure 3-16.

Range Resistor Value	K
100kΩ	10
10kΩ	1
1kΩ	0.1
100Ω	0

The above errors are applicable under the following conditions:

(1) Test Cable: HP Model 16048A (1m long) or 16048D (2m long)

- (2) Residual Capacitance to Ground:
  <100pF at HIGH terminals
  and <50pF at LOW terminals
- (3) Sample's D Value: < 0.1
- (4) Zero offset adjustment has been performed with the OPEN and SHORT terminations of the 16074A.

Additional Error of Test Fixture:

Add the following errors to the measurement accuracies listed in Table 1-1 when these test fixtures or test leads are used:

Model	Additional Error	
16047A, 16047C, 16048A, 16048B, 16048D, 16065A	-	
16048C	C<5pF L<200nH R<10mΩ	
16034B	C<0.02pF L<30nH R<50mΩ	

Accuracies in D>0.1:

Multiply the measurement accuracies for C, L, or D listed in Table 1-1 by  $(1 + D^2)$ , where D is the sample's D value.

Ranging Time: Approximately 60ms

Test Signal Settling Time:

Approximately 270ms when the test frequency is changed.

Approximately 60ms when the test signal level is changed.

The same as the dc bias settling time when the dc bias voltage is changed.

Continuous Memory: 2 weeks at 23 ° C

Table 1-2. Supplemental Performance Characteristics (Sheet 2 of 2)

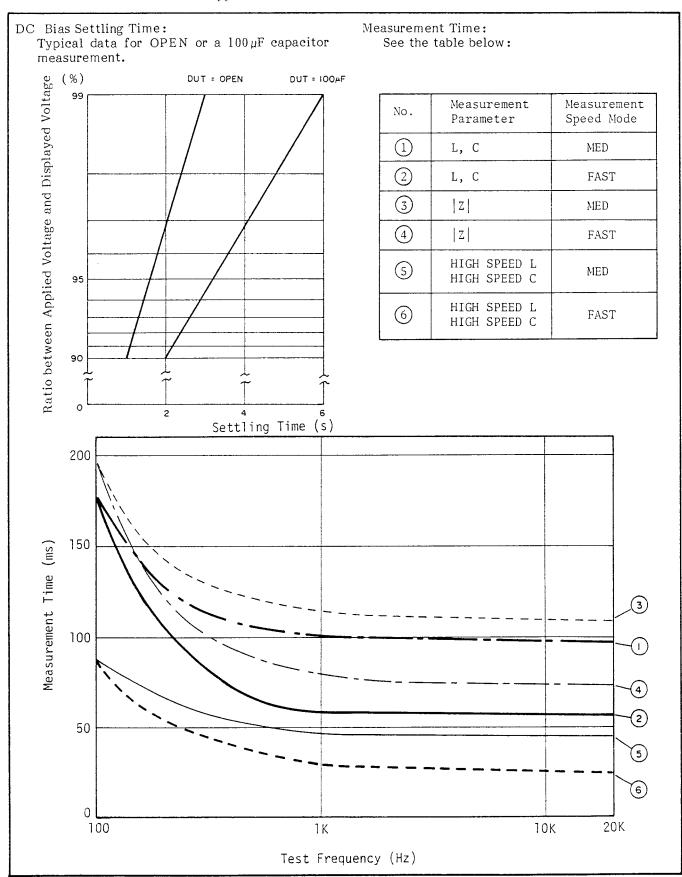


Table 1-3. Accessories Available (Sheet 1 of 3)

# Model Description Test Fixture (direct attachment type) for 16047A measurement of either axial-or radial-lead components. Three kinds of contact inserts are furnished: 1) For axial-lead components, (HP P/N: 16061-70022) (2) For general radial-lead components, (HP P/N: 16061-70021) (3) For radial short-lead components, (HP P/N: 16047-65001) DC bias up to ±40V\* can be applied. Test Fixture (direct attachment type) designed 16047C especially for high frequency measurements requiring high accuracy. Two screw knobs facilitate and ensure optimum contact between the test fixture electrodes and the sample leads. DC bias up to ±40V\* can be applied. Test Fixture (tweezer type) for measurement of 16034B miniature leadless components such as chip Employs a three configuration tweezer probe suitable for high impedance (above $50\Omega$ ) measurements. DC bias up to ±40V\* can be applied. Cable length: 1m

Model 4276A SECTION I

Table 1-3. Accessories Available (Sheet 2 of 3)

Model	Description		
16048A	Test Leads (four terminal pair) with BNC connectors for connecting user-fabricated test fixtures.		
	DC bias up to ±40V* can be applied.  Cable length: 1m		
16048B	Test Leads (four terminal pair) with miniature RF connectors suitable for connecting user-fabricated test fixtures in systems applications.  DC bias up to ±40V* can be applied.  Cable length: lm		
16048C	Test Leads with dual alligator clips for testing components of non-standard shapes and sizes at frequencies below 100kHz.  Applicable measurement ranges:  Capacitance 1000pF Inductance 100 µH  DC bias up to ±40V* can be applied.  Cable length: 1m		

Table 1-3. Accessories Available (Sheet 3 of 3)

# Model Description Double-shielded Test Leads (four terminal pair) 16048D BNC connectors for connecting user-fabricated test fixtures. DC bias up to ±40V\* can be applied. Cable length: 2m Fixture (cable connection type) 16065A measurement of either axial- or radial-lead components at frequencies between 50Hz and 2MHz. Three kinds of contact inserts are furnished (same as those for the 16047A Test Fixture). DC bias up to ±200V can be applied (a protective cover provides for operator safety). Cable length: Approximately 40cm

\* Though " $\pm 35$ V DC MAX" is indicated on the test fixtures, they are capable of handling dc bias voltages up to  $\pm 40$ V when used with the 4276A.

Model 4276A SECTION II

# SECTION II

# 2-1. INTRODUCTION

2-2. This section provides installation instructions for the Model 4276A LCZ Meter. It also includes information on initial inspection and damage claims, preparation for using the 4276A, and packaging, storage, and shipment.

### 2-3. INITIAL INSPECTION

The 4276A LCZ Meter, as shipped from the factory, meets all the specifications listed in Table 1-1. Upon receipt, inspect the shipping container for damage. If the shipping container or cushioning material is damaged, it should be kept until the contents of the shipment have checked for completeness and the instrument has been checked mechanically and electrically. The contents of the shipment should be as shown in Figure 1-1. procedures for checking the general electrical operation are given in Section III (Paragraph 3-5 SELF TEST) and the procedures for checking the 4276A LCZ Meter against its specifications are given in Section IV. First, do the self test. If the 4276A is electrically questionable, then do the Performance Tests to determine whether the 4276A has failed or not.

If the contents are incomplete, if there is mechanical damage or defects (scratches, dents, broken switches, etc.), or if the performance does not meet the self test or performance tests, notify the nearest Hewlett-Packard office (see list at back of this manual). The HP office will arrange for repair or replacement without waiting for claim settlement.

#### 2-5. PREPARATION FOR USE

# 2-6. POWER REQUIREMENTS

2-7. The 4276A requires a power source of 100, 120, 220 Volts ac  $\pm 10\%$ , or 240 Volts ac  $\pm 5\%$ -10%, 48 to 66Hz single phase; power consumption is 65VA maximum.

#### WARNING

IF THE INSTRUMENT IS TO BE ENERGIZED VIA AN EXTERNAL AUTOTRANSFORMER UNIT FOR VOLTAGE REDUCTION, BE SURE THAT THE COMMON TERMINAL IS CONNECTED TO THE NEUTRAL POLE OF THE POWER SUPPLY.

# 2-8. LINE VOLTAGE AND FUSE SELECTION

## CAUTION

BEFORE TURNING THE 4276A LINE SWITCH TO ON, VERIFY THAT THE INSTRUMENT IS SET TO THE VOLTAGE OF THE POWER TO BE SUPPLIED.

2-9. Figure 2-1 provides instructions for line voltage and fuse selection. The line voltage selection switch and the proper fuse are factory installed for the voltage appropriate to instrument destination.

# CAUTION

USE PROPER FUSE FOR LINE VOLTAGE SELECTED.

#### CAUTION

MAKE SURE THAT ONLY FUSES THE REQUIRED RATED FOR CURRENT AND OF THE SPECIFIED TYPE ARE USED FOR REPLACEMENT. THE USE OF MENDED **FUSES** AND THE SHORT-CIRCUITING OF FUSE-HOLDERS MUST BE AVOIDED.

# 2-10. POWER CABLE

2-11. To protect operating personnel, the National Electrical Manufacturer's Association (NEMA) recommends that the instrument panel and cabinet be grounded. The Model 4276A is equipped with a three-conductor power cable which, when plugged into an appropriate receptacle, grounds the instrument. The offset pin on the power cable is the ground wire.

SECTION II Model 4276A

2-12. To preserve the protection feature when operating the instrument from a two contact outlet, use a three prong to two prong adapter (HP Part No. 1251-0048) and connect the green pigtail on the adapter to power line ground.

#### CAUTION

THE MAINS PLUG MUST ONLY BE INSERTED IN A SOCKET OUTLET PROVIDED WITH A PROTECTIVE EARTH CONTACT. THE PROTECTIVE ACTION MUST NOT BE NEGATED BY THE USE OF AN EXTENSION CORD (POWER CABLE) WITHOUT PROTECTIVE CONDUCTOR (GROUNDING).

2-13. Figure 2-2 shows the available power cords, which may be used in various countries including the standard power cord furnished with the instrument. HP Part number, applicable standards for power plug, power cord color, electrical characteristics and countries using each power cord are listed in the figure. If assistance is needed for selecting the correct power cable, contact the nearest Hewlett-Packard office.

#### 2-14. INTERCONNECTIONS

2-15. When an external dc bias source is used, set the DC BIAS select switch on the rear panel to EXT. The output from the external bias source should be connected to EXT INPUT/INT MONITOR connector. The external dc bias fuse is installed in EXT DC BIAS FUSE Holder on rear panel to protect the instrument from excessive current. Fuse rating is as follows:

1/16A, 250V (HP Part No: 2110-0011)

#### CAUTION

MAKE SURE THAT ONLY FUSES OF THE REQUIRED RATED CURRENT AND OF THE SPECIFIED TYPE ARE USED FOR REPLACEMENT. THE USE OF MENDED FUSES AND THE SHORT-CIRCUITING OF FUSE-HOLDERS MUST BE AVOIDED.

#### 2-16. OPERATING ENVIRONMENT

- 2-17. Temperature. The instrument may be operated in temperatures from 0  $^{\circ}$ C to +55  $^{\circ}$ C.
- 2-18. Humidity. The instrument may be operated in environments with relative humidities to 95% at 40°C. However, the instrument must be protected from temperature extremes which cause condensation within the instrument.

#### 2-19. INSTALLATION INSTRUCTIONS

2-20. The HP Model 4276A can be operated on the bench or in a rack mount. The 4276A is ready for bench operation as shipped from the factory. For bench operation a two-leg instrument stand is used. For use, the instrument stands are designed to be pulled towards the front of instrument.

100V/120V OPERATION			
100V ~ ] 120V ~ ] 220V ~ ]	FUSE SELECTION		
	Line Voltage	Fuse Rating	HP Part No.
220V/240V OPERATION	100V/120V	1.0AT, 250V, Slow Blow	2110-0007
100V ~ ] 120V ~ ] 220V ~ ]	220V/240V	0.75AT, 250V, Slow Blow	2110-0360
240V ~			

Figure 2-1. Voltage and Fuse Selection.

Model 4276A SECTION II

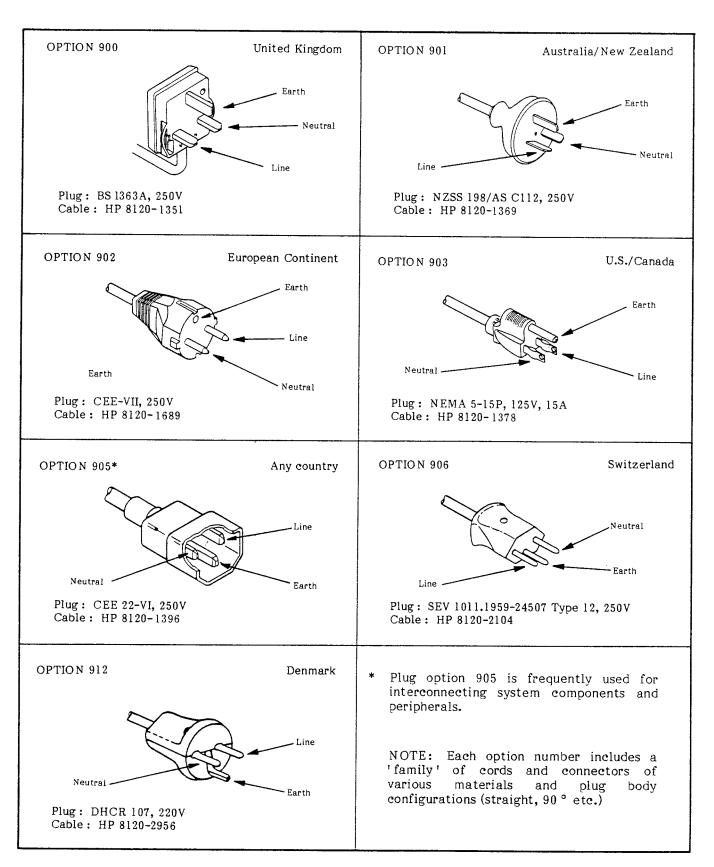


Figure 2-2. Power Cables Supplied.

- 2-21. Installation of Options 907, 908 and 909.
- 2-22. The 4276A can be installed in a rack and be operated as a component of a measurement system. Rack mounting information for the 4276A is presented in Figure 2-3.

# 2-23. STORAGE AND SHIPMENT

#### 2-24. ENVIRONMENT

2-25. The instrument may be stored or shipped in environments within the following limits:

Temperature  $\cdots$  -40 °C to +70 °C Humidity  $\cdots$  to 95% at 40 °C

The instrument must be protected from temperature extremes which cause condensation inside the instrument.

#### 2-26. PACKAGING

- 2-27. Original Packaging. Containers and materials identical to those used in factory packaging are available from Hewlett-Packard. If the instrument is being returned to Hewlett-Packard for servicing, attach a tag indicating the type of service required, return address, model number, and full serial number. Also mark the container FRAGILE to assure careful handling. In any correspondence, refer to the instrument by model number and full serial number.
- 2-28. Other Packaging. The following general instructions should be used for re-packing with commercially available materials:
  - a. Wrap instrument in heavy paper or plastic. If shipping to Hewlett-Packard office or service center, attach tag indicating type of service required, return address, model number, and full serial number.
  - b. Use strong shipping container. A double-wall carton made of 350 pound test material is adequate.
  - c. Use enough shock absorbing material (3 to 4 inch layer) around all sides of instrument to provide firm cushion and prevent movement inside container. Protect control panel with cardboard.
  - d. Seal shipping container securely.
  - e. Mark shipping container FRAGILE to ensure careful handling.

f. In any correspondence, refer to instrument by model number and full serial number.

#### 2-29. OPTION INSTALLATION

2-30. Installation procedures for DC Bias option (Option 001) and Comparator/Handler Interface option (Option 002) are given in Figure 2-4.

# 2-31. POWER FAILURE MONITOR INSTALLATION

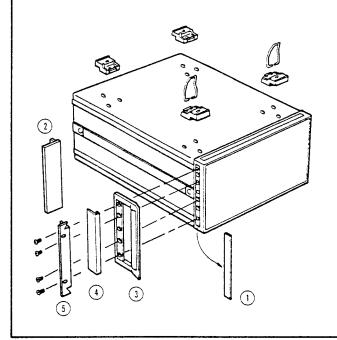
2-32. To use the power failure monitor signal, you must solder two wires to a jumper on the mother board, remove a cap from a hole on the rear panel, and bring the wires out through the hole. The procedure is given below. A simplified drawing of the open collector circuit, a timing diagram, and the locations of the jumper and hole are shown in Figure 2-6. Refer to paragraph 3-114 for a description of the power failure monitor signal.

# Procedure:

- 1. Turn off the 4276A.
- 2. Disconnect the 4276A from the ac power source.
- 3. Remove the top cover.
- 4. Disconnect the brown 4-terminal connector from the A5 board.
- 5. Remove the two screws that secure the A5 board to the chassis.
- 6. Remove the A5 board.
- 7. Solder a wire to each terminal of A6J3. The location of A6J3 is shown in Figure 2-6 (c).
- 8. Remove the cap from the access hole in the rear panel, as shown in Figure 2-6 (d).
- 9. Thread the wires first through the teflon clamp (securing the wires from A6J1) on the A6 board, and then through the access hole in the rear panel.
- 10. Reinstall the A5 board, reconnect the brown 4-terminal connector to the A5 board, and replace the top cover.

Model 4276A SECTION II

Option	Kit Part Number	Parts Included	Part Number	Q'ty	Remarks
907	Handle Kit 5061–0090	Front Handle Trim Strip X8-32 x 3/8 Screw	③ 5060-9900 ④ 5020-8897 2510-0195		9.525mm
908	Rack Flange Kit 5061-0078	Rack Mount Flange X8-32 x 3/8 Screw	② 5020-8863 2510-0193	3 2 8	9.525mm
909	Rack Flange & Handle Kit 5061-0084	Front handle Rack Mount Flange X8-32 x 3/8 Screw	③ 5060-9900 ⑤ 5020-8875 2510-0194	_	15.875mm



- l. Remove adhesive-backed trim strips (1) from side at right and left front of instrument.
- 2. HANDLE INSTALLATION: Attach front handle (3) to sides at right and left front of instrument with screws provided and attach trim (4) to handle.
- 3. RACK MOUNTING: Attach rack mount flange 2 to sides at right and left front of instrument with screws provided.
- 4. HANDLE AND RACK MOUNTING: Attach front handle 3 and rack mount flange 5 together to sides at right and left front of instrument with screws provided.
- 5. When rack mounting (3 and 4 above), remove all four feet (lift bar at inner side of foot, and slide foot toward the bar).

Figure 2-3. Rack Mount Kit.

11. Connect the pull-up resistor and external voltage source as shown in Figure 2-6 (a).

# Note

A +5V is recommended but higher voltage can be used as long as the current through AlT5 and AlQ4 does not exceed  $25\,\mathrm{mA}$ .

CAUTION: BEFORE PROCEEDING WITH INSTALLATION OF OPTION(S),

TURN OFF THE INSTRUMENT AND DISCONNECT THE

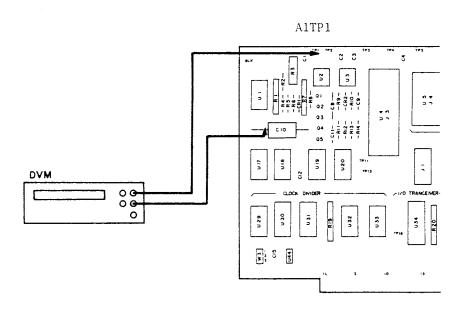
AC POWER CORD.

	OPTION 001 DC BIAS SUPPLY (0 to ±40V)	OPTION 002 COMPARATOR/ HANDLER INTERFACE
Option Parts	Board Assembly A22 04276-66522	Comparator 16064A Includes: Interface Board Assembly 16064-66502 and 36-pin male Amphenol connector 1251-0084
Installation Procedure (after removing top cover)	1. Remove the rear panel access plate shown below.  2. Insert the dc bias board (P/N: 04276-66522) into the access hole.  3. Insert the male edge connector of the interface board into the female edge connector of the 4276A mother board and push firmly until the interface board is completely seated.  4. Reinstall the screws removed in step (1).	<ol> <li>Remove the rear panel access plate shown below.</li> <li>Insert the interface board (P/N: 16064-66502) into the access hole.</li> <li>Insert the male edge connector of the interface board into the female edge connector of the 4276A mother board and push firmly until the interface board is completely seated.</li> <li>Reinstall the screws removed in step 1.</li> <li>Connect the 16064A keyboard cable to the connector on the interface board (installed in step 3).</li> <li>Adjust the power supply in accordance with the procedure given in Figure 2-5.</li> </ol>

Figure 2-4. Option Installation.

Model 4276A SECTION II

- 1. Connect the 4276A to the ac power line.
- 2. Turn on the instrument. ("16064" should be displayed on DISPLAY B.)
- 3. Connect A DVM (HP 3478A is recommended) to AlTPl and GND as shown below.
- 4. Adjust "V-ADJ" on the A4 board until the reading on the DVM is 5.10V±0.02V.



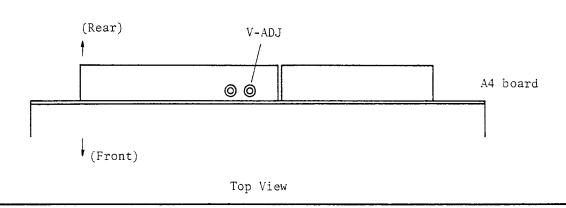


Figure 2-5. Power Supply Adjustment After Installing Option 002.

Model 4276A

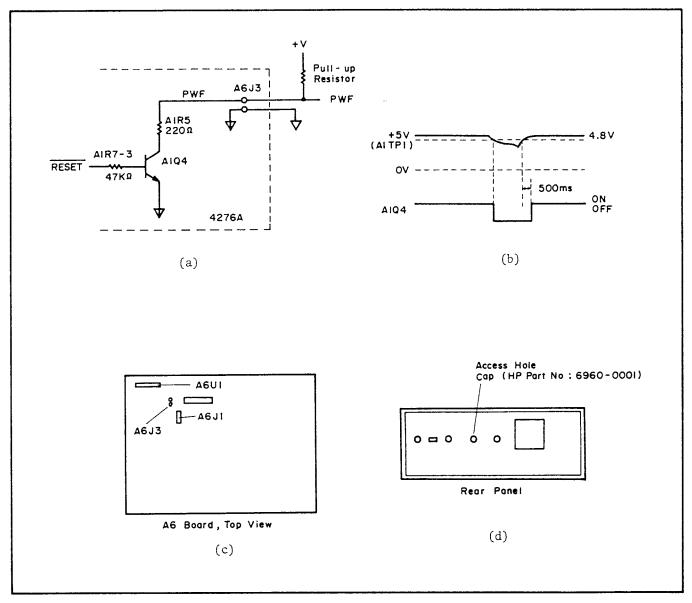


Figure 2-6. Power Failure Monitor Installation.

# SECTION III OPERATION

#### 3-1. INTRODUCTION

3-2. This section provides all the information necessary to operate the Model 4276A LCZ Meter. Included are descriptions of the front-and rear-panels, displays, lamps and connectors; discussions on operating procedures and measuring techniques for various applications; and instructions on the instrument's SELF TEST function. Warnings, Cautions, and Notes are given throughout; they should be observed to insure the safety of the operator and the serviceability of the instrument.

#### WARNING

BEFORE THE INSTRUMENT SWITCHED ON, ALL PROTECTIVE EARTH TERMINALS, EXTENSION CORDS, AUTO-TRANSFORMERS AND DEVICES CONNECTED TO SHOULD BE CONNECTED TO GROUNDED **PROTECTIVE** EARTH SOCKET. ANY INTERRUPTION OF PROTECTIVE EARTH GROUNDING WILL CAUSE POTENTIAL SHOCK HAZARD THAT COULD RESULT **SERIOUS** IN PERSONAL INJURY.

ONLY FUSES WITH THE REQUIRED RATED CURRENT AND OF THE SPECIFIED TYPE SHOULD BE USED. DO NOT USE REPAIRED FUSES OR SHORTED FUSEHOLDERS. TO DO SO COULD CAUSE A SHOCK OR FIRE HAZARD.

# CAUTION

BEFORE THE INSTRUMENT IS SWITCHED ON, IT MUST BE SET TO THE VOLTAGE OF THE POWER SOURCE (MAINS), OR DAMAGE TO THE INSTRUMENT MAY RESULT.

# 3-3. PANEL FEATURES

3-4. Figures 3-1 and 3-2 identify and briefly describe the purpose of each key, indicator, and connector on the front panel and rear panel, respectively. More detailed information on front panel displays and controls is given starting in paragraph 3-5.

#### 3-5. SELF TEST

3-6. The self test function confirms correct operation of the instrument's basic functions and facilitates troubleshooting. It consists of three parts: (1) ROM/RAM Test, (2) Display Test, and (3) Analog Circuit Test. Each is described in paragraphs 3-7 through 3-11.

#### 3-7. ROM/RAM Test

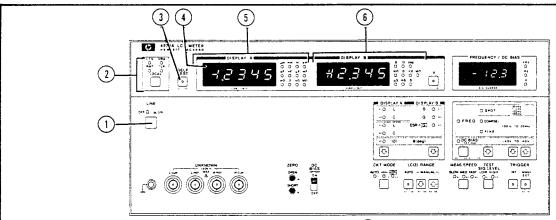
3-8. The ROM/RAM Test is performed each time the instrument is turned on. During this test, all ROMs and RAMs in the instrument's digital control section are tested using a check-sum test and a read/write test (RAMs only). If a malfunction is detected, the test will stop and an error-code will be displayed on DISPLAY A. If the ROMs and RAMs are functioning properly, the instrument will display the HP-IB address (or output data format if the HP-IB control switch is set to TALK ONLY) on DISPLAY A and the option annunciations on DISPLAY B and the FREQUENCY/DC BIAS DISPLAY. Error-codes are described paragraph 3-20.

#### Note

If a ROM/RAM test error code, E61 through E68, appears on DISPLAY A when the instrument is turned on, contact the nearest Hewlett-Packard Sales or Service Office for repairs.

# Note

ROM/RAM code E68 test error indicates that instrument's the continuous memory feature is not functioning properly. All other instrument functions. including measurement, are not affected.



# (1) LINE OFF/ON Switch:

Applied ac line power to the instrument when set to the ON ( in) position. Removes ac line power when set to the OFF ( in) out) position.

(2) HP-IB Status Indicators and LOCAL Key:

These four LED lamps—SRQ, LISTEN, TALK, and REMOTE—indicate the status of the 4276A when it is interfaced with a controller via the HP-IB.

The LOCAL key, when pressed, releases the instrument from REMOTE (HP-IB) control and enables front panel control. The LOCAL key is disabled (does not function) when the instrument is set to "local lockout" by the controller.

# (3) SELF TEST Key and Indicator:

This key initiates the instrument's SELF TEST function. During SELF TEST (when the indicator is on), nine tests that check the basic operation of the instrument are automatically performed. SELF TEST is repeated until this key is pressed again. If a fault is detected, an error code will be displayed on DISPLAY A (§). A complete description of the SELF TEST function is given in paragraph 3-5; error codes are described in paragraph 3-20.

# (4) Trigger Lamp:

Comes on each time the instrument is internally, externally, or manually triggered. Trigger mode is set by the TRIGGER keys 1.

# (5) DISPLAY A:

Displays measured values of inductance, capacitance, or impedance magnitude with a maximum 4-1/2 digits; maximum display is 19999. Number of display digits depends on instrument control settings. The nine LED lamps located to the right of the display are the engineering unit indicators for displayed values. Measurement error messages—OF, UF, CF—operation error codes, SELF TEST error codes, and the instrument's HP-IB address are also displayed on this display.

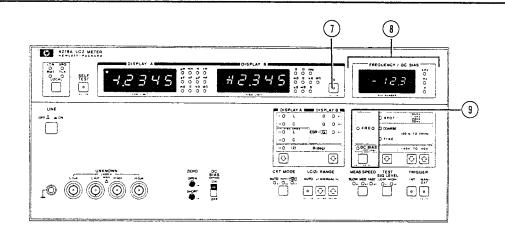
If the instrument is equipped with Option 002, Comparator/Handler Interface, the LOW LIMITS keyed in from the 16064A will be displayed on this display when the 16064A is set to ENABLE and RUN is off.

Decimal point location and engineering unit indicator lamp change when the LC  $\mid Z \mid$  RANGE (6) changes.

# (6) DISPLAYD B:

Displays measured values of dissipation factor, quality factor, equivalent series resistance, conductance or impedance phase angle with a maximum 4-1/2 digits; maximum display is 10000 for quality factor, 18000 for phase, and 19999 for all other parameters. Number of display digits depends on instrument control settings. The nine LED lamps located to the right of the display are the engineering unit indicators for displayed values. Measurement error messages-OF, UF, and CF-are also displayed on this display. When the DISPLAY A Function (18) is set to HIGH SPEED L or HIGH SPEED C, or when an error code is displayed on DISPLAY A (5), this display is blanked (turned off) by the microprocessor.

Figure 3-1. Front Panel Features (Sheet 1 of 6).



If the instrument is equipped with Option 002, Comparator/Handler Interface, and if the 16064A comparator is connected, the number 16064 will be displayed on this display when the instrument is turned on. Also, the HIGH LIMITS keyed in from the 16064A will be displayed on this display when the 16064A is set to ENABLE and RUN is off.

# $\bigcirc$ $\triangle$ Key and Indicator:

This key enables deviation ( $\triangle$ ) measurements on both displays. When this key is pressed, the values displayed on DISPLAY A ( $\Im$ ) and DISPLAY B ( $\Im$ ) are stored as reference values. The difference between values obtained in subsequent measurements and the stored reference values is calculated and displayed on each display. The formula used to calculate the deviation is

#### A-B

Where A is the measured value of the device under test and B is the stored reference value.

LC  $\mid$  Z  $\mid$  RANGE  $\bigcirc$  is set to MANUAL when this key is pressed.

Also, the deviation measurement function is turned off by pressing this key again, or by changing the DISPLAY A function (B), DISPLAY B function (S), LC | Z | RANGE (B), or CIRCUIT MODE (T). It may be turned off also if the test frequency is changed when the DISPLAY B function is ESR/G.

(8) FREQUENCY/DC BIAS Display:

Displays test frequency or DC bias voltage (Option 001 only) with 3 digits. The three LED lamps located to the right of the display are unit indicators for displayed values. On instruments equipped with Option 002, Comparator/Handler Interface, bin numbers are displayed on this display when the comparator is set to RUN. Also, on Option 001 instruments, the number 001 is briefly displayed here when the instrument is turned on.

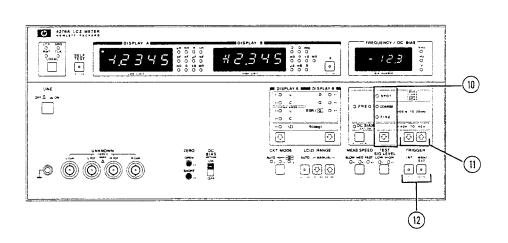
# FREQUENCY/DC BIAS Select Key and Indicators: Indicators:

This key sets the FREQUENCY/DC BIAS Display (3), the SPOT/COARSE/FINE Select key (10), and the FREQUENCY/DC BIAS Step Control Keys (11) to FREQUENCY control mode or DC BIAS control mode. The selected control mode is indicated by the corresponding LED lamp.

FREQ: When this LED lamp is on, frequency is displayed on the FREQUENCY/DC BIAS Display and is controlled by the SPOT/COARSE/FINE Key and the FREQUENCY/DC BIAS Step Control Keys.

DC BIAS: When this LED lamp is on, DC bias voltage is displayed on the FREQUENCY/DC BIAS Display and is controlled by the FREQUENCY/DC BIAS Step Control Keys.

FREQUENCY control mode and DC BIAS control mode are mutually exclusive, and DC BIAS can be selected only if the instrument is equipped with Option 001.



SPOT/COARSE/FINE Select Key and Indicators:

This key selects the SPOT, COARSE, or FINE vernier mode for frequency changes mode by the FREQUENCY/DC BIAS Step Control Keys (1). The selected vernier mode is indicated by the corresponding LED lamp. Frequencies possible in each vernier mode are listed below

SPOT: 100Hz, 120Hz, 1kHz, 10kHz

COARSE: 100Hz to 1kHz in 100Hz steps lkHz to 10kHz in 1kHz steps

10kHz to 20kHz in 10kHz steps

FINE: 100Hz to 200Hz in 1Hz steps

200Hz to 500Hz in 2Hz steps 500Hz to 1kHz in 5Hz steps 1kHz to 2kHz in 10Hz steps 2kHz to 5kHz in 20Hz steps 5kHz to 10kHz in 50Hz steps 10kHz to 20kHz in 100Hz steps

Note

When the FREQUENCY/DC BIAS Select Key (3) is set to DC BIAS mode, this key is disabled and the SPOT, COARSE, and FINE indicators are turned off.

(11) FREQUENCY/DC BIAS Step Control Keys:

These keys— ② and ③ —are used in conjunction with the FREQUENCY/DC BIAS Select Key ③ and the SPOT/COARSE/FINE Select Key ⑩ to set the test frequency and DC bias voltage (Option 001 instruments only). When

FREQUENCY mode is selected by the FREQUENCY/DC BIAS Select Key (9), test frequency is increased in accordance with the selected vernier mode (SPOT, COARSE, FINE) each time the [3] is pressed, and is decreased each time the 🖸 key is pressed. These keys control DC bias in a similar manner when DC BIAS mode is selected by the FREQUENCY/DC BIAS Select Key 3. When either of these keys is pressed and displayed on the held. the value FREQUENCY/DC BIAS Display will continuously change in the indicated direction. The actual value, however, will not change until the key is released.

# (12) TRIGGER Keys:

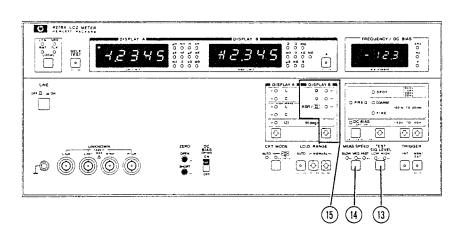
These keys select the trigger mode for triggering measurement (Internal or Manual/External):

INT: Internal trigger signal enables instrument to make repeated automatic measurements.

MAN/EXT: Measurement is triggered key is each time this pressed. Measurement data is held until the next time the key is pressed. Or in this mode measurement triggered by an external trigger signal applied to the rear panel EXT TRIGGER connector (4) in Figure 3-2).

Figure 3-1. Front Panel Features (Sheet 3 of 6).

Model 4276A SECTION III



TEST SIGNAL LEVEL Selector Key and Indicators:

This key selects two test signal levels: HIGH and LOW. HIGH level is 1Vrms and LOW level is 50mVrms. The selected test signal level is indicated by the corresponding LED lamp.

#### Note

The 4276A cannot measure L, HIGH SPEED L, or Z when TEST SIGNAL LEVEL is set to LOW.

(4) MEASUREMENT SPEED Select Key and Indicators:

This key selects three measurement speeds: SLOW, MEDIUM or FAST. Actual measurement speed depends on test frequency, LC | Z | range (§), DISPLAY A Function (§), and the value of the device under test. The selected measurement speed mode is indicated by the corresponding LED lamp.

SLOW: Measurement speed is approximately 1/4 that of medium measurement speed.

MED: Measurement speed is approximately 11 measurements per second in C-G measurement mode.

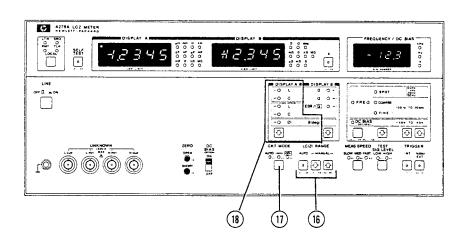
FAST: Measurement speed is approximately twice that of medium measurement speed.

(5) DISPLAY B Function Select Key and Indicators:

This key, ③, selects the measurement parameter for display on DISPLAY B ⑤. The selected parameter is indicated by the corresponding LED lamp. Pressing this key shifts the selected parameter in a top-to-bottom sequence. Selectable parameters are as follows:

- D: Measures the dissipation factor of the DUT. DISPLAY A Function

  (B) must be set to L (inductance) or C (capacitance).
- Q: Measures the quality factor of the DUT. DISPLAY A Function (8) must be set to L (inductance) or C (capacitance). Q values are calculated as the reciprocal dissipation factor.
- ESR/G: Measures the equivalent series resistance or conductance of the DUT. DISPLAY A Function ® must be set to L (inductance) or C (capacitance). ESR is selected when CIRCUIT MODE ① is set to CIRCUIT MODE ① is set to CIRCUIT MODE ① is set to



(b) LC | Z | RANGE Selector Keys and Indicator:

These keys select the measurement range and the ranging method for inductance, capacitance and impedance measurements.

AUTO (when indicator is lit):

Optimum range for the DUT's value is automatically selected.

MANUAL (when indicator is not lit):

Measurement range is fixed (even when the DUT is changed). Manual ranging is done by pressing the adjacent DOWN (  $\odot$  ) or UP (  $\odot$  ) key.

# Note

Pressing the DOWN or UP key sets the ranging mode to MANUAL even if the ranging mode was initially set to AUTO.

(17) CIRCUIT MODE Select Key and Indicators:

This key selects the measurement circuit mode to be used during measurement. The selected circuit mode is indicated by the corresponding LED lamp.

AUTO:

Automatically selects the equivalent circuit (parallel or series) most appropriate for the DUT's value. When  $LC \mid Z \mid$  RANGE (a) is set to the  $100\Omega$  range or lower, circuit mode is set to  $\square$  When  $LC \mid Z \mid$  RANGE (b) is set to the  $1k\Omega$  range or higher, circuit mode is set to  $\square$ .

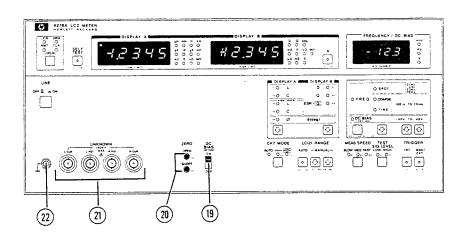
•— \*: Selects equivalent series circuit.

Selects equivalent parallel circuit.

(8) DISPLAY A Function Select Key and Indicators:

This key, ③, selects the measurement parameter for display on DISPLAY A ⑤. The selected parameter is indicated by the corresponding LED lamp. Pressing this key shifts the selected parameter in a top-to-bottom sequence. The selectable parameters are as follows:

- L: Measures inductance and—depending on the setting of DISPLAY B Function (5)—D (dissipation factor), Q (quality factor), or ESR/G (equivalent series resistance or equivalent parallel conductance).
- C: Measures capacitance and—depending on the setting of DISPLAY B Function (5)—D (Dissipation factor), Q (quality factor), or ESR/G (equivalent series resistance or equivalent parallel conductance).



HIGH SPEED L:
Measures only inductance.

HIGH SPEED C:
Measures only capacitance.

## $|Z| - \theta (\text{deg})$ :

Measures impedance magnitude and phase angle. The results are displayed on DISPLAY A (|Z|) and DISPLAY B ( $\theta$ ) to provide a polar representation ( $|Z| \perp \theta$ ) of the DUT's impedance.

# (19) DC BIAS SWITCH:

On instruments equipped with Option 001, DC BIAS, this switch turns the internal DC bias source on and off. When this switch is set to ON and the DC BIAS Select Switch (1) in Figure 3-2) on the rear panel is set to INT, the DC voltage selected by the FREQUENCY/DC BIAS Step Control Keys (n) is output from the H CUR UNKNOWN terminal (1). When set to OFF, this switch turns off the internal DC bias source; no DC voltage is output from the H CUR UNKNOWN terminal (1), and "OFF" will be briefly displayed on the FREQUENCY/DC BIAS Display 8 each time a new DC bias voltage is set by the FREQUENCY/DC BIAS Step Control Keys 11 or via the HP-IB.

#### Note

This switch controls the internal DC bias source only. It does not control external DC bias voltage applied to the EXT INPUT/INT MONITOR connector (3 in Figure 3-2) on the rear panel. Also, this switch is not HP-IB programmable.

# (20) ZERO Offset:

These buttons perform ZERO offset compensation (OPEN and SHORT) for the residuals of the test fixture, test leads, and measurement circuit. ZERO offset is performed at the following spot frequencies:

20kHz, 16kHz, 10kHz, 5kHz, 2kHz, 1kHz, 500Hz, 200Hz, and 100Hz.

OPEN: If this button is pressed when the test fixture or test leads are terminated open, measured values at this time are stored as residual admittance data.

SHORT: If this button is pressed when the test fixture or test leads are shorted, measured values at this time are stored as residual impedance data.

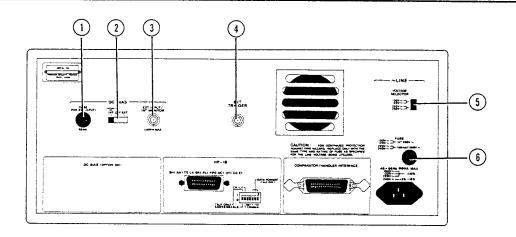
## (1) UNKNOWN Terminals:

These four BNC connectors provide the means to connect DUT's in a five-terminal configuration: High current terminal  $(H_{\text{CUR}})$ , High potential terminal  $(H_{\text{POT}})$ , Low potential terminal  $(L_{\text{POT}})$ , and Low current terminal  $(L_{\text{CUR}})$ . Four-terminal pair test fixtures attach directly to these terminals.

# (22) GUARD Terminal:

This terminal is tied to the instrument's chassis and can be used in measurements that require guarding.

Figure 3-1. Front Panel Features (Sheet 6 of 6).



(1) EXT DC BIAS FUSE Holder:

External DC bias fuse is installed in this holder. The fuse must be installed when an external bias source is used. Fuse rating is 1/16A, 250V (HP P/N: 2110-0011).

# (2) DC BIAS Select Switch:

This switch selects the DC bias source that will be used for biasing DUTs connected to the UNKNOWN terminals.

INT: On instruments equipped with Option 001, DC BIAS, the DC voltage output from the internal DC bias source will be applied to the DUT when the DC BIAS Switch (19) in Figure 3-2) is set to ON.

OFF: No DC bias voltage will be applied to the DUT.

EXT: DC voltage provided by an external voltage source connected to the EXT INPUT/INT MONITOR Connector ③ will be applied to the DUT regardless of the setting of the DC BIAS Switch (③ in Figure 3-2). Maximum allowable voltage is ±40V.

# (3) EXT INPUT/INT MONITOR Connector:

The function of this connector depends on the setting of the DC BIAS Select Switch ②. When the DC BIAS Select Switch ② is set to EXT, this connector is the input terminal for an external DC voltage source. When the DC BIAS Select Switch ③ is set to INT, this connector is the monitor output terminal for the internal DC bias source (Option 001 instruments only).

# (4) EXT TRIGGER Connector:

This connector is for external trigger input. TRIGGER key on front panel should be set to MAN/EXT. Specific information is provided in paragraph 3-70.

(5) ~ LINE VOLTAGE SELECTOR Switch:

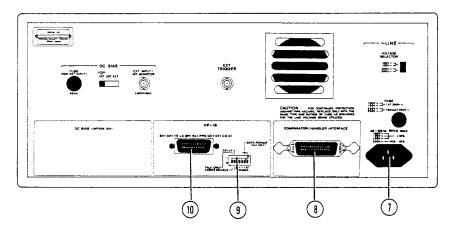
This switch selects the appropriate ac operating voltage. Selectable voltages are  $100V/120V\pm10\%$  and  $220V\pm10\%/240V\pm5\%-10\%$  (48 - 66Hz).

(6) ~ LINE FUSE Holder:

Instrument's power-line fuse is installed in this holder.

100V/120V operation: 1AT, 250V (HP P/N: 2110-0007)

220V/240V operation: 750mAT, 250V (HP P/N: 2110-0360)



↑ LINE Input Receptacle:

AC power cord connects to this receptacle.

(8) COMPARATOR/HANDLER INTERFACE Connector:

Thirty-six pin connector, Option 002 instruments only, connects the 16064A COMPARATOR/HANDLER INTERFACE to the instrument.

(9) HP-IB Control Switch:

This switch sets the instrument's HP-IB address (0 - 30), data output format (F1 - F6), and interface capability (Addressable or Talk Only). Specific information on this switch is given in paragraph 3-82.

(10) HP-IB Connector:

Twenty-four pin connector; connects the instrument to an HP-IB controller or other HP-IB instruments via an HP-IB cable.

Figure 3-2. Rear Panel Features (Sheet 2 of 2).

### 3-9. DISPLAY TEST

3-10. All LED lamps and 7-segment displays on the front panel are lit for approximately one second when the instrument's self-test function is initiated from the front panel or via the HP-IB. This test is repeated until the self-test function is turned off.

#### Note

If an LED lamp or 7-segment display fails to light during the Display test, contact the nearest Hewlett-Packard Sales or Service Office for repairs.

#### Note

If the instrument is equipped with Option 002, Comparator/Handler Interface, and if the 16064A Comparator/Handler Interface is connected to the instrument, all 16064A LED lamps except D/Q/ESR/G and LIMIT LOW lamps will be lit during the Display test.

# 3-11. ANALOG CIRCUIT TEST

3-12. The Analog Circuit test is performed when the instrument's self-test function is initiated from the front panel or via the HP-IB. It is performed after the Display test, described in paragraph 3-9, and it confirms correct operation of the instruments analog circuits. Like the Display test, this test is repeated until the self-test function is turned off. The test lasts approximately three seconds. If a malfunction is detected, an error-code will be displayed on DISPLAY A. Refer to Table 3-4.

#### Note

The Analog Circuit test must be performed with an open-terminated (no DUT) test fixture (e.g., 16047A) connected to the UNKNOWN terminals.

#### Note

If one or more of the error codes listed in Table 3-4 appear on DISPLAY A during the Analog Circuit test, contact the nearest Hewlett-Packard Sales or Service Office for repairs.

# 3-13. MEASUREMENT FUNCTIONS

3-14. Values displayed on DISPLAY A and DISPLAY B are for the parameters selected by the DISPLAY A and DISPLAY B function keys.

Inductance (L), capacitance (C), or impedance magnitude (|Z|) values are displayed on DISPLAY A; dissipation factor (D), quality factor (Q), equivalent series resistance (ESR), conductance (G), or impedance phase ( $\theta$ ) values are displayed on DISPLAY B. The DISPLAY B measurement function depends on the selected DISPLAY A function and the selected CKT MODE, as listed in Table 3-1. When DISPLAY A function is HIGH SPEED C or HIGH SPEED L, DISPLAY B is always blank.

Table 3-1. Measurement Functions

	DISPLAY B		
D. C.	Circuit Mode		
DISPLAY A	• <b>□</b> •••	<b>₹₽</b>	
L	D, Q, or ESR	D, Q, or G	
С	D, Q, or ESR	D, Q, or G	
HIGH SPEED L			
HIGH SPEED C			
Z	θ	θ	

## 3-15. DISPLAYS

3-16. The 4276A has three front panel displays: DISPLAY A, DISPLAY B, and FREQUENCY/DC BIAS. Each is described in paragraphs 3-17 through 3-19, respectively. The number of display digits depends on measurement range, test frequency, and test signal level. Refer to Figure 3-39.

3-17. DISPLAY A provides direct readout of measured C, L, or |Z|, with 41/2-digit display resolution. The actual number of display digits depends on measurement range, test frequency, and test signal level. The least significant digit may be displayed as a small zero,  $\Box$ , or may be blank,  $\Box$ , to indicate that the digit does not provide a specified value. Maximum number of counts is  $\pm 19999$ . DISPLAY A also displays error-codes, operational annunciations, and the HP-IB address or output data format (refer to paragraphs 3-72 through 104 for details).

3-18. DISPLAY B provides direct readout of measured D, Q, ESR, G, or  $\theta$ , with 4 1/2-digit display resolution. The actual number of display digits depends on measurement range, test frequency, test signal level, and number of DISPLAY A counts. The least significant digit may be displayed as a small zero, 🗖 , or may be blank,  $\blacksquare$  , to indicate that the digit does not provide a specified value. Maximum number of display counts depends on the DISPLAY B function. Refer to Table 3-2. DISPLAY B also displays error-codes, operational annunciations, and option annunciation "16064" when the instrument is equipped with option 002. When the DISPLAY A function is HIGH SPEED C or HIGH SPEED L, DISPLAY B is blank.

#### Note

Option annunciation "16064" appears only when the 16064A Comparator is connected to the rear panel.

Table 3-2. Number of Counts on DISPLAY B

Measurement Function	Display Counts	
D	Max. 1.9999	
Q	Мах. 10 о о о	
ESR/G	- 19999 to 19999 counts	
Ө	- 180.00° to 180.00°	

3-19. The FREQUENCY/DC BIAS display provides direct readout of test frequency and, if the instrument is equipped with option 001, the voltage output from the internal dc bias source. If option 001 is installed, option annunciation 001 is displayed on this display each time the instrument is turned on. If the DC BIAS ON/OFF switch is set to OFF when the dc bias voltage is changed, OFF will be briefly displayed on this display after the new value has been set. Refer to paragraph 3-24. Also, if the instrument is equipped with option 002, BIN numbers are displayed on this display when the 16064A Comparator is enabled.

#### 3-20. ERROR-CODES

3-21. Error-codes related to the ROM/RAM test (see paragraph 3-7) are listed in Table 3-3. If one of these errors is displayed on DISPLAY A when the instrument is turned on, measurements can not be made.

#### Note

If E68 is displayed, measurements can be made. The instrument's continuous memory function, however, is disabled.

3-22. Error-codes related to the Analog Circuit test (see paragraph 3-11) are listed in Table 3-4. If one or more of these errors are displayed on DISPLAY A during Self Test, the specifications listed in Table 1-1 are not guaranteed.

#### Note

If one of the error-codes listed in Table 3-3 or Table 3-4 is displayed, contact the nearest Hewlett-Packard Sales or Service Office for repairs.

3-23. Error-codes related to operator errors are listed in Table 3-5. Corrective action for each error is also given in the table.

# 3-24. OPERATIONAL ANNUNCIATION

3-25. On instruments equipped with option 001, DC BIAS, the annunciation shown in Table 3-6 may briefly appear on the FREQUENCY/DC BIAS display after a new dc bias voltage has been set. It indicates that the DC BIAS ON/OFF switch on the front panel is set to OFF. This switch must be set to ON if voltage from the internal dc bias source is to be applied to the DUT.

# Note

For applications using the internal de bias source, the DC BIAS select switch on the rear panel must be set to INT.

Table 3-3. Error-Codes for ROM/RAM Self Test

Error Code	Meaning
<b>E 5 7</b>	AlU5 ROM is faulty.
888	AlU6 ROM is faulty.
EEE	AlU7 ROM is faulty.
TEE E	AlU8 ROM is faulty.
855 5	A1U9 ROM is faulty.
888	AlUlO ROM is faulty.
EBT	AlUl2 RAM is faulty.
88	A1U12 RAM or A6BT1 is faulty.

Table 3-4. Error-Codes for Analog Circuit Self Test

Display	Meaning	
	Analog Circuit is not functioning properly.	

Table 3-6. Operation Error Codes Displayed on FREQUENCY/DC BIAS Display

DISPLAY A	DISPLAY B	FREQ/DC BIAS	Meanings	Treatment
(any reading)	(any reading)	PREQUENCY / DC BAS US O O O O O O O O O O O O O O O O O O	Illegal INTERNAL DC BIAS operation (Option 001). The internal dc bias voltage was set manually or via the HP-IB when the DC BIAS ON/OFF switch on the front panel was set to OFF.	Set the DC BIAS switch to ON.  Note  Make sure that the DC BIAS switch on the rear panel is set to INT.

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Table 3-5. Operation Error Codes Displayed on DISPLAY A/B (Sheet 1 of 3)

ERROR CODE	Meaning	Treatment
DF COCCO	Overflow - The inductance, capacitance, or impedance of the DUT is too high to be measured on the selected LC  Z   RANGE.	Select a higher LC Z  RANGE.
(any reading)	Overflow - The dissipation factor, quality factor, ESR, or conductance of the DUT is too high.	Change the DISPLAY B function, or change the DISPLAY A function to  Z .
CONTRACT A CONTRACT OF THE CON	Underflow -The inductance, capacitance, or impedance of the DUT is too low to be measured on the selected LC   Z   RANGE.	Select a lower LC Z  RANGE.
(any reading)	Underflow -The dissipation factor, quality factor, ESR, or conductance of the DUT is too low.	Change the DISPLAY B function, or change the DISPLAY A function to  Z .
1018PLAY A	Change Function -The selected parameter cannot be measured with	Change the DISPLAY A function to another parameter.
(any reading)	the present control settings.	Change the DISPLAY B function, or change the DISPLAY A function to  Z .
6 1 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0	Zero Offset Adjustment error. The residuals of the test fixture or test leads are too high to be offset, or nothing is connected to the UNKNOWN terminals. Previous Zero Offset data are unchanged.	Use a different test fixture or test leads; or, if nothing is connected to the UNKNOWN terminals, connect an appropriate test fixture or test leads. Refer to paragraph 3-48 for details on Zero Offset Adjustments.

Table 3-5. Operation Error Codes Displayed on DISPLAY A/B (Sheet 2 of 3)

ERROR CODE	Meaning	Treatment
E / / F O O O O O O O O O O O O O O O O O	TEST SIG LEVEL was set to LOW when DISPLAY A function is set to L, HIGH SPEED L, or $ z $ .	LOW TEST SIG LEVEL can be used only for C and HIGH SPEED C measurements.
6 1 8 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7 7	Illegal LC Z  RANGE, DISPLAY A, FREQ, or TEST SIG LEVEL setting.	The instrument will automatically select the correct setting.
E / 3 To	Illegal DC BIAS or COMPARATOR operation. Internal dc bias voltage was set via the HP-IB, but the instrument is not equipped with Option 001; or the comparator enable code (E1) was sent via the HP-IB; but the instrument is not equipped with Option 002.	Install the desired option. Refer to Section II.
018PLAY A 00 00 00 00 00 00 00 00 00 00 00 00 0	Illegal COMPARATOR operation. The D/Q/ESR/G key on the 16064A was pressed or was set via the HP-IB while the DISPLAY A function was set to HIGH SPEED C, HIGH SPEED L, or  Z .	D, Q, ESR, or G comparison cannot be performed. The instrument is set to HIGH SPEED L or HIGH SPEED C measurement mode.
018PLAY A 00000 0000 00000 00000 00000 00000 0000	Illegal COMPARATOR operation. One of the 4276A's front panel keys (except TRIGGER, LOCAL, or DC BIAS) was pressed or was set via the HP-IB.	To change a front panel control setting on the 4276A, first disable (turn off) the 16064A. Press the COMPARATOR ENABLE key (the lamp at the center of the key should go off).
	Illegal COMPARATOR operation. One of the 16064A's keys (except the COMPARATOR ENABLE key) was pressed or was set via HP-IB while the 16064A was disabled.	To operate the COMPARATOR, first enable (turn on) the 16064A. Press the COMPARATOR ENABLE key (the lamp at the center of the key should come on).

Table 3-5. Operation Error Codes Displayed on DISPLAY A/B (Sheet 3 of 3)

ERROR CODE	Meaning	Treatment
DISPLAY A CONTROL OF THE PLAY B CONTROL OF T	Illegal COMPARATOR operation. The 4276A's front panel control settings are different from those that existed when the present bin limits were entered.	Reset the front panel controls to the previous settings, or clear the stored bin limits by pressing the ERASE button.
5 0.15 PLAY A	Illegal COMPARATOR operation. The RUN key on the 16064A was pressed or was set via HP-IB when no bin limits were entered, or a bin's LOW LIMIT is higher than its HIGH LIMIT.	Enter LOW and HIGH limits, or correct the displayed LOW and HIGH LIMITs.
DIPLAY A DO NO	Illegal parameter setting. The test frequency setting, internal dc bias setting, or a bin limit setting is out- side the specified limits.	Reset the incorrect parameter.
018 PLAY A 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0	Illegal HP-IB address. The HP-IB address switches on the rear panel were set to 31 (11111) when the instrument was turned on.	Turn off the instrument and set the HP-IB address to one between 0 (00000) and 30 (11110).
DIBPLAY A DISPLAY B DISPLA	Illegal deviation measurement operation. The \( \Delta \) key on the front panel was pressed or was set via HP-IB when \( \Omega F \), \( \Omega F \) was displayed on DISPLAY A or DISPLAY B.	Only valid reference values can be used for deviation measurement.

# 3-26. TEST FREQUENCY

3-27. There are seven test frequency ranges, as listed in Table 3-7. Frequency accuracy is 0.01% of the value displayed on the FREQUENCY/DC BIAS display.

Table 3-7. Test Frequency

Test Frequency	Resolution
100Hz200Hz	1Hz
200Hz - 500Hz	2Hz
500kHz - 1.00kHz	5Hz
1.00kHz - 2.00kHz	10Hz
2.00kHz - 5.00kHz	20Hz
5.00kHz - 10.0kHz	50Hz
10.0kHz - 20.0kHz	100Hz

# 3-28. TEST SIGNAL LEVEL

3-29. The 4276A has two test signal levels: HIGH (1Vrms) and LOW (50mVrms). The output impedance of the test signal source is  $100\Omega\pm20\%$ , so the voltage across the DUT depends on the DUT's impedance. Refer to Figure 3-3.

#### Note

On several ranges, HIGH test signal level is 2Vrms. Refer to Table 1-1.

# Note

Low test signal level can be used only when DISPLAY A function is set to C or HIGH SPEED C.

#### 3-30. MEASUREMENT RANGE

3-31. Measurement range depends on the test frequency. The ranges which can be selected at each test frequency and the range resistor used on each range are shown in Figure 3-16. Each range allows a 100% overrange of the 10000 full scale counts (maximum 19999 counts).

Measurement range is selected by the LC  $\mid$  Z  $\mid$  RANGE keys. When the LC  $\mid$  Z  $\mid$  RANGE control is set to AUTO, the optimum range is automatically selected for each measurement. Manual ranging is also possible. When an inappropriate range is selected, OF or UF is displayed on DISPLAY A or DISPLAY B.

#### 3-32. CIRCUIT MODE

3-33. An impedance can be represented by a simple series or parallel equivalent circuit consisting of resistive and reactive elements. This is possible because both equivalent circuits have identical impedances at a given test frequency by properly establishing the values of the equivalent circuit elements. The equivalent circuit measurement mode is selected by setting the CIRCUIT MODE control. When the CIRCUIT MODE is set to AUTO, the 4276A will automatically select the circuit mode most appropriate for the range and function settings. Equivalent series circuit mode is automatically selected when the measurement range is inside the area enclosed in the dotted line in Figure Equivalent parallel circuit mode is automatically selected when the measurement range is outside the area enclosed in the dotted line in Figure 3-16. By setting CIRCUIT MODE manually, either circuit mode can be selected, regardless all measurement ranges.

3-34. Capacitance and inductance measurements can be performed in either equivalent series circuit mode or equivalent parallel circuit mode. However, measured values obtained in each mode are different. The difference in measured values is related to the loss factor of the sample being measured. The impedance of a sample measured in both series and parallel circuit mode is the same at a particular frequency. Therefore, the following equations are satisfied:

$$jX R jB G$$

$$\bullet \square W \bullet \qquad \bullet \square \square \bullet$$

$$Z = R + jX Y = \frac{1}{Z} = G + jB$$

$$G + jB = \frac{1}{R + jX}$$

$$= \frac{R}{R^2 + X^2} - j \frac{X}{R^2 + X^2}$$

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Expanding the above equation, we have

G + 
$$j\omega Cp = \frac{R}{R^2 + \frac{1}{\omega^2 Cs^2}} + j \frac{\frac{1}{\omega Cs}}{R^2 + \frac{1}{\omega^2 Cs^2}}$$

where, Cs (= -  $\frac{1}{\omega X}$ ): equivalent series circuit capacitance

Cp (= 
$$\frac{B}{\omega}$$
) : equivalent parallel circuit capacitance

Obviously, if no series resistance (R) or parallel conductance (G) are present, the equivalent series circuit capacitance (Cs) and equivalent parallel circuit capacitance (Cp) are identical. Likewise, if R and G are not present, the equivalent series circuit inductance (Ls) and equivalent parallel circuit inductance (Lp) are identical.

However, a sample value measured in a parallel measurement circuit can be correlated with that of a series circuit by a simple conversion formula which covers the effect of dissipation factor. See Table 3-8. Figure 3-4 graphically shows the relationships of parallel and series parameters for various dissipation factor values. Applicable diagrams and equations are given in the chart. For example, a parallel capacitance

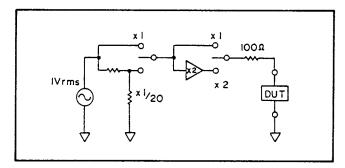


Figure 3-3. Equivalent Circuit of the Test Signal Source.

(Cp) of 1000pF with a dissipation factor of 0.5 is equivalent to a series capacitance (Cs) of 1250pF with an identical dissipation factor. As shown in Figure 3-4, inductance or capacitance values for parallel and series equivalents are nearly equal when the dissipation factor is less than 0.03. The dissipation factor of a component always has the same value at a given frequency for both parallel and series equivalents.

In ordinary LCR measuring instruments, the measurement circuit is set (automatically or manually) to a predetermined equivalent circuit with respect to either the selected range or to the dissipation factor value of the sample. The wider circuit mode selection capability of the 4276A, which is free from these restrictions. permits taking measurements in the desired circuit mode and of comparing such measured values directly with those obtained by another instrument. This obviates the inconvenience and necessity of employing instruments capable of taking measurements with the same equivalent circuit to assure measurement correspondence.

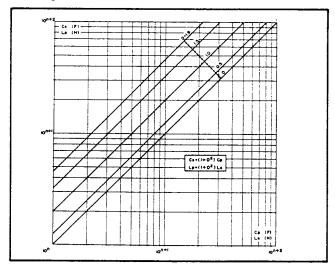


Figure 3-4. Parallel and Series Parameter Relationship.

Table 3-8. Dissipation Factor Equations and Equivalent Circuit Conversion Formulas

Circu	it Mode	Dissipation Factor	Conversion to Other Modes	
C	C₽ •∰• G	$D = \frac{G}{\omega Cp} = \frac{1}{Q}$	Cs = $(1 + D^2)$ Cp, R = $\frac{D^2}{1 + D^2} \cdot \frac{1}{G}$	
	Cs R •□····•	$D = \omega C s R = \frac{1}{Q}$	$Cp = \frac{1}{1 + D^2} Cs, G = \frac{D^2}{1 + D^2} \cdot \frac{1}{R}$	
,	₽ G G	$D = \omega LpG = \frac{1}{Q}$	Ls = $\frac{1}{1 + D^2}$ Lp, R = $\frac{D^2}{1 + D^2}$ · $\frac{1}{G}$	
L	Ls R •□•••	$D = \frac{R}{\omega Ls} = \frac{1}{Q}$	Lp = $(1 + D^2)$ Ls, $G = \frac{D^2}{1 + D^2} \cdot \frac{1}{R}$	

# 3-35. INITIAL DISPLAY AND INDICATIONS

3-36. Each time the instrument is turned on, the option codes for installed options and the HP-IB address are displayed on the front panel for approximately two seconds. The HP-IB address is displayed on DISPLAY A, as shown below. The factory set address is 17 (10001), but any address from 0 (00000) to 30 (11110) can be set. Refer to the HP-IB discussion starting in paragraph 3-72.





Note

If the instrument is set to TALK ONLY mode, the output data format number (see paragraph 3-95) will appear on DISPLAY A instead of the HP-IB address.

The following option code is displayed on DISPLAY B if the instrument is equipped with Option 002, Comparator/Handler Interface.

DISPLAY B



Note

The above option code will not be displayed if the 16064A Comparator/Handler Interface is not connected to the instrument.

The following option code is displayed on the FREQUENCY/DC BIAS display if the instrument is equipped with Option 001, Internal DC Bias.

FREQUENCY / DC BIAS



3-37. After the HP-IB address and option codes have been displayed, the continuous memory function automatically recalls the front panel control settings that existed when the instrument was turned off.

#### Note

Output from the internal dc bias source (option 001 instruments) is automatically set to 0V at instrument power on as a safety precaution.

# 3-38. INITIAL CONTROL SETTINGS

3-39. The 4276A is automatically set to the control settings listed below when the continuous memory function (refer to paragraph 3-40) is reset as described in paragraph 3-42.

DISPLAY A Function	C
	•
DISPLAY B Function	
CIRCUIT MODE	
LC   Z   RANGE	AUTO
MEASUREMENT SPEED	MED
TEST SIGNAL LEVEL	HIGH
TRIGGER	INT
SELF TEST	OFF
Δ	OFF
FREQ/DC BIAS	FREQ
SPOT/COARSE/FINE	SPOT
Frequency	1.00kHz
OPEN ZERO DATA	$\Omega$ 0
SHORT ZERO DATA	OS
	~~

When the instrument is equipped Option 001:

When the instrument is equipped Option 002, control settings of the 16064A Comparator are as follows:

ENABLE	
LC   Z   //D/Q/ESR/G ·············	
LIMIT LOW/HIGH ·····	LOW
BIN NUMBER	1
RUN	OFF
BIN LIMITS	blank

# 3-40. CONTINUOUS MEMORY

3-41. The continuous memory function of the 4276A automatically memorizes all front panel control settings when the instrument is turned off or experiences a power failure. When the instrument is turned on, the memorized settings are automatically recalled. Continuous memory powered by а rechargeable nickel-cadmium battery that lasts for approximately 2 weeks when the instrument is turned off. The battery is recharged while the 4276A is turned on.

> Note It takes approximately 24 hours to recharge up the battery.

#### Note

When turned on, the 4276A automatically performs a Check Sum Test as part of its turn-on Self Test. The Check Sum Test checks the contents of memory. If incorrect, E68 will be displayed on DISPLAY A and memory will be cleared. instrument will be set to the initial control settings (refer to paragraph 3-38).

OPEN and SHORT Zero Offset values (refer to paragraph 3-48) and reference values for deviation measurements (refer to paragraph 3-57) are also memorized by the continuous memory function. On instruments equipped with the Comparator/Handler Interface option (Option 002), all high and low limits and all 16064A control settings (except RUN) are memorized. DC bias voltage (Option 001) settings, however, are not memorized.

# 3-42. RESETTING CONTINUOUS MEMORY

3-43. To reset, or clear, continuous memory, proceed as follows:

- (1) Turn off the 4276A.
- (2) Press and hold both FREQ/DC BIAS Step Control Keys ( ᠍ ② ).
- (3) Turn on the 4276A.

# 3-44. UNKNOWN TERMINALS

3-45. The UNKNOWN terminals of the 4276A are arranged in a five-terminal configuration. The five-terminal configuration provides accurate measurements over a broad impedance range. Low impedance errors caused by residual inductance and residual resistance, are lower than those of measurements made using the three-terminal configuration. Also, high impedance measurement errors caused by residual conductance and residual capacitance are lower than those of measurements made using the four-terminal configuration.

In the five-terminal configuration, the current through the DUT is fed back to the oscillator via the outer conductors of  $L_{\text{CUR}}$  and  $H_{\text{CUR}}$  terminals so as to reduce electromagnetic coupling between the current terminals ( $H_{\text{CUR}}$  and  $L_{\text{CUR}}$ ) and the voltage terminals ( $H_{\text{POT}}$  and  $L_{\text{POT}}$ ). This feature reduces the voltage detection error when a low-impedance DUT is measured at a high frequencies. Refer to Figure 3-5.

#### Note

It is recommended that four short BNC cables be used as test leads.

#### Note

Do not connect the outer conductors of the test leads to ground. If the outer conductors are grounded, displayed values will fluctuate when a low-impedance measurement is made at high frequencies.

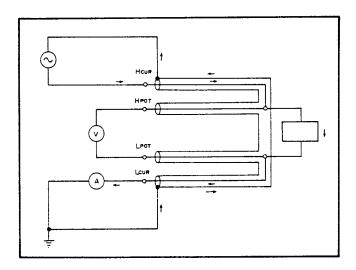


Figure 3-5. Five Terminal Configuration.

# 3-46. MEASUREMENT OF GROUNDED SAMPLES

3-47. Samples which have one terminal (except GROUND terminal) grounded to earth cannot normally be measured by the 4276A. Such measurement conditions are, for example, the distributed capacitance measurement of a coaxial cable with a grounded shield conductor or the input/output impedance measurement of a amplifier. ended When one-side-grounded sample is connected for 4276A may measurement. the display a measurement error message or incorrect measurement results. This is because the current though the DUT does not flow in the I-to-V Convertor which converts the current to the voltage, with LOW measurement terminals grounded.

#### Note

If HIGH terminals are grounded, the test signal is not supplied to the DUT.

#### 3-48, ZERO OFFSET ADJUSTMENT

3-49. The test fixtures and test leads used to connect samples to the instrument's UNKNOWN terminals have inherent residual impedance and stray admittance which, unless compensated for in some way, affect measurement accuracy. To minimize the effects of these residuals and strays, the 4276A is equipped with OPEN and SHORT Zero Offset Adjustment functions that can be executed from the front panel or via the HP-IB. Each Zero Offset Adjustment is performed at the following frequencies:

20kHz 16kHz 10kHz 5kHz 2kHz 1kHz 500Hz 200Hz 100Hz

Zero Offset data for test frequencies other than those listed above are calculated from the Zero Offset data obtained at the above test frequencies by using second degree interpolation. Thus, Zero Offset is provided for measurements made at all test frequencies. Brief descriptions of the Zero Offset Adjustments (OPEN and SHORT) are given below.

# ZERO OPEN:

The procedure for performing OPEN Zero Offset Adjustment is as follows:

(1) Connect the test fixture or test leads to the instrument's UNKNOWN terminals.

# Note

If test leads are used, you must convert the five-terminal configuration to a two-terminal configuration. Refer to paragraph 3-44 and Figure 3-5.

- (2) Connect nothing as the DUT.
- (3) Press the ZERO OPEN button.

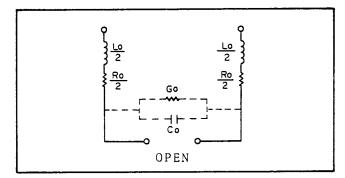


Figure 3-6. ZERO OPEN Circuit.

When the ZERO OPEN button is pressed, the instrument will be automatically set to C-G measurement mode. It will then measure the test fixture's stay admittance at each of the previously mentioned test frequencies. The measured values are stored in the instrument's internal memory. When offset adjustment is completed, DISPLAY A and DISPLAY B will be blank for 1 or 2 seconds, after which the front panel controls will be reset to the settings that existed when the ZERO OPEN button was pressed.

The purpose of OPEN Zero Offset Adjustment is to measure the test fixture's stray admittance, which, as shown in Figure 3-6 , consists of G and C . (This stray admittance is equivalent to a high impedance, which will "swamp out" a high impedance DUT connected to the test fixture.) The residual impedance of the test fixture—R  $_{0}$  and L  $_{0}$  in Figure 3-6 —is negligibly low and therefore does not affect the accuracyof OPEN Zero Offset Adjustments.

# ZERO SHORT:

The procedure for performing SHORT Zero Offset Adjustment is as follows:

(1) Connect the test fixture or test leads to the instrument's UNKNOWN terminals.

#### Note

If test leads are used, you must convert the five terminal configuration to a two-terminal configuration. Refer to paragraph 3-44 and Figure 3-5.

- (2) Connect a low impedance shorting-bar to the test fixture. If you're using test leads, simply connect the ends of the leads together.
- (3) Press the ZERO SHORT button.

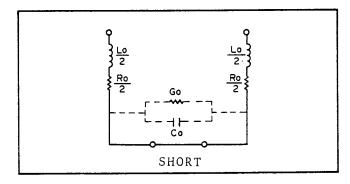


Figure 3-7. ZERO SHORT Circuit.

When the ZERO SHORT button is pressed, the instrument will be automatically set to  $|Z| - \theta$ measurement mode. It will then measure the test fixture's residual impedance at each of the previously mentioned test frequencies. The measured values are stored in the instrument's internal memory. When offset adjustment is completed, DISPLAY A and DISPLAY B will be blank for 1 or 2 seconds, after which the front panel controls will be reset to the settings that existed when the ZERO SHORT button was pressed. The purpose of SHORT Zero Offset Adjustment is to measure the test fixture's (or test lead's) residual impedance, which, as shown in Figure 3-7 , consists of Ro and Lo. This residual impedance, although small, degrades the accuracy of low impedance measurements. The stray admittance of the test fixture—Go and Co in Figure 3-7 - is shunted by the low impedance shorting-bar and therefore is not measured.

OPEN and SHORT Offset Zero Adjustments have been made, the instrument automatically compensates all measurements for the residuals and stravs of the test fixture or test leads. The values displayed on the front panel are the actual values of the DUT. Also, because the Zero Offset data is maintained by the instrument's continuous memory function, OPEN and SHORT Zero Offset Adjustments do not have to be repeated each time the instrument is turned on. You need to repeat Zero Offset Adjustments only when you change test fixtures (the residuals and strays of one test fixture are different from those of another). Maximum values that can be offset are listed below.

Capacitance: Up to 20pF OPEN Conductance: Up to 2µS

Impedancee: Up to  $2\Omega$  SHORT

Note

During Zero Offset Adjustment, OF or CF may appear on DISPLAY A or DISPLAY B. Zero Offset Adjustment, however, is performed correctly unless error code "E10" is displayed.

Note

After Zero Offset Adjustments, CF and 0000 may be alternately displayed on DISPLAY A if the measurement mode is other than C-G and nothing is connected to the test fixture. This is normal; it is not a malfunction.

Note

OPEN and SHORT Zero Offset Adjustments cannot be performed without a test fixture.

3-50. ACTUAL MEASUREMENT EQUIVALENT CIRCUIT

3-51. The test fixture or test leads used to connect a sample to the instrument's UNKNOWN terminals becomes part of the sample which the instrument measures. The five terminal configuration employed in the 4276A minimizes residual impedance circuit. The residual impedance, inherent in the test fixture or test leads, can be eliminated by the 4276A's ZERO offset function (refer to paragraph 3-48).

However, the five terminal measurement system must be converted to a two terminal configuration at the sample because most components have only two terminals. Moreover, additional stray capacitance is introduced when the sample is connected to the test fixture. Figure 3-8 illustrates lead impedance and the stray capacitances between the component's leads.

3-52. Diverse parasitic elements present between the sample and the UNKNOWN terminals will affect measurement results. These parasitic elements are series resistive and reactive elements and parallel conductive and susceptive elements. Figure 3-9 shows the equivalent circuit of the samples parasitic elements (R + jX is the sample's impedance). In Figure 3-9, Lo represents the residual inductance of the component's leads, and Ro is lead resistance. Go is the conductance between the leads, and Co is the sum of all stray capacitances shown in Figure 3-8. Reactive factors in the residual impedance and susceptive factors in the stray admittance have a greater measurements made at higher effect on frequencies.

3-53. Figure 3-10 shows the effect of residual impedance on C-G measurement and the effect of stray admittance on L-R measurement. Generally,  $L_0$  resonates with the capacitance of the sample (series resonance) and  $C_0$  resonates with the inductance of the sample (parallel resonance), respectively, at a specific high frequency. Thus, the impedance of the test sample will have a minimum value corresponding to resonant peaks, as shown in Figure 3-11. The presence of  $L_0$  and  $C_0$  causes measurement errors, as the phase of the test signal current varies over a broad frequency region around the

resonant frequencies. Additional errors, due to the resonance, increase in proportion to the square of the measurement frequency (below resonant frequency) and can be theoretically approximated as follows:

$$C_{ERROR} = \omega^2 L_0 Cx \cdot 100 (\%)$$
  
 $L_{ERROR} = \omega^2 C_0 Lx \cdot 100 (\%)$ 

where,

 $\omega$  =  $2\pi f$  (f: test frequency) Cx = Capacitance value of sample Lx = Inductance value of sample At low frequencies,  $L_0$  and  $C_0$  affect measured inductance and capacitance values, respectively, as simple additive errors. These measurement errors cannot be fully eliminated by the ZERO offset adjustment (which compensates for residual factors inherent in the test fixture). This is because  $L_0$  and  $C_0$  are peculiar to the component being measured. Their values depend on component lead length and on the distance between the sample and test fixture. The measurement results, then, are substantially the sample values including the parasitic impedances present under the conditions necessary to connect and hold the sample.

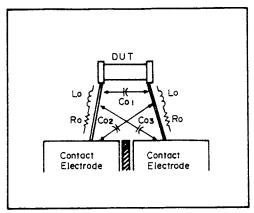


Figure 3-8. Parasitic Elements Incident to DUT Connections.

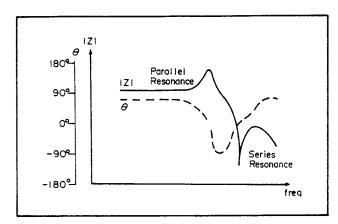


Figure 3-11. Effect of Resonance in Sample (Example).

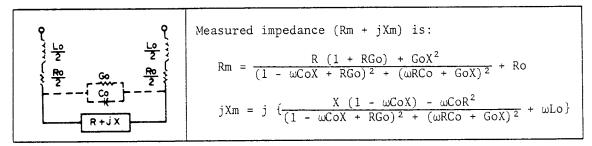


Figure 3-9. Equivalent Circuit Including Residual Impedance.

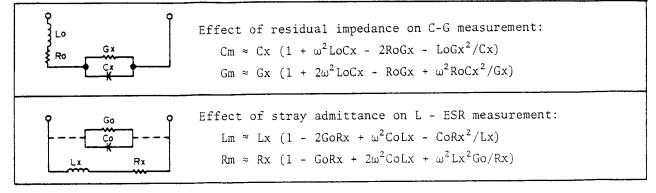


Figure 3-10. Effects of Residual Impedance.

3-54. MEASURED VALUES AND BEHAVIOR OF COMPONENTS

3-55. A component's measured value and its nominal value can, and often do, differ considerably because of various electromagnetic effects; for example, skin-effect of a conductor, the ferromagnetic properties of inductors, and the effects of dielectric materials in capacitors. Here, will discuss only the effects which result from the interaction of the reactive elements (L, C, etc.) of a component.

3-56. The impedance of a component can be graphically represented in vector form as shown in Figure 3-12. In such representation, the effective resistance and effective reactance correspond to the projections of the impedance vector  $|Z| \leftarrow \theta$ ; that is, the real (R) axis and the imaginary (jX) axis, respectively, as shown below:

Re = 
$$|Z| \cos \theta$$
  
Xe =  $|Z| \sin \theta$   
D =  $\frac{\cos \theta}{\sin \theta} = \frac{1}{\tan \theta}$ 

where, Re: Effective resistance

Xe: Effective reactance

Z: Impedance of the sample (Re + jXe).

D: Dissipation factor

When the phase angle,  $\theta$ , changes, both Re and Xe change in accordance with the definitions above. As component measurement parameters L, C, R, D, etc., are also representations of components related to the impedance vector, the phase angle dominates their values. Consider, for example, the inductance and the loss of an inductive component at frequencies around its self-resonant frequency. Figure 3-13 shows the equivalent circuit of the inductor. inductance, Lx, resonates with the distributed capacitance  $C_0$  at frequency  $f_0$ . The phase angle  $(\theta)$  of the impedance vector approaches 0 degrees (the vector approaches the R axis) when the frequency is close to the resonant frequency. Thus, the inductance of this component decreases while the resistive factor (loss) increases. At the resonant frequency, fo, this component is purely resistive. The effective resistance increases at resonance even if the inductor has no resistance (ideal inductor) at dc. Consequently, the loss factor varies sharply at frequencies around the resonant point.

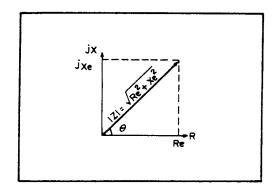


Figure 3-12. Impedance Vector Representation.

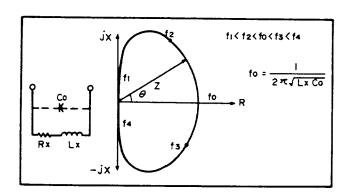


Figure 3-13. Typical Impedance Locus of an Inductor.

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# 3-57. DEVIATION MEASUREMENT FUNCTION

3-58. When many components of similar value are to be tested, it may be more practical to measure the difference between the value of the component and a predetermined, or ideal, reference value than measuring the DUT value itself. When the purpose of the measurement is to observe the change of a component's value versus changes in temperature, frequency, bias, etc., a direct measurement of this change (deviation) makes examination more meaningful and easier.

key is pressed, the values When the (measurement results) displayed on DISPLAY A and DISPLAY B are stored in the instrument's memory and are then used as the reference values for all subsequent measurements. The value displayed on each display is not the sample's measured value, it is the difference between the stored reference value and the measured value. Stored reference values are maintained by the 4276A's continuous memory function when the instrument is turned off. The deviation measurement function is automatically turned off when the DISPLAY A function, DISPLAY B function, LC | Z | RANGE, or CKT MODE is changed. It may be turned off also if the test frequency is changed when the DISPLAY B function is ESR/G, because the measurement range for ESR and G is frequency dependent.

# 3-59. CHARACTERISTICS OF TEST FIXTURES

3-60. Characteristics and applicable measurement ranges of the HP test fixtures and test leads for the 4276A are summarized in Table 3-9. To facilitate measurement and to minimize measurement errors, a test fixture appropriate for the measurement should be chosen from among HP's standard accessories. Select the test fixture or leads that have the desired performance characteristics.

Table 3-9. Typical Characteristics of Test Fixtures and Leads

· · · · · · · · · · · · · · · · · · ·		
14-1-7	Applicable Measurement Ranges	
Model Parameter Value	Measurement Frequency	
16047A	Full range	Full range
16047C	Full range	Full range
16048A 16048B	Full range	Full range
100400		
16048C	C>1000pF L>100µH	Below 100kHz
16048D	Full range	Full range
16034B	Ranges satisfied  Z >50Ω	Full range
16065A	Full range	50Hz to 2MHz

# 3-61. RANGE RESISTOR

3-62. The relation between the range resistor value and the measurement range is listed in Figure 3-16.

# 3-63. MEASUREMENT ACCURACY

3-64. The measurement reference plane for the accuracies specified in Section I is the UNKNOWN terminals. The measurement accuracy of the 4276A is guaranteed at the UNKNOWN terminals. The conditions under which accuracy is specified are described in Table 1-1. An example of the how to calculate measurement accuracy is shown in Figure 3-15.

# 3-65. MEASUREMENT EXAMPLE

3-66. The procedures for measuring general components—inductors, capacitors, resistors—are given in Figure 3-17. Almost any discrete component, except for those having special shapes or dimensions, can be measured with this setup. Special components may be measured by using test leads 16048A, 16048D, 16034B, etc., or by using specially designed user-built fixtures instead of the 16047A Test Fixture.

3-67. As an example of a typical semiconductor measurement, the procedures for measuring the base-collector juncton capacitance(C  $_{\rm ob}$ ) of an NPN transistor are given in Figure 3-18.

# 3-68. EXTERNAL DC BIAS

3-69. The special biasing circuits and procedures for using external voltage or current bias (required for certain capacitance or inductance measurements) are given in Figures 3-19, 3-20, and 3-21. The figures show sample circuits appropriate for 4276A applications. When applying a dc voltage to capacitors, be sure the applied voltage does not exceed the maximum specified voltage of the capacitor and that the capacitor is connected with correct polarity. Note that the externally applied bias voltage is present at the  $H_{\text{CUR}}$  and  $H_{\text{POT}}$  terminals.

#### 3-70. EXTERNAL TRIGGERING

3-71. The 4276A can be externally triggered by connecting an external triggering device to the EXT TRIGGER connector on the rear panel and setting the TRIGGER control on the front panel to MAN/EXT on front panel. The instrument is triggered (measurement is made) each time a positive-going TTL level pulse is applied to this connector (refer to Figure 3-14). External triggering can be also done by alternately shorting and opening the center conductor of the EXT TRIGGER connector to ground (chassis).

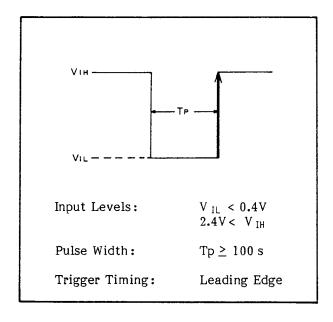


Figure 3-14. External Trigger Pulse.

Model 4276A SECTION III

[Examples of Calculating C, D, and Q [Examples of Calculating C and ESR/G Measurement Accuracies] Measurement Accuracies] Front Panel Settings: Front Panel Settings: Test Frequency: lkHz Test Frequency: 10kHz LC | Z | RANGE: InF LC | Z | RANGE: 1 uF TEST SIG LEVEL: HIGH TEST SIG LEVEL: HIGH MEAS SPEED: MED MEAS SPEED: MED Measured Values: Measured Values: C: 1.9945nF C: .852 µF ESR:  $.42\Omega$ D: .0008 Q: OF (Assume a value of Qm) G: 1.2mS Accuracies (Refer to Table 1-1): Accuracies: C: ±.1% of reading + 5 counts C: .9% of reading + 2 counts  $1.9945nF \times (.1/100) + .0005nF$  $.852 \mu F \times (.9/100) + .002 \mu F$  $= (\pm) 2.49 pF$  $= (\pm) 9.67 nF$ D: ±.1% of reading + .0006A + 5 counts ESR: .6% of reading + .6/ $\Omega$  + 5 counts  $.0008 \times (.1/100) + .0006 \times 1.248 + .0005$  $.42\Omega \times (.6/100) + .6/.852 + .05\Omega$  $= (\pm).00125$  $= (\pm).757\Omega$  $Q_{mx}$  (.00125/.0008) + .1 G:  $1.2 \text{mS} \times (.757/.42) + .1 \text{mS}$  $= (\pm) 2.16 mS$  $= \pm (Q_m x 1.5625 + .1)$ Note In this case, Q accuracy (1.5625 times  $Q_m$ ) has no meaning, because  $Q_m$  is overflow (OF).

Figure 3-15. How to Calculate Measurement Accuracies.

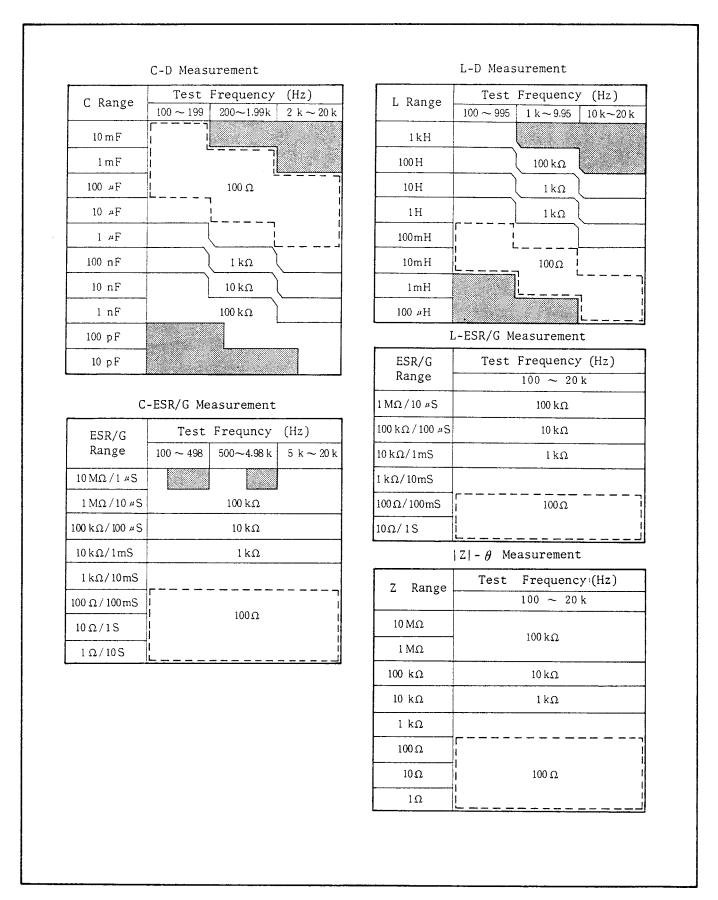
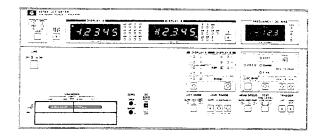


Figure 3-16. Range Resistor.

# GENERAL MEASUREMENT



- 1. Connect the 16047A Test Fixture to the UNKNOWN terminals.
- 2. Turn on the 4276A.
- 3. Verify that the HP-IB address and option codes (16064 and 001) are displayed on DISPLAY A, DISPLAY B, and the FREQUENCY/DC BIAS display, respectively.

#### Note

Option codes are displayed only if the corresponding option is installed.

#### Note

The HP-IB address is set to 17 (10001) when the instrument is shipped from the factory.

- 4. Press the SELF TEST key to verify that the instrument is functioning properly. Refer to paragraph 3-5, SELF TEST. If no error-codes are displayed, press the SELF TEST key again to turn off the SELF TEST function.
- Select the measurement functions for DISPLAY A and DISPLAY B.
- 6. Set the test frequency, test signal level, and measurement speed.

#### Note

SLOW measurement speed minimizes display fluctuation.

## Note

Best measurement accuracy is obtained when test signal level is set to HIGH and measurement speed is set to MED.

- 7. Perform OPEN and SHORT Zero Offset adjustments as described in paragraph 3-48.
- 8. Connect the device to be measured to the test fixture.
- Read the measured values from DISPLAY A and DISPLAY B.

#### Note

Refer to paragraph 3-20 for the meaning of any error-codes that may appear on DISPLAY A.

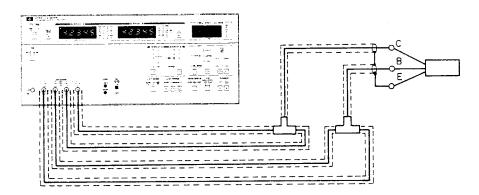
#### Note

When the instrument is set to C-D or C-Q measurement mode and nothing is connected to the measurement terminals, CF and .0000 may be alternately displayed on DISPLAY A. This is not a malfunction, however.

# Note

For C or L measurement, if the dissipation factor of the DUT is higher than 0.1, C, L, and D measurement accuracy tolerances increase by a factor of  $1 + D^2$ . If D is higher than 1, AUTO ranging cannot be performed correctly. |Z| measurement mode should be selected.

# SEMICONDUCTOR DEVICE MEASUREMENT



Parameters of semiconductor devices have a strong dependency on the applied voltage and device temperature. Because of the non-linear characteristics of semiconductor impedance devices, a semiconductor measurement is subject to exact establishment of the test conditions to make measured values meaningful. detailed analysis of the device under its operating test conditions, a low level test signal is employed in order to obtain measured values with respect to a local region around the operating test point selected for plotting characteristic parameter curves of the sample. procedure measuring for typical semiconductor junction capacitance in P-N and MOS junction devices is outlined below.

# Measurement Setup:

The figure above shows a typical test setup for measuring the base-collector junction capacitance (Cob) of an NPN transistor. For this measurement, the test fixture may be user designed. A 4276A equipped with option 001 is ideal for controlling the dc bias required for the measurement. If dc bias is not necessary, setup with procedures associated measurement may be deleted.

# PROCEDURE:

- 1. Connect the test fixture or test cables to the UNKNOWN terminals of the 4276A.
- 2. Turn on the 4276A.
- 3. Set the 4276A's front panel controls as follows:

DISPLAY A: C DISPLAY B: G Test Freq.: lkHz TEST SIG LEVEL: LOW

- adjustments as described in paragraph 3-48.
- 5. Set the DC BIAS SELECT switch on the rear panel to INT.

4. Perform OPEN and SHORT Zero Offset

#### Note

If an external voltage source is used for dc biasing, set the DC BIAS SELECT switch to EXT, and connect the voltage source output to the EXT INPUT/INT MONITOR connector on the rear panel.

#### Note

DC bias voltage, whether supplied from the internal bias source or from an external bias source, should be set to OV at this time.

#### Note

Use the HP Model 16065A EXTERNAL VOLTAGE BIAS FIXTURE for high voltage bias applications up to ±200V.

- 6. Connect the transistor to the measurement terminals.
- 7. Monitor the bias voltage actually applied to the transistor.

#### Note

If the 16065A is used, close the lid after you connect the transistor to the measurement terminals. Measurement cannot be made while the lid is open.

switch on the front 8. Set the DC BIAS panel to ON, and set the desired bias voltage.

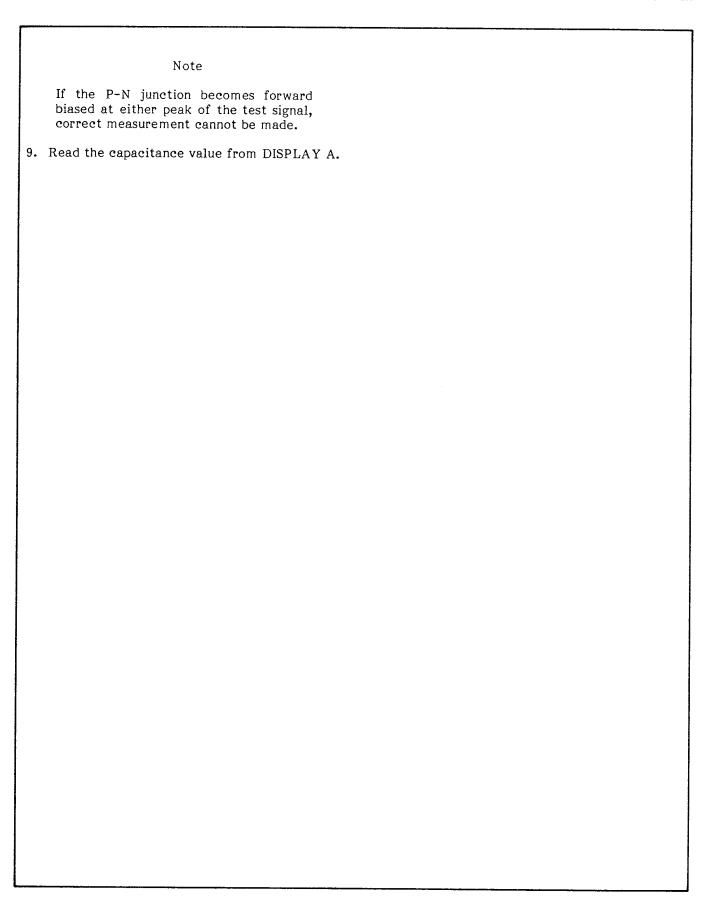
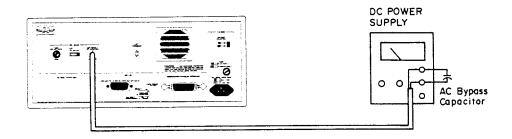


Figure 3-18. Semiconductor Device Measurement (Sheet 2 of 2).

# EXTERNAL DC BIAS OPERATION (≤±40V)



To make capacitance measurements using externally supplied dc bias voltages up tp  $\pm 40V$ , connect a dc voltage source to EXT INPUT/INT MONITOR connector on the rear panel as shown in the diagram.

### CAUTION

DO NOT APPLY GREATER THAN ±40V TO THE 4276A'S EXT INPUT/INT MONITOR CONNECTOR. IF THE APPLIED VOLTAGE EXCEEDS ±40V, THE 4276A MAY BE DAMAGED.

#### CAUTION

BE SURE THE CORRECT FUSE (HP P/N 2110-0011) IS INSTALLED IN THE DC BIAS FUSE HOLDER ON THE REAR PANEL.

### PROCEDURE:

- Set DC BIAS select switch on rear panel to EXT.
- 2. Connect the test fixture or test leads to the UNKNOWN terminals of the 4276A.
- 3. Turn on the instruments.
- 4. Set the 4276A's controls as described in steps 3 through 6 of Figure 3-17. Set the DISPLAY A function to "C" measurement mode.
- 5. Perform OPEN and SHORT ZERO offset adjustments as described in paragraph 3-48.
- Connect a sample to the test fixture or test leads.

### CAUTION

DO NOT SHORT THE HIGH AND LOW TERMINALS.

## CAUTION

WHEN A POSITIVE BIAS VOLTAGE IS USED, THE POSITIVE TERMINAL OF ELECTROLYTIC CAPACITORS MUST CONNECTED TO THE **INSTRUMENT'S** HIGH TERMINAL. WHEN USING A NEGATIVE BIAS VOLTAGE, CONNECT CAPACITOR'S NEGATIVE TERMINAL HIGH INSTRUMENT'S THE TO TERMINAL.

- 7. Set the external dc voltage source to the desired output voltage.
- 8. Read the measured values. Wait until the applied dc bias across the sample becomes stable.
- 9. Reset the external voltage source to 0V.
- Remove the sample from test fixture or test leads.

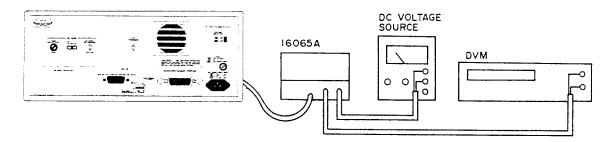
## Note

Use a stable dc voltage source.

## Note

To make stable measurements, connect an ac bypass capacitor (approximately  $l\,\mu F$ ) between positive terminal and negative terminal of the external dc voltage source.

# EXTERNAL DC BIAS OPERATION (≤±200V)



To make capacitance measurements using externally supplied dc bias voltages up to  $\pm 200 \, \text{V}$ , use the HP 16065A Test Fixture. Connect a dc voltage source to the 16065A as shown in the diagram.

#### CAUTION

DO NOT APPLY GREATER THAN ±40V TO THE 4276A'S EXT INPUT/INT MONITOR CONNECTOR. IF THE APPLIED VOLTAGE EXCEEDS ±40V, THE 4276A MAY BE DAMAGED.

### PROCEDURE:

- Set DC BIAS select switch on rear panel to OFF.
- 2. Connect the 16065A to the UNKNOWN terminals of the 4276A.
- 3. Connect the dc voltage source to DC BIAS INPUT connector of the 16065A.
- 4. Connect a DVM or an oscilloscope to the DC BIAS MONITOR connector of the 16065A.
- 5. Turn on the instruments.
- Set the 4276A's controls as described in steps 3 through 6 (Figure 3-17). Set the DISPLAY A function to "C" measurement mode.
- 7. Perform OPEN and SHORT ZERO offset adjustments.
- 8. Connect a sample to the 16065A test fixture.

## CAUTION

DO NOT SHORT THE HIGH AND LOW TERMINALS.

#### CAUTION

WHEN A POSITIVE BIAS VOLTAGE IS USED, THE POSITIVE TERMINAL OF ELECTROLYTIC CAPACITORS MUST BE CONNECTED TO THE INSTRUMENTS HIGH TERMINAL. WHEN USING A NEGATIVE BIAS VOLTAGE, CONNECT THE CAPACITOR'S NEGATIVE TERMINAL **INSTRUMENT'S** TO THE HIGH TERMINAL.

- 9. Set the external dc voltage source to the desired output voltage and close the cover of the 16065A.
- Read the measured values. Wait until the monitored voltage becomes stable.
- 11. Open the cover of the 16065A.

### Note

When the cover of the 16065A is opened, the charge on the sample is discharged through two paralleled  $20\Omega$  resistors.

12. Remove the sample from the 16065A.

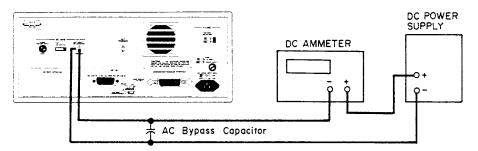
# Note

Use a stable dc voltage source.

# Note

The test signal will appear at the DC BIAS MONITOR connector. This does not affect measurement results, however.

# EXTERNAL DC CURRENT BIAS OPERATION (≤1 mA)



DC bias current can be applied to the sample through the UNKNOWN terminals by connecting a dc voltage source to the instrument. The procedure for making inductance measurements using current biasing is given below.

### PROCEDURE:

- Set the DC BIAS select switch on the rear panel to EXT.
- 2. Connect an external dc voltage source and dc ammeter (for current monitoring) to the EXT INPUT/INT MONITOR connector on the rear panel, as shown in the diagram.
- 3. Connect a test fixture or test leads to the UNKNOWN terminals of the 4276A.
- 4. Turn on the instruments.
- Set the 4276A's controls as described in steps 3 through 6 of Figure 3-17. Set the DISPLAY A function to "L" measurement mode.
- Perform OPEN and SHORT ZERO offset adjustments.
- 7. Connect the sample to the test fixture or test leads.
- 8. Gradually increase the dc voltage source output voltage until the desired bias current, as indicated on the dc ammeter, is obtained.

## CAUTION

DO NOT ALLOW THE BIAS CURRENT TO EXCEED 35mA AND DO NOT ALLOW THE OUTPUT VOLTAGE FROM THE EXTERNAL DC VOLTAGE TO EXCEED SOURCE ±40V. IF CURRENT EXCEEDS 35mA OR IF VOLTAGE EXCEEDS ±40V, THE INSTRUMENT MAY BE DAMAGED.

#### Note

DC bias current flowing through sample can be calculated by the following equation:

$$I_{DC} = \frac{E_{bias}}{Rx + 1} (mA)$$

where E bias is the bias voltage (V) applied to EXT INPUT/INT MONITOR connector and Rx is the dc resistance  $(k\Omega)$  of the sample.

- 9. Read the measured values.
- 10. Gradually decrease the dc voltage source output voltage until the dc bias current is 0mA, then remove the sample from the test fixture or test leads.

#### Note

To make stable measurements, connect an ac bypass capacitor (near  $1\,\mu F$ ) between the positive terminal and the negative terminal of the dc voltage source.

# Note

Maximum allowable current depends on the range resistor, as listed in the table below.

Range Resistor Value	Maximum Current
$100\Omega$	lmA
$1 \mathrm{k}\Omega$	0.5mA
10kΩ	50µA
100kΩ	5µA

Refer to Figure 3-16 for details on the relation between range resistor and measurement range. Note that measurement accuracies, as specified in Section I, are not guaranteed if bias current is allowed to exceed the limits given in the above table.

#### 3-72. HP-IB INTERFACE

3-73. The 4276A can be remotely controlled via the HP-IB, a carefully defined instrument interface which simplifies integration of programmable instruments and a calculator or computer into a system.

Note

HP-IB is Hewlett-Packard's implementation of IEEE Std. 488, "Standard Digital Interface for Programmable Instrumentation."

### 3-74. HP-IB INTERFACE CAPABILITIES

3-75. The 4276A has eight HP-IB interface functions, as listed in Table 3-11.

Table 3-10. HP-IB Interface Capabilities

Code	Interface Function * (HP-IB Capabilities)				
SH1**	Source Handshake				
AH1	Acceptor Handshake				
Т5	Talker (basic talker, serial poll, talk only mode, unaddress to talk if addressed to listen)				
L4	Listener (basic listener, unaddress to listen if addressed to talk)				
SR1	Service Request				
RL1	Remote/local (with local lockout)				
DC1	Device Clear				
DT1	Device Trigger				

- \* Interface functions provide the means for a device to receive, process, and tansmit messages over the bus.
- \*\* The numeric suffix of the interface code indicates the limitation of the function, as defined in Appendix C of IEEE Std. 488. 1978.

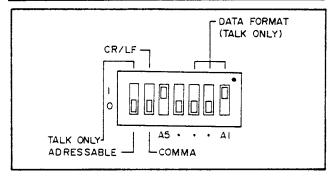


Figure 3-22. HP-IB Control Switch.

### 3-76. CONNECTION TO HP-IB

3-77. The 4276A can be connected into an HP-IB bus configuration with or without a controller (i.e., with or without an HP calculator). In an HP-IB system without a controller, the instrument functions as a "talk only" device (refer to paragraph 3-84).

#### 3-78, HP-IB STATUS INDICATORS

3-79. The HP-IB Status Indicators are four LED lamps located on the front panel. When lit, these lamps show the existing status of the 4276A in the HP-IB system as follows:

SRQ: SRQ signal from the 4276A to the controller is on the HP-IB line. Refer to paragraph 3-101.

LISTEN: The 4276A is set to listener.

TALK: The 4276A is set to talker.

REMOTE: The 4276A is under remote control.

# 3-80. LOCAL KEY

3-81. The LOCAL key releases the 4276A from HP-IB remote control and allows measurement conditions to be set from the front-panel. The REMOTE lamp will go off when this key is pressed. LOCAL control is not available when the 4276A is set to "local lockout" status by the controller.

#### 3-82. HP-IB CONTROL SWITCH

3-83. The HP-IB Control Switch, located on the rear panel, has seven bit switches. See Figure 3-22. Each bit switch has two settings: logical 0 (down position) and logical 1 (up position). The left-most bit switch, bit 7, determines whether the instrument will be addressed by the controller in a multidevice system, or will function as a "talk only" device to output measurement data and/or instructions to an external "listener," e.g., printer. The switch settings, when the instrument is shipped from the factory, are shown in Figure 3-22.

When bit switch 7 is set to 0, the instrument is in ADDRESSABLE mode and bit switches 1 through 5 determine the instrument address. When this bit switch is set to 1, however, the instrument is in TALK ONLY mode.

Bit switch 6 determines the output data delimiter. When this bit switch is set to 0, the delimiter is a comma (,); when set to 1, the delimiter is a carriage return and line feed (CR/LF).

#### Note

The HP-IB Control Switch setting is memorized only at instrument turn on. Thus, even if the HP-IB Control Switch setting is changed while the instrument is turned on, the memorized setting is not changed until the instrument is turned off and on.

### 3-84. TALK ONLY MODE

3-85. When bit switch 7 of the HP-IB Control Switch is set to TALK ONLY (i.e., set to 1), the instrument functions as a "talker," outputting data to a "listener" (e.g., printer). In TALK ONLY mode, bit switches 1, 2, and 3 determine the format in which data is output. There are six formats, Fl through F6, and the bit switch setting for each format is shown in Table 3-11. Refer to paragraph 3-95 for details on the output data formats.

## Note

If the instrument is set to TALK ONLY mode, the Output Data Format number will be briefly displayed on DISPLAY A (instead of the HP-IB address) when the instrument is turned on. The displayed number, however, will be the format number plus 50. For example, if the Output Data Format is F2, the number displayed on DISPLAY A at turn on will be 52.



## Note

When the instrument is used in TALK ONLY mode, devices connected to the instrument must be set to LISTEN ONLY mode.

3-86. In TALK ONLY mode, both bit switches 4 and 5 have no function.

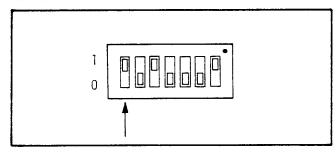


Figure 3-23. TALK ONLY Mode.

Table 3-11. Data Output Format Selection

Bit Sv	vitch Set		
Bit 3	Bit 2	Bit 1	Output Data Format
0	0	0	F1
0	0	1	F2
0	1	0	F3
0	1	1	F4
1	0	0	F5
1	0	1	F6
1	1	0	F1
1	1	1	F2

Note: Refer to paragraph 3-95 for details.

### 3-87. ADDRESSABLE MODE

3-88. When bit switch 7 of the HP-IB Control Switch is set to ADDRESSABLE (i.e., set to 0), bit switches 1 through 5 represent the HP-IB address of the instrument, in binary. These switches are set to 10001 (decimal 17) when the instrument leaves the factory but can be set to any desired address between 0 and 30.

#### Note

When the instrument is turned on, the HP-IB address is displayed, in decimal, on DISPLAY A. For example, the factory-set address (10001) is displayed as "17."



## Note

HP-IB address lllll (decimal 31) cannot be used. If this address is set, E19 will be displayed on DISPLAY A (after 31 has been displayed) when the instrument is turned on.

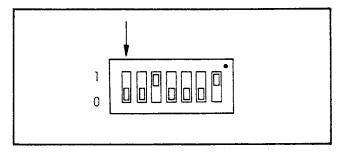


Figure 3-24. ADDRESSABLE Mode.

## 3-89. REMOTE PROGRAM CODES

3-90. Remote program codes for the 4276A are listed in Table 3-12.

### 3-91. PARAMETER SETTING

3-92. Test frequency, DC bias (option 001), and bin limits (option 002) can be set via remote programming.

# [1] Test Frequency Setting

FR 
$$\frac{XXX.X}{(1)}$$
 EN

(1) Setting value, in kHz.

## Note

When an illegal frequency that is within the instrument's frequency range is set, the frequency below the illegal setting is automatically selected. For example:

"FR7.59EN" 7.55kHz displayed on FREQUENCY/DC BIAS DISPLAY

[2] DC Bias setting

BI 
$$\frac{\pm XX.X}{(1)}$$
 EN

(1) Setting value, in volts.

# Note

If not set, polarity sign is automatically set to positive (+).

[3] Comparator Limit Setting (Option 002 only)

(Low Limit) LL 
$$\frac{XX.XXX}{(1)}$$
 EN

(High Limit) LH 
$$\frac{XX.XXX}{(1)}$$
 EN

(1) Setting value. The position of the decimal point must agree with the measurement range. Unit is in accordance with the unit indicators of DISPLAY A or DISPLAY B.

# 3-93. DATA OUTPUT

3-94. Measurement and status data are output to external devices in bit parallel, byte serial format via the eight DIO signal lines of the HP-IB. Data can be output in ASCII mode or

PACKED BINARY mode. Each mode is described below.

# [1] ASCII mode

Output data in this mode includes status data, key status (function) data, and measurement data (including range) for DISPLAY A and DISPLAY B. If the instrument is equipped with Option 002, comparison data (LOW, IN, HIGH) for L/C/|Z| and D/Q/ESR/G, and BIN number data can be output, too. The output format is shown in Figure 3-25. All characters are coded in accordance with ASCII coding conventions.

### [2] PACKED BINARY mode

Output data in this mode is output as one or two binary bytes, rather than as a character representation. This data output format is for high speed data transfer. Contents of output data, however, is less than that of ASCII mode. Output data in this mode includes status data for DISPLAY A and DISPLAY B, measurement range data as an 8-bit byte, and measurement data of DISPLAY A and DISPLAY B (not including unit and decimal point) as a 16-bit, 2's complement binaryword. If the instrument is equipped with Option 002, comparison data (LOW, IN, HIGH) for L/C/|Z| and D/Q/ESR/G, and BIN number data can be output as an 8-bit byte. The displayed data is output as the equivalent decimal values of the resulting words. The output format is shown in Figure 3-26.

# 3-95. OUTPUT DATA FORMAT

3-96. The 4276A can output measurement data to a controller or can output data directly to an external "listener" device (i.e., printer). There are six Output Data Formats, Fl through F6. The contents of the output data for each format are listed in Table 3-15.

## Note

In ADDRESSABLE MODE, only F1 through F4 can be set by HP-IB remote control. Output data can be in either ASCII mode or BINARY PACKED mode. Also, in ADDRESSABLE mode, bit switch settings have no relation to Output Data Format.

#### Note

In TALK ONLY mode, any Output Data Format, F1 through F6, can be set by HP-IB Control Switch settings (bit 1 through bit 3). Also, in TALK ONLY mode, data can be output in ASCII mode only.

SECTION III Model 4276A

Table 3-12. Remote Program Codes (Sheet 1 of 2)

Item	Control	Program Code	Description
DISPLAY A Function	L C HIGH SPEED L HIGH SPEED C  Z *	A1 A2 A3 A4 A5	DISPLAY A and DISPLAY B combinations are listed in the table below:
DISPLAY B Function	D Q ESR/G	B1 B2 B3	A  1 L-D L-Q L-ESR/G  2 C-D C-Q C-ESR/G  * When DISPLAY A is set to   z ,DISPLAY B is automatically set to θ.
CKT MODE	AUTO	C1. C2 C3	
MEAS SPEED	SLOW MED FAST	M1 M2' M3	
Auto Range	OFF ON	N1.	<ul><li>: Range is fixed.</li><li>: Range is automatically selected.</li></ul>
LC Z  Range	100mΩ 10pF/1Ω 100μH/100pF/10Ω 1mH/1nF/100Ω 10mH/10nF/1kΩ 100mH/100nF/10kΩ 1H/1μF/100kΩ 10H/10μF/1MΩ 100H/100μF/10MΩ 1kH/1mF 10mF	R1 R2 R3 R4 R5 R6 R7 R8 R9 RA	If the instrument is set to a range which cannot make the measurement, range is automatically reset to the nearest range capable of making the measurement.
Test Signal Level	LOW HIGH	V1 V2·	
Trigger Mode	INT MAN/EXT	T1. T2	This code only sets the trigger mode; it does not trigger the instrument.
Execute		EX	This code is used to trigger the instrument.
Self Test	OFF ON	S0° S1	
Deviation Measurement	OFF ON	XO' X1	
Zero Offset	OPEN SHORT	Z0 ZS	

Table 3-12. Remote Program Codes (Sheet 2 of 2)

Item	Control	Program Code	Description
Data Ready	OFF ON	D0' D1	If Data Ready is set to ON, an SRQ signal is output when the measurement is completed.
Comparator Enable	OFF ON	E0. E1	If the instrument is not equipped with Option 002, an error will result if El is sent via the HP-IB.
Comparator Run	OFF ON	GO . G1	
Comparator Limit	L/C/Z input D/Q/ESR/G input	L1' L2	
Comparator Bin Number	BIN1 BIN2 BIN3 BIN4 BIN5 BIN6 BIN7 BIN8 BIN8	N1 ° N2 N3 N4 N5 N6 N7 N8 N9	These codes are used when setting L/C/ Z  limits.
Comparator Limit Recall		LR	Refer to paragraph 3-100.
Comparator Limit Erase		ER	Comparator limits stored in all bins are cleared.
Output Data Abort		DA	HP-IB output data are erased from the output buffer.
Output Data Format	Displays A/B or Comparator Displays A/B/Comparator Display A or Comparator Display A/Comparator	F1 F2 F3 F4	Refer to paragraph 3-96 and Table 3-16.
Learn Mode		LN	Refer to paragraph 3-98.
Output Data Mode	ASCII BINARY	PO . P1	

Note: · indicates an initial control setting (Refer to paragraph 3-38.)

- [1] ASCII mode (Set using HP-IB remote program code "P0")
- l DISPLAY A/B

$$\frac{X}{(1)}\frac{X}{(2)}\frac{X}{(3)} \stackrel{\pm NN.NNN}{(4)} \stackrel{E\pm NN}{(5)} \stackrel{X}{(6)} \frac{X}{(7)} \frac{X}{(8)}$$

$$\frac{\pm \text{N.NNNN}}{(9)} \quad \frac{\text{E} \pm \text{NN}}{(10)} \quad \frac{\text{CR} \text{ CP}}{(11)}$$

- (1) Measurement circuit mode
- (2) Status of DISPLAY A
- (3) Function of DISPLAY A
- (4) Value of DISPLAY A (position of decimal point is coincident with display)
- (5) Unit of DISPLAY A
- (6) Unit of DISPLAY A
- (7) Status of DISPLAY B
- (8) Function of DISPLAY B
- (9) Value of DISPLAY B (position of decimal point is coincident with display)
- (10) Unit of DISPLAY B
- (11) Data Terminator
- 2 COMPARATOR (Option 002 only)

$$\begin{array}{ccc} X & X & N & \bigcirc \\ \hline (1) & (2) & (3) & \hline \end{array}$$

- (1) Status of L/C/|Z|
- (2) Status of D/Q/ESR/G
- (3) BIN number
- (4) Data Terminator

Note

Status and function data of DISPLAY A and DISPLAY B, and status of Comparator are each represented as one alphabetic character, as listed in Table 3-14.

#### Note

When measurement error code, OF, UF, CF or blank, is indicated on DISPLAY A or DISPLAY B, value of DISPLAY A or DISPLAY B ((4) or (9)) is output as follows:

OF	(overflow) ······	±19999E+20
UF	(underflow) ······	+00000E-20
CF	(change function)/	
bl	ank	+0000E-30

### Note

DISPLAY A and DISPLAY B ranges are expressed as an exponent as follows:

10 <sup>-12</sup> (p)······	E-12
10 <sup>-9</sup> (n)······	E-09
10 -6 (µ)·····	E-06
10 <sup>-3</sup> (m)······	E-03
10 0	E+00
10 <sup>3</sup> (k)	E+03
10 6 (M) ·····	E+06

#### Note

The data delimiter, bit switch 6 on the HP-IB Control Switch, is set at the factory to comma (,). This causes the instrument to output all data (DISPLAY A data, DISPLAY B data, and, if Comparator is used, Comparator data) as a continuous string. When the data delimiter is set to CR/LF, a carriage return and line feed signal is output after each field. This is useful when outputting data to certain peripherals, such as a printer.

# Note

The EOI signal is output with the LF signal.

- [2] PACKED BINARY mode (Set using HP-IB remote program code "P1")
- 1 DISPLAY A/B

1st byte 2nd byte 3rd byte

4th byte 5th byte

(B: 0 or 1)

- (1) Status of DISPLAY A
- (2) Status of DISPLAY B
- (3) Measurement Range
- (4) Value\* of DISPLAY A (not including decimal point and unit)
- (5) Value\* of DISPLAY B (not including decimal point and unit)
- \* Output data is the binary equivalent of the measured value.

#### Note

The first byte includes DISPLAY A status, DISPLAY B status, and measurement range. The value of the byte is output in decimal. For example, DISPLAY A status is OF (1), DISPLAY B status is "blank" (3), and measurement range is 5 (see Table 3-15), the byte will be as shown below.

01 11 0101 1 3 5

The decimal equivalent of this is 117. This is the value that will be output.

2 COMPARATOR (Option 002 only)

$$\begin{array}{c|c} BB & BB & BBBB \\ \hline (1) & (2) & \hline (3) \end{array}$$

- (1) Status of L/C/ | Z |
- (2) Status of D/Q/ESR/G
- (3) BIN number

#### Note

Status data of DISPLAY A and DISPLAY B, measurement range, and status and BIN number data of Comparator are each represented as a number as listed in Table 3-15.

# Note

Values displayed on DISPLAY A and DISPLAY B are output as number of counts. Actual measured values are obtained with measurement range and output data values.

### Note

Comparator data is output when the comparator is in RUN mode. When Fl, F3, or F5 is selected, if comparator is not in RUN mode, or if the comparator is not connected to the instrument, contents of output data is Type I.

## Note

If the instrument is set to TALK ONLY mode, the Output Data Format number will be briefly displayed on DISPLAY A (instead of the HP-IB address) when the instrument is turned on. The displayed number, however, will be the format number plus 50. For example, if the Output Data Format is F2, the number displayed on DISPLAY A at turn on will be 52.

SECTION III Model 4276A

Table 3-13. Data Output Codes for ASCII Mode

Item	Information	Code
Circuit Mode	• <del>•</del>	Р
	o- <u>CJ-</u> -W-o	S
Data Status of DISPLAY A/B	Normal Normal on Deviation Measurement Overflow Underflow Change Function Blank (used only for DISPLAY B)	N D O U C B
Function of DISPLAY A	L C HIGH SPEED L HIGH SPEED C  Z	L C L C
Function of DISPLAY B	D Q ESR G θ HIGH SPEED L*1 HIGH SPEED C*1	D Q R G T
Data Status of L/C/IZI for Comparator	Bin IN HIGH LOW Embedded Undefined	I H L E* <sup>2</sup> U* <sup>3</sup>
Data Status of D/Q/ESR/G for Comparator	Limit IN HIGH LOW Undefined	П* 3 Г І
Bin Number	Out of Bin BIN1 BIN2 BIN3 BIN4 BIN5 BIN6 BIN7 BIN8 BIN8	0 1 2 3 4 5 6 7 8

 $<sup>\</sup>ensuremath{^{*\,\text{\!1}}}$  HIGH SPEED C and HIGH SPEED L have the same output codes.

 $<sup>\</sup>ensuremath{^{\star\,2}}$  This code appears when the measurement value is between two continued bins.

<sup>\*3</sup> This code appears when DISPLAY A or B indicates "CF" or blank.

Table 3-14. Data Output Codes for PACKED BINARY Mode

Item	Information	Code
Data Status of DISPLAY A/B	Normal Overflow Underflow Change Function or Blank	0 1 2 3
Measurement Range	100mΩ 10pF/1Ω 100μH/100pF/10Ω 1mH/1nF/100Ω 10mH/10nF/1kΩ 100mH/100nF/10kΩ 1H/1μF/100kΩ 10H/10μF/1MΩ 100H/100μF/10MΩ 1kH/1mF 10mF	1 2 3 4 5 6 7 8 9 10 11
Data Status of L/C/ Z  for Comparator	Bin IN HIGH LOW Embedded or Undefined	0 1 2 3
Data Status of D/Q/ESR/G for Comparator	Bin IN HIGH LOW Undefine	0 1 2 3
Bin Number	Out of Bin BIN1 BIN2 BIN3 BIN4 BIN5 BIN6 BIN7 BIN8 BIN8	0 1 2 3 4 5 6 7 8

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Table 3-15. Data Output Format

D . 0		-	Output Da	ta	Outpo	ıt Mode
Data Ou Forma		DISPLAY A	DISPLAY B	COMPARATOR	ASCII	BINARY PACKED
F1	I	YES	YES	NO	VEC	VEC
LI	11	NO	NO	YES	YES	YES
F2	I	YES	YES	YES	VEC	VEC
F2	ΙΙ	YES	YES	NO	YES	YES
F3	I	YES	NO	NO	VEC	VEC
73	ΙΙ	NO	NO	YES	YES	YES
F4	I	YES	NO	NO	VEC	VEC
Г4	4 II YES		NO	YES	YES	YES
הר	I	NO	YES	NO	VEC	NO.
F5	ĬI NO		NO	YES	YES	NO
Γ.	I	NO	YES	NO	VEC	NO
F6	II	NO	YES	YES	YES	NO

I: Without the comparator (Model 16064A)

II: Using the comparator (Model 16064A).

### 3-97. LEARN MODE DATA

3-98. All front panel settings and comparator key settings are output from the 4276A when the program code "LN" is used (refer to Figure 3-29). The data is output in the following format:

$$\frac{\text{FRnnnnEN}}{(1)} \ \frac{\text{An}}{(2)} \ \frac{\text{Bn}}{(3)} \ \frac{\text{Cn}}{(4)} \ \frac{\text{Dn}}{(5)} \ \frac{\text{Fn}}{(6)} \ \frac{\text{Mn}}{(7)} \ \frac{\text{Pn}}{(8)}$$

$$\frac{\text{Rn}}{(9)} \frac{\text{Sn}}{(10)} \frac{\text{Tn}}{(11)} \frac{\text{Un}}{(12)} \frac{\text{Vn}}{(13)} \frac{\text{Xn}}{(14)} \frac{\text{BI} \pm \text{nnnnEN}}{(15)}$$

$$\underbrace{\frac{\operatorname{En}}{(16)}}_{(17)}\underbrace{\frac{\operatorname{Cn}}{(18)}}_{(18)}\underbrace{\frac{\operatorname{Nn}}{(19)}}_{(19)}\underbrace{\frac{\operatorname{CR}}{(20)}}_{(20)}$$

- (1) Test Frequency Setting
- (2) A1 A5: DISPLAY A Function
- (3) B1 B3: DISPLAY B Function
- (4) C1 C3: Circuit Mode
- (5) D0, D1: Data Ready
- (6) F1 F4: Output Data Format
- (7) M1 M3: Measurement Speed
- (8) P0, P1: Output Data Mode (ASCII or Packed Binary)
- (9) R1 R9, RA, RB: LC | Z | Range
- (10) S0, S1: Self Test
- (11) T1, T2: Trigger Mode
- (12) U0, U1: Auto Range
- (13) V1, V2: Test Signal Level
- (14) X0, X1: Deviation Measurement
- (15) DC Bias Setting
- (16) E0, E1: Comparator Enable
- (17) GO, Gl: Comparator Run
- (18) L1, L2: Comparator Limit Input
- (19) N1 N9: Comparator Bin Number for L/C/|Z|
- (20) Data Terminator

#### Note

DC Bias data is not output when DC Bias option (OPTION 001) is not installed. Similarly, when the comparator (OPTION 002) is not installed, comparator data is not output.

### Note

Don't open the UNKNOWN terminals no test fixture or test leads when LEARN mode data is output in AUTO range. If so, measurement range is not fixed in some cases. there is no problem when a test fixture is connected to the UNKNOWN terminals or when measurement range is set to MANUAL mode.

## 3-99. RECALL COMPARATOR LIMIT DATA

3-100. Low and high bin limits can be output from the 4276A when the program code "LR" is used (refer to Figure 3-31). The  $L/C/\mid Z\mid$  limits for the designated bin are output when code "L1" is used, D/Q/ESR/G limits are output. The data is output in the following format:

- (1) Value of Low Limit (position of decimal point is coincident with display)
- (2) Value of High Limit (position of decimal point is coincident with display)
- (3) Data Terminator

# 3-101. SERVICE REQUEST STATUS BYTE

3-102. The 4276A outputs an RQS (Request Service) signal whenever it is set to one of the five possible service request states. Figure 3-27 shows the contents of the Status Byte.

Bit	8	7	6	5	4	3	2	1
Content		RQS		Error	Trigger Too Fast	Zero Offset Self Test End	Syntax Error	Data Ready

Bit 7 (RQS) indicates whether or not a service request exists. Bits 6 and 8 are always zero (0). Bits 1 through 5 identify the type of service request. Following are the service request states of the 4276A:

- (1) Bit 1: This bit is set when measurement data is ready for output.
- (2) Bit 2: This bit is set when the remote program contains a syntax error.
- (3) Bit 3: This bit is set when Zero Offset or Self Test is completed under remote control.

- (4) Bit 4: This bit is set when the 4276A is externally triggered before the measurement has been completed.
- (5) Bit 5: 1 This bit is set when the 4276A has one of the following operation errors:

OFF, E10, E11, E13, E14, E15, E16, E17, E18, E10

2 If Self Test is set to ON, this bit is set when the instrument fails Self Test.

E37 - E39, E40 - E45

Figure 3-27. Status Byte for the 4276A.

# 3-103. PROGRAMMING GUIDE FOR 4276A

3-104. Sample programs that can be run on the HP-85,  $9835\,\mathrm{A/B}$ ,  $9845\,\mathrm{B}$ ,  $9826\,\mathrm{A}$ , or  $9836\,\mathrm{A}$  are given in Figures 3-28 through 3-31. These programs are listed in Table 3-16.

### Note

Controller-specific HP-IB programming information is given in the controller's programming manual.

# Note

Following equipment is required to run the sample programs:

- (1) 4276A LCZ Meter
- (2) HP-85 Personal Computer 00085-15003 I/O ROM
- (3) 82937A HP-IB INTERFACE

or

- (2) 9835A/B Desktop Computer 98332A I/O ROM
- (3) 98034A HP-IB INTERFACE CARD

or

- (2) 9826A Desktop Computer
- (3) 10833B HP-IB INTERFACE CABLE

or

- (2) 9836A Desktop Computer
- (3) 10833B HP-IB INTERFACE CABLE

Table 3-16. Sample Programs Using HP-85

Sample Program	Figure	Description
1	3-28	Remote control and data output program
2	3-29	How to use remote program code
3	3-30	How to input low and high bin limits for the Comparator.
4	3-31	How to use remote program code

Sample Program 1

# Description:

This program has three capabilities:

- (1) Control of the 4276A via the HP-IB
- (2) Trigger of the 4276A via the HP-IB
- (3) Data output from the 4276A via the HP-IB

- 10 REMOTE 717
- 20 CLEAR 717
- 30 DIM A\$[50]
- 40 OUTPUT  $\frac{717}{(1)(2)}$ ; " A2B1T2P0F1" (3)
- 50 OUTPUT 717; "FR10EN"
  (4)
- 60 <u>OUTPUT 717; "EX"</u> (5)
- 70 ENTER 717; A\$
- 80 DISP A\$
- 90 PRINT A\$
- 100 END

- (1) HP-IB INTERFACE Select Code (82937A or 98034A)
- (2) HP-IB Address of the 4276A
- (3) Program codes for the 4276A (refer to Table 3-13)
- (4) Program codes for parameter setting of the 4276A (refer to paragraph 3-96)
- (5) This is equivalent to: TRIGGER 717

If program code "P1" is used, refer to the following program:

# Program:

- 10 REMOTE 717
- 20 CLEAR 717
- 30 OUTPUT 717; "A2B1T2P1F1"
- 40 OUTPUT 717; "EX"
- 50 ENTER 717 USING " $\frac{\%}{(1)}$ ,  $\frac{B}{(2)}$ ,  $\frac{W}{(3)}$ ,  $\frac{W}{(3)}$ "; A, B, C
- 60 DISP A;B;C
- 70 PRINT A;B;C
- 80 END
- (1) ENTER terminator. "#" can also be used.
- (2) Specifier for entering one byte (8-bit) of binary data
- (3) Specifier for entering two bytes (16-bit) of binary data

Figure 3-28. Sample Program 1 (Sheet 2 of 2).

# Sample Program 2

# Description:

The remote program code "LN" can be used to read the front panel control settings and comparator settings. This program shows how to use "LN."

- 10 REMOTE 717
- 20 CLEAR 717
- 30 DIM A\$[60]
- 40 OUTPUT 717; "LN"
- 50 ENTER 717; A\$
- 60 DISP A\$
- 70 PRINT A\$
- 80 END

Figure 3-29. Sample Program 2.

# Sample Program 3

# Description:

This program shows how to input low and high bin limits via the HP-IB when the instrument is equipped with Option 002.

```
REMOTE 717
10
20 CLEAR 717
30 DIM A $ [ 50 ]
40 OUTPUT 717; "A2B1R4T2P0F2"
                      (1)
   OUTPUT 717; "FR10EN"
50
   OUTPUT 717; "E1G0ER"
60
    OUTPUT 717; "L1N1LL. 995ENLH. 998EN"
                   (2)
   OUTPUT 717; "N2LL1 ENLH1. 1EN"
80
   OUTPUT 717; "N3LL 1.1001ENLH1.2EN"
90
100 OUTPUT 717; "L2LL0ENLH. 001EN"
110 OUTPUT 717; "G1"
120 OUTPUT 717: "EX"
130 ENTER 717; A$
140 DISP A$
150 PRINT A$
160 END
```

- (1) Measurement range must be set.
- (2) Program codes for comparator setting
- (3) Program codes for inputting low and high bin limits

# Sample Program 4

# Description

The remote program code "LR" can be used to recall the high and low limits for each bin. This program shows how to use "LR."

```
REMOTE 717
10
20
    DIM A$[30]
    OUTPUT 717; "E1G0"
30
    FOR I = 1 TO 9
40
50
    OUTPUT 717; "L1N"; I, "LR"
60
    ENTER 717; A$
    PRINT A$
70
    NEXT I
80
90
    OUTPUT 717; "L2LR"
100
    ENTER 717; A$
110
    PRINT A$
120
    END
```

Figure 3-31. Sample Program 4.

# 3-105. OPTIONS

3-106. Options are standard modifications to the instrument that implement user's special requirements for minor functional changes. Operating instructions for the 4276A's options (except rack mount and handle installation kit options) and associated information are described in the following paragraphs.

3-107. Two options are available, as listed in the following tables:

Option No.	Option			
001	Internal Dc Bias			
002	Comparator/Handler Interface			

# Option contents are as follows:

Option NO.	Contents			
001	HP Part No.04276-66522			
002	HP 16064 A			

3-108. OPTION 001 INTERNAL DC BIAS (-40V to +40V)

3-109. Option 001 adds an internal de bias supply variable from .00 volts to ±40.0 volts. The dc bias voltage can be controlled manually from the frontpanel or remotely via the HP-IB. Manual control is described in Figure 3-32. For de bias applications under HP-IB control, refer to Figure 3-33. The internal dc bias source has two ranges and a maximum resolution of 10mV. Refer to Table 3-17. Output from the bias source is automatically set to 0V each time the instrument is turned on or when the CLEAR command is sent via the HP-IB. DC bias voltage is applied to the DUT only when the DC BIAS select switch on the rear panel is set to INT and switch on the front panel the DC BIAS is set to ON. If the DC BIAS switch is set to OFF, OFF will be briefly displayed on the FREQUENCY/DC BIAS display each time a new bias voltage is set. The dc bias voltage actually applied to the DUT depends on the impedance of the DUT and in most cases will be less than the voltage value displayed on the FREQUENCY/DC BIAS display. By connecting a DVM or an oscilloscope to the EXT INPUT/INT MONITOR connector on the rear panel, the dc bias voltage actually applied across the DUT can be monitored. Refer to Figure 3-34.

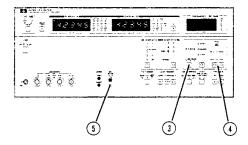
Table 3-17. Bias Voltage Resolution

Bias Voltage Range	Resolution
OV∼± 9.99 V	10 m V
$\pm 10.0 \mathrm{V} \sim \pm 40.0 \mathrm{V}$	100 m V

### Note

For the option 001 operation, set the DC BIAS select switch on the rear-panel to INT.

# OPTION 001 INTERNAL DC BIAS OPERATION



- 1. Set the DC BIAS select switch (2) to INT.
- 2. Connect the 16047A Test Fixture to the UNKNOWN terminals.

### Note

Any of the test fixtures and test leads listed in Table 1-3 can be used for measurements requiring dc bias.

- 3. Turn on the 4276A.
- 4. Perform OPEN and SHORT Zero Offset adjustments as described in paragraph 3-48.
- Set the instrument's front panel controls as appropriate for the desired measurement.
- 6. Press the FREQ/DC BIAS select key ①. The DC BIAS lamp will come on.
- Set the desired voltage by pressing the appropriate FREQ/DC BIAS control key
   The voltage value will be displayed on the FREQUENCY/DC BIAS display.

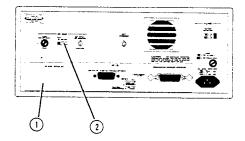
### Note

OFF will be briefly displayed on the FREQUENCY/DC BIAS display when the FREQ/DC BIAS control key is released, if the DC BIAS ON/OFF switch (5) is set to OFF.

8. Connect the DUT to the test fixture.

#### CAUTION

DO NOT CONNECT A CHARGED DUT TO THE TEST FIXTURE. DOING SO MAY DAMAGE THE INSTRUMENT.



- 9. Set the DC BIAS switch 5 to ON.
- 10. If you are measuring a capacitive DUT, allow sufficient time for the DUT to charge up to the applied voltage.
- 11. Read the measured values displayed on DISPLAY A and DISPLAY B.
- 12. Set the DC BIAS switch (5) to OFF.
- 13. Wait until the voltage across the DUT return to 0V.
- 14. Remove the DUT from the test fixture.

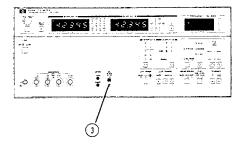
## Note

For reasons of safety and measurement accuracy, the voltage actually applied to the DUT should be monitored. Refer to Figure 3-34.

#### Note

When the DC BIAS switch on the front panel has been set to ON and the desired bias voltage is entered, the instrument automatically takes a wait time of approximately 0.8 seconds before outputting the bias voltage (after completion of the bias data input). Accordingly, takes it approximately (0.8 seconds + bias settling time) for the bias voltage to be applied to the DUT as well as to be settled after the bias data has been set. For the bias settling time, refer to Figure 1-2 Supplemental Performance Characteristics.

# OPTION 001 INTERNAL DC BIAS HP-IB OPERATION



# [HP-IB Operation]

The following procedure is an example of dc bias remote control via the HP-IB.

- 1. Set the DC BIAS select switch (2) to INT.
- 2. Connect the 16047A Test Fixture to the UNKNOWN terminals.

## Note

Any of the test fixtures and test leads listed in Table 1-3 can be used for measurements requiring dc bias.

- 3. Turn on the 4276A.
- 4. Perform OPEN and SHORT Zero Offset adjustments as described in paragraph 3-48.
- 5. Set the DC BIAS switch 3 to ON.

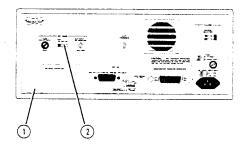
# Note

The dc bias voltage is automatically set to 0V each time the instrument is turned on.

- 6. Set the front panel control via the  $\mbox{HP-IB.}$ 
  - \* Example of setting the instrument for a C-D measurement at 10kHz, external trigger.

CLEAR 717 OUTPUT 717; "A2B1FR10ENF1T2"

- 7. Connect the DUT to the test fixture.
- 8. Set the desired dc bias voltage via the HP-IB.
  - \* Example of setting a dc bias voltage of +10V.



# OUTPUT 717; "BI10EN"

- 9. Wait until the dc bias voltage settles.
  - \* Example of programming a 2-second wait.

# WAIT 2000

10. Trigger the instrument via the HP-IB.

OUTPUT 717; "EX" or TRIGGER 717

11. Read and print the measured values.

ENTER 717; A, B PRINT A, B

12. Set the bias voltage to 0V via the HP-IB.

OUTPUT 717; "BI 0EN"

- 13. Wait until the dc bias voltage returns to 0V.
  - \* Example of programming a 1-second wait.

WAIT 1000

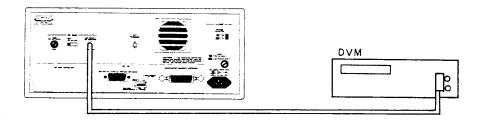
14. Remove the DUT from the test fixture.

## Note

The above remote programming examples can be used on the HP Model 85 (with 00085-15003 I/O ROM), Model 9835A, Model 9845B/C, Model 9826A, and Model 9836A.

In the above examples, HP-IB address 17 was used.

# INTERNAL DC BIAS VOLTAGE MONITOR



The internal dc bias voltage is monitored by a DVM or an oscilloscope at the EXT INPUT/INT MONITOR connector on the rear panel.

### Note

The dc bias voltage monitored at the EXT INPUT/INT MONITOR connector may contain a small ac component.

When the DUT impedance is higher than  $100 k\Omega$ , the monitored voltage is equal to the dc voltage source voltage, and to the voltage applied to the DUT. These voltages, however, are different when the DUT impedance is less than  $100 k\Omega$ . The following paragraph describes how to measure the actual bias voltage across the DUT.

- 1.  $R_1/R_2/R_L$  Detection (See Figure A.)
  - (a) Set the TEST SIG LEVEL to LOW.
  - (b) Set the LC |Z| range so that the range resistor value will be  $100\Omega$ . Refer to Figure 3-16.
  - (c) Set the DC BIAS voltage to +5V on the FREQUENCY/DC BIAS display.
  - (d) Connect nothing to the test fixture.
  - (e) Set the DC BIAS switch on the front panel to ON.
  - (f) Measure the monitor voltage (V₀) at the EXT INPUT/INT MONITOR connector.
  - (g) Connect a reference resistor (R<sub>0</sub>) (e.g.,  $100\Omega\pm1\%$ ) to the test fixture.
  - (h) Measure the dc voltages at the HIGH and LOW terminals of the test fixture and at the EXT INPUT/INT MONITOR connector ( $V_H$ ,  $V_L$ , and  $V_K$ ).

#### Note

Connect the LOW terminal of the DVM or the oscilloscope to the GUARD terminal of the instrument.

(i) Calculate the resistances, R<sub>1</sub>, R<sub>2</sub>, and R<sub>L</sub>, using the following equations:

$$R_{1} = (V_{0} - V_{K}) \cdot R / (V_{H} - V_{L})$$

$$R_{2} = (V_{K} = V_{H}) \cdot R / (V_{H} - V_{L})$$

$$R_{L} = V_{L} \cdot R_{0} / (V_{H} - V_{L})$$

- 2. Actual Bias Voltage/Current Measurement
  - (a) Connect nothing to the test fixture.
  - (b) Measure the monitor voltage (V<sub>0</sub>).
  - (c) Connect the desired sample to the test fixture.
  - (d) Measure the monitor voltage (V<sub>M</sub>).
  - (e) Calculate the actual voltage applied to the DUT (V) or the actual current through the DUT (I) using the following equations:

$$I = (V_0 - V_M)/R_1$$
  
 $V = V_0 - (R_1 + R_2 + R_L) \cdot I$ 

## Note

Repeat step 2 each time the DUT is changed since the monitor voltage  $(V_M)$  depends on the DUT impedance.

Figure 3-34. Internal DC Bias Voltage Monitor (Sheet 1 of 2).

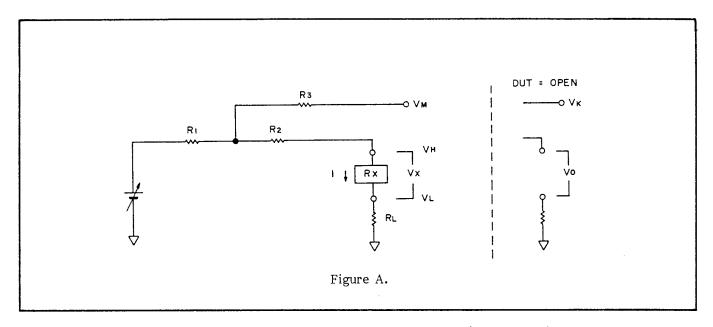


Figure 3-34. Internal DC Bias Voltage Monitor (Sheet 2 of 2).

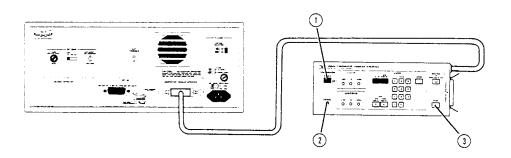
3-110. OPTION 002 COMPARATOR/HANDLER INTERFACE

3-111. Option 002 equips the standard 4276A with a comparator function and a handler (component sorter) interface capability. The comparator provides go/no-go testing and ten-bin sorting. The handler interface is for control of an automatic component handler.

3-112. Up to nine sets of high/low limits for L, C, or |Z| measurement, and one set of high/low limits for D, Q, ESR, or G measurement can be keyed in from the 16064A keyboard or entered via the HP-IB. When measurement is made, the comparator compares the measured values displayed on DISPLAY A and DISPLAY B with the stored high/low limits. If the measured values fit any set of limits, the bin number for that set is displayed on the FREQUENCY/DC BIAS display. If the measured values do not fit any of the limits, zero (0), the number for the out-of-limits bin, is displayed. Go/no-go decisions are indicated by two sets of LOW/IN/HIGH LED lamps on the 16064A Comparator/Handler Interface operation is described in Figures 3-35 through 3-38.

3-113. The 16064A has a 36-pin female Amphenol connector for interfacing with an The 16064A automatic component handler. comparison results-LOW/IN/HIGH sends decisions and bin number-to the handler, and receives control signals via a user-fabricated interface cable constructed using the furnished 36-pin male Amphenol connector 1251-0084). Pin assignments are given in Figure 3-38. For complete information, refer to the 16064A Operation Note.

### OPTION 002 COMPARATOR OPERATION



- 1. Connect the Model 16064A COMPARATOR/HANDLER INTERFACE to the COMPARATOR/HANDLER INTERFACE connector on the 4276A's rear-panel.
- Connect the desired test fixture to UNKNOWN terminals.
- 3. Turn on the instrument.
- 4. Perform OPEN and SHORT Zero Offset adjustments as described in paragraph 3-48.
- Set the front panel controls as appropriate for the desired measurement.
- Press the ENABLE key on the 16064A. The LED lamp at the center of the key should come on.

# Note

If El6 is displayed on DISPLAY A, press the ERASE button on the 16064A to erase previously stored limits.

- 7. Enter the high/low limits for L/C/|Z| or D/Q/ESR/G as described in Figure 3-36.
- 8. Press the RUN key on the 16064A. The comparator will then begin comparing all measured values with the high/low limits entered in step 7. The appropriate LED lamps--LOW, IN, HIGH--will be lit and the number of the bin whose high/low limits fit the measured values will be displayed on the FREQUENCY/DC BIAS display.

# Example:

If the bin limits listed in Tables A and B are entered, the measured values listed in Table C will cause the comparison results shown in Table D.

Table A. Limits for L/C/|Z|

BIN No.	LOW LIMIT	HIGH LIMIT		
1	1 nF	1.1 nF		
2	1.1 nF	1.2 nF		
3	1.2 n F	1.3 n F		
4	1.3 nF	1.4 nF		
5	1.4 nF	1.5 n F		
6	2 nF	2.5 n F		
7	2.5 nF	3 n F		

Table B. Limits for D/Q/ESR/G

LOW LIMIT	HIGH LIMIT
. 01	. 05

Table C. Measured Values

Sample	Measured Value		Sample		asured lue
1	С	1.22 nF	6	С	1.1 nF
1	D	.013	U	D	.02
2	С	1.08 nF	7	С	1.18 nF
۷	D .005	D	.071		
2	C -8 nF	С	4.1 nF		
3	D	.025	0	D	.033
4	С	2.75 nF	9	С	1.5 nF
4	D	•06	9	D	.029
5	С	.95 nF	10	С	1.72 n F
	D	.055	10	D	.025

Table D. Comparison Results

Sample	La	C/ mp	Z н 1 G H	La			FREQUENCY/ DC BIAS Display
1	0	•	0	0	•	0	3
2	0	•	0	•	0	0	
3	•	0	0	0	•	0	0
4	0	•	0	0	0	•	0
5	•	0	0	0	0	•	
6	0	•	0	0	•	0	1
7	0	•	0	0	0	•	
8	0	0	•	0	•	0	0
9	0	•	0	0	•	0	5
10	•	0	•	0	•	0	8

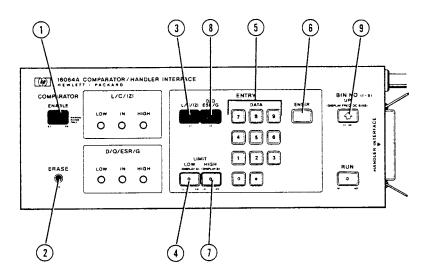
Note

LOW and HIGH limits are inclusive; that is, if the measured value is exactly equal to the LOW or HIGH limit of a bin, the measured value fits the limits for that bin. Also, if a measured value fits the limits of more than one bin (bin limits overlap), the comparator selects the bin with the lower number. An example follows.

Bin 1: 100pF to 200pF Bin 2: 150pF to 250pF Measured Value: 190pF Selected Bin: Bin 1 Note

If the LOW/HIGH limits for D/Q/ESR/G are not entered, or when the instrument is set to HIGH SPEED L or HIGH SPEED C, the IN lamp for D/Q/ESR/G will be always lit. D/Q/ESR/G comparison is not performed, however.

# COMPARATOR LIMIT SETTING



- 1. Press the ENABLE key ①. The LED lamp at the center of the key should come on.
- 2. Press the ERASE button ② to erase previously stored limits. One (1) will be displayed on the FREQUENCY/DC BIAS display.

## [L/C/|Z| Limit Entry]

- 3. Press the L/C/|Z| key 3.
- 4. Press the LIMIT LOW key 4.
- 5. Key in the desired LOW limit using the DATA keys (3). The LOW limit value will be displayed on DISPLAY A.
- 6. Press the ENTER key (3). The LOW limit will be stored for bin 1. Also, the maximum allowable value that can be entered for the HIGH limit on the present LC | Z | RANGE will be displayed on DISPLAY B.

## Note

If the LOW or HIGH limit is higher than the full scale value of the existing LC | Z | RANGE, E18 will be briefly displayed on DISPLAY A when the ENTER key is pressed. Re-enter the limits correctly.

7. Press the LIMIT HIGH key 1.

- 8. Key in the desired HIGH limit using the DATA keys (5). The HIGH limit value will be displayed on DISPLAY B.
- 9. Press the ENTER key (6). The HIGH limit will be stored for bin 1.
- 10. Press the BIN NO UP key (3). Two (2) will be displayed on the FREQUENCY/DC BIAS display.
- 11. Repeat steps 4 through 9 to enter the LOW and HIGH limits for bin 2.
- 12. Repeat steps 10 and 11 for bins 3 through 9.

# [D/Q/ESR/G Limit Entry]

13. Press the D/Q/ESR/G key 3.

# Note

When D/Q/ESR/G limits are being entered, no bin number is displayed on the FREQUENCY/DC BIAS display.

- 14. Press the LIMIT LOW key 4.
- 15. Key in the desired LOW limit using the DATA keys (5). The LOW limit value will be displayed on DISPLAY A.
- 16. Press the ENTER key (§). The LOW limit will be stored. Also, the maximum allowable value that can be entered for the HIGH limit will be displayed on DISPLAY B.

### Note

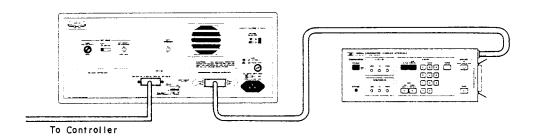
If the LOW or HIGH limit is higher than the full scale value of the existing DISPLAY B range, El8 will be briefly displayed on DISPLAY A when the ENTER key is pressed. Re-enter the limits correctly.

- 17. Press the LIMIT HIGH key 1.
- 18. Key in the desired HIGH limit using the DATA keys (5). The HIGH limit value will be displayed on DISPLAY B.
- 19. Press the ENTER Key.

Note

Press the ERASE button ②, erases the high/low limits of all bins.

# OPTION 002 COMPARATOR HP-IB OPERATION



## [HP-IB OPERATION]

- 1. Connect the Model 16064A COMPARATOR/HANDLER INTERFACE to the COMPARATOR/HANDLER INTERFACE connector on the 4276A's rear-panel.
- Connect the desired test fixture to the UNKNOWN terminals.
- 3. Turn on the instrument.
- 4. Perform OPEN and SHORT Zero Offset Adjustments.
- 5. Set the front panel controls as appropriate for the desired measurement and enable the 16064A via the HP-IB.
  - \* Example of setting C-D measurement, lnF range, and 10kHz test frequency

DIM A\$[1],B\$[1] CLEAR 717 OUTPUT 717;"A2B1R4T2FR10EN" OUTPUT 717;"E1ER"

- 6. Enter the LOW/HIGH limits for L/C/|Z| via the HP-IB.
  - \* Example of setting a low limit of .950nF and a high limit = 1.1nF

OUTPUT 717;"LL.95ENLH1.1EN"

If necessary, enter the limits for the next bin (Bin 2).

\* Example of setting bin 2's low limit to 1.1001nF and high limit to 1.2nF

OUTPUT 717;"N2"
OUTPUT 717;"LL1.1001ENLH1.2EN"

#### Note

The same setting can be made by the following program:

OUTPUT 717;"N2"
OUTPUT 717;"LH1.2EN"

- 7. Enter the limits for D/Q/ESR/G via the HP-IB.
  - \* Example of setting a low limit of .0000 and a high limit of .005

OUTPUT 717;"L2"
OUTPUT 717;"LL0 ENLH.005 EN"

# Note

The same setting can be made by the following program:

OUTPUT 717;"L2" OUTPUT 717;"LH.005EN"

# Note

Comparator operations can be done without high/low limits for D/Q/ESR/G.

Figure 3-37. Option 002 Comparator HP-IB Operation (Sheet 1 of 2).

- 8. Start the comparator operation by HP-IB program.
  - \* Example of starting the comparator operation:

OUTPUT 717;"G1"

- 9. Connect the DUT to the test fixture.
- 10. Trigger the instrument via the HP-IB.
  - \* Example of triggering the instrument:

OUTPUT 717;"EX"

Or

TRIGGER 717

If necessary, read the comparison results via the HP-IB.

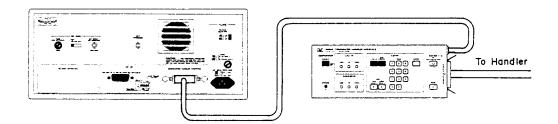
ENTER 717;A\$,B\$,N PRINT A\$,B\$,N

Note

The HP-IB address code in the above examples is 17 (10001).

Figure 3-37. Option 002 Comparator HP-IB Operation (Sheet 2 of 2).

# OPTION 002 HANDLER INTERFACE OPERATION



The 16064A outputs four types of signals to the component handler.

- (1) Comparison result signals (LCHI, LCIN, LCLO, DQHI, DQIN, DQLO)
- (2) Bin number signals (BIN1 ... BIN9, OUT-OF-BIN)
- (3) DUT change signal (INDEX)
- (4) Comparison complete signal (EOM)

Type (1) signals correspond to the LOW/IN/HIGH LED lamps on the 16064A keyboard. Type (1) signals are divided into two groups of three. When the signal line corresponding to the lit LED lamp goes LOW, the other signal lines in that group stay HIGH.

Type (2) signals correspond to the bin numbers displayed on the FREQUENCY/DC BIAS display. When the signal line corresponding to the displayed bin number goes LOW, the other signal lines stay HIGH.

The type (3) signal, INDEX, goes LOW when the 4276A has completed the analog portion of the measurement. The DUT can be disconnected from the measurement terminals and the next one can be connected. Comparison results, however, are not yet valid.

The type (4) signal, EOM, goes LOW when the 4276A has completed the measurement and the comparator has made a judgement. Comparison results are now valid.

All signals are negative true, and all are from TTL open-collector outputs. Pull-up resistors are installed. TTL voltage levels or higher voltages (up to 30V) are possible by changing a few jumper settings inside the 16064A. Refer to the 16064A Operating Note for details.

Signals sent from the external component handler to the 16064A are a trigger signal (EXT TRIG) that starts measurement and a key lock signal (KEY LOCK) that disables all control keys during comparator operation. To trigger the 4276A, apply a LOW signal (at least 100 ps duration) to the EXT TRIG line. To disable the control keys of the 4276A and 16064A, apply a LOW signal to the KEY LOCK line.

#### Note

The INDEX and KEY LOCK signals are not mandatory for comparator/handler interface applications.

# Note

More information on the Option 002 Handler Interface is given in the 16064A Operating Note.

# Note

When the RUN key on the 16064A is pressed to start the comparator's operation, the  $\overline{OUT}$  OF  $\overline{BIN}$  line, pin 10 of the Handler Interface connector, is initially set to its active level (LOW).

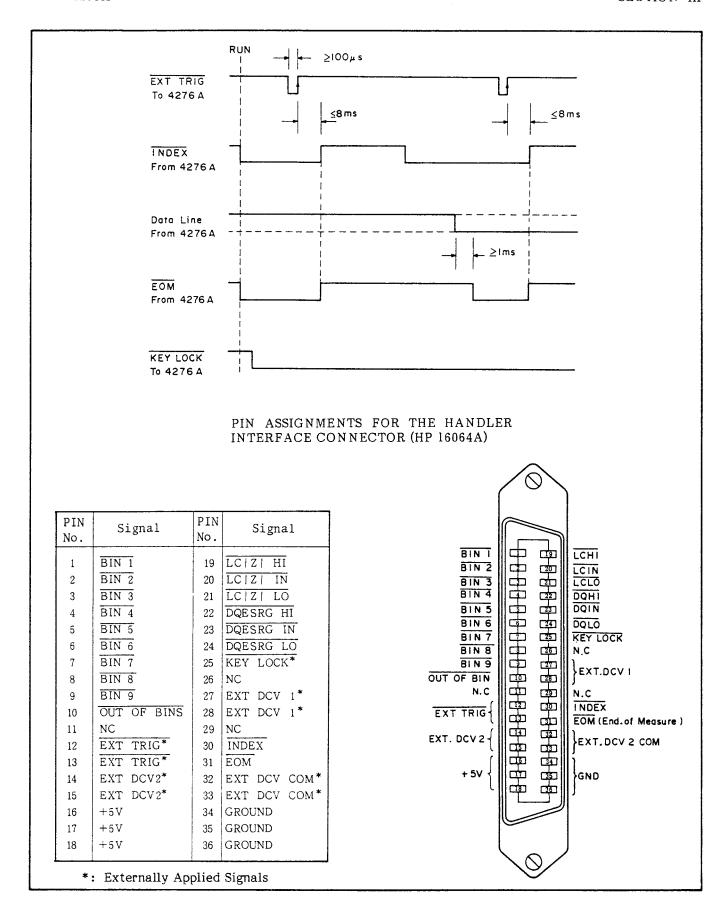


Figure 3-38. Option 002 Handler Interface Operation (Sheet 2 of 2).

SECTION III Model 4276A

3-114. POWER FAILURE MONITOR SIGNAL

3-115. If the instrument experiences a transient power failure such that the output voltage from the +5V power supply on the Al board drops below +4.8V, all measurement circuits will be reset when the voltage returns to normal. Usually the duration of a transient power failure and the time required for the reset operation are so short that the operator may not be aware that the instrument has experienced a power failure. In most applications, this is not a problem, because the instrument's continuous memory function restores all settings (except the DC bias setting on option 001 instruments) after the reset operation. In some applications, though, a transient power failure, however brief, can adversely affect measurement results. In such applications it is important to know whether a power failure has occurred.

The 4276A is equipped with a power failure monitor signal (PWF) which can be used to inform the operator or peripherals of a transient power failure. The PWF signal is of the open collector type, and therefore requires an external (user-supplied) voltage source (+5V max.) and a pull-up resistor. Also, two wires must be connected to a jumper inside the 4276A and brought out through a hole in the rear panel. Complete instructions are given in paragraph 2-31.

Model 4276A SECTION III

## Number of Display Digits

The number of display digits on DISPLAY A/B for C-D measurement, C-ESR/G measurement, L-D measurement, L-ESR/G measurement, and  $|Z| - \theta$  measurement is listed in Tables 1 through 5. The number of display digits for a Q measurement depends on the D value and is listed in Table 6.

#### Note

Tables 1 through 5 are valid under the following conditions:

(1) Circuit Mode: AUTO

(2) Test Signal Level: HIGH

#### Note

When the test signal level is set to LOW, the number of display digits is one less than the number of display digits when the test signal level is set to HIGH.

## Note

Alphabetic characters used in the Tables represent the number of display digits as follows:

Symbol	Display
A	88888
В	8888
С	888.
D	8800

SECTION III Model 4276A

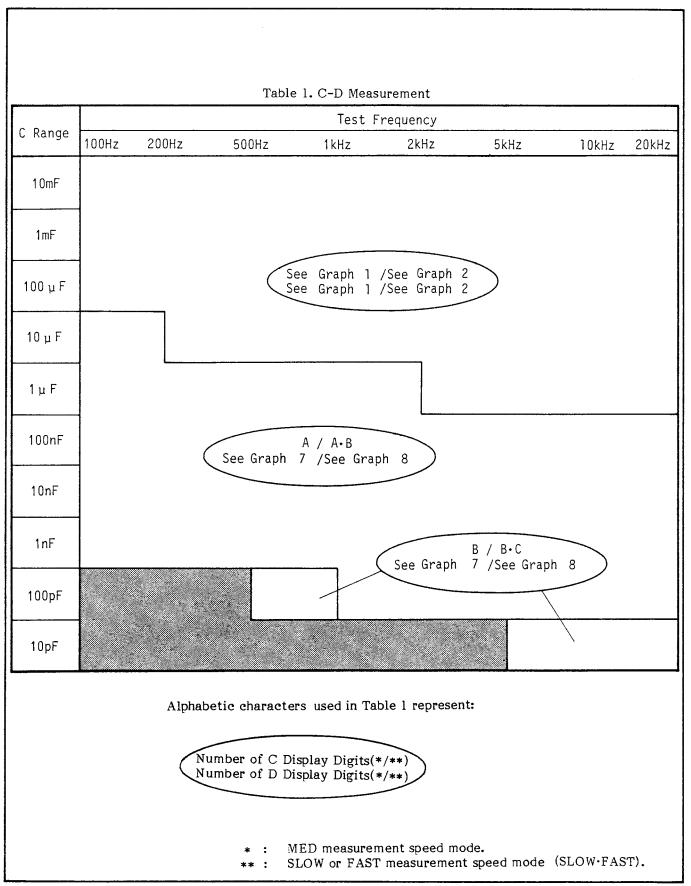


Figure 3-39. Number of Display Digits (Sheet 2 of 13).

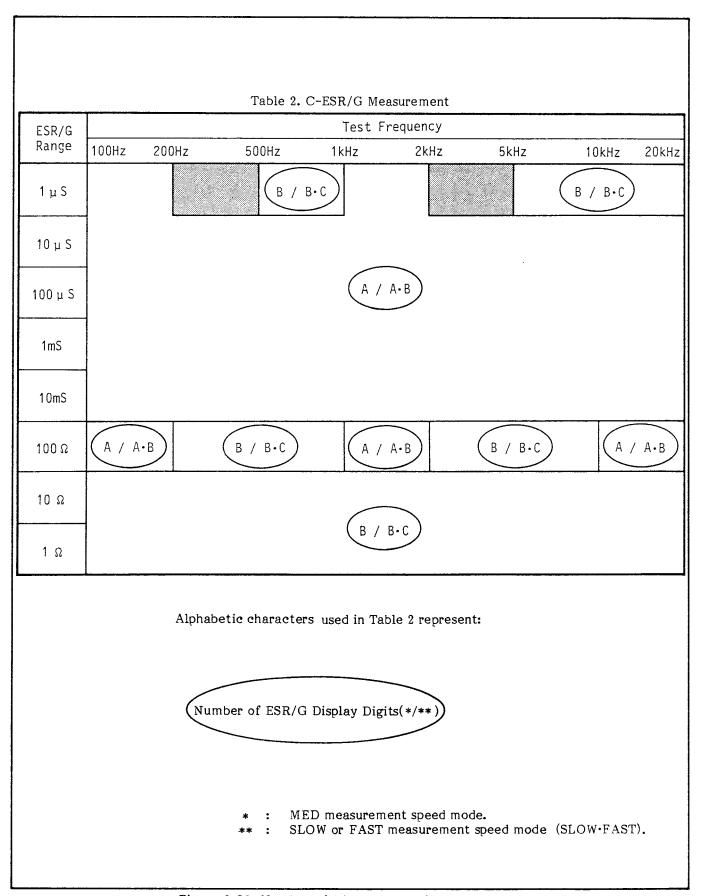


Figure 3-39. Number of Display Digits (Sheet 3 of 13).

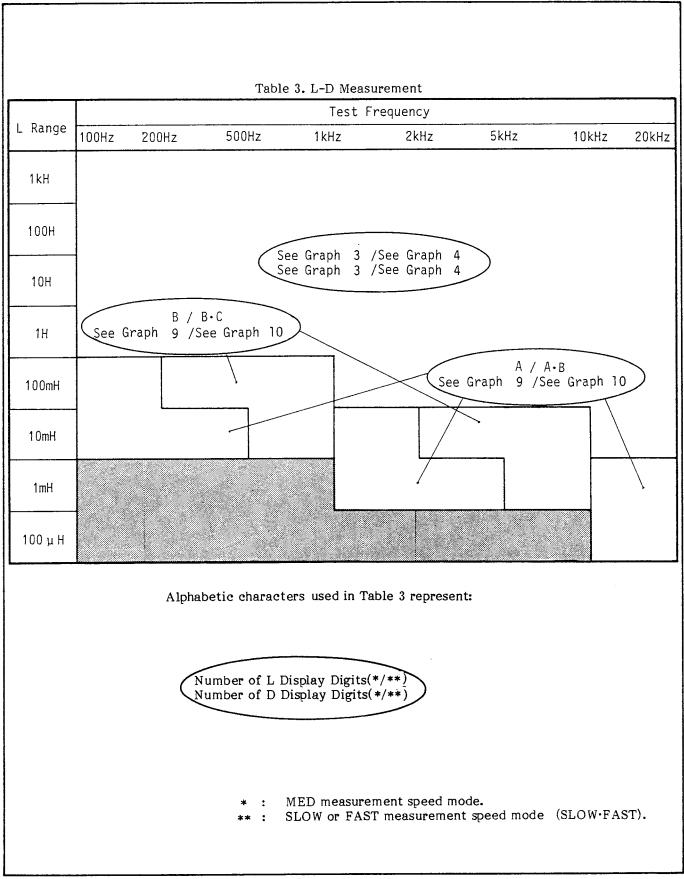
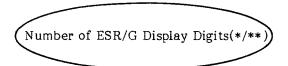


Figure 3-39. Number of Display Digits (Sheet 4 of 13).

Table 4. L-ESR/G Measurement Test Frequency ESR/G Range 2kHz 100Hz 200Hz 500Hz 1kHz 5kHz 10kHz 20kHz B / B · C 10 μ S 100 μ S 1mS A / A · B 10mS C / C • D C / C • D 100 Ω A / A • B A / A•B A / A · B B / B·C B / B·C 10 Ω

Alphabetic characters used in Table 4 represent:



\* : MED measurement speed mode.

\*\* : SLOW or FAST measurement speed mode (SLOW.FAST).

Figure 3-39. Number of Display Digits (Sheet 5 of 13).

Table 5. Z  $-\theta$  Measurement

7 0				Test F	requency			
Z Range	100Hz	200Hz	500Hz	1kHz	2kHz	5kHz	10kHz	20kHz
10M Ω								
1Μ Ω								
100k Ω			Se Se	e Graph 5 e Graph 5	/See Graph 6 /See Graph 6			
10k Ω								
1kΩ								
100 Ω				A /	A•B			
10 Ω			See	e Graph 11 /	See Graph 12	<i></i>		
1 Ω				B /	B·C			
100m Ω			See		See Graph 12			

Alphabetic characters used in Table 5 represent:



\*: MED measurement speed mode.

\*\*: SLOW or FAST measurement speed mode (SLOW.FAST).

Figure 3-39. Number of Display Digits (Sheet 6 of 13).

Table 6. Q Measurement.

D Display	Q Display	D Display	Q Display
.0001 to .0006	OF	.001 to .006	OF
.0007 to .0010	1000.	.007 to .010	loo.
.0011 to .0033	900. to 300.	.011 to .033	8 o . to 3 O .
.0034 to .0099	29 a. to 10 a.	.034 to 1.999	to
.0100 to .0333	100. to 30.		
.0334 to 1.9999	2		
D F	Ð		

Figure 3-39. Number of Display Digits (Sheet 7 of 13).

SECTION III Model 4276A

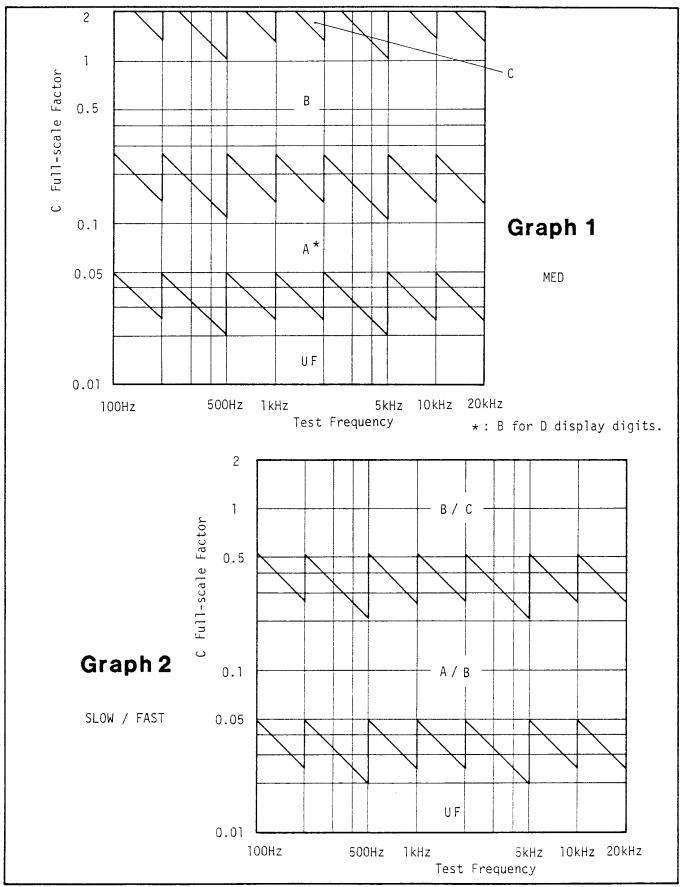


Figure 3-39. Number of Display Digits (Sheet 8 of 13).

Model 4276A SECTION III

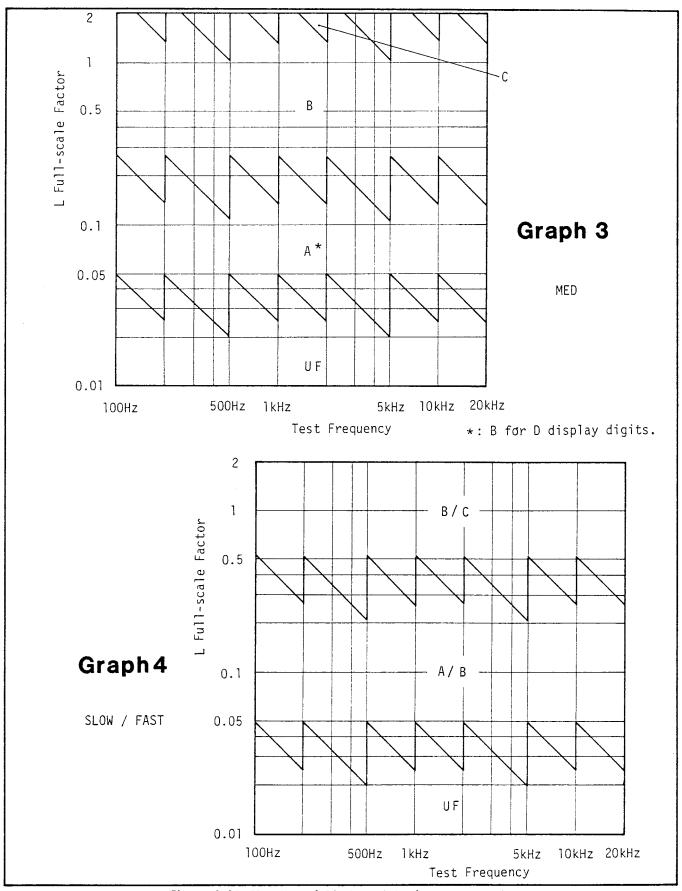


Figure 3-39. Number of Display Digits (Sheet 9 of 13).

SECTION III Model 4276A

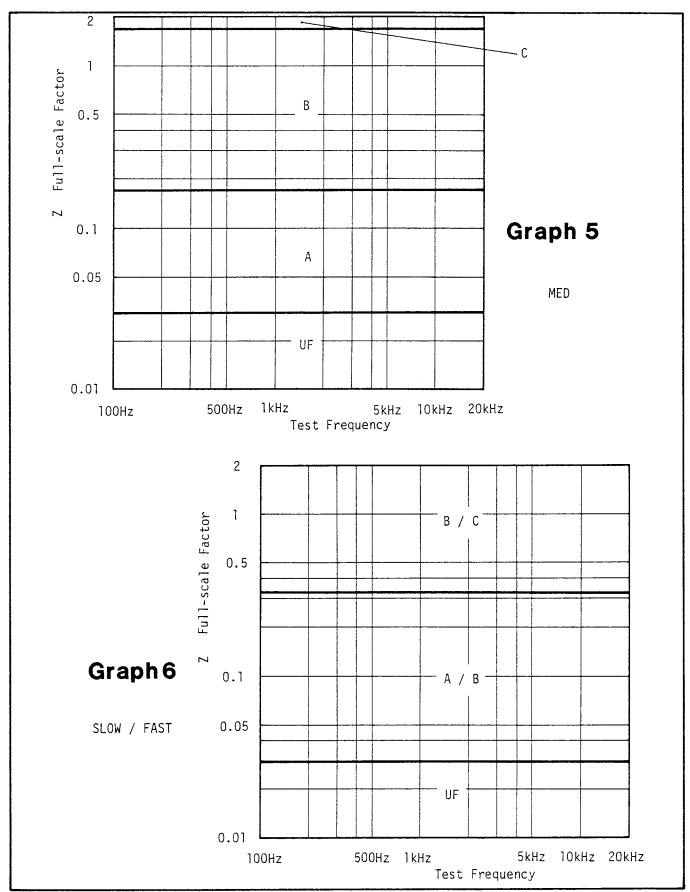


Figure 3-39. Number of Display Digits (Sheet 10 of 13).

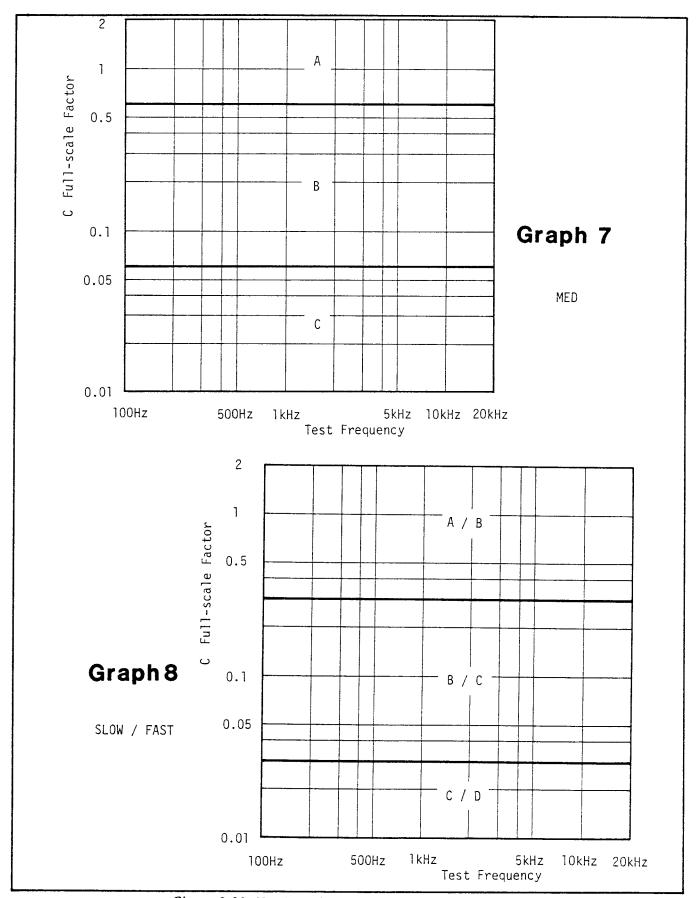


Figure 3-39. Number of Display Digits (Sheet 11 of 13).

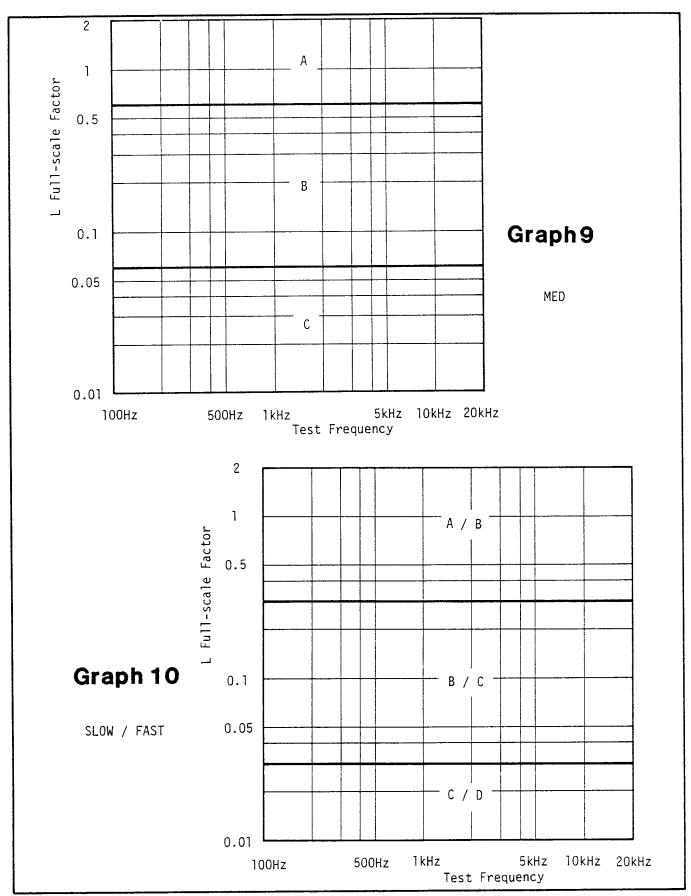


Figure 3-39. Number of Display Digits (Sheet 12 of 13).

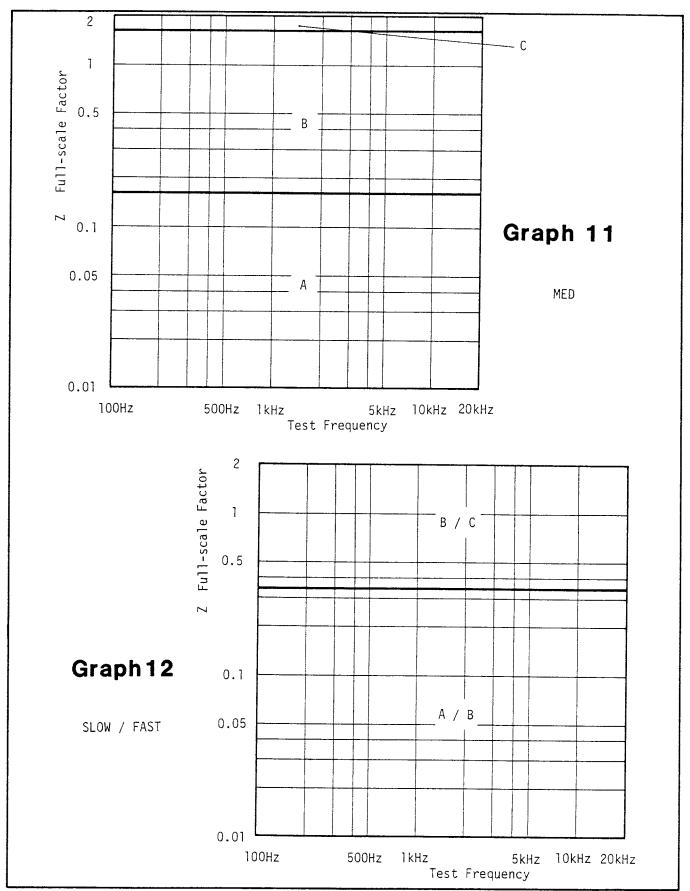


Figure 3-39. Number of Display Digits (Sheet 13 of 13).

SECTION IV Model 4276A

Table 4-1. Recommended Performance Test Equipment

Equipment	Critical Specifications	Recommended Model	Use
Capacitance Standards	10pF±0.4% 100pF±0.1% 1000pF±0.03% Usable frequencies: Up to 20kHz	HP 16382A HP 16383A HP 16384A	Р, А
Resistance Standards	$0.1\Omega$ $1\Omega$ $1\Omega$ $10\Omega$ $10\Omega$ $10\Omega$ $100\Omega_{\pm}0.03\%$ $1k\Omega_{\pm}0.06\%$ $10k\Omega_{\pm}0.06\%$ $100k\Omega_{\pm}0.06\%$ Usable frequencies: Up to $20k$ Hz	HP 16074A Standard Resistor Set	P, A, T
Terminations	0Ω SHORT OPEN Usable frequencies: Up to 20kHz	HP 16074A Standard Resistor Set	Р, А
Frequency Counter	Frequency range: 100Hz to 100kHz f.s. Resolution: 1 x 10 <sup>-5</sup> of f.s. Accuracy: 0.001%	HP 5314A	Р, А
Digital Multimeter	DCV: Voltage range: 10mV to 100V f.s. Resolution: 1 x 10 <sup>-5</sup> of f.s. Accuracy: 0.1%  ACV: Voltage range: 100mV to 1Vrms f.s. Resolution: 1 x 10 <sup>-2</sup> of f.s. Accuracy: 3% in 100Hz to 20kHz	НР 3478А	P, A, T
RC Oscillator	Frequency: lkHz Level: lmV	HP 652A	Т
Oscilloscope	Bandwidth: 100MHz Storage capability	HP 1741A	Т
Signature Analyzer		HP 5004A	Т
Test Cables	BNC-to-BNC cable (<1m) 2EA	HP 11170A	P, A, T
	BNC-to-dual banana plug cable (<1m)	HP 11035A	P, A
	Dual banana plug-to-alligator clip cable	HP 11002A	A
	BNC-to-dual alligator clip cable Alligator clip-to-alligator clip cable (<20cm)	(HP 11002A/w 10110B)	A
BNC Adapter	BNC-T-Adapter	HP 1250-0781	т
Test Fixture	Four terminal pair configuration design	HP 16047A	Р
HP-IB Controller		HP 85/ w0085-15003/ w82936A/ w82937A	Р
Extender Board	Large extender board	HP 04276-66561	Т
	small extender board (for INTERNAL DC BIAS Adjustment)	HP 04276-66562	A

P: Performance Test, A: Adjustment, T: Troubleshooting

# SECTION IV

## **PERFORMANCE TESTS**

#### 4-1. INTRODUCTION

This section provides the tests and the to verify procedures used the 4276A specifications listed in Table 1-1. All tests can be performed without access to the interior of the instrument. A simpler operational test is presented in Section III under Self Test. The performance tests can be used when performing incoming inspection of the instrument and when verifying that the instrument meets performance specifications after troubleshooting or adjustment or both. If the performance tests indicate the instrument is operating outside specified limits, check to see if the controls on the instrument used in the test and the test setup itself are correct and then proceed with adjustments or troubleshooting or both.

#### Note

To ensure proper test results and instrument operation, Hewlett-Packard recommends a 30-minute warm-up and stabilization period before performing any of the performance tests.

## Note

All performance tests except for the HP-IB Interface Test should be performed in an ambient temperature range of 23  $^{\circ}$ C±5  $^{\circ}$ C.

## 4-3. EQUIPMENT REQUIRED

4-4. Equipment required to perform all of the performance tests is listed in Table 4-1. Any equipment that satisfies or exceeds the critical specifications listed in the table may be used as a substitute for the recommended models.

Accuracy checks described in this section use the HP 16380A series standard capacitors (16382A, 16383A and 16384A) and 16074A Standard Resistor Set. The characteristics of the equipment satisfy the performance requirements for the accuracy checks and are especially suited for use as the 4276A's accuracy test standards.

#### Note

Components used as standards should be calibrated by an instrument whose specifications are traceable to NBS or an equivalent standards group; or calibrated directly by an authorized calibration organization such as NBS. The calibration cycle should be in accordance with the stability specifications for each component.

#### 4-5. TEST RECORD

4-6. Performance test results can be recorded on the Test Record at the completion of the test. The Test Record is at the end of this section and it lists all the tested specifications and their acceptable limits. Test results recorded at incoming inspection can be used for comparison in periodic maintenance, trouble-shooting, and after repair or adjustment.

## 4-7. CALIBRATION CYCLE

4-8. This instrument requires periodic verification of performance. Depending on the conditions under which the instrument is used, e.g., environmental conditions or frequency of use, the instrument should be checked with the performance tests described here at least once a year. To keep instrument down-time to a minimum and to insure optimum operation, preventive maintenance should be performed at least twice a year.

## **ACCURACY TEST CONSIDERATIONS**

This paragraph discusses how the 4276A accuracy is tested and verified. As the 4276A has wider measurement capabilities in regard to the selectable measurement parameters, frequency, measurement range and accuracy, the performance tests include some critical measuring regions where accuracy is difficult to verify directly by measuring available standards.

Measurement accuracy is tested by measuring standard capacitors, resistors and other reference devices. The standards must have been calibrated and certified by transfer of values of national standards. However, a portion of the measurement range of the 4276A is out of the applicable ranges of the available standards. The method then, is to check accuracies by comparison with references on the specific ranges at which the standards are applicable, and to apply alternative tests for verification of accuracies on the other ranges.

## Theoretical Background of Accuracy Checks

The 4276A, in accordance with its measurement principles, determines the vector impedance (or its reciprocal value: admittance) of the unknown device under test. The various measurement data provided, with respect to the 8 selectable measurement parameters (L, C, D, etc.), are arithmetically derived from measured values of the orthogonal vector components (resistance and reactance). For example, the capacitance value of a DUT is calculated by the following equation relative to the capacitance-to-reactance values:

$$Cx = \frac{1}{2\pi f Xm}$$

where, Cx is capacitance value of DUT, f is test frequency, Xm is measured reactance value of DUT.

As stated above, each measurement parameter is interrelated with the impedance (or admittance) value; consequently, the accuracies on all ranges can be verified if the instrument satisfies specified accuracies for each one of its resistive and reactive measurement parameters; that is, resistance and capacitance from the lowest through the highest test frequencies.

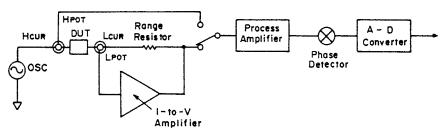
The technician should note that accuracy here is based on arithmetic relationships as are the parameter relationships. Therefore, the accuracy tests can be done by simplified procedures instead of time-consuming tests on the approximately 250000 possible combinations of the fundamental test parameters such as measurement parameter, frequency and range.

## Verification Check Considerations

The measurement accuracy test can be made by using calibrated standards on specific ranges only. On other ranges, which would be uncertifiable because of the limitations of the standards, the test takes the method proven to be theoretically and experimentally practicable for verification of accuracy. If the results of these checks meet all the individual test limits, the instrument should satisfy its specified accuracy across its entire range. How then can these methods be explained? Let us look at the performance test articles.

Accuracy test procedures include checks for the following circuit sections:

- 1) Range Resistors
- 2) Process Amplifier
- 3) I-to-V Amplifier
- 4) Phase Detector
- 5) A-D (Analog to Digital) Converter



4276A Measurement Section

CAPACITANCE ACCURACY TEST verifies Range Resistor accuracy for reactive impedance measurements from the lowest through the highest test frequencies. I-to-V Amplifier linearity and normal operation of the Phase Detector and A-D Converter are also verified.

IMPEDANCE ACCURACY TEST is similar to the Capacitance Accuracy test, but for resistive impedance measurements. Thus, accuracy for both reactive and resistive components of the vector impedance is verified. In this test, phase-flatness characteristics (minimum phase shift) of the overall measurement section, and Phase Detector accuracy from the lowest through highest test frequencies are verified.

SELF-OPERATING TEST verifies the accuracy of the Process Amplifier which extends the measurement ranges. The A-D Converter accuracy is also checked by this combined self-test function which enables automatic check of each one of these circuits.

#### Note

A set of detection phases, each different by 90 degrees, is used in the Phase Detector. If the relative phase difference between the detection phases is exactly 90 degrees, the Phase Detector is operated at the maximum detection accuracy.

The accuracy of the right-angle detection phases is verified by both this test and dissipation factor checks associated with the Capacitance Accuracy Test.

## ACCURACY TEST STANDARDS

## 1) Standard Capacitors

The HP 16380A Series Standard Capacitors, featuring the four terminal pair configuration, are recommended for use as performance test standards. The four standard capacitors, 16381A (lpF), 16382A (l0pF), 16383A (l0pF) and 16384A (l00pF) are calibrated at 0.01% accuracy at lkHz (and have capacitances within 0.1% of their nominal values).

#### 2) Standard Resistors

The standard resistors used for accuracy checks should be nearly pure resistances and should maintain an extremely low residual reactance at frequencies to 1MHz. The HP 16074A Standard Resistor Set, especially designed as standards useable over a broad frequency region, with thin film resistors and four terminal pair configurations, is suitable for the accuracy checks. Because of low residual inductance and less skin effect of the thin film resistors, the 16074A provides the standard resistance values of 0 $\Omega$ , 0.1 $\Omega$ , 1 $\Omega$  and 10 $\Omega$  at ±10% and 100 $\Omega$ , 1k $\Omega$ , 10k $\Omega$  and 100k $\Omega$  at ±0.01% calibration accuracies to 10MHz (1MHz at 100k $\Omega$ ). Open (OS) and Short terminations, which facilitate optimum zero offset adjustment, and two quasi-inductors are included in the 16074A.

#### Note

The 0Ω, 0.1Ω,  $1\Omega$  and  $10\Omega$  resistors are used as the (pure resistance) reference device in the Impedance Accuracy Test. Two quasi-inductors are not used in the 4276A performance tests.

## 3) The principle of Inductance Accuracy Test

The 4276A inductance accuracy is theoretically certified if the capacitance accuracy meets the specifications. Generally, inductors have unwanted parasitic impedances such as coil resistance and distributed capacitance. As these residuals significantly affect the inductance values at high frequencies, inductance standards useable in the RF region above 100kHz are substantially unavailable. Inductors with higher inductance values have lower frequency limits.

If it is desired to check inductance measurement accuracy, use standard capacitors as a substitution test device. The capacitors act as negative inductors when measured in inductance measurement function of the 4276A. The equivalent inductance value of capacitor is calculated by the following equation:

$$Z = \frac{1}{j\omega C} = j\omega L$$

$$L = \frac{1}{-\omega^2 C}$$

C: Calibrated value of standard capacitor

 $\omega$ :  $2\pi$  [test frequency]

## GENERAL

The standards should be of four terminal pair configuration design to provide compatibility with the instrument. This minimizes reduction in reliability of the values due to the effects of the residuals associated with cabling and connections.

## 4-9. TEST FREQUENCY ACCURACY TEST

4-10. This test verifies that the test signal frequencies for the 4276A meet the specified frequency accuracy of 0.01%.

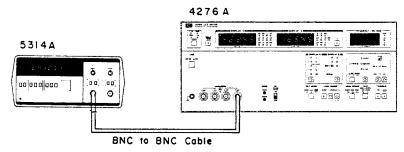


Figure 4-1. Test Frequency Accuracy Test Setup.

## **EQUIPMENT:**

Frequency Counter	Н	P	5314A
BNC-to-BNC cable	H	P	11170A

#### PROCEDURE:

- 1. Connect the BNC-to-BNC cable to the 4276A UNKNOWN  $H_{\text{CUR}}$  terminal and to the 5314A's input as shown in Figure 4-1.
- 2. Set the 4276A's controls as follows:

Test Frequency	1.00kHz
DC BIAS	
TEST SIG LEVEL	HIGH
TRIGGER	MAN/EXT
Other controls	Any setting

- 3. Verify that the frequency reading on the 5314A is 1.000kHz±0.1Hz.
- 4. Set test frequency in accordance with Table 4-2 and confirm that the frequency readings on the 5314A are within the test limits given in the table.

## Note

- 1) Test limit values in the table do not account for the tolerance dependent on the specified accuracy of the 5314A.
- 2) If this test fails, the instrument requires troubleshooting.

Table 4-2. Test Frequency Accuracy Test

Frequency Setting	Test Limits
100Hz	99.99 - 100.01Hz
200Hz	199.98 - 200.02Hz
500Hz	499.95 - 500.05Hz
1.00kHz	999.9 - 1000.1Hz
2.00kHz	1.9998 - 2.0002kHz
5.00kHz	4.9995 - 5.0005kHz
10.0kHz	9.999 - 10.001kHz
20.0kHz	19.998 - 20.002kHz

## 4-11. TEST SIGNAL LEVEL ACCURACY TEST

4-12. This test verifies that the test signal level for the 4276A meets the specified test signal level accuracy.

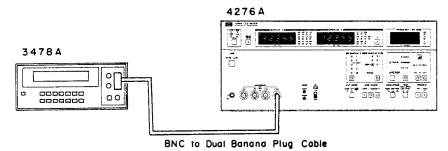


Figure 4-2. Test Signal Level Accuracy Test Setup.

## **EQUIPMENT:**

Note

Use a digital multimeter calibrated for a frequency response of 100Hz to 20kHz.

## PROCEDURE:

- 1. Connect the BNC-to-dual banana plug cable to the UNKNOWN  $H_{\text{CUR}}$  terminal of the 4276A and to the 3478A's input as shown in Figure 4-2.
- 2. Set the 4276A's controls as follows:

DISPLAY A/B functions	C-D
Test Frequency	1.00kHz
DC BIAS	OFF
LC   Z   RANGE	lnF
TEST SIG LEVEL	HIGH
TRIGGER	MAN/EXT
Other controls	Any setting

- 3. Set the function of the 3478A to ACV.
- 4. Verify that the voltage reading on the 3478A is 1.0Vrms±0.1Vrms.
- 5. Change test frequency and test signal level settings in accordance with Table 4-3. Verify that the voltage readings on the 3478A are within the test limits given in the table.

Contro	l Settings	
Test Level	Test Frequency	Test Limits
HIGH	100Hz 200Hz 500Hz 1.00kHz 2.00kHz 5.00kHz 10.0kHz 20.0kHz	.7 - 1.3Vrms .7 - 1.3Vrms .7 - 1.3Vrms .9 - 1.1Vrms .7 - 1.3Vrms .7 - 1.3Vrms .7 - 1.3Vrms .7 - 1.3Vrms
LOW	100Hz 1.00kHz 10.0kHz	35 - 65mVrms 40 - 60mVrms 35 - 65mVrms

Table 4-3. Test Signal Level Accuracy Test

## 4-13. SELF-OPERATING TEST

4-14. The Self-operating Test checks operating conditions of the circuits which are critical to sustaining specified accuracies. To verify that these circuits satisfy the performance requirements for ensuring the specified accuracies, the values displayed in the SELF TEST are compared with test limits. Because basic circuit operating conditions related to the accuracy are verified in this test, the instrument should be initially checked with this test.

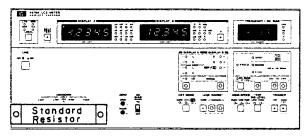
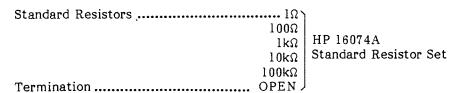


Figure 4-3. Self Operating Test Setup.

## **EQUIPMENT:**

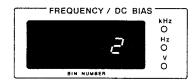


## PROCEDURE:

- 1. Connect the  $100k\Omega$  Standard Resistor directly to UNKNOWN terminals of the 4276A as shown in Figure 4-3.
- 2. Set the 4276A's controls as follows:

DISPLAY A/B functions	.   Z   - θ
Test Frequency	.1.00kHz
DC BIAS	
CKT MODE	. Any setting
LC   Z   RANGE	
MEAS SPEED	
TEST SIG LEVEL	HIGH
TRIGGER	

3. Press the SELF TEST key and then press the FREQUENCY/DC BIAS Select key. Press the FREQUENCY/DC BIAS Step Control ⚠ key until Self test item number "2" is displayed on the FREQUENCY/DC BIAS display as shown below.



- 4. Verify that the value displayed on DISPLAY A is  $10.00\,\mu\text{S}\pm0.01\,\mu\text{S}$ .
- 5. Change Standard Resistor to  $10k\Omega$ ,  $1k\Omega$  and  $100\Omega$  in that order and change LC | Z | RANGE  $\bigoplus$  key each time. Verify that the display outputs are within the test limits given in the table below.

Standard Resistor	Range	Limits		
100kΩ	10μS	1/C.V. ± 0.01µS		
10kΩ	100μS	1/C.V. ± 0.1µS		
1kΩ	1mS	1/C.V. ± 0.001mS		
100Ω	10mS	1/C.V. ± 0.01mS		

1/C.V. = Reciprocal of Calibrated Value

6. Press SELF TEST to resume measurement mode and set the 4276A's controls as follows:

DISPLAY A/B functions	C-G
Test Frequency	1.00kHz
DC BIAS	
CKT MODE	Any setting
LC   Z   RANGE	
MEAS SPEED	MED
TEST SIG LEVEL	HIGH
TRIGGER	INT

- 7. Connect the  $1\Omega$  Standard Resistor directly to the UNKNOWN terminals as shown in Figure 4-3.
- 8. Repeat step 3.

9. Verify that the values displayed on DISPLAY A and DISPLAY B are within the following test limits.

DISPLAY A:  $C.V.\pm.02\Omega$  DISPLAY B:  $0\pm.012\Omega$ 

- 10. Set the MEAS SPEED to SLOW and connect the OPEN termination to the UNKNOWN terminals in place of the  $1\Omega$  standard.
- 11. Select SELF TEST "8" by pressing the FREQUENCY/DC BIAS Step Control key until "8" appears on the FREQUENCY/DC BIAS display.
- 12. Confirm that the values displayed on DISPLAY A and DISPLAY B are within the following test limits:

DISPLAY A: .0020±.0003 DISPLAY B: -.0020±.0003

13. Perform the test for the self test steps and the test frequencies shown in the table below in the same way. Confirm that the values displayed on DISPLAY A and DISPLAY B are within the test limits given in the table.

#### Note

To change the test frequency setting while the 4276A is in SELF TEST mode, press the FREQUENCY/DC BIAS Select key (FREQ lamp should light), set the frequency with the FREQUENCY/DC BIAS Control keys ( $\odot$  and  $\odot$ ), and then press the FREQUENCY/DC BIAS Select key again (DC BIAS lamp should light).

SELF TEST NUMBER	1.0	00kHz	20.0kHz		
	DISP A	DISP B	DISP A	DISP B	
8	.0020±.0003	0020±.0003			
9	0±.0020	0±.0013	0±.0080	0±.0052	
12	0±.0012	0±.0020	0±.0048	0±.0043	
13	0±.0012	0±.0020	0±.0048	0±.0043	
14	0±.0024	0±.0020	0±.0096	0±.0036	
15	0±.0048	0±.0020	0±.0192	0±.0360	

## 4-15. CAPACITANCE ACCURACY TEST

4-16. This test checks capacitance measurement accuracies for various combinations of test frequency and test signal level. The checks are made by connecting a standard capacitor to the instrument and comparing measurement results with the calibrated values of the standard. Accuracies for dissipation factors near zero are also checked in this test.

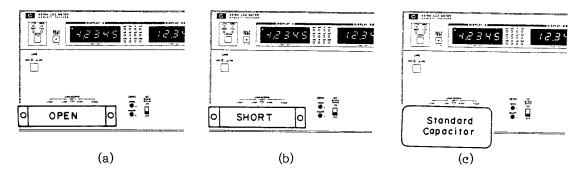


Figure 4-4. Capacitance Accuracy Test Setups.

## **EQUIPMENT:**

Standard Capacitors	10pF: HP 16382A
-	100pF: HP 16383A
	1000pF: HP 16384A
Terminations	SHORT ) HP 16074A
	OPEN   Standard Resistor Set

## PROCEDURE:

1. Set the 4276A's controls as follows:

DISPLAY A/B functions	C-D
DC BIAS	
CKT MODE	AUTO
LC   Z   RANGE	AUTO
TRIGGER	INT
Other controls	Any setting

 Perform OPEN and SHORT Zero Offset Adjustments as described in paragraph 3-48.

#### Note

Use the OPEN and SHORT terminations of the 16074A for Zero offset Adjustment.

- 3. Connect the 10pF Standard Capacitor directly to the UNKNOWN terminals as shown in Figure 4-4 (c).
- 4. Set the test frequency in accordance with Table 4-4. Verify that the capacitance and dissipation factor readings on the 4276A are within the test limits given in the table.
- 5. Change the standard capacitor to 100pF and 1000pF in that order and perform the test. Verify that the values displayed on the 4276A are within the test limits given in the table.

Table 4-4. Capacitance Accuracy Tests

Standard Capacitance	Function	Speed	Level	Frequency	Test Lim	its
10pF	C-D	MED	HIGH	9.95kHz 10.0kHz 20.0kHz	C.V.±.16pF C.V.±.16pF C.V.±.24pF	0±.010 0±.010 0±.012
100pF	C-D	MED	HIGH	995Hz 1.00kHz 1.99kHz 4.98kHz 9.95kHz 10.0kHz 20.0kHz	C.V.±.6pF C.V.±.35pF C.V.±.65pF C.V.±.25pF C.V.±.70pF C.V.±.70pF C.V:±2.20pF	0±.004 0±.0035 0±.0065 0±.0017 0±.0046 0±.0046 0±.0140
1000pF	C-D	MED	HIGH	100Hz 120Hz 199Hz 498Hz 995Hz 1.00kHz 1.99kHz 4.98kHz 9.95kHz 10.0kHz	C.V.±5.0pF C.V.±5.0pF C.V.±6.5pF C.V.±2.5pF C.V.±1.5pF C.V.±2.5pF C.V.±2.5pF C.V.±2.5pF C.V.±7.0pF C.V.±7.0pF C.V.±7.0pF	0±.0050 0±.0050 0±.0065 0±.0017 0±.0017 0±.0017 0±.0017 0±.0046 0±.0046 0±.0070
		SLOW	HIGH	100Hz 1.00kHz 10.0kHz	C.V.±5.0pF C.V.±1.50F C.V.±7.0pF	0±.0050 0±.0011 0±.0046
		FAST	HIGH	100Hz 1.00kHz 10.0kHz	C.V.±10pF C.V.±6pF C.V.±16pF	0±.010 0±.006 0±.014
		MED FAST MED	LOW	1.00kHz 1.00kHz 10.0kHz	C.V.±12pF C.V.±100pF C.V.±32pF	0±.011 0±.100 0±.027
	HIGH SPEED	MED	HIGH LOW HIGH	1.00kHz 1.00kHz 10.0kHz	C.V.±1.5pF C.V.±12pF C.V.±7.0pF	

C.V. = Calibrated Value of Standard Capacitor

## 4-17. IMPEDANCE ACCURACY TEST

4-18. This test checks impedance measurement accuracies at four spot test frequencies. The checks are made by connecting a standard resistor to the instrument and comparing the measurement readouts with the calibrated values of the standard.

Range Freq.	0.1Ω	1Ω	10Ω	100Ω	1kΩ	10kΩ	100kΩ
100Hz	Δ	Δ	Δ	0	0	0	0
120Hz	Δ	Δ	Δ	0	0	0	0
1.00kHz	Δ	Δ	Δ	0	0	0	0
10.0kHz	Δ	Δ	Δ	0	0	0	0

 $\bigcirc$  : tested for both  $\mid$  Z  $\mid$  and  $\theta$ .

 $\triangle$ : tested for  $\theta$  only.

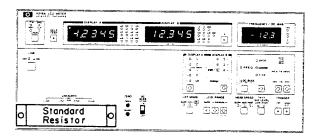


Figure 4-5. Impedance Accuracy Test Setup.

## **EQUIPMENT:**

Standard Resistors	0.1Ω	1
	$1\Omega$	
	10Ω	HP 16074A Standard
	100Ω	Resistor Set
	lkΩ	
	10kΩ	
	100kΩ	
Terminations	0Ω	
	OPEN	

## PROCEDURE:

1. Set the 4276A's controls as follows:

DISPLAY A/B functions	$ Z  - \theta$
Test Frequency	
DC BIAS	OFF
CKT MODE	
LC   Z   RANGE	
MEAS SPEED	MED
TEST SIG LEVEL	
TRIGGER	

2. Perform OPEN and SHORT Zero Offset Adjustment as described in paragraph 3-48.

#### Note

Use the OPEN and 0 $\Omega$  terminations of the 16074A for Zero Offset Adjustment. DO NOT use the SHORT termination.

- 3. Connect the  $0.1\Omega$  standard resistor directly to UNKNOWN terminals as shown in Figure 4-5 and set test frequency to  $100\,\mathrm{Hz}$ ,  $120\,\mathrm{Hz}$ ,  $1k\,\mathrm{Hz}$  and  $10k\,\mathrm{Hz}$ . Verify that the impedance and phase angle readings on the  $4276\mathrm{A}$  are within the test limits given in Table 4-5.
- 4. Repeat step 3 with the each of the standard resistors listed in Table 4-5. Verify that the values displayed on the 4276A meet the test limits given in the table.

Z  Ra	nge.		Test Limits					
Freq.	_	0.1Ω	$1\Omega$	10Ω	100Ω	1kΩ	$10 \mathrm{k}\Omega$	100k $\Omega$
100Hz	Z				C.V.±.15Ω	C.V.±6Ω	C.V.±.06kΩ	C.V.±.6kΩ
100112	9	0±1.2°	0±.7°	0±.35°	0±.15°	0±.6°	0±.6°	0±.6°
120Hz	Z				C.V.±.15Ω	C.V.±6Ω	C.V.±.06kΩ	C.V.±.6kΩ
120112	θ	0±1.2°	0±.7°	0±.35°	0±.15°	0±.6°	0±.6°	0±.6°
1.00kHz	Z				C.V.±.15Ω	C.V.±6Ω	C.V.±.06kΩ	C.V.±.6kΩ
1.00KHZ	θ	0±1.2°	0±.7°	0±.35°	0±.15°	0±.6°	0±.6°	0±0.6°
10.0kHz	Z				C.V.±.25Ω	C.V.±10Ω	C.V.±.1kΩ	C.V.±1kΩ
10.0K12	θ	0±2.2°	0±1.2°	0±.65°	0±.25°	0±1.0°	0±1.0°	0±1.0°

Table 4-5. Impedance Accuracy Tests

C.V. = Calibrated Value of Standard Resistors

## 4-19. INDUCTANCE ACCURACY TEST

4-20. Inductance accuracy is verified if the instrument meets both capacitance and impedance accuracy test limits. If it is desired to confirm the inductance accuracy on at least at one range, perform the following test:

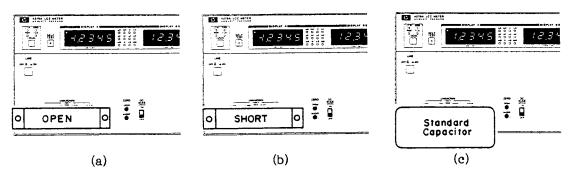


Figure 4-6. Inductance Accuracy Test Setup.

## **EQUIPMENT:**

Terminations	.SHORT `	HP 16074A Standard
	OPEN /	Resistor Set
Standard Capacitor	1000pF:	HP 16384A

## PROCEDURE:

1. Set the 4276A's controls as follows:

DISPLAY A/B functions	L-D
Test Frequency	
DC BIAS	
CKT MODE	
LC   Z   RANGE	
MEAS SPEED	
TEST SIG LEVEL	
TRIGGER	

2. Perform OPEN and SHORT Zero Offset Adjustment as described in paragraph 3-48.

Note

Use the OPEN and SHORT terminations of the 16074A for Zero Offset Adjustment.

- 3. Connect the 1000pF standard capacitor directly to the UNKNOWN terminals. See Figure 4-6 (c).
- 4. Verify that the inductance and dissipation factor readings on the 4276A are within the following test limits:

DISPLAY A: -25.33±0.18 DISPLAY B: 0±0.012

## 4-21. INTERNAL DC BIAS ACCURACY TEST (OPTION 001)

4-22. This test verifies that the Option 001 Internal DC BIAS Supply applies the specified bias voltages to the device under test.

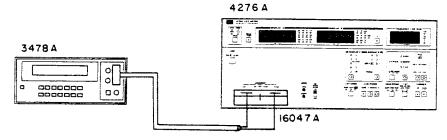


Figure 4-7. Option 001 Internal DC Bias Accuracy Test Setup.

## EQUIPMENT:

#### PROCEDURE:

1. Set the 4276A's controls as follows:

DISPLAY A/B functions	·C-D
Test Frequency	.20.0kHz
LC   Z   RANGE	
TFST SIG LEVEL	.LOW
TRIGGER	.MAN/EXT
DC BIAS (Front Panel)	
DC BIAS (Rear Panel)	
Other controls	

- 2. Interconnect the 4276A, 16047A, and 3478A as shown in Figure 4-7.
- 3. Set dc bias voltage to +40V with the FREQUENCY/DC BIAS Select key and the FREQUENCY/DC BIAS Step Control ② key. Wait approximately 10 seconds after setting the voltage.
- 4. Set DC bias voltage in accordance with Table 4-6. After the wait time specified in the table, read the voltage displayed on the 3478A. Verify that the voltage readings are within the test limits given in the table.

Table 4-6. Internal DC Bias Accuracy Test

DC BIAS Setting	Wait Time	Test Limits
+40.0V .00 +.01 +9.99 +10.0 +40.0	10 seconds 30 20 10 10 15 30	Precharge 0±.01V .01±.01V 9.99±.04V 10.0±.085V 40.0±.24V 01±.01V -9.99±.11V
-9.99 -10.0	10 10	-10.0±.135V
-40.0	15	-40.0±.44V

## 4-23. 16064A COMPARATOR/HANDLER INTERFACE TEST (OPTION 002)

4-24. The test in this paragraph, verifies the functions of the 16064A Comparator/Handler Interface.

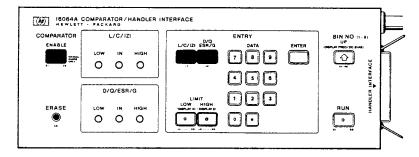


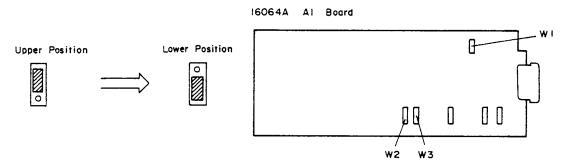
Figure 4-8. 16064A Comparator/Handler Interface.

## **EQUIPMENT:**

Digital Multimeter	HP	3478A
100kΩ Standard resistor	HP	16074A
1000pF Standard capacitor		

## PROCEDURE:

1. Set jumpers Al W1/W2/W3 in the 16064A to the lower position as shown below:



- 2. Connect the 16064A to the COMPARATOR/HANDLER INTERFACE connector on the rear panel of the 4276A.
- 3. Turn on the 4276A. "16064" should be displayed on DISPLAY B.
- 4. Set the 4276A's controls as follows:

DISPLAY A/B functions	.C-G
Test Frequency	1.00kHz
DC BIAS	
CKT MODE	_
LC   Z   RANGE	777
MEAS SPEED	MED
TEST SIG LEVEL	HIGH
TRIGGER	INT

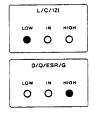
5. Set the 3478A's controls as follows:

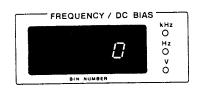
Function	DCV
RANGE	300V
Display	3 1/2 digits

- 6. Connect the 3478A's LO input to the 4276A's GUARD terminal.
- 7. Press the ERASE key on the 16064A and set the following comparator limits:

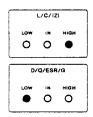
```
L/C/|Z| LOW LIMIT (BIN1): .3
L/C/|Z| HIGH LIMIT (BIN1): .9
D/Q/ESR/G LOW LIMIT: 2
D/Q/ESR/G HIGH LIMIT: 8
```

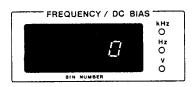
- 8. Connect the  $100 k\Omega$  standard resistor directly to the 4276 A's UNKNOWN terminals.
- 9. Press the RUN key on the 16064A's control panel.
- 10. Verify that the L/C/|Z| LOW and D/Q/ESR/G HIGH lamps light, and that "0" is displayed on the 4276A's FREQUENCY/DC BIAS DISPLAY.





- 11. Check the states of the comparison data output (TTL) at the HANDLER INTERFACE connector using the 3478A. The pin assignments and the data states are shown in Figure 4-9 and Table 4-7.
- 12. Disconnect the  $100k\Omega$  resistor and connect the 1000pF standard capacitor.
- 13. Verify that the L/C/|Z| HIGH and D/Q/ESR/G LOW lamps light, and that "0" is displayed on the 4276A's FREQUENCY/DC BIAS DISPLAY.





14. Check the comparison data output at the HANDLER INTERFACE connector by comparing it with the Data States shown in Table 4-7.

Model 4276A SECTION IV

## PERFORMANCE TESTS

15. Press the ERASE key and set the following comparator limits:

L/C/|Z| HIGH LIMIT:

BIN1: .9

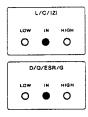
BIN2: 1.1 BIN3: 1.9999

D/Q/ESR/G HIGH LIMIT:

.01

16. Press the RUN key on the 16064A's control panel.

17. Verify that the L/C/|Z| IN and D/Q/ESR/G IN lamps light, and that "2" is displayed on the 4276A's FREQUENCY/DC BIAS DISPLAY.





18. Check the comparison data output at the HANDLER INTERFACE connector by comparing it with the Data States shown in Table 4-7.

Table 4-7. Handler Interface Output Data States

TEST	Connector Pin Numbers			
STEP	1 2 3 4 5 6 7 8 9 10	19 20 21 22 23 24		
11	нннннннн	HHLLHH		
14	нниннинн	LHHHHL		
18	нгининини	HLHHLH		

H: Approximately 5VL: Approximately 0V

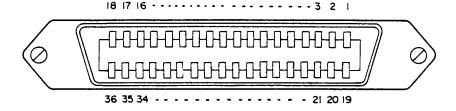


Figure 4-9. Handler Interface Connector Pin Assignments.

## 4-25. HP-IB INTERFACE TEST

4-26. This test verifies the instrument's HP-IB capabilities.

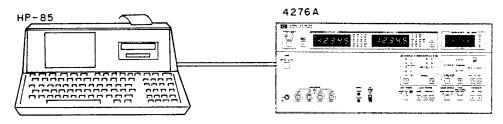


Figure 4-10. HP-IB Interface Test Setup.

## EQUIPMENT:

Personal Computer	HP-85
I/O ROM	HP 00085-15003
ROM Drawer	
HP-IB Interface	HP 82937A
100pF Standard	HP 16383A

## PROCEDURE:

- 1. Turn off the 4276A and the HP-85.
- 2. Connect the 82937A HP-IB Interface between the HP-85 and the 4276A as shown in Figure 4-8, and install the I/O ROM into the HP-85.
- 3. Set the 4276A's HP-IB Control switch, located on the rear panel, as follows:

bits 5-1 : 10001 (17<sub>10</sub>) bit 6 : 0 bit 7 : 0

- 4. Turn on the 4276A and the HP-85.
- 5. Load one of the three test programs into the personal computer. Test programs are listed on pages 4-21, 4-23 and 4-25.
- 6. Execute the program and follow the prompts and instructions output by the HP-85. Details on the controller's (personal computer) instructions and the appropriate operator response are given in Tables 4-8 through 4-10.

#### TEST PROGRAM 1

#### PURPOSE:

This test verifies that the 4276A has the following HP-IB capabilities:

- (1) Remote/Local Capability
- (2) Local Lockout
- (3) Talk Disable
- (4) Listen Disable

#### PROGRAM LISTING:

```
10 ! 42768 HP-IB TEST No.1
20 | REMOTE/LOCAL TEST
30 DIM A$[1]
40 N=0 @ M=7 @ M1=717
50 S=SPOLL(M1)
80 CLEAR
70 PRINT "*** 4276A HP-IB TEST No.1 ***"
80 DISP "REMOTE/LOCAL TEST"
90 REMOTE M
100 OUTPUT M1 ;"T1"
110 DISP "LISTEN=1, TALK=0, REMOTE=1"
120 GOSUB 580
130 ABORTIO M
140 DISP "LISTEN=O,TALK=O,REMOTE=1"
150 GOSUB 580
160 LOCAL M
170 DISP "LISTEN=0, TALK=0, REMOTE=0"
180 GOSUB 580
190 REMOTE M1
200 DISP "LISTEN=1, TALK=0, REMOTE=1"
210 GOSUB 580
220 LOCAL LOCKOUT M
230 DISP "PRESS LOCAL KEY"
240 DISP "LISTEN*1, TALK*0, REMOTE*1"
250 G0SUB 580
260 LOCAL M1
270 DISP "LISTEN=1, TALK=0, REMOTE=0"
280 GOSUB 580
290 REMOTE M1
300 OUTPUT M1 ;"T1"
310 DISP "LISTEN=1, TALK=0, REMOTE=1"
320 GOSUB 580
330 IF N=1 THEN 340 ELSE 370
340 PRINT "REMOTE/LOCAL TEST FAIL"
350 DISP "REMOTE/LOCAL TEST FAIL"
360 GOTO 390
370 PRINT "REMOTE/LOCAL TEST PASS"
380 DISP "REMOTE/LOCAL TEST PASS"
390 N*0
400 DISP "LISTEN/TALK TEST"
410 ENTER M1 ; A
420 DISP "LISTEN=0, TALK=1, REMOTE=1"
430 GOSUB 580
440 OUTPUT M1
                  ;"T1"
450 DISP "LISTEN*1, TALK =0, REMOTE *1"
460 GOSUB 580
470 IF N=1 THEN 480 ELSE 510
480 PRINT "LISTEN/TALK TEST FAIL"
490 DISP "LISTEN/TALK TEST FAIL"
500 GDT0 530
510 PRINT "LISTEN/TALK TEST PASS" 520 DISP "LISTEN/TALK TEST PASS"
530 PRINT "END"
540 DISP "END"
550 CLEAR M
560 LOCAL M
570 END
580 INPUT AS
590 IF A$ ="N" THEN N=1
600 RETURN
```

Table 4-8. Controller Instructions and Operator Responses for Test Program 1

Controller Instructions		Operator Response	
Displays	Printout	operator Response	
	*** 4276A HP-IB TEST NO.1 ***		
REMOTE/LOCAL TEST			
LISTEN=1*, TALK=0, REMOTE=1 LISTEN=0, TALK=0, REMOTE=1 LISTEN=0, TALK=0, REMOTE=0 LISTEN=1, TALK=0, REMOTE=1		If the 4276A HP-IB Status Indicators and Controller Display are the same, press Y, and END LINE. If not, press N, and END LINE.	
PRESS LOCAL KEY		Press Local Key.	
LISTEN=1, TALK=0, REMOTE=1 LISTEN=1, TALK=0, REMOTE= LISTEN=1, TALK=0, REMOTE=1		If the 4276A HP-IB Status Indicators and Controller Display are the same, press Y, and END LINE. If not, press N, and END LINE.	
REMOTE/LOCAL TEST PASS	REMOTE/LOCAL TEST PASS	If all steps are correct, this message is output.	
REMOTE/LOCAL TEST FAIL	REMOTE/LOCAL TEST FAIL	If any step fails, this message is output.	
LISTEN/TALK TEST			
LISTEN=0, TALK=1, REMOTE=1 LISTEN=1, TALK=0, REMOTE=1		If the 4276A HP-IB Status Indicators and Controller Display are the same, press Y, and END LINE. If not, press N, and END LINE.	
LISTEN/TALK TEST PASS	LISTEN/TALK TEST PASS	If both steps are correct, this message is output.	
LISTEN/TALK TEST FAIL	LISTEN/TALK TEST FAIL	If any step fails, this message is output.	
END	END		

<sup>\*1</sup> indicates ON; 0 indicates OFF.

TEST PROGRAM 2

#### PURPOSE:

This test verifies that the 4276A has the following HP-IB capabilities:

- (1) Talker
- (2) Device Trigger

#### PROGRAM LISTING:

```
10 ! 4276A HP-IB TEST No.2
 20 | TALKER TEST
 30 DIM A$ [100], B$ [1]
 40 M=7 @ M1=717
 50 S*SPOLL(M1)
 60 PRINT "*** 4276A HP-IB TEST No.2 ***"
 70 CLEAR
 80 DISP "TALKER TEST"
 90 DISP "CONNECT 100pF"
 100 BEEP
 120 DISP "DATA CUTPUT TEST"
 130 REMOTE M
 140 ABORTIO M
 150 CLEAR M1
160 OUTPUT M1 ;"A2B1F1T2"
170 DISP "TEST FREQUENCY IN kHz ";
 180 INPUT F
 190 OUTPUT M1 ;"FR",F,"EN"
200 TRIGGER M1
210 ENTER M1 ; A,B
220 DISP A*1.E12;"pF",B
230 DISP "IS OUTPUT DATA CORRECT (Y or N) ";
240 INPUT B$
250 IF B$="N" THEN 250 ELSE 290
250 PRINT "DATA OUTPUT TEST FAIL"
270 DISP "DATA OUTPUT TEST FAIL"
280 GOTO 310
290 PRINT "DATA OUTPUT TEST PASS"
300 DISP "DATA OUTPUT TEST PASS"
310 DISP "COMPLETE DATA OUTPUT TEST"
320 TRIGGER M1
330 ENTER M1 ; A$
340 DISP A$
350 DISP "IS OUTPUT DATA CORRECT (Y or N) ";
360 INPUT B$
370 IF B$ = "N" THEN 380 ELSE 410
380 PRINT "COMPLETE DATA OUTPUT TEST FAIL"
390 DISP "COMPLETE DATA OUTPUT TEST FAIL"
400 GDTB 430
410 PRINT "COMPLETE DATA DUTPUT TEST PASS"
420 DISP "COMPLETE DATA OUTPUT TEST PASS"
430 PRINT "END"
440 DISP "END"
450 CLEAR M
460 LOCAL M
470 END
```

Table 4-9. Controller Instructions and Operator Responses for Test Program 2

Controller Instr	Controller Instructions					
Displays	Printout	Operator Responses				
	*** 4276A HP-IB TEST No.2 ***					
TALKER TEST						
CONNECT 100pF		Connect the 16383A (100pF Standard) to the UNKNOWN terminals.				
DATA OUTPUT TEST TEST FREQUENCY IN kHz ?		Key in the desired test frequency value, from 0.1 to 20, and press END LINE.				
[Capacitance] [Dissipation Factor] IS OUTPUT DATA CORRECT (Y or N) ?		If the output data is the same as the values displayed on each 4276A display, press Y and END LINE. If not, press N and END LINE.				
	DATA OUTPUT TEST PASS	DATA OUTPUT TEST result.				
	DATA OUTPUT TEST FAIL					
COMPLETE DATA OUTPUT TEST						
PNC[Capacitance],ND[Dissipation Factor] IS OUTPUT DATA CORRECT (Y or N) ?		If the output data is the same as the left values, press 🗹 and ENO LINE . If not, press M and ENO LINE .				
	COMPLETE DATA OUTPUT TEST PASS	COMPLETE DATA OUTPUT				
	COMPLETE DATA OUTPUT TEST FAIL	TEST result.				
	END					

TEST PROGRAM 3

#### PURPOSE:

This test program verifies that the 4276A has the following HP-IB capabilities:

- (1) Service Request
- (2) Serial Poll

#### PROGRAM LISTING:

```
10 ! 4276A HP-IB TEST No.3
20 ! SRQ TEST
30 S=0 @ M=7 @ M1=717
40 ON INTR 7 GOSUB 560
50 CLEAR
50 PRINT "*** 4276A HP-IB TEST No.3 ***"
70 PRINT "SRQ TEST"
80 DISP "SRQ TEST"
90 REMOTE M
100 ABORTIO M
110 CLEAR M1
120 DISP "DATA READY SRG TEST"
130 OUTPUT M1 ;"D1T2"
140 TRIGGER M1
150 GOSUB 480
160 PRINT "DATA READY SRG TEST PASS"
170 5=0
180 DISP "SYNTAX ERROR SRG TEST"
190 OUTPUT M1 ;" BOA6DA"
200 GOSUB 480
210 PRINT "SYNTAX ERROR SRG TEST PASS"
220 S=0
230 DISP "SELF TEST END SRQ TEST"
240 OUTPUT M1 ;"S1"
250 DISP "SELF TEST in progress"
250 DISP "SELF TEST in progress"
260 GOSUB 480
270 IF BIT(5,2)=0 THEN GOSUB 480
280 OUTPUT M1 ;"SO"
290 PRINT "SELF TEST END SRQ TEST PASS"
300 S=0
310 DISP "TRIGGER TOO FAST SRQ TEST"
320 DISP "MOMENTARILY GROUND"
330 DISP " EXT TRG CONNECTOR"
 340 GOSUB 510
350 GOSUB 480
360 PRINT "TRG TOO FAST SRG TEST PASS"
370 S±0
380 DISP "OPERATIONAL ERROR SRQ TEST"
390 DISF UPERATIONAL ERROR SRQ TEST PASS"
400 GOSUB 480
410 PRINT "OPERATIONAL ERROR SRQ TEST PASS"
420 PRINT "SRQ TEST END"
430 CLEAR M1
440 ABORTIO M
450 LOCAL M
460 DISP "END"
 470 END
 480 ENABLE INTR 7;8
 490 IF S>0 THEN DISP S @ RETURN
 500 G0T0 480
 510 OUTPUT M1 ;"F1T2DA"
 520 TRIGGER M1
 530 ENTER M1 ; A,B
540 IF S=0 THEN 510
550 RETURN
 560 S*SPOLL(M1) @ STATUS 7,1 ; Z
 570 IF BIT(S,5)=1 THEN 590
580 DISP "OTHER DEVICE SRO"
 590 ENABLE INTR 7;8
 600 RETURN
```

Table 4-10. Controller Instructions and Operator Responses for Test Program 3

Controller Insti	ructions	Operator Response
Displays	Printout	operator nespense
SRQ TEST	*** 4276A HP-IB TEST No.3 *** SRQ TEST	
DATA READY SRQ TEST 65	DATA READY SRQ TEST PASS	SRQ Status Byte data should be 65 [=01000001].
SYNTAX ERROR SRQ TEST 66	SYNTAX ERROR SRQ TEST PASS	SRQ Status Byte data should be 66 [=01000010].
SELF TEST END SRQ TEST SELF TEST in progress 68	SELF TEST END SRQ TEST PASS	SRQ Status Byte data should be 68 [=01000100]. If the instrument fails SELF TEST, it should be 84 [=01010100].
TRIGGER TOO FAST SRQ TEST MOMENTARILY GROUND EXT TRG CONNECTOR 72*1	TRG TOO FAST SRQ TEST PASS	Ground EXT TRG Connector on rear panel momentarily. SRQ Status Byte data should be 72 [=01001000].
OPERATIONAL ERROR SRQ TEST 80* <sup>2</sup>	OPERATIONAL ERROR SRQ TEST PASS	SRQ Status Byte data should be 80 [=010100000].
	SRQ TEST END	

 $<sup>\</sup>ensuremath{^*}_1\colon$  SRQ Status Byte data may be 73 [=01001001] due to the timing of connecting the EXT TRG pin to ground.

 $<sup>{}^{\</sup>star}{}_{2}$ : SRQ Status Byte data may be 81 [=01010001] due to the timing of connecting the EXT TRG pin to ground.

Model 4276A LCZ METER

Serial Number \_\_\_\_

Tested	by	
Date		

Paragraph	TEST	TEST		Results		
Number			Minimum	Actua1	Maximum	
4-9	TEST FREQUENCY ACCURACY	TEST				
	Frequency setting					
		00Hz	99.99Hz		100.01H	
		00Hz 00Hz	199.98Hz 499.95Hz		200.02H	
		0kHz	999.9Hz		500.05H 1000.1H	
	2.0	0kHz	1.9998kHz		2.0002kH	
		0kHz	4.9995kHz		5.0005kH	
		OkHz OkHz	9.999kHz 19.998kHz		10.001kH	
	20.1	ORI12	19.990KH2		20.002kH	
4-11	TEST SIGNAL LEVEL ACCUR	ACY TEST				
	Test Signal Level: HI	GH				
	Frequency 10	00Hz	Vrms		Vr 1.3	
		00Hz	0.7		1.3	
		00Hz	0.7		1.3	
		OkHz	0.9		1.1	
		OkHz OkHz	0.7		1.3	
	10.0	OkHz	0.7		1.3	
	20.0	OkHz	0.7		1.3	
	Test Signal Level: LOV	V				
	Frequency 1(	OOHz	mVrms 35		mVrr 65	
	1.00		40		60	
	10.0	)kHz	35		65	
4-13	SELF-OPERATING TEST					
	SELF TEST #2	00kΩ	1/C.V01µS		1/C V . 03	
	1	.0kΩ	1/C.V1μS		1/C.V.+ .01 1/C.V.+ .1	
		1kΩ	1/C.V001mS		1/C.V.+.001	
	1	.000.	1/C.V01mS		1/C.V.+ .01	
	1Ω DISPLA		C.V02Ω		C.V. + .02	
	13" [ DISPLA	Y B	- 0.012Ω		0.012	

1/C.V. = Reciprocal of Calibrated Value C.V. = Calibrated Value

Paragraph	ТF	ST			Results	
Number	11-			Minimum	Actual	Maximum
4-13	SELF-OPERATING TE	ST (Cont'd)	)			
	1.00	kHz				
	SELF TEST #8	DISPLAY A DISPLAY B		0.0017 -0.0023		0.0023 -0.0017
	SELF TEST #9	DISPLAY A DISPLAY B		-0.0020 -0.0013		0.0020 0.0013
	SELF TEST #12	DISPLAY A DISPLAY B		-0.0012 -0.0020		0.0012 0.0020
	SELF TEST #13	DISPLAY A DISPLAY B		-0.0012 -0.0020		0.0012 0.0020
	SELF TEST #14	DISPLAY A DISPLAY B		-0.0024 -0.0020		0.0024 0.0020
	SELF TEST #15	DISPLAY A DISPLAY B		-0.0048 -0.0020		0.0048 0.0020
	20.0	(Hz			:	
	SELF TEST #9	DISPLAY A DISPLAY B		-0.0080 -0.0052		0.0080 0.0052
	SELF TEST #12	DISPLAY A DISPLAY B		-0.0048 -0.0043		0.0048 0.0043
	SELF TEST #13	DISPLAY A DISPLAY B		-0.0048 -0.0043		0.0048 0.0043
	SELF TEST #14	DISPLAY A DISPLAY B		-0.0096 -0.0036		0.0096 0.0036
	SELF TEST #15	DISPLAY A DISPLAY B		-0.0192 -0.0360		0.0192 0.0360
4-15	CAPACITANCE ACCUR	ACY TEST				
	10pF Range Speed: MED Level: HIGH		_			
		9.95kHz	C D	C.V16pF 010		C.V. + .16pF .010
		10.0kHz	C D	C.V16pF 010		C.V. + .16pF .010
		20.0kHz	C D	C.V24pF 012		C.V. + .24pF .012

Paragraph	TEST			Results	
Number			Minimum	Actual	Maximum
4-15	CAPACITANCE ACCURACY TEST  100pF Range Speed: MED	(Cont'd)			
	Level: HIGH 995Hz	C D	C.V6pF 004		C.V. + .6pF .004
	1.00kHz	C D	C.V35pF 0035		C.V. + .35pF .0035
	1.99kHz	C D	C.V65pF 0065		C.V. + .65pF .0065
	4.98kHz	C D	C.V25pF 0017		C.V. + .25pF .0017
	9.95kHz	C D	C.V70pF 0046		C.V. + .70pF .0046
	10.0kHz	C D	C.V70pF 0046		C.V. + .70pF .0046
	20.0kHz	C D	C.V 2.20pF 0140		C.V. + 2.20pF .0140
	1000pF Range Speed: MED Level: HIGH				
	100Hz	C D	C.V 5.0pF 0050	-	C.V. + 5.0pF .0050
	120Hz	C D	C.V 5.0pF 0050		C.V. + 5.0pF .0050
	199Hz	C D	C.V 6.5pF 0065		C.V. + 6.5pF .0065
	498Hz	C D	C.V 2.5pF 0017		C.V. + 2.5pF .0017
	995Hz	C D	C.V 2.5pF 0017		C.V. + 2.5pF .0017
	1.00kHz	C D	C.V 1.5pF 0011		C.V. + 1.5pF .0011
	1.99kHz	C D	C.V 2.5pF 0017		C.V. + 2.5pF .0017
	4.98kHz	C D	C.V 2.5pF 0017		C.V. + 2.5pF .0017
	9.95kHz	C D	C.V 7.0pF 0046		C.V. + 7.0pF .0046

Paragraph					Results	
Number	TES	T		Minimum	Actual	Maximum
4-15	CAPACITANCE ACCUR	ACY TEST	(Cont'd)			
		10.0kHz	C D	C.V 7.0pF 0046		C.V. + 7.0pF .0046
		20.0kHz	C D	C.V 11.0pF 0070		C.V. + 11.0pF .0070
	1000pF Range Speed: SLOW Level: HIGH					
	Level. High	100Hz	C D	C.V 5.0pF 0050		C.V. + 5.0pF .0050
		1.00kHz	C D	C.V 1.5pF 0011		C.V. + 1.5pF .0011
		10.0kHz	C D	C.V 7.0pF 0046		C.V. + 7.0pF .0046
	1000pF Range Speed: FAST					
	Level: HIGH	100Hz	C D	C.V 10pF 010		C.V. + 10pF .010
		1.00kHz	C D	C.V 6pF 006		C.V. + 6pF .006
		10.0kHz	C D	C.V 16pF 014		C.V. + 16pF .014
	1000pF Range Level: LOW					
	Speed: MED	1.00kHz	C D	C.V 12pF 011		C.V. + 12pF .011
	Speed: FAST	1.00kHz	C D	C.V 100pF 100		C.V. + 100pF .100
	Speed: MED	10.0kHz	C D	C.V 32pF 027		C.V. + 32pF .027
	1000pF Range HIGH SPEED C Speed: MED					
	Level: HIGH	1.0	0kHz	C.V 1.5pF		C.V. + 1.5pF
	Level: LOW	1.0	0kHz	C.V 12pF		C.V. + 12pF
	Level: HIGH	10.	0kHz	C.V 7.0pF		C.V. + 7.0pF

Paragraph	т	EST			Results	
Number	*			Minimum	Actual	Maximum
4-17	IMPEDANCE ACCUR	ACY TEST				
	100mΩ Range	100Hz	θ	- 1.2°		1.2°
		120Hz	θ	- 1.2°		1.2°
		1.00kHz	θ	- 1.2°		1.2°
		10.0kHz	θ	- 2.2°		2.2°
	lΩ Range	100Hz	θ	7°		.7°
		120Hz	θ	7°		.7°
		1.00kHz	θ	7°		.7°
		10.0kHz	θ	- 1.2°		1.2°
	10Ω Range	100Hz	θ	35°		.35°
		120Hz	θ	35°		.35°
		1.00kHz	θ	35°		.35°
		10.0kHz	θ	65°		.65°
	100Ω Range	100Hz	Z   0	C.V15Ω 15°		C.V. + .15Ω .15°
		120Hz	Z   Θ	C.V15Ω 15°		C.V. + .15Ω .15°
		1.00kHz	Z   Θ	C.V15Ω 15°		C.V. + .15Ω .15°
		10.0kHz	Z   0	C.V25Ω 25°		C.V. + .25Ω .25°
	lkΩ Range	100Hz	Z   0	C.V 6Ω 6°		C.V. + 6Ω .6°
		120Hz	Z   0	C.V 6Ω 6°		C.V. + 6Ω .6°
		1.00kHz	Z   0	C.V 6Ω 6°		C.V. + 6Ω .6°
		10.0kHz	Z   0	C.V 10Ω - 1.0°		C.V. + 100 1.0°
	10kΩ Range	100Hz	Z   0	C.V06kΩ 6°		C.V. + .06kΩ .6°
		120Hz	Z   Θ	C.V06kΩ 6°		C.V. + .06kΩ .6°

Paragraph	TEST			Results	
Number	1201		Minimum	Actual	Maximum
4-17	IMPEDANCE ACCURACY T	EST (Cont'd)			
	1.0	OOkHz  Z  θ	C.V06kΩ 6°		C.V. + .06kΩ .6°
	10	.OkHz   Z   0	C.V1kΩ - 1.0°		C.V. + .1kΩ 1.0°
	100kΩ Range	100Hz  Z  Ө	C.V6kΩ 6°		C.V. + .6kΩ .6°
		120Hz   Z   0	C.V6kΩ 6°		C.V. + .6kΩ .6°
	1.0	00kHz   Z   0	C.V6kΩ 6°		C.V. + .6kΩ .6°
	10	.0kHz  Z  θ	C.V 1kΩ - 1.0°		C.V. + 1kΩ 1.0°
4-19	INDUCTANCE ACCURACY	TEST			
	Frequency: 1.00 Speed: MED Level: HIGH	kHz DISPLAY A DISPLAY B	- 25.51 012		- 25.15 .012
4-21	INTERNAL DC BIAS ACCI	URACY TEST PTION 001 ONLY)			
	Precharge + 40V	Wait: 10sec.			
	.00V	Wait: 30sec.	01V		.01V
	+ .01V	Wait: 20sec.	ov		.02V
	+ 9.99V	Wait: 10sec.	9.95V		10.03V
	+ 10.0V	Wait: 10sec.	9.015V		10.85V
	+ 40.0V	Wait: 15sec.	39.76V		40.24V
	01V	Wait: 30sec.	02V		ov
	- 9.99V	Wait: 10sec.	- 10.1V		- 9.88V
	- 10.0V	Wait: 10sec.	- 10.135V		- 9.865V
	- 40.0V	Wait: 15sec.	- 40.44V		- 39.56V

Model 4276A SECTION V

Table 5-1. Adjustable Components (Sheet 1 of 2)

Reference Designator	Name of Control	Purpose
Al R3 (Para. 5-18)		Sets the reset voltage level.
A2 R1 (Para. 5-19)	OSC/LEVEL	Sets the test signal level.
A2 R4 A2 R5 . A2 R6 A2 R7 (Para. 5-24)	10K/MAG 1K/MAG 100/MAG 100K/MAG	Fine adjustment for range resistor.
A2 R13 (Para. 5-20)	ER/OFS	Eliminates dc offset voltage from ERR buffer amplifier.
A2 R14 (Para. 5-23)	LF TRK	Compensates for phase offset between the $E_{\rm DUT} and \ E_{\rm RR}$ buffer amplifiers.
A2 R15 (Para. 5-20)	ED/OFS	Eliminates dc offset voltage from the $E_{\rm DUT} buffer \ amplifier. \\$
A2 R16 (Para. 5-20)	LC/OFS	Eliminates dc offset voltage from the I/V amplifier.
A2 R17 (Para. 5-20)	DIF/OFS	Eliminates dc offset voltage from the differential amplifier and the process amplifier.
A2 R25 (Para. 5-20)	AM1/OFS	Eliminates dc offset voltage from the AM1 amplifier.
A2 R26 A2 R27 A2 R35 (Para. 5-22)	AF 1/2 AF 1/5 AM 1/10	Sets AM/AF attenuation.
A2 R39 (Para. 5-20)	AF/OFS	Eliminates dc offset voltage from the AF amplifier.
A2 R40 (Para. 5-20)	AM2/OFS	Eliminates dc offset voltage from the AM2 amplifier.
A2 R99 A2 R100 (Para. 5-21)	ZERO ZERO SHIFT	Sets the zero detect voltage level.
A2 C2 C3 (Para. 5-25)	10K/PH 100K/PH	Eliminates range resistor phase offset.
A2 C7 (Para. 5-23)	HF TRK	Compensates for phase offset between the $E_{\rm DUT}  {\rm and}   E_{\rm RR}$ buffer amplifiers.
A2 C12 A2 C13 A2 C18 (Para. 5-22)	PH 1/2 PH 1/5 PH 1/10	Eliminates AM/AF attenuator phase offset.

Table 5-1. Adjustable Components (Sheet 2 of 2)

Reference Designator	Name of Control	Purpose
A4 R14 (Para. 5-17) (Para. 5-18)	V-ADJ	Sets power supply voltage.
A4 R15 (Para. 5-17)	F-ADJ	Sets power supply switching frequency.
A22 R6 A22 R7 A22 R8 (Para. 5-26)	ZERO FS	Eliminates bias voltage offset and gain error.

Model 4276A SECTION V

# SECTION V

# **ADJUSTMENTS**

#### 5-1. INTRODUCTION

This section describes the adjustments and checks required to return the 4276A to the specifications listed in Table 1-1 after repairs have been made. These adjustments and checks can also be performed along with periodic maintenance to keep the instrument in optimum condition. The recommended operating adjustment cycle for the 4276A is twice a year. All adjustable components referred to in the adjustment procedures are listed in Table 5-1. If proper performance cannot be achieved after adjustment, refer to the troubleshooting procedures described in Section VIII.

#### Note

All options that the instrument is normally equipped with must be installed before the adjustments described in this section are made. If the options are installed after the instrument is adjusted, the specifications listed in Table 1-1 are not guaranteed.

### Note

To ensure proper results and correct instrument operation, Hewlett- Packard suggests a 30-minute warm-up and stabilization period before performing any of the adjustments described here.

# 5-3. SAFETY REQUIREMENTS

5-4. Although the 4276A was designed in accordance with international safety standards, this manual contains information, cautions, and warnings which must be followed to ensure operator safety and to keep the instrument in a safe and serviceable condition. Adjustments described in this section should be performed by qualified service personnel only.

### WARNING

ANY INTERRUPTION OF THE PROTECTIVE (GROUNDED) CONDUCTOR (INSIDE OR OUTSIDE THE INSTRUMENT) OR DISCONNECTION OF THE PROTECTIVE EARTH TERMINAL IS LIKELY TO MAKE THE

INSTRUMENT DANGEROUS. INTENTIONAL INTERRUPTION, FOR ANY REASON, IS PROHIBITED.

- 5-5. The removal or opening of covers for removal or adjustment of parts, other than those which are accessible by hand, will expose live parts.
- 5-6. Capacitors in the instrument may still be charged even if the instrument has been disconnected from the power source (AC line) for an extended period of time.

#### WARNING

ADJUSTMENTS DESCRIBED IN THIS SECTION ARE PERFORMED WITH POWER SUPPLIED AND PROTECTIVE COVERS REMOVED. ENERGY EXISTING AT MANY POINTS MAY, IF CONTACTED, RESULT IN SERIOUS PERSONAL INJURY.

# 5-7. EQUIPMENT REQUIRED

5-8. All the equipment required to perform the adjustments described in this section are listed in Table 4-1 on page 4-0. Each piece of equipment listed in Table 4-1 should be calibrated to satisfy its own specifications, as well as those of the required characteristics. If the recommended model is not available, any instrument whose specifications equal or surpass those of the recommended model may be used instead.

#### 5-9. FACTORY SELECTED COMPONENTS

5-10. Factory selected components are identifiable by an asterisk (\*) adjacent to the reference designator on the schematic diagrams in Section VIII (only nominal values are given). Table 5-2 lists the reference designators of all factory selected components. Also listed in Table 5-2 are the nominal value range of each component and a brief description of how each component affects instrument performance.

Adjustable components, with reference designators, are listed in Table 5-1. This table also lists the name of the adjustment and its purpose.

SECTION V Model 4276A

#### 5-11. ADJUSTMENT RELATIONSHIPS

5-12. The adjustment procedures described in this section, beginning with paragraph 5-17, are interactive and therefore should be performed in the sequence given. Ignoring or changing the order of the procedures may make it impossible to obtain optimum instrument performance. Table 5-3 lists the necessary adjustment procedures to follow after the instrument has been repaired.

# 5-13. ADJUSTMENT LOCATIONS

5-14. To help locate the appropriate adjustment points, the locations of the components to be adjusted are illustrated throughout the adjustment procedures. The locations of factory selected components, connectors, and other components related to the adjustments are shown in the individual board assembly-component illustrations (fold-out service sheets) in Section VIII.

#### 5-15. INITIAL OPERATING PROCEDURE

5-16. Before proceeding with the adjustments described starting in paragraph 5-17, perform the following preliminary procedure. This procedure provides access to the various adjustment points and facilitates a thoroughgoing adjustment.

# [BASIC OPERATING CHECK]

Check that the instrument's line voltage selector switch, located on the rear panel, is set to the position appropriate for the local line voltage. This should be performed before proceeding with any of the adjustments.

After the recommended 30-minute warm-up period, the instrument should pass the SELF TEST (no error message should appear). If the instrument displays an error message or does not have the correct control settings, refer to the troubleshooting procedures given in Section VIII.

# [TOP COVER REMOVAL]

- Fully loosen the top-cover retaining screw located at the rear of the top cover.
- b. Slide the top cover towards the rear and lift off.

#### WARNING

DC VOLTAGES—MAXIMUM ±16V—ARE PRESENT AT TEST POINTS ON THE A1, A2, AND A4 BOARDS. AS A SAFETY PRECAUTION AGAINST POSSIBLE ELECTRICAL SHOCK HAZARDS AND RESULTANT INJURY, USE INSULATED TOOLS FOR ALL ADJUSTMENTS.

#### Note

- 1. To select a numbered SELF TEST:
- a) Press the SELF TEST key and the FREQUENCY/DC BIAS Select key in order.
- b) The number displayed on the FREQUENCY/DC BIAS DISPLAY is the SELF TEST number.
- c) The SELF TEST number can be changed by pressing the FREQUENCY/ DC BIAS Step Control Key.
- 2. To obtain initial control settings (erase continuous memory):
- a) Turn off the instrument.
- b) Press and hold both FREQUENCY/
  DC BIAS Step Control keys (②
  and ③).
- c) Turn on the instrument.
- d) Release both FREQUENCY/DC BIAS Step Control Keys after the displays appears.

Model 4276A SECTION V

Table 5-2. Factory Selected Components

Reference Designator	Nominal Value Range	Effect on Performance
A2 R58 (Para. 5-19)	HP P/N: 0757-0278 , R: FXD 1.78kΩ ► HP P/N: 0698-0084 , R: FXD 2.15kΩ HP P/N: 0698-0085 , R: FXD 2.61kΩ	Changes test signal level. If signal level is too high, use more resistance; if too low, use less resistance.
A2 C72 (Para. 5-21)	►HP P/N: 0160-5597 , C: FXD 5pF HP P/N: 0160-5593 , C: FXD 12pF	Compensates for junction capacitance of FET switch. If [DISP A] <72, remove C72; if [DISP A]>88, use more capacitance.
A2 C14 (Para. 5-22)	HP P/N: 0140-0191 , C: FXD 56pF ► HP P/N: 0160-2202 , C: FXD 75pF HP P/N: 0160-2204 , C: FXD 100pF HP P/N: 0140-0196 , C: FXD 150pF	Eliminates phase offset in the AF attenuator. If [DISP B]<-4, use more capacitance; if [DISP B]>+4, use less capacitance.
A2 C15 (Para 5-22)	HP P/N: 0160-4794 , C: FXD 5.6pF ► HP P/N: 0160-4789 , C: FXD 15pF HP P/N: 0160-4787 , C: FXD 22pF	Eliminates phase offset in the AF attenuator. If [DISP B]<-4, use more capacitance; if [DISP B]>+4, use less capacitance.
A2 C16 (Para 5-22)	HP P/N: 0160-2204 , C: FXD 100pF ► HP P/N: 0140-0196 , C: FXD 150pF HP P/N: 0140-0198 , C: FXD 200pF	Eliminates phase offset in the AM2 attenuator. If [DISP B]<-4, use less capacitance; if [DISP B]>+4, use more capacitance.
A2 R93 (Para. 5-23)	HP P/N: 0757-0463 , R: FXD 82.5kΩ ► HP P/N: 0757-0464 , R: FXD 90.9kΩ HP P/N: 0757-0465 , R: FXD 100kΩ	Compensates for buffer amplifier phase offset. If [DISP B]<-1, use more resistance; if [DISP B]>+1, use less resistance.
A2 C55 (Para. 5-23)	HP P/N: 0160-4806 , C: FXD 39pF ► HP P/N: 0160-4803 , C: FXD 68pF HP P/N: 0160-4801 , C: FXD 100pF	Compensates for buffer amplifier phase offset. If [DISP B]<-1, use more capacitance; if [DISP B]>+1, use less capacitance.
A2 C5 (Para. 5-25)	HP P/N: 0160-4795 , C: FXD 4.7pF ► HP P/N: 0160-4788 , C: FXD 18pF HP P/N: 0160-4786 , C: FXD 27pF HP P/N: 0160-4806 , C: FXD 39pF HP P/N: 0160-4804 , C: FXD 56pF	Eliminates phase offset of range resistor. If [DISP B]<-2, use less capacitance; if [DISP B]>+2, use more capacitance.
A2 C4 (Para. 5-25)	HP P/N: 0160-4795 , C: FXD 4.7pF ► HP P/N: 0160-4791 , C: FXD 10pF HP P/N: 0160-4789 , C: FXD 15pF HP P/N: 0160-4787 , C: FXD 22pF	Eliminates phase offset of range resistor. If [DISP B]<-2, use less capacitance; if [DISP B]>+2, use more capacitance.

Note:  $\blacktriangleright$  indicates the component usually used.

Table 5-3. Adjustment Requirements

Assembly	Repaired or Replaced	Required Adjustments
A1	(04276-66501)	Para. 5-17 and 5-18
A2	(04276-66502)	Para. 5-19 through 5-25
A4	(04276-66504)	Para. 5-17 and 5-18
A5	(04276-66505)	None
A6	(04276-66506)	None
A21	(04276-66521)	None
A22	(04276-66522)	Para. 5-17, 5-18 and 5-26 (only if A22 is added or removed)
16064A	A2 (16064-66502)	Para. 5-17 and 5-18 (only if 16064 A A2 is added or removed)

Model 4276A SECTION V

#### **ADJUSTMENTS**

# 5-17. A4 POWER SUPPLY ADJUSTMENT

### PURPOSE:

Adjusts the output voltages and the switching frequency of the switching power supply.

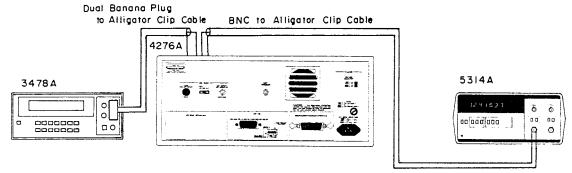


Figure 5-1. A4 Power Supply Adjustment Setup.

### EQUIPMENT:

Digital Voltmeter	HP	3478A
Frequency Counter	ΗP	5314A
BNC-to-Alligator Clip Cable	HP	11000A
Dual Banana Plug-to-Alligator Clip Cable	HP	11002

### PROCEDURE:

- 1. Connect the 3478A HI and LOW inputs to A4TP1 (+5V DIG) and A4TP2 (DIG GND), respectively, with the dual banana plug-to-alligator clip cable as shown in Figure 5-1.
- 2. Set the 3478A function to DCV.
- 3. Turn on the 4276A.
- 4. Adjust A4R14 (V-ADJ) until the reading on the 3478A is +5.10V±0.01V.
- 5. Connect the 3478A LOW input to A4TP10 (GND).
- 6. Verify that the voltages at A4TP5 (+16V)/TP6(-16V)/TP7(+8V) are within test limits in the table below:

Test Point	Test Limits
A4TP5 (+16V)	+15V to +17V
A4TP6 (-16V)	-15V to -17V
A4TP7 (+8V)	+8V to +10V

- 7. Connect the 5314A to A4TP9 (DIG PWM) and A4TP10 (GND) with a BNC-to-alligator clip cable as shown in Figure 5-1.
- 8. Adjust A4R15 (F-ADJ) until the reading on the 5314A is 21kHz±0.1kHz.

Note

If necessary, adjust the trigger level on the 5314A.

# 5-18. A1 RESET VOLTAGE ADJUSTMENT

### PURPOSE:

Adjusts the threshold voltage for instrument reset to +4.80V.

#### Note

Perform the adjustment described in paragraph 5-17 before performing this adjustment.

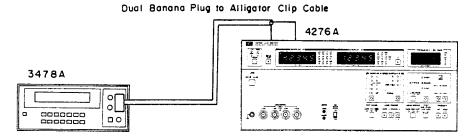
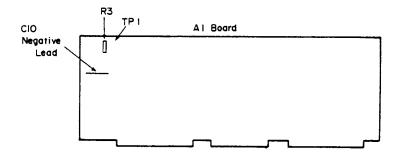


Figure 5-2. Al Reset Voltage Adjustment Setup.

# EQUIPMENT:

### PROCEDURE:

- 1. Connect the HI and LOW inputs of the 3478A to AlTPl and the negative lead of AlCl0, respectively, with a dual banana-to-alligator clip cable as shown in Figure 5-2.
- 2. Set the 3478A function to DCV.
- 3. Turn on the 4276A.
- 4. Adjust AlR3 fully CCW.
- 5. Adjust A4R14 (V-ADJ) until the reading on the 3478A is +4.80V±0.02V.
- 6. Gradually adjust A1R3 CW until the 4276A display lamps go on and off.



### [Reset Voltage Level Check]

- 7. Adjust A4R14 (V-ADJ) until the reading on the 3478A is +5V.
- 8. Gradually adjust A4R14 (V-ADJ) until the reading on the 3478A is +4.83V. Verify that the 4276A display lamps have not gone off.
- 9. Gradually adjust A4R14 (V-ADJ) until the reading on the 3478A is +4.77V. Verify that the 4276A display lamps have gone on and off.
- 10. If the state of display lamps is different from the state described in steps 8. and 9., repeat steps 3. through 9.
- 11. Adjust A4R14 (V-ADJ) until the reading on the 3478A is +5.10V±0.02V.

#### Note

When the Comparator/Handler Interface board assembly (HP P/N 16064-66502) has been installed, connect the model 16064A to the connector on the rear panel before making this adjustment.

# 5-19. A2 TEST SIGNAL LEVEL ADJUSTMENT

#### PURPOSE:

Adjusts the test signal level.

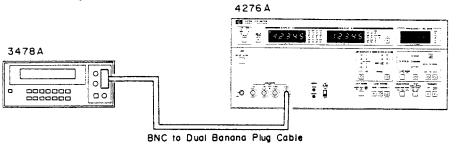


Figure 5-3. A2 Test Signal Level Adjustment Setup.

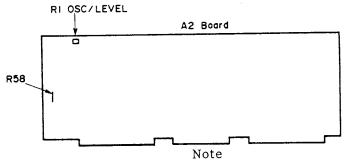
### EQUIPMENT:

# PROCEDURE:

- 1. Connect the 3478A to the 4276A H<sub>CUR</sub> terminal with a BNC-to dual banana plug cable, as shown in Figure 5-3.
- 2. Set the 3478A function to ACV.
- 3. Turn on the 4276A and set the controls as follows:

DISPLAY A/B functions	C-ESR/G
Test Frequency	1.00kHz
DC BIAS	OFF
LC   Z   RANGE	100nF
TEST SIG LEVEL	HIGH
Other Controls	Any Settings

- 4. Adjust A2R1 (OSC/LEVEL) until the reading on the 3478A is 1Vrms ±10mVrms.
- 5. Set the test signal level to LOW, and verify that the reading on the 3478A is  $50 \text{mVrms} \pm 10 \text{mVrms}$ .



If this adjustment cannot be performed successfully, replace A2R58 in accordance with the instructions given in Table 5-2.

# 5-20. A2 DC OFFSET ADJUSTMENT

#### PURPOSE:

Eliminates residual dc offset voltages from amplifier output voltages on the A2 board.

Dual Banana Plug to Alligator Clip Cable

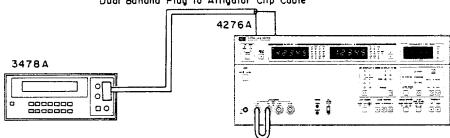


Figure 5-4. A2 DC Offset Adjustment Setup.

### EQUIPMENT:

# PROCEDURE:

1. Connect the 4276 A's LCUR and LPOT terminals to each other with a BNC-to-BNC cable.

[Differential Amplifier and Process Amplifier DC Offset Adjustment]

- 2. Connect the 3478A HI and LOW inputs to A2TP7 (PD/IN) and A2TP4 (GND), respectively, as shown in Figure 5-4.
- 3. Set the 3478A function to DCV.
- 4. Turn on the 4276A.
- 5. Set the 4276A to SELF TEST 16. (Refer paragraph 5-15 for the procedure.)
- 6. Adjust A2R17 (DIF/OFS) until the reading on the 3478A is  $0V \pm 0.2mV$ .

# [AM-l Amplifier DC Offset Adjustment]

- 7. Connect the 3478A HI input to A2TP8 (AM1).
- 8. Adjust A2R25 (AM1/OFS) until the reading on the 3478A is  $0V\pm0.5mV$ .

# [AF Amplifier DC Offset Adjustment]

- 9. Connect the 3478A HI input to A2TP9 (AF).
- 10. Adjust A2R39 (AF/OFS) until the reading on the 3478A is  $0V \pm 1mV$ .

### [AM-2 Amplifier DC Offset Adjustment]

- 11. Connect the 3478A HI input to A2TP10 (PD/IN).
- 12. Adjust A2R40 (AM2 OFS) until the reading on the 3478A is 0V±5mV.

# [I/V Amplifier DC Offset Adjustment]

- 13. Set the 4276A to SELF TEST 17. (Refer to paragraph 5-15 for the procedure.)
- 14. Connect the 3478A HI input to A2TP5 (I/V OUT).
- 15. Adjust A2R16 (LC/OFS) until the reading on the 3478A is  $0V \pm 0.5 mV$ .

# [Err Buffer Amplifier DC Offset Adjustment]

- 16. Connect the 3478A HI input to A2TP7 (VRD/IN).
- 17. Adjust A2R13 (ER/OFS) until the reading on the 3478A is 0V±0.2mV.

# [HPOT Buffer Amplifier DC Offset Adjustment]

- 18. Set the 4276A to SELF TEST 18. (Refer to paragraph 5-15 for the procedure.)
- 19. Adjust A2R15 (ED/OFS) until the reading on the 3478A is 0V±0.2mV.
- 20. Press the SELF TEST key to release the self test mode.

Procedure	SELF TEST	Digital Voltme	ter Inputs	Adjustment	r:_:_	
. 10004470	Number	HI	LO	Component	Limit	
(1)	16	A2TP7 (VRD/IN)	A2TP4 (GND)	A2R17 (DIF/OFS)	0±0.2mV	
(2)	16	A2TP8 (AM1)	A2TP4 (GND)	A2R25 (AM1/OFS)	0±0.5mV	
(3)	16	A2TP9 (AF)	A2TP4 (GND)	A2R39 (AF/OFS)	0±1mV	
(4)	16	A2TP10 (PD/IN)	A2TP4 (GND)	A2R40 (AM2 OFS)	0±5mV	
(5)	17	A2TP5 (IV OUT)	A2TP4 (GND)	A2R16 (LC/OFS)	0±0.5mV	
(6)	17	A2TP7 (VRD IN)	A2TP4 (GND)	A2R13 (ER/OFS)	0±0.2mV	
(7)	18	A2TP7 (VRD IN)	A2TP4 (GND)	A2R15 (ED/OFS)	0±0.2mV	

Note

Use a BNC-to-BNC coaxial cable to interconnect the  $L_{\text{CUR}}$  and  $L_{\text{POT}}$  UNKNOWN terminals.

# [[DC Offset Check]]

# [SHORT Offset Check]

1. Set the 4276A's controls as follows:

DISPLAY A/B functions	Z  - θ
Test Frequency	1.00kHz
DC BIAS	OFF
LC   Z   RANGE	$100$ m $\Omega$
MEAS SPEED	MED
TRIGGER	INT
Other Controls	Any Settings

- 2. Connect the 3478A HI and LOW inputs to A2TP10 (PD/IN) and A2TP4 (GND), respectively, with a dual banana plug-to-alligator clip cable.
- 3. Set the 4276A to SELF TEST 23. (Refer to paragraph 5-15 for the procedure.)
- 4. Press the MAN/EXT key on the front-panel once.
- 5. Connect the SHORT termination to the UNKNOWN terminals.
- 6. Verify that the reading on the 3478A is 0mV±100mV.

### [OPEN Offset Check]

- 7. Press the SELF TEST key to release the self test mode.
- 8. Set the 4276 A's controls as follows:

DISPLAY A/B functions	C - D
Test Frequency	5.00kHz
DC BIAS	OFF
CKT MODE	
LC   Z   RANGE	
MEAS SPEED	•
TEST SIG LEVEL	LOW
TRIGGER	

- 9. Set the 4276A to SELF TEST 23.
- 10. Press MAN/EXT key on the front-panel once.
- 11. Connect the OPEN termination to the UNKNOWN terminals.
- 12. Verify that the reading on the 3478A is 0mV±100mV.

### Note

If the SHORT and OPEN Offset checks cannot be performed successfully, repeat steps  $1\ \mathrm{throgh}\ 20\ \mathrm{of}\ \mathrm{the}\ \mathrm{main}\ \mathrm{procedure}.$ 

# 5-21. A2 ADC ZERO ADJUSTMENT

#### PURPOSE:

Eliminates residual dc offset voltage from the integrator output.

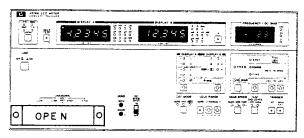


Figure 5-5. A2 ADC Zero Adjustment Setup.

#### Note

DO NOT extend A2 board assembly in this adjustment.

# EQUIPMENT:

Termination ...... OPEN: HP 16074A

#### PROCEDURE:

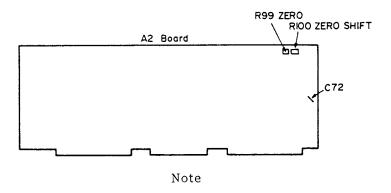
- 1. Turn on the 4276A.
- 2. Connect the OPEN termination of the 16074A directly to the UNKNOWN terminals.
- 3. Set the 4276A's controls as follows:

Test Frequency	1 006Hz
DC BIAS	OFF
MEAS SPEED	MED
TEST SIG LEVEL	HIGH
TRIGGER	INT
Other Controls	Any setting

- 4. Set the 4276A to SELF TEST 8. (Refer to paragraph 5-15 for the procedure.)
- 5. Set A2R100 (ZERO SHIFT) at the midway point.
- 6. Adjust A2R99 (ZERO) until [A] equals [B].

### Note

- $[\mbox{\sc A}]$  and  $[\mbox{\sc B}]$  are defined as the values displayed on DISPLAY  $\mbox{\sc A}$  and DISPLAY  $\mbox{\sc B},$  respectively.
- 7. Adjust A2R100 (ZERO SHIFT) until
  - [A] = 20 counts  $\pm 1$  count and [B] = -20 counts  $\pm 1$  count.
- 8. Set the measurement speed mode to FAST. Verify that
  - [A] =  $80 \text{ counts} \pm 8 \text{ count}$  and [B] =  $-80 \text{ counts} \pm 8 \text{ counts}$ .



If the test in step 8 cannot be performed successfully, replace A2C72 in accordance with the instructions given in Table 5-2.

# 5-22. A2 AM/AF ATTENUATOR ADJUSTMENT

# PURPOSE:

Adjusts the attenuation and phase of the attenuators in the AM and AF circuits.

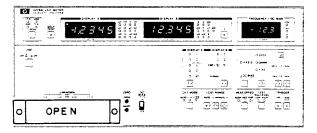


Figure 5-6. A2 AM/AF Attenuator Adjustment Setup.

# **EQUIPMENT:**

Termination ...... OPEN: HP 16074A PROCEDURE:

- 1. Connect the OPEN termination of the 16074A directly to the UNKNOWN terminals.
- 2. Turn on the 4276A.
- 3. Set the 4276A's controls as follows:

Test Frequency	l.00kHz
DC BIAS	
MEAS SPEED	
TEST SIG LEVEL	
TRIGGER	
Other Controls	Any settings

### [AF 1/2 Attenuator Adjustment]

- 4. Set the 4276A to SELF TEST 12. (Refer to paragraph 5-15 for the procedure.)
- 5. Adjust A2R26 (AF 1/2) until [A] is 0 counts±4 counts. Verify that [B] is 0 count±10 counts.

#### Note

[A] and [B] are defined as the values displayed on DISPLAY A and DISPLAY B, respectively.

# [AF 1/5 Atteunator Adjustment]

- 6. Set the 4276A to SELF TEST 13.
- 7. Adjust A2R27 (AF 1/5) until [A] is 0 counts±4 counts. Verify that [B] is 0 counts±10 counts.

# [AM2 1/10 Attenuator Adjustment]

- 8. Set the 4276A to SELF TEST 14.
- 9. Adjust A2R35 (AM 1/10) until [A] is 0 counts±4 counts. Verify that [B] is 0 counts±10 counts.

### [AMl Attenuator Check]

- 10. Set the 4276A to SELF TEST 15.
- 11. Verify that [A] and [B] are 0 counts ±40 counts and 0 counts ±15 counts, respectively.

### [VRD Full Scale Check]

- 12. Press the SELF TEST key to release the SELF TEST mode.
- 13. Set the test frequency to 20kHz.
- 14. Set the 4276A to SELF TEST 9
- 15. Verify that [A] and [B] are 0 counts ± 20 counts and 0 counts ± 10 counts, respectively.

# [AF 1/2 Attenuator Phase Adjustment]

- 16. Set the 4276A to SELF TEST 12.
- 17. Adjust A2Cl2 (PH 1/2) until [B] is 0 counts±20 counts. Verify that [A] is 0 counts±4 counts.

#### Note

If this adjustment cannot be performed successfully, replace A2C14 in accordance with the instructions given in described in Table 5-2.

# [AF 1/5 Attenuator Phase Adjustment]

- 18. Set the 4276A to SELF TEST 13.
- 19. Adjust A2Cl3 (PH 1/5) until [B] is 0 counts±4 counts. Verify that [A] is 0 counts±20 counts.

#### Note

If this adjustment cannot be performed successfully, replace A2Cl5 in accordance with the instructions given in Table 5-2.

# [AM2 1/10 Attenuator Phase Adjustment]

- 20. Set the 4276A to SELF TEST 14.
- 21. Adjust A2C18 (PH 1/10) until [B] is 0 counts±4 counts. Verify that [A] is 0 counts±40 counts.

#### Note

If this adjustment cannot be performed successfully, replace A2Cl6 in accordance with the instructions given in Table 5-2.

#### [AMI Attenuator Phase Check]

- 22. Set the 4276A to SELF TEST 15.
- 23. Verify that [A] and [B] are 0 counts  $\pm 120$  counts and 0 counts  $\pm 200$  counts, respectively.

Procedure	SELF TEST	Test	Adjustment Component	Adjustmer (count		Check I (count	ı
	Number	Frequency	Component	DISPLAY A	DISPLAY B	DISPLAY A	DISPLAY B
(1)	12	1kHz	A2R26 (AF 1/2)	0±4			0±10
(2)	13	1kHz	A2R29 (AF 1/5)	0±4			0±10
(3)	14	1kHz	A2R35 (AM 1/10)	0±4			0±10
(4)	15	1kHz				0±40	0±15
(5)	9	20kHz				0±20	0±10
(6)	12	20kHz	A2C12 (PH 1/2)		0±4	0±20	
(7)	13	20kHz	A2C13 (PH 1/5)		0±4	0±20	
(8)	14	20kHz	A2C18 (PH 1/10)		0±4	0±40	
(9)	15	20kHz				0±120	0±200

### 5-23. A2 LF/HF TRACKING ADJUSTMENT

#### PURPOSE:

Compensates for the difference between the phase shift caused by the Err circuit and the phase shift caused by the Edut circuit in the low frequency range (LF) and high frequency (HF) ranges.

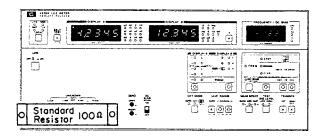


Figure 5-7. A2 LF/HF Tracking Adjustment Setup.

### EQUIPMENT:

### PROCEDURE:

[LF Adjustment]

- 1. Connect the  $100\Omega$  standard resistor directly to the UNKNOWN terminals.
- 2. Reset continuous memory (see Note 2 in paragraph 5-15) and turn on the 4276A.
- 3. Set the 4276A's controls as follows:

DISPLAY A/B functions	Z   - θ
Test Frequency	100Hz
DC BIAS	OFF
CKT MODE	Any Setting
LC   Z   RANGE	lkΩ
MEAS SPEED	MED
TEST SIG LEVEL	HIGH
TRIGGER	INT

- 4. Set the 4276A to SELF TEST 2. (Refer to paragraph 5-15 for the procedure.)
- 5. Adjust A2R14 (LF TRK) until the 4276A displays 0 counts +1 count on DISPLAY B.

### Note

If this adjustment cannot be paformed successfully, replace A2R93 in accordance with the instructions given in Table 5-2.

SECTION V Model 4276A

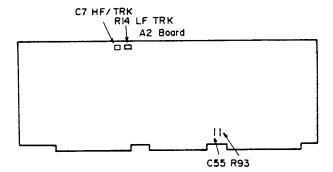
### **ADJUSTMENTS**

# [HF Adjustment]

- 6. Press the SELF TEST key to release the SELF TEST mode.
- 7. Set the test frequency to 20kHz.
- 8. Set the 4276A to SELF TEST 2.
- 9. Adjust A2C7 (HF/TRK) until the 4276A displays 0 counts±1 count on DISPLAY B.

### Note

If this test cannot be performed successfully, replace A2C55 in accordance with instructions given in Table 5-2.



# 5-24. A2 RANGE RESISTOR MAGNITUDE ADJUSTMENT

# PURPOSE:

Adjusts the range resistor values.

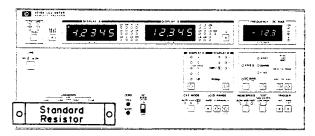


Figure 5-8. A2 Range Resistor Magnitude Adjustment Setup.

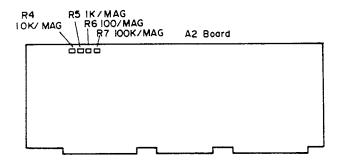
# EQUIPMENT:

#### PROCEDURE:

- l. Calculate the reciprocal of the calibrated values of the  $100\Omega/lk\Omega/10k\Omega$  and  $100k\Omega$  standard resistors; [l/c.v.].
- 2. Reset continuous memory (see Note 2 in paragraph 5-15).
- 3. Connect the  $100k\Omega$  standard resistor directly to the UNKNOWN terminals.
- 4. Set the 4276A's controls as follows:

DISPLAY A/B functions	Z   - θ
Test Frequency	1.00kHz
DC BIAS	OFF
CKT MODE	Any Setting
LC   Z   RANGE	
MEAS SPEED	MED
TEST SIG LEVEL	HIGH
TRIGGER	INT

5. Set the 4276A to SELF TEST 2. (Refer to paragraph 5-15 for the procedure.)



### [100K/MAG Adjustment]

6. Adjust A2R7 (100K/MAG) until the 4276A displays [1/c.V.]±1 count on DISPLAY A. Verify that the DISPLAY B value is 0±10 counts.

Note

The units for both DISPLAY A and DISPLAY B values are  $\mu S$ .

# [10K/MAG Adjustment]

- 7. Set the DISPLAY A range to the 100 µS range. (Use the LC | Z | RANGE key.)
- 8. Connect the  $10k\Omega$  standard resistor directly to the UNKNOWN terminals.
- 9. Adjust A2R4 (10k/MAG) until the 4276A displays [1/c.v.] ±1 count on DISPLAY A. Verify that the DISPLAY B value is 0±5 counts.

### [1K/MAG Adjustment]

- 10. Set the DISPLAY A range to the 1mS range.
- 11. Connect the  $lk\Omega$  standard resistor directly to the UNKNOWN terminals.
- 12. Adjust A2R5 (1K/MAG) until the 4276A displays  $[1/c.v.]\pm 1$  count on DISPLAY A. Verify that the DISPLAY B value is  $0\pm 5$  counts.

# [100/MAG Adjustment]

- 13. Set the DISPLAY A range to the 10mS range.
- 14. Connect the  $100\Omega$  standard resistor directly to the UNKNOWN terminals.
- 15. Adjust A2R6 (100/MAG) until the 4276A displays [1/c.v.] ±1 count on DISPLAY A. Verify that the DISPLAY B value is 0±5 count.

### 5-25. A2 RANGE RESISTOR PHASE ADJUSTMENT

### PURPOSE:

Eliminates the phase shift caused by the range resistor circuit.

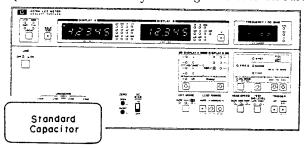
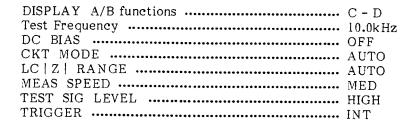


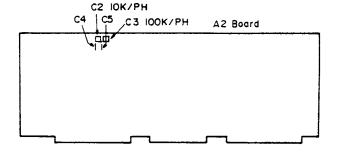
Figure 5-9. A2 Range Resistor Phase Adjustment Setup.

### **EQUIPMENT:**

### PROCEDURE:

- 1. Connect the 100pF standard capacitor directly to the UNKNOWN terminals.
- 2. Reset continuous memory (see Note 2 in paragraph 5-15) and turn on the 4276A.
- 3. Set the 4276A's controls as follows:





4. Adjust A2C3 (100K/PH) until the 4276A displays 0 counts±2 counts on DISPLAY B. Verify that the 4276A displays C.V. ±30 counts on DISPLAY A.

### Note

If this test cannot be performed successfully, replace A2C5 in accordance with the instructions given in Table 5-2.

- 5. Connect the 1000pF standard capacitor directly to the UNKNOWN terminals.
- 6. Adjust A2C2 (10K/PH) until the 4276A displays 0 counts±2 counts on DISPLAY B. Verify that the 4276A displays C.V.±30 counts on DISPLAY A.

#### Note

If this test cannot be performed successfully, replace A2C4 in accordance with the instructions given in Table 5-2.

# 5-26. INTERNAL DC BIAS ADJUSTMENT (OPTION 001 ONLY)

#### PURPOSE:

Adjusts the output voltage of the internal dc bias voltage source.

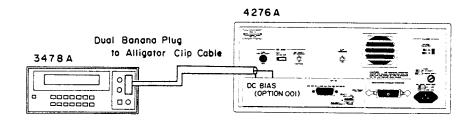


Figure 5-10. Internal DC Bias Adjustment Setup.

### EQUIPMENT:

 Digital Voltmeter
 HP 3478A

 Extender Board
 HP P/N 04276-66562

### PROCEDURE:

- 1. Extend the A22 DC BIAS board from the rear-panel with the extension board.
- 2. Set the DC BIAS select switch on the rear panel to INT.
- 3. Turn on the 4276A.
- 4. Set the DC BIAS switch on the front panel to ON.
- 5. Set the 3478A function to DCV.
- 6. Connect the 3478A HI and LOW inputs to A22TP2 and A22 GND, respectively.
- 7. Adjust A22R8 until the reading on the 3478A is 0V±0.05mV.
- 8. Connect the 3478A HI input to A22TP3.
- 9. Adjust A22R6 (ZERO) until the reading on the 3478A is  $0V\pm0.2mV$ .
- 10. Connect the 3478A HI input to the center conductor of the EXT INPUT/INT MONITOR connector on the rear panel.
- 11. Set the dc bias voltage to +9.99V.
- 12. Adjust A22R7 (FS) until the reading on the 3478A is +9.99V±0.002V.

Model 4276A SECTION VI

# SECTION VI REPLACEABLE PARTS

#### 6-1. INTRODUCTION

6-2. This section contains information for ordering parts. Table 6-1 lists abbreviations used in the parts list and throughout the manual. Table 6-3 lists all replaceable parts in reference designator order. Table 6-2 contains the names and addresses that correspond to the manufacturer's code numbers.

#### 6-3. ABBREVIATIONS

6-4. Table 6-1 lists abbreviations used in parts list, schematics and throughout the manual. In some cases, two forms of abbreviations are used, one in all capital letters, and one in partial capitals or no capitals. This occurs because the abbreviations in parts list are always all capitals. However, in the schematic and in other parts of the manual, other abbreviation forms with both lower case and upper case letters are used.

#### 6-5. REPLACEABLE PARTS LIST

- 6-6. Table 6-3 is a list of replaceable parts and is organized as follows:
  - a. Electrical assemblies and their components in alphanumerical order by reference designation.
  - b. Chassis-mounted parts in alphanumerical order by reference designation.
  - c. Miscellaneous parts.
  - d. Illustrated parts breakdowns, if appropriate.

The information for each part includes:

- a. The Hewlett-Packard part number.
- b. The total quantity (Qty) in the instrument.
- c. A description of the part.
- d. A typical manufacturer of the part in a five-digit code.
- e. The manufacturer's number for the part.

Table 6-1. List of Reference Designators and Abbreviations

			REFERENCE DESIG	GNATORS			
A	= assembly	E	= misc electronic part	P	= plug	U	= integrated circuit
В	= niotor	F	= fuse	Q	= transistor	V	= vacuum, tube, neon
BT	= battery	FL	= filter	R	= resistor		bulb, photocell, etc
C	= capacitor	J	= jack	RT	⇒ thermistor	VR	= voltage regulator
ĊР	= coupler	K	= relav	S	= switch	w	= cable
CR	= diode	Ĺ	= inductor	Ť	= transformer	х	= socket
DL	= delay line	M	= meter	тв	= terminal board	Ÿ	= crystal
DS	= device signaling (lamp)	MP	= mechanical part	TP	≈ test point	<del>-</del>	**,****
			ABBREVIATI	ONS			
A	= amperes	н	= benries	NPN	= negative-positive-	RWV	= reverse working
A. F. C.	= automatic frequency control	HEX	= hexagonal		negative		voltage
	= amplifier	HG	= mercury	NRFR	= not recommended for		-
	·	HR	= hour(s)		field replacement		
	= beat frequency oscillator	Hz	= hertz	NSR	= not separately	S-B	= slow-blow
	= beryllium copper				replaceable	SCR	= screw
вн	= binder head	ĮF.	= intermediate freq.			SE	⇒ selenium
BP	⇒ bandpass	IM PG	= impregnated			SECT	= section(s)
BRS	= brass	INCD	= incandescent	OBD	= order by description	SEMICON	= semiconductor
BWO	<ul> <li>backward wave oscillator</li> </ul>	INCL	= include(s)	OH	= oval head	SI	= silicon
CCW	= counter-clockwise	INS	= insulation(ed)	ox	= oxide	SIL	= silver
CER	= ceramic	INT	= internal			SL.	= slide
СМО	= cabinet mount only	k	= kilo = 1000			SPG	= spring
	= coefficient			P	= peak	SPL	= special
COM	= common	LH	= left hand	PC	= printed circuit	SST	= special = stainless steel
	= common = composition	LIN	= linear taper	p	= pico = 10 <sup>-12</sup>	SR	= staintess steer = split ring
	= composition = complete	LK WASH	= lock washer	PH BRZ	<ul> <li>phosphor bronze</li> </ul>	STL	= spilt ring = steel
		LOG	= logarithmic taper	PHL	= Phillips	217	= Steel
	= connector	LPF	= low pass filter	PIV	= peak inverse voltage	<b></b>	
CP_	= cadmium plate		•	PNP	= positive-negative-	TA	= tantalum
CRT	= cathode-ray tube	m	= milli = 10 <sup>-3</sup>		positive	TD	= time delay
CW	= clockwise	M	= meg = 10 <sup>6</sup>	P/O	= part of	TGL	= toggle
DEPC	= deposited carbon		= metai film	POLY	= polystyrene	THD	= thread
DR	= drive	MET OX	= metallic oxide	PORC	= porystyrene = porcelain	TI	= titanium
		MFR	= manufacturer	POS	= position(s)	TOL	= tolerance
	= electrolytic	MINAT	= miniature	POT	= position(s) = potentiometer	TRIM	= trimmer
	= encapsulated	MOM	= miniature = momentary	PP	= potentiometer = peak-to-peak	TWT	= traveling wave tube
EXT	= external	MTG	= momentary = mounting	PT	= peak-to-peak = point		æ
F	= farads	MY	= 'mylar'	PWV		μ	= micro = 10 <sup>-6</sup>
	= femto = 10 <sup>-15</sup>	.VI I	*	rwv	≠ peak working voltage	VAR	= variable
Н	= flat head	n	= nano = 10 <sup>+9</sup>			VAR	= dc working voits
	= filister head	N C	= normally closed				- GC WOLKING VOILS
		NE	= neon	RECT	= rectifier	w/	= with
	= fixed	NI PL	= nickel plate	RF	= radio frequency	w	= watts
5	= g1ga = 10 <sup>3</sup>	N O	= normally open	RH	= round head or	WIV	= working inverse
GE.	= germanium	NPO	= negative positive zero	••••	right hand		voltage
GL.	= glass		zero temperature	RMO	= rack mount only	ww	= wirewound
CRD	= ground(ed)		coefficient)	RMS	= root-mean square	w o	= without

The total quantity for each part is given only once—at the first appearance of the part number in the list.

### 6-7. ORDERING INFORMATION

- 6-8. To order a part listed in the replaceable parts table, give the Hewlett-Packard part number, indicate the quantity required, and address the order to the nearest Hewlett-Packard office.
- 6-9. To order a part that is not listed in the replaceable parts table, state the full instrument model and serial number, and description and function of the part, and the number of parts required. Address your order to the nearest Hewlett-Packard office.

#### 6-10. DIRECT MAIL ORDER SYSTEM

- 6-11. Within the USA, Hewlett-Packard can supply parts through a direct mail order system. Advantages of using the system are:
  - a. Direct ordering and shipment from the HP Parts Center in Mountain View, California.
  - b. No maximum or minimum on any mail order (there is a minimum order amount for parts ordered through a local HP Office when the orders require billing and invoicing).
  - c. Prepaid transportation (there is a small handling charge for each order).
  - d. No invoices—to provide these advantages, a check or money order must accompany each order.
- 6-12. Mail order forms and specific ordering information are available through your local HP Office. Addresses and phone numbers are located at the back of this manual.

Table 6-2. Manufacturers Code Lists

MFR NO.	MANUFACTURER NAME	ADDRESS		ZIP CODE
00000 01121 01295 02111 03508 03888 04713 05574 07263 14936 24355 24546 27014 27167 28480 3L585 30983 34649 52763 56289 72136 75042 75915	ANY SATISFACTORY SUPPLIER ALLEN-BRADLEY CO TEXAS INSTR INC SEMICOND CMPNT DIV SPECTROL ELECTRONICS CORP GE CO SEMICONDUCTOR PROD DEPT K D I PYROFILM CORP MOTOROLA SEMICONDUCTOR PRODUCTS VIKING INDUSTRIES INC FAIRCHILD SEMICONDUCTOR DIV GENERAL INSTR CORP SEMICON PROD GP ANALOG DEVICES INC CORNING GLASS WORKS (BRADFORD) NATIONAL SEMICONDUCTOR CORP CORNING GLASS WORKS (WILMINGTON) HEWLETT-PACKARD CO CORPORATE HQ RCA CORP SOLID STATE DIV MEPCO/ELECTRA CORP INTEL CORP STETTNER-TRUSH INC SPRAGUE ELECTRIC CO ELECTRO MOTIVE CORP TRW INC PHILADELPHIA DIV LITTELFUSE INC	MILWAUKEE DALLAS CITY OF IND AUBURN WHIPPANY PHOENIX CHATSWORTH MOUNTAIN VIEW HICKSVILLE NORWOOD BRADFORD SANTA CLARA WILMINGTON PALO ALTO SOMERVILLE SAN DIEGO MOUNTAIN VIEW CAZENOVIA NORTH ADAMS FLORENCE PHILADELPHIA DES PLAINES	WIXAY JZAAYAAACAJAAYACAN MACAYAAAAACA JAAAYAACAI	53204 75222 91745 13201 07981 85008 91311 94042 11802 02062 16701 95051 28401 94304 92121 95051 13035 01247 06226 19108 60016
98291	SEALECTRO CORP	MAMARONECK	NY	10544

Table 6-3. Replaceable Parts

lable b-3. Replaceable Parts							
Reference Designation	HP Part Number	C D	Qty	Description	Mfr Code	Mfr Part Number	
A 1	04276-66501	٥	1	LOGIC BOARD ASSEMBLY	28480	0.4276-66501	
Alc1 C2 C3 C4 C5	0180-1085 0180-1085 0180-1085 0180-1085 0180-1085	ម្មាលមុខ	13	CAPACITOR-FXD 4.7UF 16VDC TA CAPACITOR-FXD 4.7UF 16VDC TA CAPACITOR-FXD 4.7UF 16VDC TA CAPACITOR-FXD 4.7UF 16VDC TA CAPACITOR-FXD 4.7UF 16VDC TA	28480 28480 28480 28480 28480	0180-1085 0180-1085 0180-1085 0180-1085 0180-1085 0180-1085	
C6 C7 C8 C9 C16	0180-1085 0180-1085 0180-0197 0160-4832 0180-3219	5 5 8 4 1	1 2 1	CAPACITOR-FXD 4.7UF 16VDC TA CAPACITOR-FXD 4.7UF 16VDC TA CAPACITOR-FXD 2.2UF+-10% 20VDC TA CAPACITOR-FXD 0.1UF +-10% 100VDC CER CAPACITOR-FXD 2.2 UF 63VDC	28480 28480 56289 28480 28480	0180-1085 0180-1085 150D225X9020A2 0160-4832 0180-3219	
C11 C12 C13 C14 C15	0160-4801 0180-1085 0180-1085 0180-1085 0180-2951	7 5 5 5 6	1	CAPACITOR-FXD 100PF +-5% 100VDC CER CAPACITOR-FXD 4.7UF 16VDC TA CAPACITOR-FXD 4.7UF 16VDC TA CAPACITOR-FXD 4.7UF 16VDC TA CAPACITOR-FXD 33UF+-20% 16VDC AL	28480 28480 28480 28480 28480	0160-4801 0180-1085 0180-1085 0180-1085 0180-1085	
C16 C17 C18 C19 C20	0180-1085 0180-1085 0180-1085 0160-4832 0180-3217	5 5 5 4 9	2	CAPACITOR-FXD 4.7UF 16VDC TA CAPACITOR-FXD 4.7UF 16VDC TA CAPACITOR-FXD 4.7UF 16VDC TA CAPACITOR-FXD .01UF +-10% 100VDC CER CAPACITOR-FXD 470UF	28480 28480 28480 28480 28480	0180-1085 0180-1085 0180-1085 0160-4832 0180-3217	
C21	0180-3217	9		CAPACITOR-FXD 470UF	28480	0180-3217	
AICR1 CR2	1901-0539 1901-0539	3	2	DIODE-SM SIG SCHOTTKY DIODE-SM SIG SCHOTTKY	28490 29480	1901-0539 1901-0539	
Alj1 J2 J3 J4 J5	1200-0607 1200-0607 1200-0654 1200-0541 1200-0541	0 7 1	2 10	SOCKET-IC 16-CONT DIP DIP-SLDR SOCKET-IC 16-CONT DIP DIP-SLDR SOCKET-IC 40-CONT DIP DIP-SLDR SOCKET-IC 24-CONT DIP DIP-SLDR SOCKET-IC 24-CONT DIP DIP-SLDR	28480 28480 28480 28480 28480	1200-0607 1200-0607 1200-0654 1200-0541 1200-0541	
J6 J7 J8 J9 J10	1200-0541 1200-0541 1200-0541 1200-0541 1200-0541	1 1 1 1 1		SOCKET-IC 24-CONT DIP DIP-SLDR	28488 28488 28488 28488 28480 28480	1200-0541 1200-0541 1200-0541 1200-0541 1200-0541	
J11 J12 J13 J14	1200-0541 1200-0541 1200-0541 1200-0654	1 1 1 7		SOCKET-IC 24-CONT DIP DIP-SLDR SOCKET-IC 24-CONT DIP DIP-SLDR SOCKET-IC 24-CONT DIP DIP-SLDR SOCKET-IC 40-CONT DIP DIP-SLDR	28480 28480 28480 28480 28480	1200-0541 1200-0541 1200-0541 1200-0654	
A]Q1 Q2 Q3 Q4 Q5	1854-0810 1854-0810 1853-0281 1854-0810 1853-0015	2 2 9 2 7	3 1 1	TRANSISTOR NPN SI PD=625MW FT=200MHZ TRANSISTOR NPN SI PD=625MW FT=200MHZ TRANSISTOR PNP SI PD=625MW FT=200MHZ TRANSISTOR NPN SI PD=625MW FT=200MHZ TRANSISTOR PNP SI PD=200MW FT=500MHZ	28480 28480 94713 28480 28480	1854-0810 1854-0810 2N2907A 1854-0810 1853-0015	
Alrı R2	1810-0488 0757-0199	8	1 1	NETWORK-RES 8-SIP4.7K OHM X 4 RESISTOR 21.5K 1% ,125W F TC=0+-100	28480 24546	1810-0488 C4-1/8-T0-2152-F	
R3 R4	2100-3103 0757-0440	6 7	1 1	RESISTOR-TRMR 10K 10% C SIDE-ADJ 17-TRN RESISTOR 7.5K 1% .125W F TC=0+-100	02111 24546	43P103 C4-1/8-T0-7501-F	
R5 R6 R7 R8 R9	0683-2215 0683-4715 1810-0607 0683-1045 0683-2245	1 0 3 3 7	1 1 1 1	RESISTOR 220 5% .25W FC TC=-400/+600 RESISTOR 470 5% .25W FC TC=-400/+600 RESISTIVE NETWORK- SIP RESISTOR 100% 5% .25W FC TC=-400/+800 RESISTOR 220K 5% .25W FC TC=-800/+900	01121 01121 28480 01121 01121	CB2215 CB4715 1818-0607 CB1045 CB2245	
R10 R11 R12 R13 R14	0483-1005 0483-1025 0483-5405 0483-0565 0483-1025	5 9 9 9	2 2 1 1	RESISTOR 10 5% .25W FC TC=-400/+500 RESISTOR 1K 5% .25W FC TC=-400/+600 RESISTOR 56 5% .25W FC TC=-400/+500 RESISTOR 5.6 5% .25W FC TC=-400/+500 RESISTOR 1K 5% .25W FC TC=-400/+600	01121 01121 01121 01121	CB1005 CB1025 CB5605 CB1025	
R15 R16 R17 R18 R19	1810-0305 1810-0305 1810-0269 1810-0305 1810-0305	8 3 8 8	5	NETWORK-RES 9-SIP4.7K OHM X 8 NETWORK-RES 9-SIP4.7K OHM X 8 NETWORK-RES 9-SIP10.0K OHM X 8 NETWORK-RES 9-SIP4.7K OHM X 8 NETWORK-RES 9-SIP4.7K OHM X 8	28480 28480 28480 28480 28480	1810-0305 1810-0305 1810-0269 1810-0305 1810-0305	
R20 R21 R22 R23 R24	1810-0305 0683-4725 0683-4725 0683-4725 0683-3325	8 2 2 2 6	5	NETWORK-RES 9-SIP4.7K OHM X 8 RESISTOR 4.7K 5% .25W FC TC=-400/+700 RESISTOR 4.7K 5% .25W FC TC=-400/+700 RESISTOR 4.7K 5% .25W FC TC=-400/+700 RESISTOR 3.3K 5% .25W FC TC=-400/+700	28480 61121 01121 01121 01121	1810-0305 CB4725 CB4725 CB4725 CB4325 CB3325	
R25 R26 R27	0683-6825 0683-4725 0683-4725	7 2 2	1	RESISTOR 6.8K 5% .25W FC TC=-400/+700 RESISTOR 4.7K 5% .25W FC TC=-400/+700 RESISTOR 4.7K 5% .25W FC TC=-400/+700	01121 01121 01121	CB6825 CB4725 CB4725	
A151	3101-1973	7	1	SWITCH-SL 7-1A DIP-SLIDE-ASSY .1A 50VDC	28480	3101-1973	

Table 6-3. Replaceable Parts

Reference Designation	HP Part Number	C D	Qty	Description	Mfr Code	Mfr Part Number
A 101 U2 U3 U4 U5	1813-0291 1826-0978 1826-0180 1820-2649 04276-85001	7 4 0 8 5	1 1 1 1	IC-CRYSTAL 11.52 M IC- (MISC) IC TIMER TTL MONO/ASTBL IC- Z88B-CPU IC-PROM, PROGRAMMED	28480 28480 01295 28480 28480	1813-0291 1826-0978 NE555P 1820-2649 04276-85001
บ6 บ7 บ8 บ9 บ1 จ	04276-85002 04276-85003 04276-85004 04276-85005 04276-85006	6 7 8 9	1 1 1 1	IC-PROM, PROGRAMMED IC-PROM, PROGRAMMED IC-PROM, PROGRAMMED IC-PROM, PROGRAMMED IC-PROM, PROGRAMMED	28480 28480 28480 28480 28480	04276-85002 04276-85003 04276-85004 04276-85005 04276-85006
U12 U13 U14 U15 U16	1818-1974 1820-2024 1820-2024 1820-1730 1820-1217	5 3 6 4	1 2 1 1	IC-HSM512B-15 IC DRVR TTL LS LINE DRVR OCTL IC DRVR TTL LS LINE DRVR OCTL IC FF TTL LS D-TYPE POS-EDGE-TRIG COM IC HUXR/DATA-SEL TTL LS B-TO-1-LINE	28480 01295 01295 01295 01295	1818-1974 SN74LS244N SN74LS244N SN74LS273N SN74LS273N
U17 U18 U19 U20 U21	1820-1197 1820-1112 1820-1197 1820-0682 1820-1197	9 8 9 5 9	4 5 1	IC GATE TTL LS NAND QUAD 2-INP IC FF TTL LS D-TYPE POS-EDGE-TRIG IC GATE TTL LS NAND QUAD 2-INP IC GATE TTL S NAND QUAD 2-INP IC GATE TTL LS NAND QUAD 2-INP	01295 01295 01295 01295 01295	SN74LS00N SN74LS74AN SN74LS00N SN74S03N SN74S03N
U22 U23 U24 U25 U26	1820-1216 1820-1199 1820-0681 1820-2150 1820-1216	3 1 4 6 3	4 3 1 1	IC DCDR TTL LS 3-TD-8-LINE 3-INP IC INV TTL LS HEX 1-INP IC GATE TTL S NAND QUAD 2-INP IC HICPROC-ACCESS HMOS IC DCDR TTL LS 3-TO-8-LINE 3-INP	01295 01295 01295 01295 34649 01295	SN74LS138N SN74LS04N SN74S00N D8279-5 SN74LS138N
U27 U28 U29 U30 U31	1820-1112 1820-1112 1820-1420 1820-1432 1820-1432	8 1 5 5	1 2	IC FF TTL LS D-TYPE POS-EDGE-TRIG IC FF TTL LS D-TYPE POS-EDGE-TRIG IC CNTR TTL LS DIV-X-12 ASYNCHRO IC CNTR TTL LS BIN SYNCHRO POS-EDGE-TRIG IC CNTR TTL LS BIN SYNCHRO POS-EDGE-TRIG	01295 01295 01295 01295 01295	SN74LS74AN SN74LS74AN SN74LS92N SN74LS163AN SN74LS163AN
U32 U33 U34 U35 U36	1820-1112 1820-1199 1820-2075 1820-1216 1820-1624	8 1 4 3 7	1	IC FF TTL LS D-TYPE POS-EDGE-TRIG IC INV TTL LS HEX 1-INP IC MISC TTL LS IC DCDR TTL LS 3-TO-8-LINE 3-INP IC BFR TTL S OCTL 1-INP	01295 01295 01295 01295 01295	SN74LS74AN SN74LS04N SN74LS245N SN74LS138N SN74S241N
U37 U38 U39 U40 U41	1820-1199 1820-1197 1820-2873 1820-2873 1820-1216	1 9 0 0 3	2	IC INV TTL LS HEX 1-INP IC GATE TTL LS NAND QUAD 2-INP IC-UPD8253-5 IC-UPD8253-5 IC DCDR TTL LS 3-TO-8-LINE 3-INP	01295 81295 28480 28480 01295	SN74LS14N SN74LS10N 1820-2873 1820-2873 SN74LS138N
U42 U43 U44	1820-1112 1820-1425 1826-0122	B 6 0	1 1	IC FF TTL LS D-TYPE POS-EDGE-TRIG IC SCHMITT-TRIG TTL LS NAND QUAD 2-INP IC 7805 V RGLTR TO-220	81295 01295 07263	SN74LS74AN SN74LS132N 7805UC
ATW1 W2 W3 W4	1251-4822 1251-4822 1251-4822 1251-4787	6 6 2	3	CONNECTOR 3-PIN M POST TYPE CONNECTOR 3-PIN M POST TYPE CONNECTOR 3-PIN M POST TYPE SHUNT-DIP 8-POSITION	28488 28480 28480 28480	1251-4822 1251-4822 1251-4822 1251-4787
				MISCELLANEOUS PARTS		
	1258-0141	8	3	JUMPER-REM	28480	1258-01 <b>4</b> 1

Table 6-3. Replaceable Parts

Reference Designation	HP Part Number	CD	Qty	Description	Mfr Code	Mfr Part Number
A 2						
A2C1	04276-66502 0160-5499	1	1 2	ANALOG BOARD ASSEMBLY  CAPACITOR- 0.22UF 100VDC F	28480 28480	04276-66502 0160-5499
C2 C3 C4 C5	0121-0036 0121-0036 0160-4791 0160-4788	0 4 9	3 1 1	CAPACITOR-V TRMR-CER 5.5-18PF 350V CAPACITOR-V TRMR-CER 5.5-18PF 350V CAPACITOR-FXD 18PF +-5X 180VDC CER 0+-30 CAPACITOR-FXD 18PF +-5X 180VDC CER 0+-30	52763 52763 28480 28480	304324 5.5/18PF NPO 304324 5.5/18PF NPO 0160-4791 0160-4788
C6 C7 C8 C9 C10	0160-3402 0121-0105 0160-4835 0160-4835 0160-4835	2 4 7 7 7	2 3 3	CAPACITOR-FXD 1UF +-5% 50VDC MET-POLYC CAPACITOR-V TRMR-CER 9-35PF 200V PC-MTG CAPACITOR-FXD .1UF +-10% 50VDC CER CAPACITOR-FXD .1UF +-10% 50VDC CER CAPACITOR-FXD .1UF +-10% 50VDC CER	28480 52763 28480 28480 28480	0160-3402 304324 9/35PF N650 0160-4835 0160-4835 0160-4835
C11 C12 C13 C14 C15	0160-4822 0121-0105 0121-0105 0121-0105 0160-2202 0160-4789	2 4 8 0	2 1 1	CAPACITOR-FXD 1000PF +-5% 100VDC CER CAPACITOR-V TRNR-CER 9-35PF 200V PC-HTG CAPACITOR-V TRNR-CER 9-35PF 200V PC-HTG CAPACITOR-FXD 75PF +-5% 300VDC HICA CAPACITOR-FXD 15PF +-5% 100VDC CER 0+-30	28480 52763 52763 28480 28480	0160-4822 304324 9/35PF N650 304324 9/35PF N650 0160-2202 0160-4789
C16 C17 C18 C19 C20	0140-0196 0160-4822 0121-0036 0160-0127 0160-1674	3 2 0 2 6	1 1 1	CAPACITOR-FXD 150PF +-5% 300VDC MICA CAPACITOR-FXD 1000PF +-5% 100VDC CER CAPACITOR-V TRMR-CER 5.5-16PF 350V CAPACITOR-FXD 1UF +-20% 25VDC CER CAPACITOR-FXD 1UF +-20% 25VDC CER CAPACITOR33UF 5% 200VDC	72136 28480 52763 28480 28480	DM15F151J0300WY1CR 0160-4822 304324 5.5/18PF NPO 9160-0127 0160-1674
C21 C22 C23 C24 C25	0180-2951 0180-2951 0180-2951 0160-4806 0150-0121	66605	24 1 1	CAPACITOR-FXD 33UF+-20% 16VDC AL. CAPACITOR-FXD 33UF+-20% 16VDC AL. CAPACITOR-FXD 33UF+-20% 16VDC AL. CAPACITOR-FXD 39PF +-5% 100VDC CER 0+-30 CAPACITOR-FXD .1UF +80-20% 50VDC CER	28480 28480 28480 28480 28480	0180-2951 0180-2951 0180-2951 0160-4806 0150-0121
C26 C27 C28 C29 C30	0160-5499 0180-2951 0180-2951 0160-5502 0180-3223	1 6 7 7	1 2	CAPACITOR- 0.22UF 100VDC F CAPACITOR-FXD 33UF+-20% 16VDC AL CAPACITOR-FXD 33UF+-20% 16VDC AL CAPACITOR- 1 UF 63 VDC F CAPACITOR- 1 UF 63 VDC F	28480 28480 28480 28480 28480	0160-5499 0180-2951 0180-2951 0160-25502 0180-3223
C31 C32 C33 C34 C35	0180-3223 0180-3233 0180-2951 0180-3233 0180-2951	7 9 6 9 6	6	CAPACITOR-FXD 22 UF 63VDC CAPACITOR-FXD 22 UF 25VDCW CAPACITOR-FXD 33UF+-20% 16VDC AL CAPACITOR-FXD 22 UF 25VDCW CAPACITOR-FXD 33UF+-20% 16VDC AL	28480 28480 28480 28480 28480	0180-3223 0180-3233 0180-2951 0180-3233 0180-2951
C36 C37 C38 C39 C40	0180-2951 0180-2951 0180-2951 0180-2951 0180-2951	6 6 6 6		CAPACITOR-FXD 33UF+-20% 16VDC AL	28480 28480 28480 28480 28480	0180-2951 0180-2951 0180-2951 0180-2951 0180-2951
C41 C42 C43 C44 C45	0180-2951 0160-0194 0180-0197 0180-0197 0180-2951	6 3 8 8 6	1 2	CAPACITOR-FXD 33UF+-20% 16VDC AL CAPACITOR-FXD .015UF +-10% 200VDC POLYE CAPACITOR-FXD 2.2UF+-10% 20VDC TA CAPACITOR-FXD 2.2UF+-10% 20VDC TA CAPACITOR-FXD 33UF+-20% 16VDC AL	28480 28480 56289 56289 28480	0180~2951 0160-0174 1500225X9020A2 1500225X9020A2 0180~2951
C46 C47 C48 C49 C50	0180-2951 0160-4814 0160-4803 0160-4786 0160-4792	6 2 9 7 5	1 2 1 1	CAPACITOR-FXD 33UF+-20% 16VDC AL CAPACITOR-FXD 150PF +-5% 100VDC CER CAPACITOR-FXD 68PF +-5% 100VDC CER 0+-30 CAPACITOR-FXD 27PF +-5% 100VDC CER 0+-30 CAPACITOR-FXD 9.2PF +5PF 100VDC CER	28480 28480 28480 28480 28480	0180-2951 0160-4814 0160-4803 0160-4786 0160-4796
C51 C52 C53 C54 C55	0180-2951 0180-2951 0180-3233 0180-3233 0160-4803	6 6 9 9 9		CAPACITOR-FXD 33UF+-20% 16VDC AL CAPACITOR-FXD 33UF+-20% 16VDC AL CAPACITOR-FXD 22 UF 25VDCW CAPACITOR-FXD 22 UF 25VDCW CAPACITOR-FXD 68PF +-5% 100VDC CER 0+-30	28480 28480 28480 28480 28480	0180-2951 0180-2951 0180-3233 0180-3233 0160-4803
C56 C57 C58 C59 C60	0160-1603 0180-2951 0180-2951 0180-2951 0180-2951	1 6 6 6 6	1	C:FXD MY 1 UF 10% 100VDCW CAPACITOR-FXD 33UF+-20% 16VDC AL CAPACITOR-FXD 33UF+-20% 16VDC AL CAPACITOR-FXD 33UF+-20% 16VDC AL CAPACITOR-FXD 33UF+-20% 16VDC AL	28480 28480 28488 28480 28480	0160-1603 0180-2951 0180-2951 0180-2951 0180-2951
C61 C62 C63 C64 C65	0180-2951 0160-5498 0160-2009 0160-3901 0160-3901	6 0 3 6 6	1 1 3	CAPACITOR-FXD 33UF+-20X 16VDC AL CAPACITOR- 0.01UF S0VDC F CAPACITOR-FXD 820FF +-5X 300VDC MICA CAPACITOR-FXD 2.2UF +-20X 25VDC CER CAPACITOR-FXD 2.2UF +-20X 25VDC CER	28480 28480 28480 28480 28480	0180-2951 0160-5498 0160-2009 0160-3901 0160-3901
C66 C67 C68 C69 C70	0160-3901 0180-3233 0180-3233 0160-5501 0160-5596	69969	1	CAPACITOR-FXD 2.2UF +-20% 25VDC CER CAPACITOR-FXD 22 UF 25VDCW CAPACITOR- 0.1 UF 100VDC F CAPACITOR- 0.1 UF 100VDC F CAPACITOR- 3 PF +/5 PF	28480 28480 28480 28480 28480	0160-3901 0180-3233 0180-3233 0180-5501 0160-5596

Table 6-3. Replaceable Parts

Reference Designation	HP Part Number	C D	Qty	Description	Mfr Code	Mfr Part Number
A2C71 C72 C73 C74 C75	0160-5594 0160-5597 0160-3402 0180-2951 0180-2951	7 0 2 6 6	1 1	CAPACITOR- 1 PF +/5 PF CAPACITOR- 5 PF +/5 PF CAPACITOR-FXD 1UF +-5X 50VDC MET-POLYC CAPACITOR-FXD 33UF+-20X 16VDC AL CAPACITOR-FXD 33UF+-20X 16VDC AL	28480 28480 28480 28480 28480	0160-5594 0160-5597 0160-3402 0180-2951 0180-2951
C76	0160-4830	2	1	CAPACITOR-FXD 2280PF +-18% 180VDC CER	28480	0160-4830
A2CR1 CR2 CR3 CR4 CR5	1901-0376 1901-0033 1901-0033 1901-0033 1902-3059	6 2 2 2 0	5 7 2	DIODE-GEN PRP 35V 50MA DO-35 DIODE-GEN PRP 180V 200MA DO-7 DIODE-GEN PRP 180V 200MA DO-7 DIODE-GEN PRP 180V 200MA DO-7 DIODE-ZNR 3.83V 5% DO-35 PD=.4W	28480 28480 28480 28480 28480	1901-0376 1901-0033 1901-0033 1901-0033 1902-3059
CR6 CR7 CR8 CR9 CR10	1902-3059 1901-0040 1901-0040 1901-0033 1901-0033	0 1 2 2	2	DIODE-ZNR 3.83V 5% DO-35 PD=.4W DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-GEN PRP 180V 200MA DO-7 DIODE-GEN PRP 180V 200MA DO-7	28480 28480 28480 28480 28480	1902-3059 1901-0040 1901-0040 1901-0033 1901-0033
CR11 CR12 CR13 CR14 CR15	1901-1065 1901-1065 1901-1065 1901-1065 1901-0376	22226	4	DIODE-PWR RECT 1N4936 400V 1A 200NS DIODE-PWR RECT 1N4936 400V 1A 200NS DIODE-PWR RECT 1N4936 400V 1A 200NS DIODE-PWR RECT 1N4936 400V 1A 200NS DIODE-GEN PRP 35V 50MA DO-35	14936 14936 14936 14936 28480	1 N4936 1 N4936 1 N4936 1 N4936 1 N4936 1 9 01 - 0 376
CR16 CR17 CR18 CR19 CR20	1901-0376 1901-0033 1901-0033 1901-0376 1901-0376	90009		DIODE-GEN PRP 35V 50MA DO-35 DIODE-GEN PRP 180V 200MA DO-7 DIODE-GEN PRP 180V 200MA DO-7 DIODE-GEN PRP 35V 50MA DO-35 DIODE-GEN PRP 35V 50MA DO-35	28480 28480 28480 28480 28480	1901-0376 1901-0033 1901-0033 1901-0376 1901-0376
A2K1 K2 K3	0490-1269 0490-1269 0490-1269	4 4 4	3	RELAY 1C 12VDC-COIL .66A 30VDC RELAY 1C 12VDC-COIL .66A 30VDC RELAY 1C 12VDC-COIL .66A 30VDC	28480 28480 28480	0490-1269 0490-1269 0490-1269
A2Q1 Q2 Q3 Q4 Q5	1855-0119 1855-0119 1855-0119 1855-0119 1854-0810 1853-0459	66623	4 1 1	TRANSISTOR-FET 25K43 TRANSISTOR-FET 25K43 TRANSISTOR-FET 25K43 TRANSISTOR PN SI PD=625MW FT=200MHZ TRANSISTOR PNP SI PD=625MW FT=200MHZ	28480 28480 28480 28480 28480	1855-0119 1855-0119 1855-0119 1854-0810 1853-0459
Q6	1855-0119	6		TRANSISTOR-FET 2SK43	28480	1855-0119
A2R1 R2 R3 R4 R5	2100-3352 1810-0374 0698-3157 2100-2520 2100-2583	7 1 3 9 4	1 2 1 1	RESISTOR-TRMR 1K 10% C SIDE-ADJ 1-TRN NETWORK-RES 8-SIP1.0K 0HM X 4 RESISTOR 19.6K 1% .125W F TC=0+-100 RESISTOR-TRMR 50 20% C SIDE-ADJ 1-TRN RESISTOR-TRMR 10 20% C SIDE-ADJ 1-TRN	28480 81121 24546 30983 30983	2100-3352 2008102 C4-1/0-T0-1962-F ET50X500 ET50X100
R6 R7 R8 R9 R10	2100-2489 2100-2574 1810-0624 0699-1018 0757-0346	9 3 4 3 2	2 1 1 1 2	RESISTOR-TRMR 5K 102 C SIDE-ADJ 1-TRN RESISTOR-TRMR 500 102 C SIDE-ADJ 1-TRN RESISTIVE NETWORK RESISTOR-100.95 OHM 0.5W RESISTOR 10 12 .125W F TC=0+-100	30983 30983 28480 28480 24546	ET50X502 ET50X501 1810-0624 0699-1018 C4-1/8-T0-10R0-F
R11 R12 R13 R14 R15	0683-5605 0683-1835 2100-3274 2100-2514 2100-3274	9 2 1 2	7	RESISTOR 56 5% .25W FC TC=-400/+500 RESISTOR 18K 5% .25W FC TC=-400/+800 RESISTOR-TRMR 10K 10% C SIDE-ADJ 1-TRN RESISTOR-TRMR 20K 10% C SIDE-ADJ 1-TRN RESISTOR-TRMR 10K 10% C SIDE-ADJ 1-TRN	01121 01121 28480 30983 28480	CB5605 CB1835 2100-3274 ET50W203 2100-3274
R16 R17 R18 R19 R20	2100-3274 2100-3274 0698-0084 0757-0463 0683-1835	2 9 4 9	2	RESISTOR-TRMR 10K 10% C SIDE-ADJ 1-TRN RESISTOR-TRMR 10K 10% C SIDE-ADJ 1-TRN RESISTOR 2.15K 1% .125W F TC=0+-100 RESISTOR 82.5K 1% .125W F TC=0+-100 RESISTOR 18K 5% .25W FC TC=-400/+800	28480 28480 24546 24546 01121	2100-3274 2100-3274 C4-1/8-T0-2151-F C4-1/8-T0-8252-F C81835
R21 R22 R23 R24 R25	0683-1835 0683-1835 1810-0621 0683-1835 2100-3274	9 9 1 9 2	2	RESISTOR 18K 5% .25W FC TC=-400/+800 RESISTOR 18K 5% .25W FC TC=-400/+800 RESISTIVE NETWORK RESISTOR 18K 5% .25W FC TC=-400/+800 RESISTOR-TRMR 10K 10% C SIDE-ADJ 1-TRN	01121 01121 28480 01121 28480	CB1835 CB1835 1810-0621 CB1835 2100-3274
R26 R27 R28 R29 R30	2100-2583 2100-2583 0698-4343 0698-5453 0757-0278	4 4 1 6 9	1 1	RESISTOR-TRMR 10 20% C SIDE-ADJ 1-TRN RESISTOR-TRMR 10 20% C SIDE-ADJ 1-TRN RESISTOR 100 .1% .125W F TC=0+-50 RESISTOR 900 .1% .125W F TC=0+-50 RESISTOR 1.78K 1% .125W F TC=0+-100	30983 30983 28480 83888 24546	ET50X100 ET50X100 0698-4343 PME35 T-2-900R-B C4-1/8-T0-1781-F
R31 R32 R33 R34 R35	1810-0623 0698-3440 0683-1835 0683-1835 2100-2583	37994	1	RESISTIVE NETWORK RESISTOR 196 1% .125W F TC=0+-100 RESISTOR 18K 5% .25W FC TC=-400/+800 RESISTOR 18K 5% .25W FC TC=-400/+800 RESISTOR-TRHR 10 20% C SIDE-ADJ 1-TRN	28488 24546 81121 01121 30983	1810-0623 C4-1/8-T0-196R-F CB1835 CB1835 ET50X106
R36 R37 R38 R39 R40	0757-0280 0757-0416 0757-0416 2100-3274 2100-3274	3 7 7 2 2	2	RESISTOR 1K 1% .125W F TC=0+-100 RESISTOR 511 1% .125W F TC=0+-100 RESISTOR 511 1% .125W F TC=0+-100 RESISTOR-TRMR 10K 10% C SIDE-ADJ 1-TRN RESISTOR-TRMR 10K 10% C SIDE-ADJ 1-TRN	24546 24546 24546 28480 28480	C4-1/8-T0-1001-F C4-1/8-T0-511R-F C4-1/8-T0-511R-F 2100-3274 2100-3274

Table 6-3. Replaceable Parts

Reference Designation	HP Part Number	C D	Qty	Description	Mfr Code	Mfr Part Number
A2R41 R42 R43 R44 R45	0757-0401 1810-0622 1810-0374 1810-0607 0698-3155	0 2 1 3	6 1 1 1	RESISTOR 100 1% .125W F TC=0+-100 RESISTIVE NETWORK NETWORK-RES B-SIP1.0K OHM X 4 RESISTIVE NETWORK- SIP RESISTOR 4.64K 1% .125W F TC=0+-100	24546 28480 01121 28480 24546	C4-1/8-T0-101-F 1810-0622 208B102 1810-0607 C4-1/8-T0-4641-F
R 46 R 47 R 48 R 49 R 50	0683-4715 0757-0279 1810-0305 0683-3335 0757-0346	880	- 1 1 5 1	RESISTOR 470 5% .25W FC TC=-400/+600 RESISTOR 3.16K 1% .125W F TC=0+-100 NETWORK-RES 9-SIP4.7K OHM X 8 RESISTOR 33K 5% .25W FC TC=-400/+800 RESISTOR 10 1% .125W F TC=0+-100	01121 24546 28480 01121 24546	CB4715 C4-1/8-T0-3161-F 1810-0305 CB3335 C4-1/8-T0-10R0-F
R51 R52 R53 R54 R55	0757-1094 0757-0442 0757-0442 0698-3152 0757-0444	9 9 9 8 1	2 7 2 1	RESISTOR 1.47K 1% .125W F TC=0+-100 RESISTOR 10K 1% .125W F TC=0+-100 RESISTOR 10K 1% .125W F TC=0+-100 RESISTOR 3.48K 1% .125W F TC=0+-100 RESISTOR 12.1K 1% .125W F TC=0+-100	24546 24546 24546 24546 24546	C4-1/8-T0-1471-F C4-1/8-T0-1002-F C4-1/8-T0-1002-F C4-1/8-T0-3481-F C4-1/8-T0-1212-F
R56 R57 R58 R59 R60	0757-0442 0757-1094 0698-0084 0757-0280 0757-0438	9 9 9 3 3	2	RESISTOR 10K 1% .125W F TC=0+-100 RESISTOR 1.47K 1% .125W F TC=0+-100 RESISTOR 2.15K 1% .125W F TC=0+-100 RESISTOR 1K 1% .125W F TC=0+-100 RESISTOR 5.11K 1% .125W F TC=0+-100	24546 24546 24546 24546 24546	C4-1/8-T0-1002-F C4-1/8-T0-1471-F C4-1/8-T0-2151-F C4-1/8-T0-1001-F C4-1/8-T0-5111-F
R61 R62 R63 R64 R65	1810-0478 0683-2265 0698-3152 0699-1022 0683-1005	6 1 8 9 5	1 1 1	NETWORK-RES 8-SIP22.OK DHM X 4 RESISTOR 22M 5% .25W FC TC=-900/+1200 RESISTOR 3.48K 1% .125W F TC=0+-100 RESISTOR FUSE-47 OHM 1/4W RESISTOR 10 5% .25W FC TC=-400/+500	28480 01121 24546 28480 01121	1810-0478 CB2265 C4-1/8-T0-3481-F 0699-1022 CB1005
R66 R67 R68 R69 R70	0683-1025 0699-1020 1810-0305 0757-0401 0757-0442	9 7 8 0 9	1	RESISTOR 1K 5% .25W FC TC=-400/+600 RESISTOR-470 OHM 1W NETWORK-RES 9-SIP4.7K OHM X 8 RESISTOR 100 1% .125W F TC=0+-100 RESISTOR 10K 1% .125W F TC=0+-100	01121 28480 28480 24546 24546	CB1025 0699-1020 1810-0305 C4-1/8-T0-101-F C4-1/8-T0-1002-F
R71 R72 R73 R74 R75 R76 R77 R78 R79 R80 R81	0757-0401 0757-0442 0698-0083 0698-3161 0757-0280 0757-0280 0757-0465 0683-5605 0683-5605	0 9 8 9 3 3 6 9 3	1 1 1	RESISTOR 100 1% .125W F TC=0+-100 RESISTOR 18K 1% .125W F TC=0+-100 RESISTOR 1.96K 1% .125W F TC=0+-100 RESISTOR 1.96K 1% .125W F TC=0+-100 RESISTOR 38.3K 1% .125W F TC=0+-100 RESISTOR 1K 1% .125W F TC=0+-100 RESISTOR 1K 1% .125W F TC=0+-100 RESISTOR 100K 1% .125W F TC=0+-100 RESISTOR 100K 1% .125W F TC=0+-100 RESISTOR 56 5% .25W FC TC=-400/+500 RESISTOR 56 5% .25W FC TC=-400/+500 RESISTOR 162K 1% .125W F TC=0+-100	24546 24546 24546 24546 24546 24546 24546 24546 01121 01121 24546	C4-1/8-T0-101-F C4-1/8-T0-1002-F C4-1/8-T0-1961-F C4-1/8-T0-3832-F C4-1/8-T0-3832-F C4-1/8-T0-1001-F C4-1/8-T0-1001-F C4-1/8-T0-1003-F CB5605 CB5605 C4-1/8-T0-1623-F
R82 R83 R84 R85 R86	0698-3156 0757-0421 0757-0399 0698-3153 0757-0277	2 4 5 9 8	1 1 1 1	RESISTOR 14.7K 1% .125W F TC=0+-100 RESISTOR 825 1% .125W F TC=0+-100 RESISTOR 82.5 1% .125W F TC=0+-100 RESISTOR 3.83K 1% .125W F TC=0+-100 RESISTOR 49.9 1% .125W F TC=0+-100	24546 24546 24546 24546 24546	C4-1/8-T0-1472-F C4-1/8-T0-825R-F C4-1/8-T0-8285-F C4-1/8-T0-331-F C4-1/8-T0-4992-F
R87 R88 R89 R90 R91	0757-0401 0757-0442 1810-0305 1810-0305 1810-0305	9 8 8		RESISTOR 100 1% .125W F TC=0+-108 RESISTOR 10K 1% .125W F TC=0+-100 NETWORK-RES 9-SIP4.7K OHM X 8 NETWORK-RES 9-SIP4.7K OHM X 8 NETWORK-RES 9-SIP4.7K OHM X 8	24546 24546 28480 28480 28480	C4-1/8-T0-101-F C4-1/8-T0-1002-F 1810-0305 1810-0305 1810-0305
R92 R93 R94 R95 R96	0757-0280 0757-0464 0757-0280 0699-1019 0757-0438	3 5 3 4 3	1 2	RESISTOR 1K 1% .125W F TC=0+-100 RESISTOR 90.9K 1% .125W F TC=0+-100 RESISTOR 1K 1% .125W F TC=0+-100 RESISTOR-7.071K 0HM 0.1W RESISTOR 5.11K 1% .125W F TC=0+-100	24546 24546 24546 28480 24546	C4-1/8-T0-1001-F C4-1/8-T0-9092-F C4-1/8-T0-1001-F 0699-1019 C4-/8-T0-5111-F
R97 R98 R99 R100 R101	0699-1019 1810-0621 2100-2517 2100-2489 0757-0401	4 1 4 7	1	RESISTOR-7.071K OHM 0.1W RESISTIVE NETWORK RESISTOR-TRMR 50K 10% C SIDE-ADJ 1-TRN RESISTOR-TRMR 5K 10% C SIDE-ADJ 1-TRN RESISTOR-TRMR 5K 10% C SIDE-ADJ 1-TRN RESISTOR 100 1% .125W F TC=0+-100	28480 28480 30983 30983 24546	0699-1019 1810-0621 ET50X503 ET50X502 C4-1/8-T0-101-F
R102 R103 R104 R105 R196	0757-0280 0757-0439 0757-0442 0698-3160 0757-0401	3 4 9 8 0	1	RESISTOR 1K 1% .125W F TC=0+-108 RESISTOR 6.81K 1% .125W F TC=0+-100 RESISTOR 10K 1% .125W F TC=0+-100 RESISTOR 31.6K 1% .125W F TC=0+-100 RESISTOR 100 1% .125W F TC=0+-100	24546 24546 24546 24546 24546	C4-1/8-T0-1001-F C4-1/8-T0-6811-F C4-1/8-T0-1002-F C4-1/8-T0-3162-F C4-1/8-T0-101-F
R107	0683-3315	4	1	RESISTOR 330 5% .25W FC TC=-400/+600	01121	CB3315
2T1 2U1 U2 U3	9100-4252 1813-0303 1813-0303 1813-0303	2 2 2	3	TRANSFORMER-PULSE  IC (MISC) IC (MISC) IC (MISC)	28480 28480 28480 28480	7100-4252 1813-0303 1813-0303 1813-0303
บ4 บร	1813-0298 1826-0519	4 9	2 13	IC (MISC) IC OP AMP LOW-BIAS-H-IMPD 8-DIP-P PKG	29480 01295	1813-0303 1813-0298 TL071CP
บ6 บ7 บ8 บ9 บ10	1820-1545 1820-2111 1820-1546 1826-0519 1826-0519	1 9 2 9 9	6 1 3	IC MULTIPLXR 2-CHAN-ANLG TRIPLE 16-DIP-C IC DRVR TTL INV IC MULTIPLXR 4-CHAN-ANLG DUAL 16-DIP-C IC OP AMP LOW-BIAS-H-IMPD 8-DIP-P PKG IC OP AMP LOW-BIAS-H-IMPD 8-DIP-P PKG	3L585 01295 04713 01295 01295	CD4053BY SN75468N MC14052BCL TL071CP TL071CP

Table 6-3. Replaceable Parts

Reference Designation	HP Part Number	C D	Qty	Description	Mfr Code	Mfr Part Number
2 U11 U12 U13 U14 U15	1820-1545 1820-1546 1826-0519 1826-0547 1826-0519	1 2 9 3 9	2	IC MULTIPLXR 2-CHAN-ANLG TRIPLE 16-DIP-C IC MULTIPLXR 4-CHAN-ANLG DUAL 16-DIP-C IC OP AMP LOW-BIAS-H-IMPD 8-DIP-P PKG IC OP AMP LOW-BIAS-H-IMPD DUAL 8-DIP-P IC OP AMP LOW-BIAS-H-IMPD B-DIP-P PKG	3L585 04713 01295 01295 01295	CD4053BY MC14052BCL TL071CP TL072ACP TL071CP
U16 U17 U18 U19 U20	1820-1545 1820-1546 1820-1545 1826-0519 1826-0519	1 2 1 9		IC MULTIPLXR 2-CHAN-ANLG TRIPLE 16-DIP-C IC MULTIPLXR 4-CHAN-ANLG DUAL 16-DIP-C IC MULTIPLXR 2-CHAN-ANLG TRIPLE 16-DIP-C IC OP AMP LOW-BIAS-H-IMPD 8-DIP-P PKG IC OP AMP LOW-BIAS-H-IMPD 8-DIP-P PKG	3L585 04713 3L585 01295 01295	CD4053BY MC14052ECL CD4053BY TL071CP TL071CP
U21 U22 U23 U24 U25	1820-1545 1826-0519 1826-0175 1820-1212 1826-0138	1 9 3 9 8	1 1 1	IC MULTIPLXR 2-CHAN-ANLG TRIPLE 16DIP-C IC OP AMP LOW-BIAS-H-IMPD 8-DIP-P PKG IC COMPARATOR GP DUAL 14-DIP-P PKG IC FF TTL LS J-K NEG-EDGE-TRIG IC COMPARATOR GP QUAD 14-DIP-P PKG	3L585 01295 27014 01295 01295	CD4053BY TL071CP LH319N SN74LS112AN LH339N \
U26 U27 U28 U29 U30	1826-0519 1826-0248 1813-0297 1816-1533 1820-1442	9 1 3 B 7	1 1 1	IC OP AMP LOW-BIAS-H-IMPD 8-DIP-P PKG IC CONV 8-8-D/A 16-DIP-C PKG IC (HISC) IC-H87051 IC CNTR TTL LS DECD ASYNCHRO	01295 04713 28480 28480 01295	TL071CP MC1408L-6 1813-0297 1816-1533 SN74LS290N
U31 U32 U33 U34 U35	1826-0547 1820-1443 1820-0683 1820-0630 1820-1730	3 8 6 3 6	1 1 1 4	IC OP AMP LOW-BIAS-H-IMPD DUAL 8-DIP-P IC CNTR TIL LS BIN ASYNCHRO IC INV TIL S HEX 1-INP IC MISC TIL IC FF TIL LS D-TYPE POS-EDGE-TRIG COM	01295 01295 01295 01295 04713 01295	TL072ACP SN74L5293N SN74S04N MC4044P SN74L5273N
U36 U37 U38 U39 U40	1826-0147 1826-0221 1826-0122 1826-0122 1813-0298	9 0 0 4	5 5 4	IC 7012 V RGLTR TO-220 IC V RGLTR TO-220 IC 7005 V RGLTR TO-220 IC 7005 V RGLTR TO-220 IC (MISC)	04713 04713 07263 07263 28480	MC7012CF MC7912CT 7805UC 7805UC 1813-0298
U41 U42 U43 U44 U45	1826-0519 1826-0519 1826-0519 1826-0147 1826-0221	9 9 9 9		IC OP AMP LOW-BIAS-H-IMPD 8-DIP-P PKG IC OP AMP LOW-BIAS-H-IMPD 8-DIP-P PKG IC OP AMP LOW-BIAS-H-IMPD 8-DIP-P PKG IC 7812 V RGLTR TO-220 IC V RGLTR TO-220	01295 01295 01295 04713 04713	TL071CP TL071CP TL071CP HC7812CP HC7912CT
U46 U47 U48 U49 U50	1826-0221 1826-0147 1820-1545 1826-0519 1820-1730	9 1 9 6		IC V RGLTR TO-220 IC 7812 V RGLTR TO-220 IC MULTIPLXR 2-CHAN-ANLG TRIPLE 16-DIP-C IC OP AMP LOW-BIAS-H-IMPD 8-DIP-P PKG IC FF TTL LS D-TYPE POS-EDGE-TRIG COM	04713 04713 3L585 01295 01295	MC7912CT MC7812CP CD4053BY TL071CP SN74LS273N
บ51 บ52 บ53 บ54 บ55	1820-1730 1820-1730 1826-0147 1826-0221 1826-0122	6 9 0		IC FF TTL LS D-TYPE POS-EDGE-TRIG COM IC FF TTL LS D-TYPE POS-EDGE-TRIG COM IC 7812 V RGLTR TO-220 IC V RGLTR TO-220 IC 7805 V RGLTR TO-220	01295 01295 04713 04713 07263	SN74LS273N SN74LS273N MC7812CP MC7912CT 7805UC
U56 U57 U58 U59 U60	1826-0122 1820-1416 1826-0522 1813-0296 1813-0296	5 4 2 2		IC 7805 V RGLTR TO-220 IC SCHMITT-TRIG TTL LS INV HEX 1-INP IC OP AMP LOW-BIAS-H-IMPD QUAD 14-DIP-P IC (MISC) IC (MISC)	07263 01295 01295 28480 28480	7805UC SN74LS14N TL074CN 1813-0296 1813-0296
U61 U62 U63 U64 U65	1813-0296 1820-1975 1820-1975 1820-1975 1813-0302	1 1 1 1	3	IC (MISC) IC SHF-RGTR TTL LS NEG-EDGE-TRIG PRL-IN IC SHF-RGTR TTL LS NEG-EDGE-TRIG PRL-IN IC SHF-RGTR TTL LS NEG-EDGE-TRIG PRL-IN IC (MISC)	29480 01295 01295 01295 28480	1813-0296 SN74LS165N SN74LS165N SN74LS165N 1813-0302
U66 U67 U68	1826-0221 1826-0147 1820-0475	0 9 4		IC V RGLTR TO-220 IC 7812 V RGLTR TO-220 IC COMPARATOR HS TO-99 PKG	04713 04713 27014	MC7912CT MC7812CP LM386H
A2 U1 U2 U3 U4 U5	8159-0005 8159-0005 8159-0005 8159-0005 8159-0005	0 0		RESISTOR-ZERO OHMS 22 AWG LEAD DIA RESISTOR-ZERO OHMS 22 AWG LEAD DIA	28480 28480 28480 28480 28480	8159-0005 8159-0005 8159-0005 8159-0005 8159-0005
N6 N7 N8 N9 N10	8159-0005 8159-0005 8159-0005 8159-0005 8159-0005	0 0		RESISTOR-ZERO OHMS 22 AWG LEAD DIA RESISTOR-ZERO OHMS 22 AWG LEAD DIA	28480 28480 28480 28480 28480	8159-0005 8159-0005 8159-0005 8159-0005 8159-0005
				MISCELLANEDUS PARTS		
	0340-0060 0340-0092 0340-0220 0340-1049	8 1	28	TERMINAL-STUD SPCL-FDTHRU PRESS-MTG TERMINAL-STUD SPCL-FDTHRU PRESS-MTG SPACER-T0220	98291 28480 28480 28480	011-6809 000 209 0340-0092 0340-0220 0340-1049

Table 6-3. Replaceable Parts

Reference Designation	HP Part Number	СД	Qty	Description	Mfr Code	Mfr Part Number
A 4	04276-66584	3	1	POWER SUPPLY BOARD ASSEMBLY	28480	04276-66504
A4C1 C2 C3 C4	0180-1075 0180-2980 0180-2980 0180-1075	3 1 1 3	3	CAPACITOR-FXD 2200 UF 16VDC AL CAPACITOR-FXD 1000UF+-20% 35VDC AL CAPACITOR-FXD 1000UF+-20% 35VDC AL CAPACITOR-FXD 2200 UF 16VDC AL	28480 28480 28488 28480	0180-1075 0180-2980 0180-2980 0180-1075
cs	0180-3221	5	6	CAPACITOR-FXD 10 UF 100VDC	28480	0180-3221
C6 C7 C8 C9 C10	0180-3221 0180-3221 0180-3221 0180-3221 0180-3221	55555		CAPACITOR-FXD 10 UF 100VDC	28480 28480 28480 28480 28480	0180-3221 0180-3221 0180-3221 0180-3221 0180-3221
C11 C12 C13 C14 C15 C16 C17 C18 C19 C20	0180-1050 0180-1050 0180-1050 0160-1050 0160-2055 0180-1075 0160-3456 0190-0197 9160-4822 0180-0291 0160-3094 0180-1704	4 4 4 9 36 82 3 8 5	1 1 1 1 1 1 1 1	CAPACITOR-FXD 100UF 25VDC CAPACITOR-FXD 100UF 25VDC CAPACITOR-FXD 100UF 25VDC CAPACITOR-FXD 100UF 25VDC CAPACITOR-FXD 01UF +80-20% 100VDC CER CAPACITOR-FXD 2200 UF 16VDC AL CAPACITOR-FXD 1000PF +-10% 16VDC CER CAPACITOR-FXD 1000PF +-10% 26VDC TA CAPACITOR-FXD 1000PF +-5% 100VDC CER CAPACITOR-FXD 10F+10% 35VDC TA CAPACITOR-FXD 1UF+-10% 35VDC TA CAPACITOR-FXD 1UF+-10% 35VDC TA CAPACITOR-FXD 1VF+-10% 6VDC TA	28480 28480 28480 28480 28480 28480 56289 28480 56289 28480 56289	0180-1050 0100-1050 0180-1050 0160-2055 0180-1075 0160-3456 150P222879020A2 0160-4822 150P105879035A2 0160-3094 150P47689006B2
C22 C23 C24 C25 C26	0180-0228 0160-0127 9160-0127 0160-4593 0160-0127	6 2 2 4 2	2	CAPACITOR-FXD 22UF+-10% 15VDC TA CAPACITOR-FXD 1UF +-20% 25VDC CER CAPACITOR-FXD 1UF +-20% 25VDC CER CAPACITOR-FXD 1.5UF +-20% 400VDC CAPACITOR-FXD 1UF +-20% 25VDC CER	56289 28480 28480 28480 28480	150D226X9015B2 0160-0127 0160-0127 0160-4593 0160-0127
C27 C28 C29 C30 C31	0180-1746 0160-4593 0180-3231 0180-3231 0180-3231	5 4 7 7 7	1	CAPACITOR-FXD 15UF+-10% 20VDC TA CAPACITOR-FXD 1.5UF +-20% 400VDC CAPACITOR-FXD 4.7 UF 450VDC CAPACITOR-FXD 4.7 UF 450VDC CAPACITOR-FXD 4.7 UF 450VDC	56289 28480 28480 28480 28480	150D156X9020B2 0160-4593 0180-3231 0180-3231 0180-3231
C32 C33 C34 C35 C36	0180-3231 0180-3253 0180-3253 0160-3969 0160-3969	7 6 6 6	2	CAPACITOR-FXD 4.7 UF 458VDC CAPACITOR-FXD 470 UF 200VDC CAPACITOR-FXD 470 UF 200VDC CAPACITOR-FXD .815UF +-20PF 250VAC(RMS) CAPACITOR-FXD .015UF +-20PF 250VAC(RMS)	28480 28480 28480 28480 28480	0180-3231 0180-3253 0180-3253 0160-3769 0160-3769
C37	0180-0228	6		CAPACITOR-FXD 22UF+-18% 15VDC TA	56289	150D226X901582
A4CR1 CR2 CR3 CR4 CR5	1902-1217 1902-3234 1902-3234 1901-0025 1901-0025	3 3 2	1 2 11	DIODE-ZNR 6.2V 5% DO-4 PD=10W TC=+.035% DIODE-ZNR 19.6V 5% DO-35 PD=.4W DIODE-ZNR 19.6V 5% DO-35 PD=.4W DIODE-GEN PRP 100V 200MA DO-7 DIODE-GEN PRP 130V 200MA DO-7	28480 28480 28480 28480 28480	1902-1217 1902-3234 1902-3234 1901-0025 1901-0025
CR6 CR7 CR8 CR9 CR10	1901-0025 1901-0025 1901-0691 1901-0691 1901-0691	8 8 5 5	6	DIODE-GEN PRP 180V 200MA DO-7 DIODE-GEN PRP 180V 200MA DO-7 DIODE-PWR RECT 180V 3A 200NS DIODE-PWR RECT 180V 3A 200NS DIODE-PWR RECT 180V 3A 200NS	28480 28480 03508 03508 03508	1901-0025 1901-0025 A115A A115A A115A
CR11 CR12 CR13 CR14 CR15	1901-0691 1901-0691 1901-0691 1901-0691 1901-0969	8 8 3 3	2	DIODE-PWR RECT 100V 3A 200NS DIODE-PWR RECT 100V 3A 200NS DIODE-PWR RECT 100V 3A 200NS DIODE-POWER RECT. DIODE-POWER RECT.	03508 03508 03508 28480 28480	A115A A115A A115A 1901-0969 1901-0969
CR16 CR17 CR18 CR19 CR20	1902-3182 1901-0025 1901-0025 1902-3203 1901-0025	0 2 2 6 2	1	DIODE-ZNR 12.1V 5% DO-35 PD=,4W DIODE-GEN PRP 100V 200MA DO-7 DIODE-GEN PRP 100V 200MA DO-7 DIODE-ZNR 14.7V 5% DO-35 PD=,4W DIODE-GEN PRP 100V 200MA DO-7	28480 28480 28480 28480 28480	1902-3182 1901-0025 1901-0025 1902-3203 1901-0025
CR21 CR22 CR23 CR24 CR25	1901-0025 1902-0555 1901-0025 1901-0025 1901-1065	2 2 2 2 2	1 3	DIODE-GEN PRP 100V 200MA DO-7 DIODE-ZNR 13V 5% PD=1W IR=5UA DIODE-GEN PRP 100V 200MA DO-7 DIODE-GEN PRP 100V 200MA DO-7 DIODE-PWR RECT 1N4936 400V 1A 200NS	28480 28480 28480 28480 14936	1701-0025 1902-0555 1901-0025 1901-0025 1901-0025 1N4936
CR26 CR27 CR28 CR29 CR30	1902-3191 1901-0025	2 1 2 9	1	DIODE-PWR RECT 1N4936 400V 1A 200NS DIODE-PWR RECT 1N4936 400V 1A 200NS DIODE-ZNR 13V 2% DO-35 PD=,4W TC=+.06% DIODE-SEN PRP 100V 200MA DO-7 DIODE-FW BRDG 600V 10A	14936 14936 28480 28480 28480	1N4936 1N4936 1982-3191 1901-0025 1966-0088
94F1 F2 F3	2110-0305	1 5 4	1 1 1	FUSE .25A 250V NTD 1.25X.25 UL FUSE 1.25A 250V TD 1.25X.25 UL FUSE 1A 250V TD 1.25X.25 UL	28480 75915 25915	2110-0004 3131.25 313001

Table 6-3. Replaceable Parts

Reference Designation	HP Part Number	C D	Qty	Description	Mfr Code	Mfr Part Number
A4J1 J2	1251-4938 1251-38 <b>3</b> 7	5	1	CONNECTOR 3-PIN M METRIC POST TYPE CONNECTOR 4-PIN M UTILITY	28480 28480	1251-4938 1251-3037
A4L1 L2 L3 L4 L5	9100-3139 9140-0758 9140-0171 9140-0171 9140-0462	5 3 3 3 5	1 1 4	INDUCTOR 75UH 15% .5DX.875LG INDUCTOR 787 UH INDUCTOR RF-CH-HLD 40UH 10% .296DX.968LG INDUCTOR RF-CH-HLD 40UH 10% .296DX.968LG INDUCTOR 355UH	28480 28480 28480 28480 28480	9100-3139 9140-0758 9140-0171 9140-0171 9140-0171
L6 L7 L8 L9 L10	9140-0757 9140-0171 9140-0463 9140-0171 9140-0210	0 3 6 3 1	1 1	INDUCTOR- 980 UH INDUCTOR RF-CH-MLD 43UH 10% .296DX.968LG INDUCTOR 10MH 62 INDUCTOR RF-CH-MLD 43UH 10% .296DX.968LG INDUCTOR RF-CH-MLD 100UH 5% .166DX.395LG	28480 28480 28480 28480 28480	9140-0757 9140-0171 9140-0463 9140-0171 9140-0210
A4R1 G2 G3 G4 Q5	1853-0281 1854-0477 1854-0477 1854-0477 1853-0281	9 7 7 7 9	3 5	TRANSISTOR PNP 2N2707A SI TO-18 PD=400MW TRANSISTOR NPN 2N2222A SI TO-18 PD=500MW TRANSISTOR NPN 2N2222A SI TO-18 PD=500MW TRANSISTOR NPN 2N2222A SI TO-18 PD=500MW TRANSISTOR PNP 2N2707A SI TO-18 PD=400MW	94713 04713 04713 04713 04713	2N2907A 2N2222A 2N2222A 2N2222A 2N2222A 2N2707A
ធ.6 	1854-0477 1653-0281 1854-0624 1854-0624 1854-0935	7 9 6 6 2	2	TRANSISTOR NPN 2N2222A SI TO-18 PD=500MW TRANSISTOR PNP 2N2907A SI TO-18 PD=400MW TRANSISTOR NPN 2N630B SI TO-3 PD=125W TRANSISTOR NPN 2N630B SI TO-3 PD=125W TRANSISTOR-NPN	04713 04713 04713 04713 04713 28480	2N2222A 2N2907A 2NG308 2NG30B 1854-0935
Q11 Q12	1654-0936 1854-0477	3 7	1	TRANSISTOR-NPN TRANSISTOR NPN 2N2222A SI TO-18 PD=500MW	28480 04713	1854-0936 2N2222A
A4R1 R2 R3 R4 R5 R6	0683-2235 0683-4705 0683-1005 0683-1515 0683-4715 0683-4715	5050000	1 3 3 1 3	RESISTOR 22K 5% .25W FC TC=-400/+800 RESISTOR 47 5% .25W FC TC=-400/+500 RESISTOR 10 5% .25W FC TC=-400/+500 RESISTOR 150 5% .25W FC TC=-400/+600 RESISTOR 470 5% .25W FC TC=-400/+600 RESISTOR 470 5% .25W FC TC=-400/+600	01121 01121 01121 01121 01121 01121	CB2235 CB4705 CB1005 CB1515 CR4715 CB4715
R7 RB R9 R10 R11	0.683-4735 9683-4735 9683-4715 9683-1525 0683-1525	4 4 8 4	ય	RESISTOR 47K 5% .25W FC TC=-400/+800 RESISTOR 47K 5% .25W FC TC=-400/+800 RESISTOR 470 5% .25W FC TC=-400/+600 RESISTOR 1.5K 5% .25W FC TC=-400/+700 RESISTOR 1.5K 5% .25W FC TC=-400/+700	01121 01121 01121 01121 01121	CB4735 CB4735 CB4715 CB1525 CB1525
R12 R13 R14 R15 R16	9683-4795 0683-4795 2100-3352 2100-3274 0683-1025	8 7 2 9	1 1 1	RESISTOR 47 5% .25W FC TC=-400/+500 RESISTOR 47 5% .25W FC TC=-400/+500 RESISTOR-TRNR 1K 10% C STDE-ADJ 1-TRN RESISTOR-TRNR 10K 10% C SIDE-ADJ 1-TRN RESISTOR 1K 5% .25W FC TC=-400/+600	01121 01121 28480 28480 01121	CB4705 2100-3352 2100-3274 CB1025
R17 R18 R19 R20 R21	0764-0015 8683-0335 0683-1005 0683-0335 0683-1005	7 2 5 2 5	1 2	RESISTOR 560 5% 2W MO TC=0+-280 RESISTOR 3.3 5% .25W FC TC=-400/+500 RESISTOR 10 5% .25W FC TC=-400/+500 RESISTOR 3.3 5% .25W FC TC=-400/+500 RESISTOR 10 5% .25W FC TC=-400/+500	28480 01121 01121 01121 01121	0764-0015 CB3365 CB1005 CB3365 CR1005
R 22 R 23 R 24 R 25	0683-5615 0683-1035 0683-0275 0683-0275	1 1 9 7	1 1 4	RESISTOR 560 5% .25W FC TC=-400/+600 RESISTOR 10K 5% .25W FC TC=-400/+700 RESISTOR 2.7 5% .25W FC TC=-400/+500 RESISTOR 2.7 5% .25W FC TC=-400/+500	01121 01121 01121 01121	CB5615 CB1035 CB27G5 CB27G5
R26 R27 R28 R29 R30 R31 R32 R33 R34 R35 R35	0683-0275 0683-0275 0766-0033 0761-0004 0699-1057 0686-3945 0683-5635 0686-1055 0698-3657 0698-3657 0698-3657	9 9 3 B 4 2 5 1 B B 3	1 1 1 1 1 2	RESISTOR 2.7 5% .25W FC TC=-400/+500 RESISTOR 2.7 5% .25W FC TC=-400/+500 RESISTOR 2K 2% 3W MO TC=0+-250 RESISTOR 15 10% 3W RESISTOR 15 10% 3W RESISTOR 390K 5% .5W CC TC=0+882 RESISTOR 56K 5% .25W FC TC=-400/+800 RESISTOR 1M 5% .5W CC TC=0+1000 RESISTOR 1M 5% .5W MO TC=0+-200 RESISTOR 68K 5% 2W MO TC=0+-200 RESISTOR 68K 5% 2W MO TC=0+-200 RESISTOR 68K 5% 2W PW TC=0+-400	01121 01121 27167 28480 28480 01121 01121 01121 27167 27167 75042	CB27G5 CB27G5 FP3-3-250-2001-G 0761-0004 0699-1057 EB3745 CB5635 EB1055 FP42-2-T00-6802-J FP42-2-T00-6802-J BWH2-2R2-J
A4RT1	0839-0006	5		THERMISTOR DISC	28480	0839-0006 0837-0237
A4RV1 RV2	0837-0237 0837-0106	5	1	VARISTOR VARISTOR	28480 28480	0837-0106
A4T1 T2 T3	9100-4287 9100-0857 9100-4293	8 2	1 1 1	TRANSFORMER-POWER TRANSFORMER-PULSE 114H1 TRANSFORMER-PULSE	28480 28480 28480	9100-4287 9100-0857 9100-4293
A4U1	1813-0255	3	1	IC-REGULATOR HYPRID	28480	1813-0255
A4 14 t	8159-0005	0	1	RESISTOR-ZERO OHMS 22 AWG LEAD DIA MISCELLANEOUS PARTS	28480	8159-0005
	0340-0060 0340-0092 0340-0220 0380-0465 2110-0269	4 2 8 7 0		TERMINAL-STUD SPCL-FDTHRU PRESS-MTG TERMINAL-STUD SPCL-FDTHRU PRESS-MTG SPACER-RND .188-IN-LG .194-IN-ID FUSEHOLDER-CLIP TYPE.25D-FUSE	98291 28480 28480 28480 28480 28480	011-6809 000 209 0349-0092 0340-0220 0380-0445 2110-0269

Table 6-3. Replaceable Parts

				Table 6-3. Replaceable Parts		
Reference Designation	HP Part Number	C D	Qty	Description	Mfr Code	Mfr Part Number
	2360-0115 2740-0003 04276-01204	4 5 4	6 1 1	SCREW-MACH 6-32 .312-IN-LG PAN-HD-POZI NUT-HEX-W/LKWR 10-32-THD .125-IN-THK ANGLE (HEATSINK)	00000 00000 28480	ORDER BY DESCRIPTION ORDER BY DESCRIPTION 04276-01204
		-				

Table 6-3. Replaceable Parts

### 1   DISPLAY BOARD   A   DISPLAY BOARD ANSEMBLY   29-00   A   DISPLAY BOARD ANSEMBLY   29-00   D	Reference Designation	HP Part Number	C D	Qty	Description	Mfr Code	Mfr Part Number
1	Λ5						
	7.5	04276-66505	4	1	DISPLAY BOARD ASSEMBLY	28480	04276-66505
CAPACITOR-LOS   1	5C1	0180-1085	5	4	CAPACITOR-FXD 4.7UF 16VDC TA		
Sign=1005   1	C2			,			
1996-0541   3	C4	0180-1085	5	• 1	CAPACITOR-FXD 4.7UF 16VDC TA	28480	
Dec	C5	0180-1085	5		CAPACITOR-FXD 4,7UF 16VDC 1A		
1998				10	DISPLAY-NUM-SEG 1-CHAR .43-H		
1998   1999-1950   3			3		DISPLAY-NUM-SEG 1-CHAR .43-H	28480	5082-7650
			3		DISPLAY-NUM-SEG 1-CHAR .43-H DISPLAY-NUM-SEG 1-CHAR .43-H		
1999-0540   3   015FLAT-NUM-SEC 1-CHAR .43-H   20140   5082-7650			7		DISPLAY-NUM-SEG 1-CHAR .43-H	28480	5082-7650
1998-0548   3	DS7	1990-0540	3		DISPLAY-NUM-SEG 1-CHAR .43-H	28480	5082-7650
1991-0540   1991-0541   2					DISPLAY-NUM-SEG 1-CHAR .43-H		5082-7650
1998-1331   1998-1331   2   DISPLAY-NUM-SEC 1-CHAR 3-H   20480   5082-7610   5082-7610   50814   1998-1048   6   DISPLAY-NUM-SEC 1-CHAR 3-H   20480   5082-7610					DISPLAY-NUM-SEG 1-CHAR .43-H	28480	5992-7650
1991-1531   2   DISPLAY-NUM-SEG 1-CHAR 3-H   20480   5002-7619				4			
1979-8531   2						28480	5082-7610
1799-0486   1799-0486   6	DS14	1990-8531	2		DISPLAY-NUM-SEG 1-CHAR .3-H		
1990-0486   6	DS15	1990-0486		51			
DESIGN   1990-0466					LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V	28480	5082-4684
1998-0486   6	DS18	1990-0486	6	_	LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V	28480	5082-4684
1990-0486				5			
Dec			1 1		LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V		
1990-0486   6   LED-LAP LUN-INT-INC)   F-20MA-MAX BVR-50   20480   5082-4684   5082-4694	DS22	1990-0486	6		LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V		
DB226					LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V	28480	5082-4684
1990-0486   6					LED-LAMP LUM-INT=1MCD IF=20MA-MAX BUR=5V	28488	5082-4684
1998-0486   6   LED-LAMP LUM-INT-INCD   FF-20MA-MAX   BUR-50   20480   5382-4684   5832-					LED-LAMP LUM-INT=IMCD IF=20MA-MAX BVR=5V		
1990-0486   6					LED-LAMP LUM-INT=1MCD TF=20MA-MAX BVR=5V	28488	5082-4684
1990-0486   6	DS29	1990-0486	6		LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V		
1970-0486   6						1	
DS33					LED-LAMP LUM-INT=1MCD TF=20MA-MAX BVR=5V	28480	5082-4684
DS34	DS33	1990-0486	6		LED-LAMP LUM-INT=1MCD IF=20MA-MAX BUR=5V		
19337					LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V		
DS37 1990-0486 6 LED-LAMP LUM-INT=IMCD IF=20MA-MAX 8VK=5V 28480 5082-4684 50	DS36	1990-0486	6		LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V		
DS39	DS37				LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V		
DS41		1990-0486	5		LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V		
1770-0486	DS48	1990-0486	6				
DS42				]	ISD-LAMP LUM-INT=1MCD IF=28MA-MAX BUR=5U		
DS44 1970-0486 6 LED-LAMP LUM-INT=IHCD IF=20MA-MAX BVR=SV 28480 DS47 1990-0486 6 LED-LAMP LUM-INT=IHCD IF=20MA-MAX BVR=SV 28480 DS48 1990-0486 6 LED-LAMP LUM-INT=IHCD IF=20MA-MAX BVR=SV 28480 DS49 DS49 DS50 DS50 DS50 DS51 1990-0486 6 LED-LAMP LUM-INT=IHCD IF=20MA-MAX BVR=SV 28480 DS51 DS52 DS52 DS53 1990-0486 6 LED-LAMP LUM-INT=IHCD IF=20MA-MAX BVR=SV 28480 DS55 DS56 1990-0486 6 LED-LAMP LUM-INT=IHCD IF=20MA-MAX BVR=SV 28480 DS57 DS58 1990-0486 6 LED-LAMP LUM-INT=IHCD IF=20MA-MAX BVR=SV 28480 DS59 1990-0486 6 LED-LAMP LUM-INT=IHCD IF=20MA-MAX BVR=SV 28480 DS60 DS61 1990-0486 6 LED-LAMP LUM-INT=IHCD IF=20MA-MAX BVR=SV 28480 DS62 1990-0486 6 LED-LAMP LUM-INT=IHCD IF=20MA-MAX BVR=SV 28480 DS64 DS64 DS64 DS66 DS64 DS66 DS66 DS66	DS43	1990-0665	3		LED-LAMP LUM-INT=1MCD IF=28MA-MAX BUR=5V	28480	1990-0665
DS46 DS47 DS48 DS48 DS48 DS49 DS49 DS50 DS51 DS51 DS52 DS52 DS53 DS53 DS53 DS54 DS53 DS54 DS55 DS55 DS55 DS55 DS55 DS55 DS55					LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V		
DS47					LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V		
1770-0486	DS47	1990-0486	6		LED-LAMP LUM-INT=1MCD IF=28MA-MAX BVR=5V		
DS50					LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V	29488	5082-4684
DS51					LED-LAMP LUM-INT=1MCD IF=20MA-MAX BUR=5U	28480	
1990-0486   6   LED-LAMP LUM-INT=INCD IF=20MA-MAX BVR=5V   28480   5082-4684   290-0486   6   LED-LAMP LUM-INT=INCD IF=20MA-MAX BVR=5V   28480   5082-4684   28480					LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V		
1990-0486					LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V	28480	5082-4684
DS56	DS54	1990-0486	6		LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V		
DS56 1770-0486 1970-0486				1	i	1	5092-4684
DS58			6	l	LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V	28480	5082-4684
DS60 1990-0486 6 LED-LAMP LUM-INT=IMCD IF=20MA-MAX BVR=5V 28480 5002-4684  LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V 28480 5002-4684  DS62 1990-0486 6 LED-LAMP LUM-INT=IMCD IF=20MA-MAX BVR=5V 28480 5002-4684  LED-LAMP LUM-INT=IMCD IF=20MA-MAX BVR=5V 28480 1990-0665  DS64 1990-0486 6 LED-LAMP LUM-INT=IMCD IF=20MA-MAX BVR=5V 28480 5002-4684  LED-LAMP LUM-INT=IMCD IF=20MA-MAX BVR=5V 28480 5002-4684  DS64 1990-0486 6 LED-LAMP LUM-INT=IMCD IF=20MA-MAX BVR=5V 28480 5002-4684	DS58	1990-0486			LED-LAMP LUM-INT=1MCD IF=20MA-MAX BUR=5V		
DS62					LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V		5082-4684
DS62 1970-0486 6 LED-LAMP LUM-INT=INCD IF=20MA-MAX BVR=5V 28480 1970-0665 1970-0665 1970-0665 1970-0486 6 LED-LAMP LUM-INT=INCD IF=20MA-MAX BVR=5V 28480 5382-4684 1970-0486 6 LED-LAMP LUM-INT=INCD IF=20MA-MAX BVR=5V 28480 5382-4684	DS61				LED-LAMP LUM-INT=1MCD IF=20MA-MAX BUR=5V		
DS64 1990-0486 6 LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V 28480 5082-4684	DS62	1990-0486			LED-LAMP LUM-INT=1MCD IF=28MA-MAX BUR=5V	28480	1998-0665
D262 1991-1992   6   CED-CHILL COLL-191-11000 It SCORN-1994 BAK-24   5000 1995	DS64	1990-0486	6	ļ	LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V	28480	5082-4684
	DS65	1990-0486	16		FEN-FMUL FOULTMINION INSCREMENT DAK-24	2.5700	

Table 6-3. Replaceable Parts

1990-0486 1990-0486 1990-0486 1990-0665 1990-0665 1200-0638 1310-0301 1810-0321 1810-0827 1810-0827 1810-0627 1810-0627 1810-0627 0683-4725 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436	66633 7 47477 2 77777 77	14 2 3 1 17	LED-LAMP LUM-INT=1MCD IF=20MA-MAX 5VR=5V LED-LAMP LUM-INT=1MCD IF=20MA-MAX BVR=5V LUM-INT=1MCD IF=20MA-MAX BVR=5V LUM-INT=20MA-MAX BVR=5V	78480 28480 28480 28480 28480 28480 28480 31121 28480 31121 28480 28480	5082-4684 5082-4684 5082-4684 1798-0665 1590-0665 1209-0638 3168510 1810-0627 3168510 1810-0627 1810-0627
1810-0301 1810-0627 1810-0627 1810-0301 1810-0627 1810-0627 1810-0627 1810-0627 1810-0627 1810-0627 1810-0627 1810-09436 1810-09436 1810-09436 1810-09436 1810-09436 1810-09436 1810-09436 1810-09436 1810-09436 1810-09436 1810-09436 1810-060-9436 1810-060-9436 1810-060-9436	47477 2 77777 7	2 3	NETWORK-RES 16-D1P51.0 OHM X 8 RESISTIVE NETWORK NETWORK-RES 16-D1P51.0 OHM X 8 RESISTIVE NETWORK RESISTIVE NETWORK RESISTIVE NETWORK RESISTOR 4.7K 5% .25W FC TC=-400/+700 PUSHBUTTON SWITCH P.C. MOUNT PUSHBUTTON SWITCH P.C. MOUNT	91121 28480 91121 28480 28480	3168510 1810-0627 3168510 1810-0627 1810-0627
1810-0627 1810-0301 1810-0627 1810-0627 0683-4725 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436	7 4 7 7 2 7 7 7 7 7	3	RESISTIVE NETWORK NETWORK-RES 16-DIPS1.0 DHM X B RESISTIVE NETWORK RESISTIVE NETWORK RESISTOR 4.7K 5% .25W FC TC=-400/+700 PUSHBUTTON SWITCH P.C. MOUNT PUSHBUTTON SWITCH P.C. MOUNT	28480 01121 28480 28480 01121	1810-0627 316B510 1810-0627 1810-0627
5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436	7 7 7 7 7		PUSHBUTTON SWITCH P.C. MOUNT PUSHBUTTON SWITCH P.C. MOUNT	j	CB4725
5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436	7 7 7 7	17	PUSHBUTTON SWITCH P.C. MOUNT		
5060-9436 5060-9436 5060-9436	7		PUSHBUTTON S₩ITCH P.C. MOUNT PUSHBUTTON SWITCH P.C. MOUNT PUSHBUTTON SWITCH P.C. MCUNT	28480 28480 28480 28480 28480	5060-9436 5060-9436 5060-9436 5060-9436 5060-9436
3101-1074	7 7 9	æ	PUSHBUTTON SWITCH P.C. MOUNT PUSHBUTTON SWITCH P.C. MOUNT PUSHBUTTON SWITCH P.C. MOUNT PUSHBUTTON SWITCH P.C. MOUNT SWITCH-PUSHBUTTON SPST NO	28480 28480 28480 28480 28480	5060-9436 5060-9436 5060-9436 5060-9436 3101-1074
3101-1074 3101-2046 5060-9436 5060-9436 5060-9436	9 7 7 7 7	1	SWITCH-PUSHEUTTON SPST NO SWITCH-SLIDE DPDT-NS PUSHEUTION SWITCH P.C. MOUNT PUSHBUTTON SWITCH P.C. MOUNT PUSHBUTTON SWITCH P.C. MOUNT	29480 28480 28480 28480 28480	3101-1074 3101-2046 5060-9436 5060-9436 5060-9436
5060-9436 5060-9436 5060-9436 5060-9436 5060-9436	7 7 7 7 7		PUSHBUTTON SWITCH P.C. MOUNT	28480 28480 28480 28480 28480	5060-9435 5060-9436 5060-9436 5060-9436 5060-9436
1958-0038 1820-0495 1820-1624 1820-1624 1858-0038	4 8 7 7 4	4 1 2	TRANSISTOR ARRAY IC DCDR TTL 4-TO-16-LINE 4-INP IC BFR TTL S CCTL 1-INP IC BFR TTL S CCTL 1-INP IC BFR TTL S OCTL 1-INP TRANSISTOR ARRAY	28480 01295 01295 01295 28480	1658-0038 SN74154N SN746241N SN746241N 1958-0038
1858-0038 1858-0038 1820-1216 1816-1533	4 4 3 8	1 1	TRANSISTOR ARRAY TRANSISTOR ARRAY IC DCDR TTL LS 3-TO-8-LINE 3-INP IC-MB7051	28480 23480 81295 28480	1858-0038 1858-0038 SN74L5138N 1816-1533
0360-1901 5041-0309 5041-0318 5041-0375 5041-0384	6 5 6 5 6	1 3 3 1 2	MISCELLANEOUS PARTS  CABLE TRANSISTION KEY CAP KEY CAP KEY CAP-QUARTER(SMOKE) KEY CAP-QUARTER(SMOKE)	28480 28480 28480 28480 28480	0360-1901 5041-0309 5041-0318 5041-0375 5041-0384
04262-40001	5	8 1 6 3 1	KEY CAP-QUARTER(EBY PEARL) INSULATOR INSULATOR INSULATOR CABLE ASSEMBLY-FLAT	28480 28480 28480 28480 28480	5041-0922 04191-40002 04262-40001 04274-40003 04276-61641
5040-3323		1	INSULATOR		
	5060-9436 5060-9436 3101-1074 3101-2046 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 5060-9436 1058-0038 1820-1624 1820-1624 1820-1624 1820-1624 1858-0038 1858-0038 1858-0038 1858-0038 5060-9436	5060-9436 7 5060-9436 7 3101-1074 9 3101-2046 7 5060-9436 7 5060-9	5060-9436     7       5060-9436     7       5101-1074     9       3101-2046     7       5060-9436     7       5060-9436     7       5060-9436     7       5060-9436     7       5060-9436     7       5060-9436     7       5060-9436     7       5060-9436     7       5060-9436     7       5060-9436     7       5060-9436     7       5060-9436     7       5060-9436     7       5060-9436     7       5060-9436     7       5060-9436     7       5060-9436     7       2080-0495     8       1820-1624     7       1820-1624     7       1858-0038     4       1858-0038     4       1858-0038     4       1858-038     4       1820-1216     3       1816-1533     1       0360-1901     5       5041-0318     3       5041-0329     5       5041-0318     3       5041-0329     6       204191-40002     0       04262-40001     5       04262-40001     6       04266-616	S060-9436   7	S060-9436   7

Table 6-3. Replaceable Parts

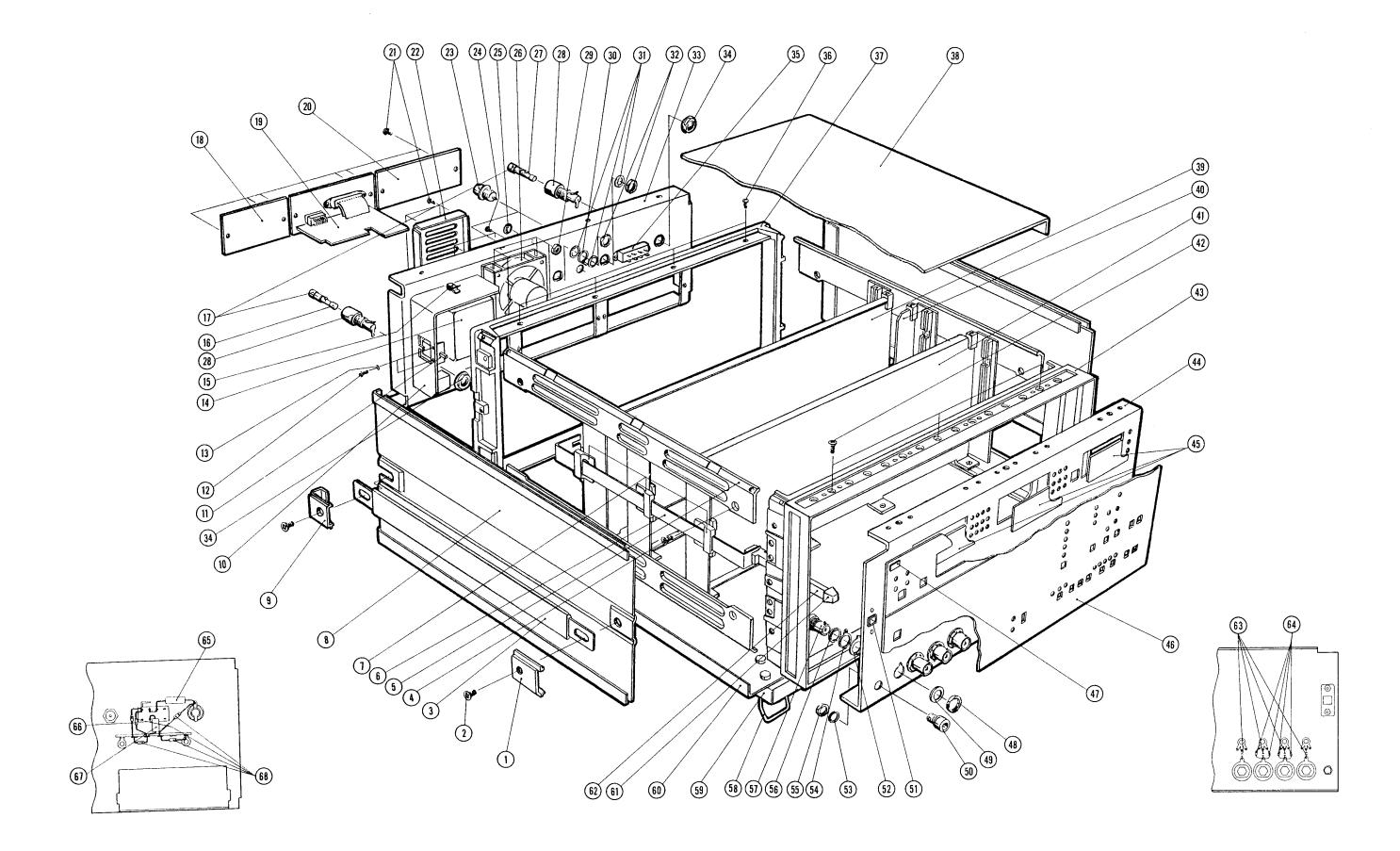
Reference Designation	HP Part Number	C D	Qty	Description	Mfr Code	Mfr Part Number
A 6						
7.0	04276-66506	5	1	MOTHER BOARD ASSEMBLY	28480	04276-66506
5BT1	1420-0306	2	1	BATTERY- 2.4V	28480	1420-0306
5 <b>J1</b> J2 J4	1251-7845 1251-5382 1251-0541	9 5 8	1 1 1	CONNECTOR- 6 PIN, MALE CONNECTOR 4-PIN M METRIC POST TYPE CONNECTOR 34-PIN M RECTANGULAR	28480 28480 28480	1251-7845 1251-5382 1251-0541
5 <b>U1</b>	1813-0304	3	1	IC (MISC) SIP	28480	1813-0304
5XA1L XA1R XA2C XA2L XA2R	1251-2582 1251-2582 1251-2026 1251-2582 1251-2582	1 9 1	5 2	CONNECTOR-PC EDGE 24-CONT/ROW 2-ROWS CONNECTOR-PC EDGE 24-CONT/ROW 2-ROWS CONNECTOR-PC EDGE 18-CONT/ROW 2-ROWS CONNECTOR-PC EDGE 24-CONT/ROW 2-ROWS CONNECTOR-PC EDGE 24-CONT/ROW 2-ROWS	28480 28480 28480 28480 28480	1251-2582 1251-2582 1251-2026 1251-2582 1251-2582
XA4C XA4R XA21 XA22 XA23	1251-2026 1251-2582 1251-4978 1251-4978 1251-4978	8 1 3 3 3	3	CONNECTOR-PC EDGE 18-CONT/ROW 2-ROWS CONNECTOR-PC EDGE 24-CONT/ROW 2-ROWS CONNECTOR-PC EDGE 18-CONT/ROW 2-ROWS CONNECTOR-PC EDGE 18-CONT/ROW 2-ROWS CONNECTOR-PC EDGE 18-CONT/ROW 2-ROWS	28480 28480 05574 05574 05574	1251-2026 1251-2582 000231-3944 000231-3944 000231-3944
				MISCELLANEOUS PARTS		
	0360-1244	0	4	TERMINAL-SPECIAL-FEEDTHRU	28480	0360-1244

Table 6-3. Replaceable Parts

Reference Designation	HP Part Number	C D	Qty	Description	Mfr Code	Mfr Part Number
A 2 1						
	04276-66521	4	1	HP-IB BOARD ASSEMBLY	28480	0 4276-66521
421 c1 ca	3180-2981 0180-1085	2 5	1 1	CAPACITOR-FXD 223UF+-23% 13VDC AL CAPACITOR-FXD 4.7UF 16VDC TA	28480 28480	0180-2981 0180-1085
A21J1 J2	1200-0485 1200-0654	2 7	1 1	SOCKETHIC 14-CONT DIP DIP-SLDR SOCKETHIC 40-CONT DIP DIP-SLDR	28480 28480	1200-0485 1200-0654
A21R1 R2 R3 R4	1810-9395 0683-4725 0683-4725 0683-4725	8 22 23	1 3	NETWORK-RES 9-SIP4.7K OHM X 8  RESISTOR 4.7K 5% .25W FC TC=-400/+700  RESISTOR 4.7K 5% .25W FC TC=-400/+700  RESISTOR 4.7K 5% .25W FC TC=-400/+700	28480 01121 01121 01121	1810-0305 CR4725 CB4725 CB4725
27 51	3101-1973	7	1	SWITCH-SL 7-1A DIP-SLIDE-ASSY .1A 50VDC	28480	3101-1973
21 U1 U2 U3 U4 U5	1820-2024 1820-2058 1820-2058 1820-2549 1820-1199	3 3 3 7 1	1 -4 1 1	IC DRVR TTL LS LINE DRVR OCTL IC MISC TTL S QUAD IC MISC TTL S QUAD IC-8291A P HPIB IC INV TTL LS MEX 1-INP	01295 07263 07263 28480 01295	SN74L5244N MC344BAL MC344BAL 1820-2549 SN74L504N
ช6 ย <b>7</b> ช8	1820-2058 1820-2058 1820-2075	3 3 4	1	IC MISC TTL S QUAD IC MISC TTL S QUAD IC MISC TTL LS	07263 07263 01295	MC3448AL MC3448AL SN74LS245N
21w1	8159-0005	0	1	RESISTOR-ZERO OHMS 22 AWG LEAD DIA MISCELLANEOUS PARTS	28480	8159-0005
	2360~0113 04276~00604 04276~61661	2 6 3	2 1 1	SCREW-MACH 6-32 .25-IN-LG PAN-HD-POZI PLATE (HP-IB) CABLE ASSEMBLY	00000 20480 28480	ORDER BY DESCRIPTION 04276-00604 04276-61661
A 2 2	04276-66522	5	1	OPTION 001 Internal DC Bias Board Assembly	28480	04276-66522
.22 C1 C2 C3 C4 C5	0180-2951 0160-5498 0180-2951 0180-2951 0180-2951	6 9 6 6 6 6	4	CAPACITOR-FXD 33UF+-20% 16VDC AL CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD 33UF+-20% 16VDC AL CAPACITOR-FXD 33UF+-20% 16VDC AL CAPACITOR-FXD 33UF+-20% 16VDC AL	28480 28480 28480 28480	0180-2951 0180-2951 0180-2951 0180-2951
C6 C7 C8 C9 C10	0180-3220 0180-3220 0150-5599 0160-5498 0160-1631	4 4 5 4 6	2 1 2 1	CAPACITOR-FXD 10 UF 63VDC AL CAPACITOR-FXD 10 UF 63VDC AL CAPACITOR-FXD .1UF +80-20% 50VDC CER CAPACITOR-FXD .01UF +-10% 100VDC CER CAPACITOR-FXD 1000FF +-10% 1KVDC CER	28480 28480	0180-3220 0180-3220
C11	0160-5498	4		CAPACITOR-FXD .01UF +-10% 100VDC CER		
22 CR 1 CR 2 CR 3	1902-0692 1901-0040 1901-0040	1 1 1	1 2	DIODE-ZNR 6.3V 1% DO-7 PD=.4W TC=+.001% DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35	28480 28480 28480	1902-0692 1901-0040 1901-0040
22 Q1 Q2 Q3 Q4 Q5	1854-0358 1853-0080 1853-0080 1854-0358 1854-0523	3 6 6 3 4	2 2	TRANSISTOR NPN SI PD=310MW FT=60MHZ TRANSISTOR PNP SI PD=300MW FT=30MHZ TRANSISTOR PNP SI PD=300MW FT=30MHZ TRANSISTOR NPN SI PD=310MW FT=60MHZ TRANSISTOR NPN SI TO-39 PD=1W FT=150MHZ	28480 28480 28480 28480 28480	1854-0358 1853-0080 1853-0080 1854-0358 1854-0523
ଘ୍ର	1853-0037	3	1	TRANSISTOR PNP SI TO-39 PD=1W FT=100MHZ	28480	1853-0037
22 R1 R2 R3 R4 R5	1810-0629 1810-0625 1810-0302 0699-1020 0683-2255	9 5 5 7 9	1 1 1 1	RESISTIVE NETWORK- DIP RESISTIVE NETWORK- DIP NETWORK-RES 8-SIP47.0 OHM X 4 RESISTOR- 470 OHM 1X RESISTOR 2.2M 5% ,25W FC TC=-900/+1100	28480 28480 01121 28480 01121	1810-0629 1810-0625 2088470 0699-1020 CB2255
R 6 R7 R8 R9 R10	2100-3214 2100-0567 2100-3214 0683-3355 0683-1035	0 0 0 2	2 1 1 1	RESISTOR-TRMR 100K 10% C TOP-ADJ 1-TRN RESISTOR-TRMR 2K 10% C TOP-ADJ 1-TRN RESISTOR-TRMR 100K 10% C TOP-ADJ 1-TRN RESISTOR 3.3M 5% .25W FC TC=-900/+1100 RESISTOR 10K 5% .25W FC TC=-400/+700	28480 28480 28480 01121	2100-3214 2100-0567 2100-3214 C93355 CB1035
22 U1 U2 U3 U4 U5 U6 U7	1820-1730 1820-1730 1826-0485 1826-0416 1826-0522 1826-0275 1826-0282	6685445	2 1 1 1 1 1 1 1	IC FF TTL LS D-TYPE POS-EDGE-TRIG COM IC FF TTL LS D-TYPE POS-EDGE-TRIG COM IC CONV 10-B-D/A 16-DIP-P PKG IC SWITCH ANLG QUAD 16-DIP-C PKG IC OP AMP LOW-BIAS-H-IMPD QUAD 14-DIP-P IC 78L12A V RGLTR TO-92 IC V RGLTR TO-92	01295 01295 24355 27014 01295 04713	SN74LS273N SN74LS273N AD7530LN LF13331D TL074CN MC78L12ACP MC79L12ACP
				MISCELLANEOUS PARTS		
	2360-0113 04276-00605	2	2	SCREW-MACH 6-32 .25-IN-LG PAN-HD-POZI PLATE (DC BIAS)	00080 28480	ORDER BY DESCRIPTION 04276-00605

Table 6-3. Replaceable Parts

Reference Designation	HP Part Number	C D	Qty	Description	Mfr Code	Mfr Part Number
1 2 3 4 5	5040-7219 2680-0172 5060-9803 2510-0192 5020-8836		2 4 2 16 4	STRAP HANDLE CAP (FRONT) SCREW STRAP HANDLE SCREW STRUT		
6 7 8 9	04276-01202 04274-40002 5060-9941 5040-7220 04276-01201		1 3 2 2 1	ANGLE (POWER SWITCH) GUIDE (ANGLE) SIDE COVER STRAP HANDLE CAP (REAR) ANGLE		
11 12 13 14 15	3101-2216 0515-0150 3050-0235 9135-0084 1400-0866		1 2 2 1 1	LINE SWITCH SCREW WASHER LINE FILTER CABLE CLAMP		
16 17 18 19 20	2110-0360 2100-0007 2110-0565 04276-00603 04276-66521 04276-00602		1 1 2 1 1	FUSE .75A 250V (220/240V) SLOW BLOW FUSE 1A 250V (100/120V) SLOW BLOW FUSEHOLDER CAP BLANK PANEL (COMPARATOR/HANDLER INTERFACE) HP-IB BOARD BLANK PANEL (INTERNAL DC BIAS)		
21 22 23 24 25	2360-0113 04276-04001 1250-0118 2200-0105 6960-0001		10 1 2 4 1	SCREW FAN COVER CONNECTOR-BNC SCREW CAP		
26 27 28 29 30	3160-0266 2110-0011 2110-0564 2260-0009 0360-1190		1 1 2 4 1	FAN FUSE 1/16A 250V FUSEHOLDER BODY NUT SOLDER TERMINAL		
31 32 33 34 35	2190-0016 2950-0001 04276-00204 2110-0569 3101-1877		3 2 1 2	WASHER NUT REAR PANEL FUSEHOLDER NUT DC BIAS SELECT SWITCH		
36 37 38 39 40	2360-0113 5020-8806 5060-9834 04276-00102 04276-00103		8 1 1 1	SCREW REAR FRAME TOP COVER CHASSIS (YELLOW) CHASSIS (RED)		
41 42 43 44 45	04276-00101 2360-0333 5020-8805 04276-00203 04276-25001		1 6 1 1 3	CHASSIS (BROWN) SCREW FRONT FRAME SUB PANEL WINDOW		
46 47 48 49 50	04276-00201 04276-00202 7120-1254 7120-0478 2950-0035 5040-3324 1510-0038		1 1 1 1 4 4	FRONT PANEL (HP) FRONT PANEL (YHP) NAME PLATE (HP) NUT INSULATOR-BNC BINDING POST		
51 52 53 54 55	04191-40001 5040-3325 2190-0084 5000-4212 2950-0006		1 4 1 4 1	GUIDE INSULATOR-BNC WASHER SOLDER TERMINAL NUT		
56 57 58 59 60	2190-0054 1250-0252 1460-1345 5040-7201 5060-9846		4 4 2 4 1	WASHER CONNECTOR-BNC STAND FOOT (BOTTOM) BOTTOM COVER		
61 62 63 64 55	5041-0564 04274-40001 0160-4297 1901-1065 0698-3634		1 1 4 4 1	KEY CAP ROD (POWER SWITCH) CAPACITOR 0.022μF DIODE RESISTOR 470Ω		
66 67 68	0683-2245 0764-0016 1902-0657		1 1 4	RESISTOR 220kΩ RESISTOR 1kΩ DIODE		



SECTION VII		
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Model 4276A

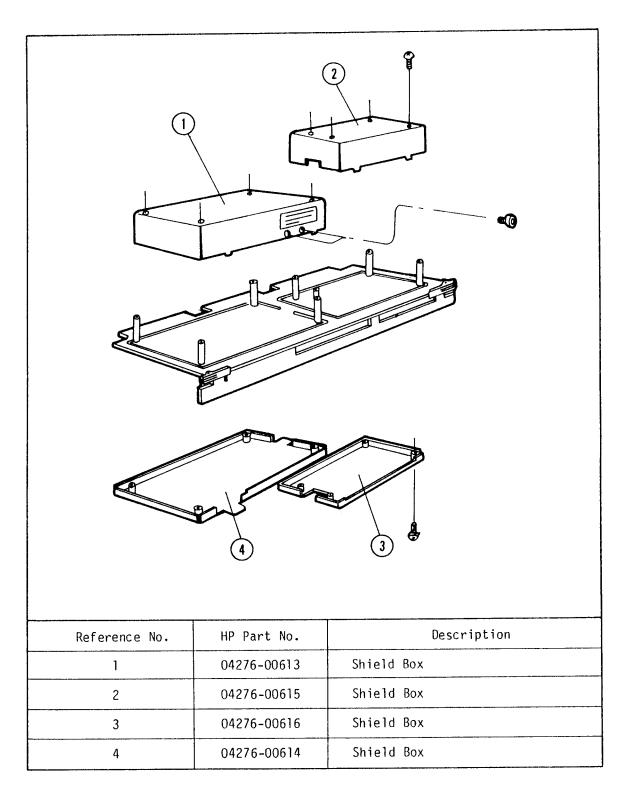


Figure 6-1. Shields on the A4 Board.

Model 4276A SECTION VII

# SECTION VII MANUAL CHANGES

## 7-1. INTRODUCTION

7-2. This section contains information for adapting this manual to instruments for which the content does not directly apply. The following paragraphs explain how to adapt this manual to apply to an older instrument with a serial prefix lower than that given on the title page.

## 7-3. MANUAL CHANGES

- 7-4. To adapt this manual to your instrument, refer to Table 7-1 and make all of the manual changes listed opposite your instrument serial number. Perform these changes in the sequence listed.
- 7-5. If your instrument serial number is not listed on the title page of this manual or in Table 7-1, it may be documented in a yellow MANUAL CHANGES supplement. For additional information about serial number coverage, refer to INSTRUMENTS COVERED BY MANUAL in Section I.

Table 7-1. Manual Changes by Serial Number

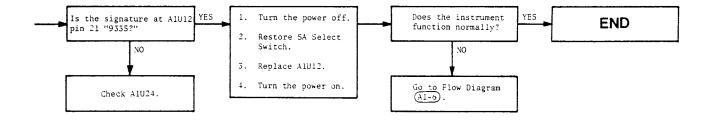
Serial Prefix or Number	Make Manual Changes
or Number 2227J00101 thru 2227J00155	1

CHANGE 1

Section VIII, Troubleshooting Flow Diagram Al-5:
Change Signature Sets 5-4, 5-6, and 5-11 as follows:

Signature Set 5-4 Signature Set 5-11 Signature Set 5-6 AlJl pin 9 3388 A1U10 pin 9 8415 AlU5 pin 9 0512 pin 10 A9FP 5193 pin 10 pin 10 5UA2 pin 11 3190 U083 pin 11 pin 11 AU44 pin 12 AF81 pin 13 2H2F 9693 pin 13 pin 13 50A4 pin 14 7A72 U762 pin 14 pin 14 4H45 pin 15 3PU8 9911 pin 15 pin 15 A07C pin 16 62A2 F79P pin 16 pin 16 6UC0 pin 17 CFF2 pin 17 UOPO

Partially change the flow diagram as follows :



## SERVICE

#### 8-1. INTRODUCTION

This section provides the information and instructions required to service the Model 4276A LCZ Meter. Included are Theory of Operation Troubleshooting Guide with Circuit Schematics. The Theory of Operation describes fundamental principles and circuit operating theory of the 4276A with block diagrams. schematics. locator illustrations. troubleshooting guide and other technical data necessary for repairs are integrated into the service sheet foldouts. An illustration of the instrument interior is shown in Figure 8-27.

#### 8-3. SAFETY CONSIDERATIONS

8-4. This section contains warnings and cautions that must be followed for your protection and to avoid damage to the instrument.

### WARNING

MAINTENANCE DESCRIBED HEREIN PERFORMED WITH POWER TO THE INSTRUMENT SUPPLIED PROTECTIVE AND COVERS SUCH MAINTENANCE REMOVED. SHOULD BE PERFORMED ONLY BY SERVICE-TRAINED PERSONNEL AWARE OF THE HAZARDS INVOLVED (FOR EXAMPLE, FIRE AND ELECTRICAL SHOCK). WHERE MAINTENANCE CAN BE WITHOUT PERFORMED POWER APPLIED, THE POWER SHOULD BE REMOVED. BEFORE ANY REPAIR IS COMPLETED, ENSURE THAT ALL FEATURES ARE INTACT SAFETY AND FUNCTIONING AND THAT ALL NECESSARY PARTS ARE CON-NECTED TO THEIR MEANS OF PROTECTIVE GROUNDING.

## 8-5. THEORY OF OPERATION

8-6. The theory of operation discussion is organized into two sections: basic theory and block diagram discussion. The basic theory, beginning with paragraph 8-13, explains the concepts and fundamental theory of the 4276A adapted for accurately measuring the DUT and for achieving automated measurements. The

block diagram discussion describes the overall circuit operating theory of the 4276A with block-to-block signal flow. Also included are block and timing diagrams.

## 8-7. RECOMMENDED TEST EQUIPMENT

8-8. The test equipment required to the perform operations outlined in this section is listed in Table 4-1. The table includes type of instrument required, critical specifications, use, and recommended model. If the recommended model is not available, equipment which meets or exceeds the critical specifications listed may be substituted.

## 8-9. TROUBLESHOOTING

8-10. The troubleshooting guide provides instructions and information for locating a faulty circuit component. All instructions consider the safety of service personnel performing the procedures. The diagnostic guides are in the flow diagrams. The board troubleshooting diagrams are used to isolate failures to an individual malfunctioning circuit board assembly. The guides for locating a defective component are given on the individual board service-sheets and integrate service support data—test point locations, waveform illustrations, voltage data, timing diagrams, and other technical information in addition to providing schematic diagrams for each board. To facilitate troubleshooting of the 4276A Digital Section, the troubleshooting guide for the logic circuits uses signature analysis.

## 8-11. REPAIR

8-12. Repair explanations tell how to replace defective circuit components. The recommended replacement procedures for components and parts which require special repair, replacement tools, or test equipment should be observed. To prevent damage resulting from improper repair procedure, refer to the appropriate manual section before proceeding with repair.

#### 8-13. BASIC THEORY

8-14. The 4276A applies a sinusoidal voltage to the device under test and detects the resulting complex voltage,  $\dot{V}$ , and complex current,  $\dot{I}$ . The instrument then converts  $\dot{V}$  and  $\dot{I}$  into their individual orthogonal components to obtain the DUT's resistance, R, and reactance, X.

$$\dot{V} = a + jb$$

$$\dot{I} = c + jd$$

$$R = \frac{ac + bd}{c^2 + d^2}$$

$$X = \frac{bc - ad}{c^2 + d^2}$$

Once the values of R and X are known, all other impedance parameters-C, L, D, Q, ESR, G,  $\mid Z \mid$ , and  $\theta$  can be calculated. Refer to Table 8-1.

8-15. A simplified drawing of the circuit used to detect  $\dot{V}$  and  $\dot{I}$  is shown in Figure 8-1. In the figure,  $\dot{V} = \mathbf{e}_{\text{DUT}}$  and  $\dot{I} = \mathbf{e}_{\text{RR}} / - \mathbf{e}_{\text{RR}}$ , where  $\mathbf{e}_{\text{DUT}}$  is the voltage across the DUT,  $\mathbf{e}_{\text{RR}}$  the output voltage from the I/V converter, and  $\mathbf{e}_{\text{RR}}$  is the value of the range resistor.

All measurement parameters are calculated from the two vector voltages  $\boldsymbol{e}_{\text{DUT}}$  and  $\boldsymbol{e}_{\text{RR}}.$ 

Table 8-1. Impedance Parameter Conversion Equations

Parameter	Equation
С	-1/2πfX
L	X/2πf
D	R/X
Q	X/R
ESR	R
G	$R/(R^2 + X^2)$
Z	$R^2 + X^2$
θ	tan <sup>-1</sup> (X/R)

f = test frequency

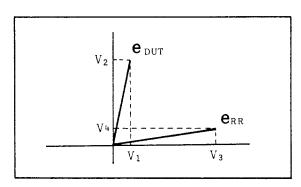


Figure 8-2. Vector Voltages.

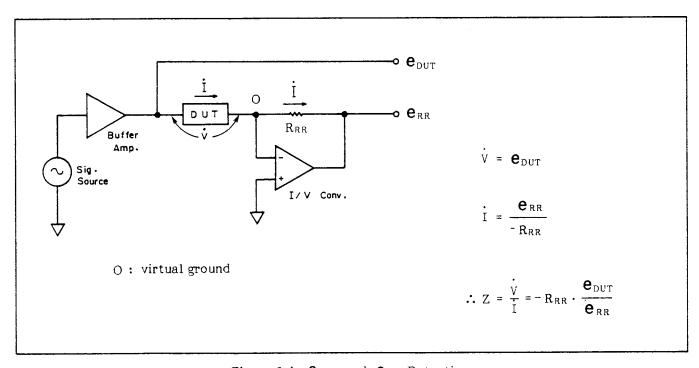


Figure 8-1.  $e_{\text{DUT}}$  and  $e_{\text{RR}}$  Detection.

Model 4276A SECTION VIII

8-16. The 4276A alternately detects  $\mathbf{e}_{\text{DUT}}$  and  $\mathbf{e}_{\text{RR}}$ , and converts them into their orthogonal (real and imaginary) voltage components.

$$\mathbf{e}_{\text{DUT}} = V_1 + jV_2$$
  
 $\mathbf{e}_{\text{RR}} = V_3 + jV_4$ 

To obtain R and X, the 4276A measures three voltage ratios:  $\alpha$ ,  $\beta$ , and  $\gamma$ . Refer to Table 8-2. R and X, the primary impedance parameters, are calculated from  $\alpha$ ,  $\beta$ , and  $\gamma$  using the equations listed in Table 8-3.

Table 8-2. Voltage Ratios

Voltage Ratio	e DUT / eRR Component
α	V <sub>4</sub> /V <sub>3</sub>
β	V <sub>2</sub> /V <sub>3</sub>
γ	V <sub>1</sub> /V <sub>2</sub> *

\* For ESR and G measurements,  $\gamma$  is  $V_1/V_3$ .

- 8-17. As long as the DUT's D value, which nearly equals  $\gamma$ , is less than 0.01, the product of  $\alpha$  and  $\gamma$  will be approximately zero,  $\alpha\gamma \simeq 0$ , because  $\alpha$  is always small. Thus, reactance, X, can be obtained by measuring  $\beta$  only. All three voltage ratios— $\alpha$ ,  $\beta$ , and  $\gamma$ —are measured when the 4276A is set to C, L, or |Z| measurement mode. When HIGH SPEED C or HIGH SPEED L is selected, however,  $\alpha$  and  $\gamma$  are not measured, thereby shortening the time required for measurement.
- 8-18. The voltage ratios are measured using dual-slope integration. Refer to Figure 8-3. The integrator is charged by voltage  $V_A$  for a constant time  $T_A$  (=25ms), and is then discharged by voltage  $V_B$ . The ratio  $V_A/V_B$  is obtained by measuring the time required to discharge the integrator. In this example, voltages  $V_A$  and  $V_B$  can be any of the orthogonal voltages of  $\mathbf{e}_{RR}$  or  $\mathbf{e}_{DUT}$ ; that is,  $V_1, V_2, V_3$ , or  $V_4$ .
- 8-19. When measurement is made on the high impedance ranges—those ranges enclosed by the bold line in Figure 8-4--the 4276A measures admittance parameters G and B instead of R and X in order to provide optimum accuracy.

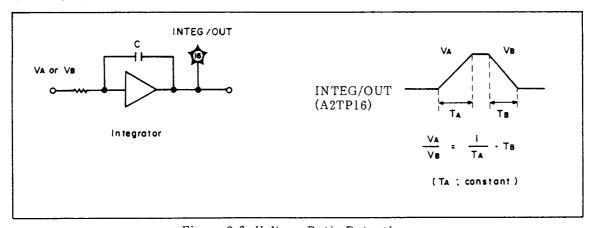


Figure 8-3. Voltage Ratio Detection

Table 8-3. Voltage Ratio Equations

DISPLAY A Function	DISPLAY B Function	Impedance Parameter	Voltage Ratio Equation
C, L	D, Q	X	- $R_{RR} \cdot \beta (1 - \alpha \gamma)/(1 + \alpha^2)$
σ, Ξ	D, Q		$(\alpha + \gamma)/(1 - \alpha\gamma)$
C, L,  Z	ESR, G, θ	Х	- $R_{RR} \cdot (\beta - \alpha \gamma)/(1 + \alpha^2)$
o, -, <sub> - </sub>		R	- $R_{RR} \cdot (\alpha \beta + \gamma)/(1 + \alpha^2)$

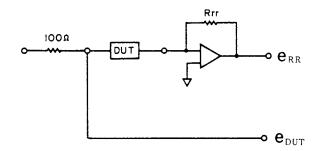
## $e_{\text{RR}}$ and $e_{\text{DUT}}$

 $\mathbf{e}_{\text{RR}}$  and  $\mathbf{e}_{\text{DUT}}$  are voltages across the range resistor and the DUT, respectively. Each voltage is calculated from the following equations:

$$\mathbf{e}_{\text{DUT}} = \frac{|Z|}{100 + |Z|} \cdot 1 \text{Vrms}$$

$$\mathbf{e}_{\text{RR}} = \frac{\mathbf{e}_{\text{DUT}}}{|Z|} \cdot R_{\text{RR}} = \frac{R_{\text{RR}}}{100 + |Z|} \cdot 1 \text{Vrms}$$

where, |Z| and  $R_{RR}$  are the DUT's impedance value and the range resistor value, respectively. It is obvious that both  $\mathbf{e}_{\text{DUT}}$  and  $\mathbf{e}_{\text{RR}}$  depend on the DUT's impedance value. The instrument amplifies the voltage which is smaller,  $\mathbf{e}_{\text{RR}}$  or  $\mathbf{e}_{\text{DUT}}$ , so as to input an appropriate level signal to the VRD. This is necessary because a small signal cannot be detected in the VRD with high accuracy and resolution.



DUT Impedance	<b>e</b> rr	<b>e</b> <sub>DUT</sub>
Ζ  100Ω	$\sim$	$\sim$
z  100Ω	$\sim$	

When the DUT's impedance is less than  $100\Omega$ ,  $\mathbf{e}_{\mathrm{DUT}}$  is smaller than  $\mathbf{e}_{\mathrm{RR}}$ , which is amplified by the AM1 and AM2 circuits in the Process Amplifier. Conversely, when the DUT's impedance is more than  $100\Omega$ ,  $\mathbf{e}_{\mathrm{RR}}$  is smaller than  $\mathbf{e}_{\mathrm{DUT}}$ , which is amplified in the range resistor circuit by changing the range resistor value, and also in the AM1 and AM2 circuits if necessary.

#### Z/Y Measurement Modes

When the measurement range is set to a range suitable for measuring a DUT whose impedance value is less than  $100\Omega$ , the instrument measures the impedance parameters R and X (Z measurement mode). When the measurement range is set to a higher range, however, the instrument measures the admittance parameters G and B (Y measurement mode). Refer to Figure A. In either measurement mode, the instrument calculates the selected measurement parameter, C, L, etc., from the measured parameter values.

#### Note

A DUT whose impedance is  $100\Omega\:\text{is}$  measured in Y measurement mode on some ranges.

Test Frequency (100Hz to 20kHz) C Range 200Hz 2kHz 10mF 1mF 100uF Z Measurement 10uF 1µF 100µF 10mF Y Measurement 1µF 100pF 10pF 1pF

Table A. Capacitance Measurement

Table B. Inductance Measurement

L Range	Test Frequency (100Hz to 20	kHz)
	1kHz 10kH	z
1 kH		
100H	Y Measurement  Z Measurement	
10H		
IH		
100mH		
10mH		
1 mH		
100µН		
10μН		

Figure 8-5. Z/Y Measurement Modes (Sheet 1 of 4).

Table C. Impedance Measurement Test Frequency Z Range 100Hz to 20kHz  $10 M\Omega$ Y Measurement Z Measurement  $1M\Omega$ (Y = G + jB)(Z = R + jX)Y Measurement  $100k\Omega$  $10k\Omega$ 100 Ω 0Ω 1k $\Omega$ DUT's Impedance Value  $100\Omega$ Figure A  $10\Omega$ Z Measurement  $1\Omega$  $100 \text{m}\Omega$ Rs = 100 a  $\mathbf{e}_{\text{DUT}} = \frac{Z}{100\Omega + Z} \text{ (Vrms)}$ IVrms O- $I = \frac{1 \text{Vrms}}{100\Omega + Z}$ - **e**rr - **e**<sub>DUT</sub>  $e_{DUT}$  $e_{\text{DUT}}$ 1 Ī i V rms IOm Arms Constant Constant Voitage Current

Figure 8-5. Z/Y Measurement Modes (Sheet 2 of 4).

1000

Figure B

OΩ

20MΩ

## (1) Z Measurement Mode

When the DUT's impedance value is lower than  $100\Omega$ , the current through the DUT, I in Figure B, is approximately constant. Thus, the voltage,  $\boldsymbol{e}_{RR}$ , across the range resistor, is also approximately constant because  $\boldsymbol{e}_{RR}$  is the product of  $R_{RR}$  and I.

In this case, impedance parameter measurement is more accurate than admittance parameter measurement. Therefore, the instrument measures R and X.

$$Z = \mathbf{e}_{DUT}/I = R_{RR} \cdot (\mathbf{e}_{DUT}/\mathbf{e}_{RR})$$

In D or Q measurement mode, the instrument measures D (= R/X) directly.

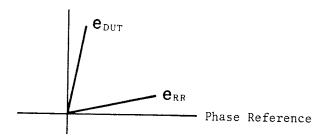


Table D. Measurement Parameters Relation in Z Mode

Measurement Functions	CKT MODE	DISPLAY A Value	DISPLAY B Value	Measured Parameters
L-D	<u> </u>	$L = X/2\pi f$	D = D	X, D
	<b>₽</b>	$L = X (1 + D^2)/2\pi f$	D = D	X, D
L-Q	• <del>□</del> ₩••	L = X/2πf	Q = 1/D	X, D
	<b>~₩</b>	$L = X (1 + D^2)/2\pi f$	Q = 1/D	X, D
L-ESR/G	<b>-□-</b> ₩••	$L = X/2\pi f$	ESR = R	X, R
	• <del></del>	$L = (X^2 + R^2)/2\pi f X$	$G = R/(R^2 + X^2)$	X, R
C-D	<b>⊶⊡-₩</b> •	$C = -1/2\pi f X$	D = -D	X, D
	<b>←</b>	$C = -1/2\pi f X (1 + D^2)$	D = -D	X, D
C-Q	• <del>□ •</del> ••	$C = -1/2\pi f X$	Q = -1/D	X, D
	<b>←</b>	$C = -1/2\pi f X (1 + D^2)$	Q = -1/D	X, D
C-ESR/G	• <del>□ ₩•</del>	$C = -1/2\pi fX$	ESR = R	X, R
,	<b>-₽</b>	$C = -X/2\pi f (R^2 + X^2)$	$G = R/(R^2 + X^2)$	X, R
Z  - 0	«□» •□•••	$ Z  = \sqrt{R^2 + X^2}$	$\theta = \tan^{-1} (X/R)$	X, R
<u> i</u>	~~~~			Í

Figure 8-5. Z/Y Measurement Modes (Sheet 3 of 4).

## (2) Y Measurement Mode

When the DUT's impedance value is higher than  $100\Omega$ , the voltage across the DUT,  $\boldsymbol{e}_{\text{DUT}}$ , is approximately constant, as shown in Figure B.

In this case, admittance parameter measurement is more accurate than impedance parameter measurement. Therefore, the instrument measures G and B.

$$Y = I/e_{DUT} = (1/R_{RR}) \cdot (e_{RR}/e_{DUT})$$

In D or Q measurement mode, the instrument measures D (= R/X) directly.

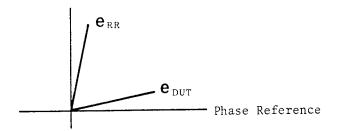


Table E Measurement Parameters Relation in Y Mode

Measurement Functions	CKT MODE	DISPLAY A Value DISPLAY B Value		Measured Parameters
	• <del>□</del> •••	$L = -1/2\pi fB (1 + D^2)$	D = D	B, D
L-D	٠٠.	L = -1/2πfB	D = D	B, D
1.0	• <del>□**</del> •	$L = -1/2\pi fB (1 + D^2)$	Q = 1/D	B, D
L-Q	٠ <del>٠</del>	$L = -1/2\pi fB$	Q = 1/D	B, D
I ECD/C	• <del>□</del> •••	$L = -B/2\pi f (G^2 + B^2)$	$ESR = G/(G^2 + B^2)$	B, G
L-ESR/G	-CD-	L = -1/2πfB	G = G	B, G
C D	• <del>□-</del> ₩•	$C = B (1 + D^2)/2\pi f$	D = D	B, D
C-D	• <del>C</del>	C = B/2πf	D = D	B, D
C 0	<b>-⊡-</b> ₩-•	$C = B (1 + D^2)/2\pi f$	Q = 1/D	B, D
C-Q	<b>₽</b>	$C = B/2\pi f$	Q = 1/D	B, D
C-ESR/G	<b>~□·</b> ~~	$C = (G^2 + B^2)/2\pi fB$	$ESR = G/(G^2 + B^2)$	B, G
G-ESR/G	<b>-₽</b>	$C = B/2\pi f$	G = G	B, G
Z  - 9	• <del>□ •</del> •	$ Z  = 1/\sqrt{G^2 + B^2}$	$\theta = \tan^{-1} (-B/G)$	B, G
111	•€ <b>₩</b> >•	1-1 1/ V G . 3	5 52. (5/6)	2, 0

Figure 8-5. Z/Y Measurement Modes (Sheet 4 of 4).

#### eref and etest

The instrument measures the ratio between two vector voltages:  $\mathbf{e}_{\text{DUT}}$  (voltage across the DUT) and  $\mathbf{e}_{\text{RR}}$  (voltage across the range resistor). In Z measurement mode, the vector voltage ratio is  $\mathbf{e}_{\text{DUT}}/\mathbf{e}_{\text{RR}}$ , and in Y measurement mode,  $\mathbf{e}_{\text{RR}}/\mathbf{e}_{\text{DUT}}$ .

 $\mathbf{e}_{\text{test}}$  and  $\mathbf{e}_{\text{ref}}$  are defined as the numerator and the denominator of the detected ratio, respectively. The relations among  $\mathbf{e}_{\text{test}}$ ,  $\mathbf{e}_{\text{ref}}$ ,  $\mathbf{e}_{\text{DUT}}$ , and  $\mathbf{e}_{\text{RR}}$  are described in Table A.

**e**<sub>ref</sub> Phase Reference

Measurement Mode	<b>e</b> <sub>test</sub>	<b>e</b> ref
Z	<b>e</b> <sub>DUT</sub>	<b>e</b> <sub>RR</sub>
Y	<b>e</b> rr	<b>e</b> <sub>DUT</sub>

Table A.

Figure 8-6. e ref and etest.

## HIGH SPEED C/L MEASUREMENT PRINCIPLE

The instrument measures the selected impedance parameters—C, L, D, Q, etc.—by detecting three voltage ratios— $\alpha$ ,  $\beta$ , and  $\gamma$ —and calculating each parameter from the equations listed in Table A of Figure 8-7. Voltage ratio  $\alpha$  is always close to zero because the phase reference signal used by the phase detector is nearly in-phase with or 90° out of phase with  $\mathbf{e}_{\text{ref}}$ . Also, when  $\gamma$  is much smaller than 1, only  $\beta$  is necessary to obtain C or L. Therefore, if the dissipation factor, D, of the DUT is much lower than

l, voltage ratios  $\alpha$  and  $\gamma$  need not be detected to obtain C or L, significantly reducing total measurement time. In HIGH SPEED C or HIGH SPEED L measurement mode, the instrument detects only  $\beta$ .

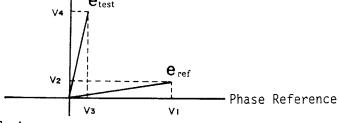


Table A.

Measurement	Measured	Parameter Equation	Vc	ltage Rati	LO
Mode	Parameters	ratameter Equation	α	β	Υ
	X, D	$X = -R_{RR} \cdot \beta (1 - \alpha \gamma) / (1 + \alpha^2)$	,	V4/V1	N /N
Z	Α, Β	$D = (\alpha + \gamma)/(1 - \alpha\gamma)$	V <sub>2</sub> /V <sub>1</sub>		V3/V4
-	X, R	$X = -\bar{R}_{RR} \cdot (\beta - \alpha \gamma) / (1 + \alpha^2)$	N /N	V4/V1	V <sub>3</sub> /V <sub>1</sub>
	Α, Κ	$R = -R_{RR} \cdot (\alpha \beta + \gamma) / (1 + \alpha^2)$	V <sub>2</sub> /V <sub>1</sub>	V4/V1	v3/ v1
	B, D	$B = -G_{RR} \cdot \beta (1 - \alpha \gamma) / (1 + \alpha^2)$		/	77 /77
γ , υ	Β, Β	$D = -(\alpha + \gamma)/(1 - \alpha\gamma)$	$V_2/V_1$	V <sub>4</sub> /V <sub>1</sub>	V3/V4
B, G	$B = -G_{RR} \cdot (\beta - \alpha \gamma) / (1 + \alpha^2)$	, , , , , , , , , , , , , , , , , , , ,	/* .	/	
	2, 0	$G = -G_{RR} \cdot (\alpha \beta + \gamma)/(1 + \alpha^2)$	$V_2/V_1$	V <sub>4</sub> /V <sub>1</sub>	V <sub>3</sub> /V <sub>1</sub>

 $(\dot{Y} = 1/\dot{Z}, \dot{Z} = R + jX, \dot{Y} = B + jX, D = R/X, D = -G/B)$ 

Figure 8-7. HIGH SPEED C/L Measurement Principle

## 8-20. ANALOG SECTION BLOCK LEVEL THEORY

8-21. The following paragraphs describe the structure and operation of the 4276A's Analog Section. The Analog Section consists of the Signal Source, the Transducer, the Process Amplifier, and the Vector Ratio Detector (VRD). The block diagram of the Analog Section is shown in Figure 8-8. The simplified block diagram of the Analog Section is shown in Figure 8-8.

## 8-22. SIGNAL SOURCE

8-23. The block diagram of the Signal Source is shown in Figure 8-9. The Signal Source consists of a crystal oscillator, a phase-locked loop (PLL), a quasi-sinewave oscillator, low-pass filters, and an attenuator.

8-24. The crystal oscillator (located on the Alboard) outputs a precise 11.5200MHz signal which is divided down to provide an 8kHz reference signal for the phase detector in the PLL, a 5.760MHz clock signal for the Z80 microprocessor, and other clock signals for the various digital operations performed by the instrument.

8-25. The voltage-controlled oscillator (VCO) in the PLL outputs an 8MHz to 20MHz signal, which is divided down to a 16F signal for test frequency generation. The output frequency of the VCO is determined by the phase detector, the loop filter, and the  $\div$ N divider. If the frequency of the signal output from the  $\div$ N divider is different from the 8kHz of the reference signal from the crystal oscillator, the phase detector will output an "unlock" signal to the loop filter, which will then adjust the VCO

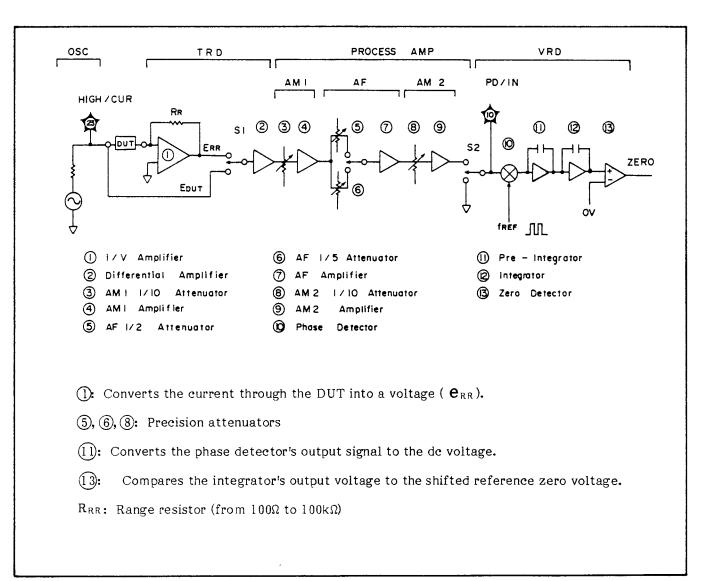


Figure 8-8 . Simplified Analog Section Block Diagram.

control voltage up or down until the PLL is locked. The VCO control voltage is between -10V and +10V; the N divisor is from 100 to 250. The PLL is locked when NFVCO is 8kHz.

8-26. The quasi-sinewave oscillator generates a digital sinewave whose fundamental frequency is the same as the test frequency. This oscillator consists of a 16-bit counter, a ROM, and a DA converter. The 16-bit counter counts the pulses of the 16F signal sent from the PLL. The counter's outputs are connected to the address-select lines of the ROM. The ROM contains digital data which determines the output from the DA converter. The digital data and the addresses are arranged so that, as the ROM is addressed by the 16-bit counter, the output from the DA converter will be a rising

and falling staircase waveform. The DA converter's output is filtered by a low-pass filter, leaving a clean lVrms sinewave. Signal level is controlled by an attenuator and a buffer amplifier. When HIGH SIG LEVEL is selected on the front panel, the lVrms signal is not attenuated; when LOW SIG LEVEL is selected, the test signal level is attenuated to 50mVrms. On certain measurement ranges, the lVrms signal level is amplified to 2Vrms.

8-27. The source resistor, Rs in Figure 8-9 , is a  $47\Omega$  fuse resistor which protects the instrument from damage if a charged capacitor is connected to the UNKNOWN terminals.

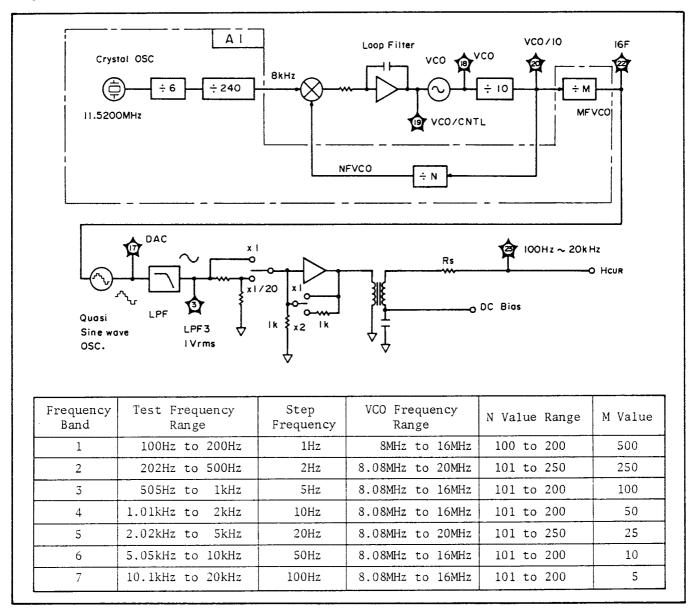


Figure 8-9 . Signal Source Block Diagram.

## Quasi-Sinewave Oscillator

The quasi-sinewave oscillator consists of a modulo-16 counter, a ROM, and a digital-to-analog converter, as shown in Figure A. The oscillator outputs a digital sinewave, which, once filtered, becomes the instrument's test signal. The circuit works as follows.

The counter, A2U32, counts the pulses of a 16F squarewave sent from the phase-locked loop. The counter's outputs are connected directly to the ROM's lower-four address control lines. As the counts changes, a different address is selected and the binary number at the selected address is output to the digital-to-analog converter. Output from the digital-to-analog converter is a dc voltage proportional to the binary number output from the ROM. The address sequence and the data stored in the ROM are such that the output from the digital-to-analog resembles a sine wave. The relationships among the ROM addresses, stored data, and output voltage are listed in Table A. Figure B shows the timing diagram of the circuit.

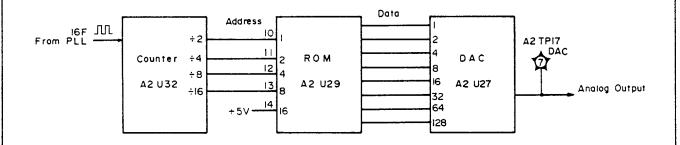


Figure A. Quasi-Sinewave Oscillator Circuit.

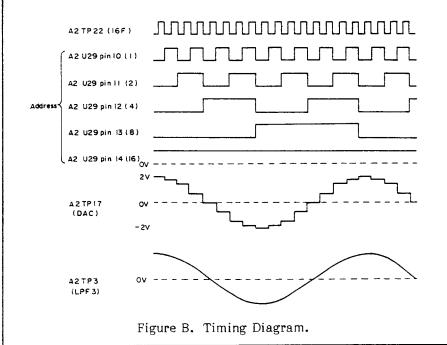


Table A. ROM Data

Data	A2TP17
255	2V
245	1.92V
218	1.71V
176	1.38V
128	ov
79	-1.38V
38	-1.71V
10	-1.92V
0	- 2V
10	-1.92V
38	-1.71V
79	-1.38V
128	ov
176	1.38V
218	1.71V
245	1.92V
	255 245 218 176 128 79 38 10 0 10 38 79 128 176 218

Figure 8-10. Quasi-Sinewave Oscillator.

Model 4276A SECTION VIII

## 8-28. TRANSDUCER

8-29. A simplified schematic Transducer is shown in Figure 8-11. The transducer detects two vector voltage, eDUT and  $\boldsymbol{e}_{\text{RR}}$  , and alternately selects each for output to the voltage ratio detector (VRD) through the process amplifier. e DUT is the voltage across the DUT. erR is the voltage across the range resistor in the feedback loop of the I/V converter amplifier. Since the current flowing through the DUT must also flow through the range resistor,  ${f e}_{RR}$  is proportional to the current. There are four range resistors:  $100\Omega$ ,  $1k\Omega$ ,  $10k\Omega$ , and  $100k\Omega$ . Measurement range is determined by the range resistor and the gain of the AM circuit in the process amplifier.  $\mathbf{e}_{\text{DUT}}$  and  $\mathbf{e}_{\text{RR}}$  are both detected by a precision differential amplifier.

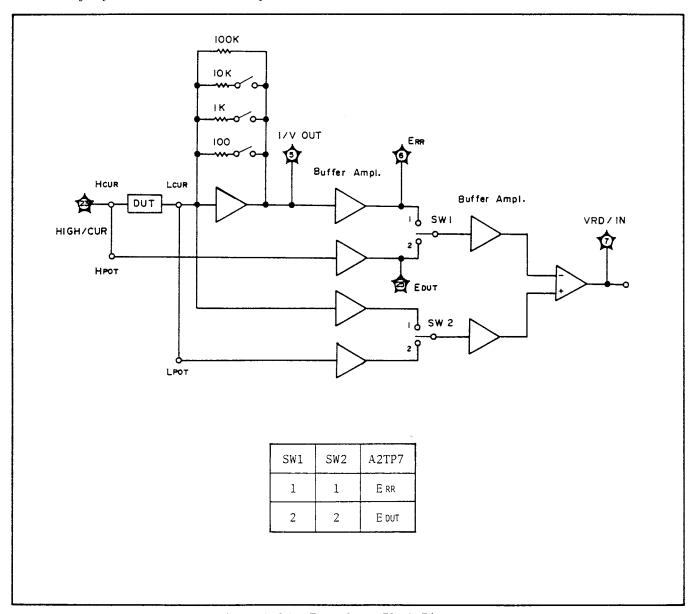


Figure 8-11. Transducer Block Diagram.

## 8-30. PROCESS AMPLIFIER

8-31. When the DUT's impedance is much lower than 100 $\Omega$ , the level of the  $e_{\text{DUT}}$  signal is much lower than that of the  $e_{\text{RR}}$  signal (  $\simeq$  1Vrms). When the DUT's impedance is much higher than 100 $\Omega$ , the level of the  $e_{\text{RR}}$  signal is much lower than that of the  $e_{\text{DUT}}$  signal ( $\simeq$ 1Vrms).

The process amplifier controls the signal levels of  $\mathbf{e}_{\text{DUT}}$  and  $\mathbf{e}_{\text{RR}}$  so that they are roughly equal; thereby improving resolution in the VRD section. The process amplifier consists of three stages--AM1, AF, and AM2--as shown in Figure 8-12. The AM1 and AM2 stages compensate for signal level differences between  $\mathbf{e}_{\text{DUT}}$  and  $\mathbf{e}_{\text{RR}}$  caused by measurement range selection. The AF stage compensates for signal level differences caused by test frequency selection. Each stage contains an amplifier to roughly magnify the signal level and an attenuator to precisely attenuate the signal. In one measurement cycle,

the amplifiers magnify both  $\mathbf{e}_{\text{DUT}}$  and  $\mathbf{e}_{\text{RR}}$  by the same gain factor. On the other hand, however, the attenuators attenuate  $\mathbf{e}_{\text{DUT}}$  and  $\mathbf{e}_{\text{RR}}$  by different attenuation factors, depending on the DUT's impedance. Therefore, the difference of amplitudes for  $\mathbf{e}_{\text{DUT}}$  and  $\mathbf{e}_{\text{RR}}$  is determined by the precise attenuators contained in AM1/AF/AM2 circuits. It is important that the common amplification of  $\mathbf{e}_{\text{DUT}}$  and  $\mathbf{e}_{\text{RR}}$  have no effect on voltage detection because the instrument detects the voltage ratios of the quadrature components of  $\mathbf{e}_{\text{DUT}}$  and  $\mathbf{e}_{\text{RR}}$  in the VRD section.

8-32. Figures 8-13 and 8-14 show the AM and the AF controls, respectively.

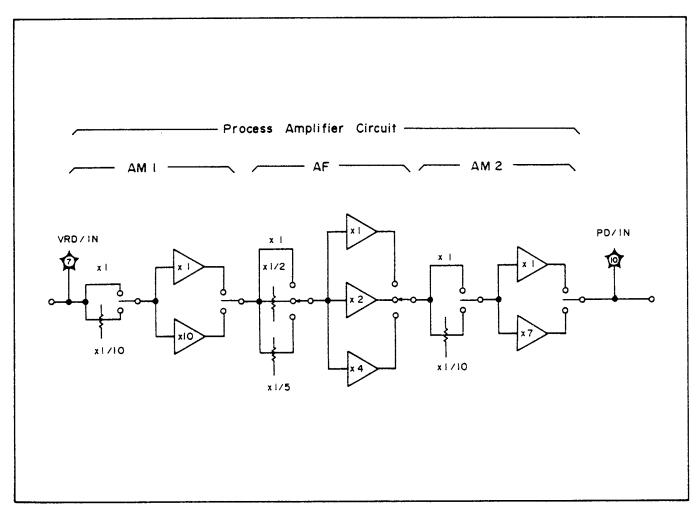


Figure 8-12. Process Amplifier Block Diagram

Model 4276A

Table AM Control (L Measurement) Table AM Control (Z Measurement)						
L Range	Test Frequency Range			Z Range	Test Frequency Range	
	100Hz to 995Hz	1kHz to 9.95kHz	10kHz to 20kHz		100Hz to 20kHz	
1kH				10ΜΩ	Y, 100kΩ, M3	
100H		Y, 100kΩ, M2		1ΜΩ	Υ, 100kΩ, M2	
10H		Y, 10kΩ, M2		100kΩ	Y, 10kΩ, M2	
1H		Y, 1kΩ, M2		10kΩ	Υ, 1kΩ, M2	
100mH		Y, 100Ω, M2		1kΩ	Υ, 100Ω, Μ2	
10mH		Z, 100Ω, M2		100Ω	Z, 100Ω, M2	
l mH		Z, 100Ω, M3		10Ω	Z, 100Ω, M3	
100µН				1Ω	Z, 100Ω, M4	

Table AM Control (C Measurement)

			,				
	Test Frequency Range						
C Range	100Hz to 199Hz	200Hz to 1.99kHz	2kHz to 20kHz				
10mF			-				
1mF		Ζ, 100Ω, ΜΟ					
100µF		Z, 100Ω, M1					
10μF		Z, 100Ω, M2					
lμF		Υ, 100Ω, Μ2					
100nF		Y, 1kΩ, M2					
10nF		Υ, 10kΩ, M2	}				
1nF	Y, 100kΩ, M3	Y, 100kΩ, M2	\				
100pF		Y, 100kΩ, M3					
10pF			Y, 100kΩ, M3				
		500Hz	5kHz				

500Hz

Note: Each description represents

Measurement, Range Resistor Value, AM Code

Table AM Gain (Gt) (Gr) Gain Ratio AM1 AM Code AM2 AM1 AM2 (Gt/Gr) ATT AMP ATT AMP ATT AMP ATT AMP x 1 x 1 x 1 x 10 x 1/10 x 1 x 1/10 100 MO x 10 x 1 M1x 1/10x 1 x 10 x 1/10x 1x 1/10 x 10 10 M2 x 1 x 1 x 1 x 1 x 1 x 1 x 1 x 1 1 М3 x 1 x 1 x 1 x 10 x 1 x 1x 1/10x 10 10 x 10 x 1 x 1/10M4 x l x 10 x 1/10x 10 x 10 100 M5 x 1 x 1 x 1/10x 10 x 1 x 1 x 1 x 10 x 1/10 x 1 x 1 M6 x 10 x 1 x 1 x 1 x 10 . 1 x 1 M7 x l x 1 x 1/10 x 1 x 1 x 1/101 x 1

Figure 8-13. AM Control.

## Table AF Control

Test Frequency	C Measurement		L Measurement		Z Measurement	
Range [Hz]	Z Mode	Y Mode	Z Mode	Y Mode	Z Mode	Y Mode
100 to 199	F0	F0	F0	F0		
200 to 498	F5	F2	F4	F1		
500 to 995	F4	F1	F5	F2		
lk to 1.99k	F0	F0	F0	F0	F0	F0
2k to 4.98k	F5	F2	F4	F1		
5k to 9.95k	F4	F1	F5	F2		
10k to 20k	F0	F0	F0	F0		

Table AF Gain

	(Gt)		(Gr)		Gain Ratio
AF Code	ATT	AMP	ATT	AMP	(Gt/Gr)
F0	x 1	x 1	x 1	x 1	1
F1	x 1	x 2	x 1/2	x 2	2
F2	x 1	x 4	x 1/5	x 4	5
F3	x 1	x 4	x 1	x 4	1
F4	x 1/2	x 1	x 1	x 1	1/2
F5	x 1/5	x 1	x 1	x 1	1/5
F6	x 1/5	x 2	x 1/2	x 2	2/5
<b>F</b> 7	x 1	x 1	x 1/5	x 1	5

Figure 8-14. AF Control.

Model 4276A SECTION VIII

## 8-33. VECTOR RATIO DETECTOR (VRD)

8-34. A simplified circuit diagram of the VRD is shown in Figure 8-15. The VRD consists of a phase detector, a pre-integrator, an integrator, and a zero detector. The VRD's function is to measure the three voltage ratios— $\alpha$ ,  $\beta$ , and  $\gamma$ —from which all measurement parameters are derived.

8--35. The phase detector is switched by two signals—  $e_{\text{ref}}$  (0°) and  $e_{\text{ref}}$  (90°)—which convert the input signals—  $e_{\text{ref}}$  and  $e_{\text{test}}$ —into their orthogonal components. Thus, the phase detector outputs four voltages:  $e_{\text{ref}}$  (0°),  $e_{\text{ref}}$  (90°),  $e_{\text{test}}$  (0°), and  $e_{\text{test}}$  (90°). The pre-integrator integrates the phase detector output voltage for 5ms at test frequencies of 200Hz and above, or for one test signal period up to 10ms at frequencies below 200Hz. The de

voltage output from the pre-integrator charges the integrator for 5ms. The next dc voltage output from the pre-integrator discharges the integrator. When the integrator is completely discharged, the output from the zero detector goes HIGH (+5V) or LOW (0V) depending on the polarity of the integrator output voltage. During the discharge period, a counter on the Al board counts the pulses of a 3.84MHz clock signal. The number of pulses counted represents the ratio of the charge and discharge voltages.

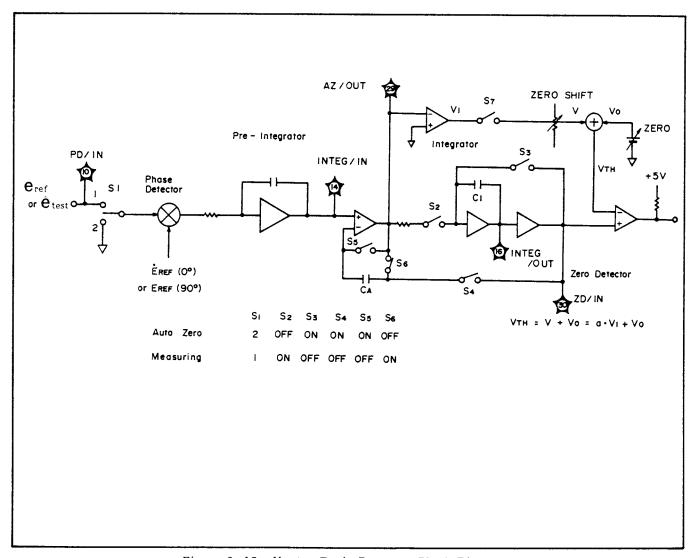


Figure 8-15. Vector Ratio Detector Block Diagram.

## Reference Detection Signal (REFDET)

The phase relationship between  $\mathbf{e}_{\text{ref}}$  and the main phase reference signal, X, must be a precise 0° when the real vector components of  $\mathbf{e}_{\text{ref}}$  and  $\mathbf{e}_{\text{test}}$  are measured. When the imaginary vector components of  $\mathbf{e}_{\text{ref}}$  and  $\mathbf{e}_{\text{test}}$  are measured, however, X must be phase shifted 90° in reference to  $\mathbf{e}_{\text{ref}}$ . Refer to Figure A.

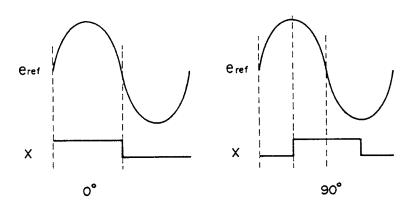


Figure A.

To establish the 0° phase relationship, the instrument detects the negative-to-positive zero crossover of the  $\mathbf{e}_{\text{ref}}$  signal. Refer to Figure B. At the zero crossover, the REFDET signal goes HIGH, starting the 8flck signal and enabling three shift registers—A2U62, U63, and U64. These shift registers generate the X, Y, and Z phase reference signals shown in Figure C. To shift the X phase reference signal 90° in reference to  $\mathbf{e}_{\text{ref}}$ , the instrument adds two narrow pulses to the 8flck signal, as shown in Figure D. The phase relationships among X, Y, and Z are constant.

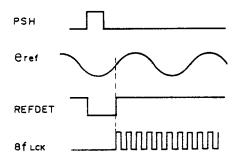


Figure B

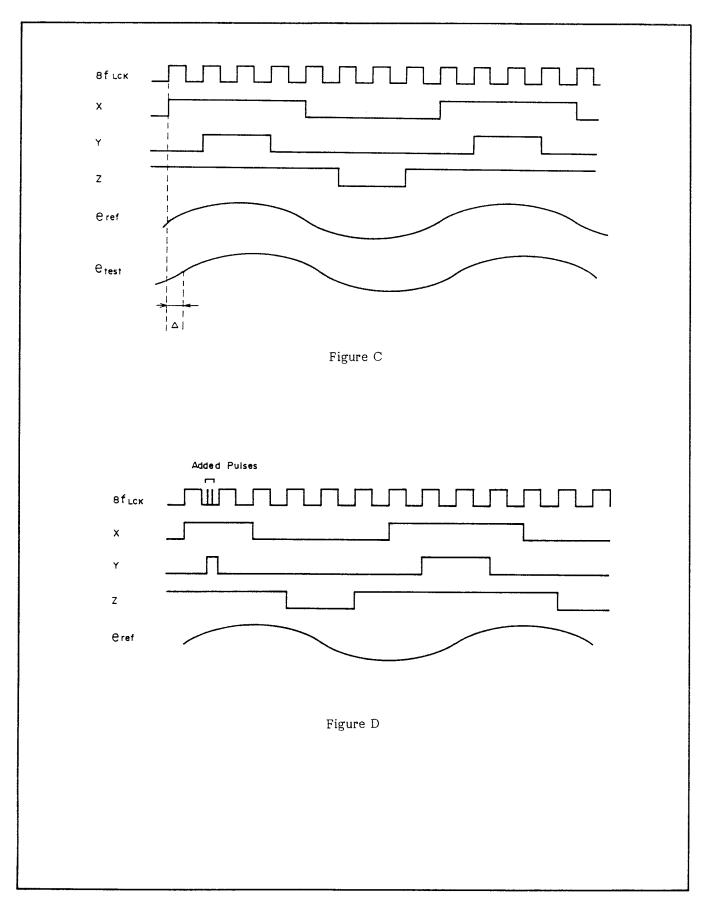


Figure 8-16. REFDET (Sheet 2 of 2).

#### Phase Detector

The phase detector is basically an analog switch controlled by a square wave signal. Refer to Figure A. The square wave is the phase reference signal, and it is either in-phase with or 90° out of phase with the  $\mathbf{e}_{\text{ref}}$  signal. When the phase reference signal is HIGH, the signal applied to the INPUT- terminal is selected for output; when the phase reference signal is LOW, the signal applied to the INPUT+ terminal is selected for output. The signals applied to the INPUT+ and INPUT- terminals are  $\mathbf{e}_{\text{ref}}$  (or  $\mathbf{e}_{\text{test}}$ ) and an inverted (shifted 180°)  $\mathbf{e}_{\text{ref}}$  (or  $\mathbf{e}_{\text{test}}$ ), respectively.

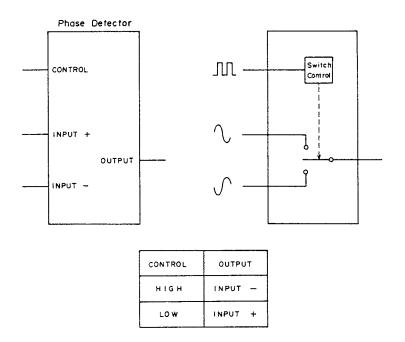


Figure A.

The function of the phase detector is to resolve  $e_{ref}$  and  $e_{test}$  into their orthogonal (real and imaginary) components. See Figure B.

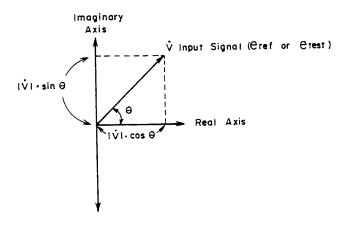


Figure B.

Figure 8-17. Phase Detector (Sheet 1 of 2)

To obtain the real component of  $e_{\text{ref}}$  or  $e_{\text{test}}$ , the phase reference signal, which controls switching of the phase detector, must be in-phase with  $e_{\text{ref}}$  (see Figure C). Similarly, to obtain the imaginary component of  $e_{\text{ref}}$  or  $e_{\text{test}}$ , the phase reference signal must be 90° out of phase with  $e_{\text{ref}}$  (see Figure D).

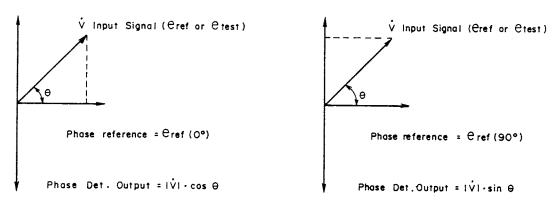


Figure C.

To minimize phase detection errors caused by low order odd harmonics which may be present at the output, three parallel-connected phase detectors (A2U59, U60, and U61) are used in the 4276A, as shown in the simplified circuit schematic of Figure E. Figure F shows the phase reference signals for the three phase detectors.

Figure D.

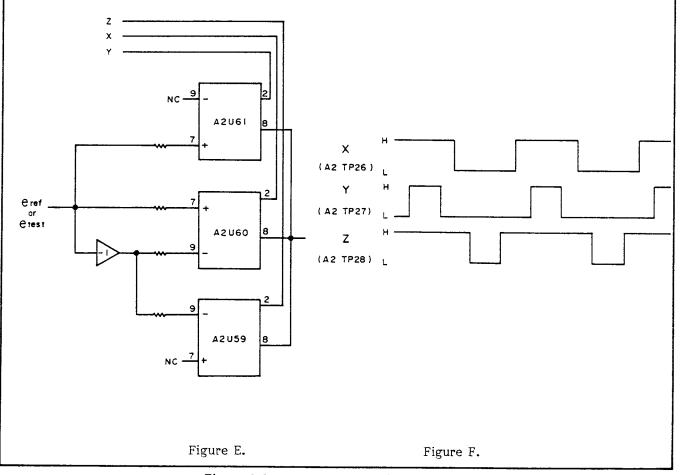


Figure 8-17. Phase Detector (Sheet 2 of 2)

#### Phase Detection

The phase detector is basically a network of analog switches which are controlled by the square waves shown in Figure A. When the switches are closed, the current through each switch charges the integrator. The integrator output voltage is given as:

$$V_{OUT} = \frac{6a}{CR} \cos \Delta \propto |V_{IN}| \cdot \cos \Delta$$
 (\*\*  $|V_{IN}| = a$ )

When the control signals X, Y, and Z are phase-shifted by 90 degrees, however, the integrator output voltage is as:

$$V_{OUT} = \frac{6a}{CR} \sin \Delta \propto |V_{IN}| \cdot \sin \Delta$$

 $|V_{IN}|\cos\Delta$  and  $|V_{IN}|\sin\Delta$  are the in-phase and the 90° out of phase component of the input vector voltage  $V_{IN}$  with X,respectively, where X is the phase reference signal.

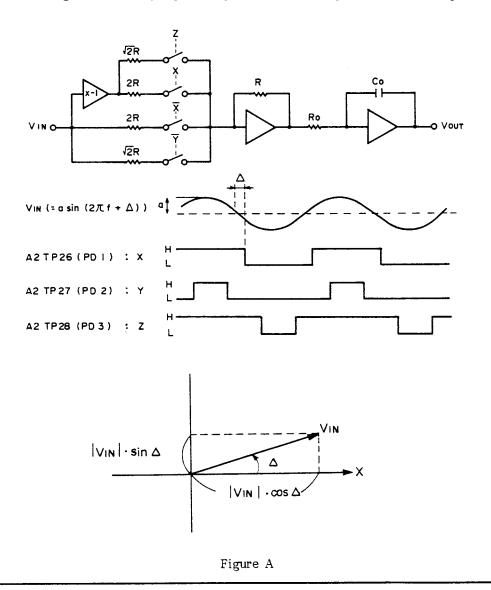


Figure 8-18. Phase Detector (Sheet 2 of 2)

ATU16 Output

- TCLK ----+5V

#### Voltage Ratio Detection

The voltage ratio is easily measured by an integrator and a counter. Refer to Figure A. The integrator is first charged by voltage Ea for a known time,  $T_0$  (determined by the instrument). It is then discharged by voltage Eb. At the start of the discharge cycle, a counter (not shown) is enabled and begins counting the pulses of a reference clock signal. The counter is disabled when the integrator's output reaches 0V. The number of pulses counted by the counter represents the integrator discharge time, T. The voltage ratio Ea/Eb is represented by the following equation:

$$Ea/Eb = (1/T_0) \cdot T \quad (: Ea \cdot T_0 = Eb \cdot T)$$

where  $T_{\text{0}}$  is a known constant.

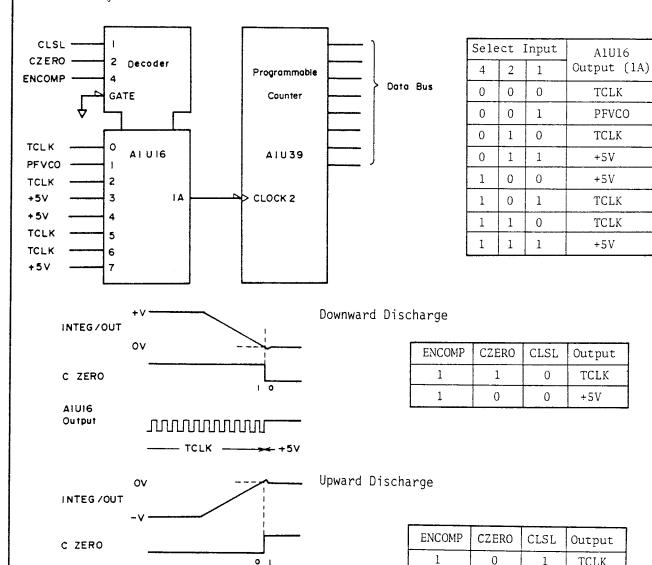


Figure 8-19. Voltage Ratio Detection.

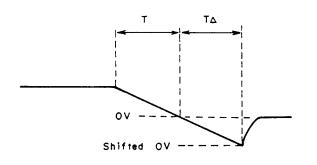
Figure A

1

TCLK +5V

#### ZERO SHIFT

The zero reference input to the Zero Detector (A2U68) is slightly shifted from zero volts so as to measure the discharge time of the Integrator (A2U65) accurately even when the discharge time is very short. The additional time required for the zero shift operation, T, is compensated for in the digital section.



Integrator Output Voltage

Figure 8-20. ZERO SHIFT.

#### 8-36. DIGITAL SECTION THEORY

8-37. Digital section block diagram is shown in Figure 8-30. A simplified digital section block diagram, including an analog section block diagram, is shown in Figure 8-21.

8-38. Overall instrument operation is controlled by a high speed Z80 microprocessor driven by a 5.76MHz clock. AlU22 controls selection of the required ROM (U5 through U11)

by decoding four address lines—Al2 through Al5—into seven ROM gate signals—ROMG1 through ROMG7. Ull and ROM gate signal ROMG7 are not used in normal operation. The correspondence between address lines Al2 through Al5 and ROMS U5 through Ul0 is given in Table 8-4. Addressing of data stored in the selected ROM is handled by the remaining address lines—A0 through Al1. Data read from the selected ROM is sent to the microprocessor via data bus lines D0 through D7. The

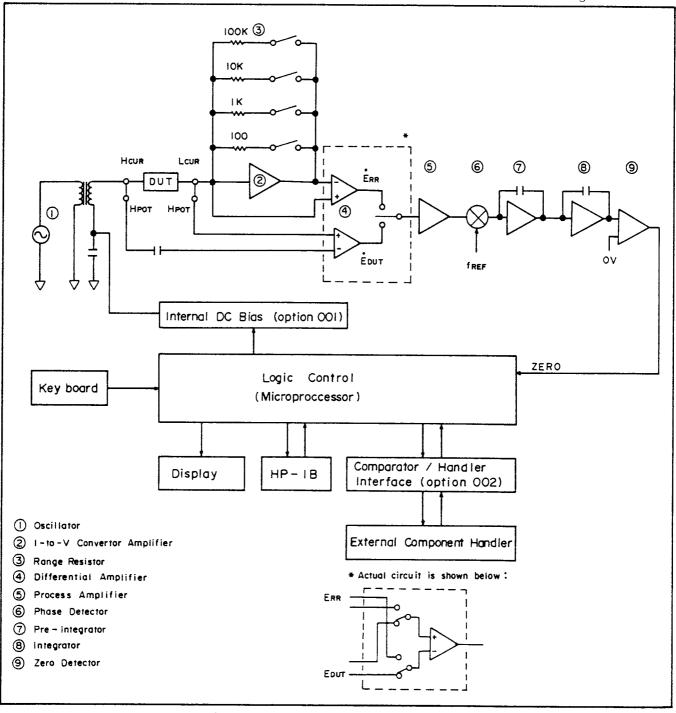


Figure 8-21. Digital Section Block Diagram.

microprocessor operates in accordance with the instructions and data stored in the ROMs.

Alu35 provides five signallines—ANACTL1 through ANACTL5—which control the data latches in the analog circuits. Similarly, Alu41 provides six signal lines—IOEN0 through IOEN5—which control data transfer to and from other boards via the data bus. For example, when the IOEN0 line is LOW, data is transmitted between the microprocessor and the HP-IB circuits on the A21 board. Refer to Table 8-5 for the correspondence between address lines A3 through A5 and I/O lines IOEN0 through IOEN5.

8-39. Operation of the microprocessor is interrupted by any one of three interrupt signals: IBINT, TRIGINT and KEYINT. The IBINT line is active (LOW) when an interrupt request is on the HP-IB: TRIGINT is active (LOW) when the instrument is externally triggered; KEYINT is active (HIGH) when a key on the front panel is pressed. These interrupts are detected at the beginning of a measurement cycle or before each voltage-ratio measurement period.

8-40. The microprocessor, the HP-IB circuits on the A21 board, and the data latches on the A22 board (option 001) are reset each time the RESET signal goes LOW. RESET goes LOW if the +5V supply on the A1 board drops below +4.8V. Refer to Figure 8-22. The active (LOW) time for the RESET signal is approximately 500ms.

8-41. Perhaps the most important function of the section is to measure the time required for the main integrator on the A2 board to discharge during each voltage-ratio measurement. This is done by a counter, A1U39. At the start of the integrator discharge period, A1U39 is enabled and begins counting the pulses of a 3.84MHz clock signal (TCLK) output from A1U16. When the integrator is completely discharged (output reaches zero volts), the ZERO DETECTOR on

the A2 board sends the ZERO signal, which stops the clock. The time required for the integrator to discharge is indicated by the number in A1U39. See Figures 8-19 and 8-20.

8-42. The Al board contains an ll.5200MHz crystal oscillator, AlUl. Output from the oscillator is counted down to provide the 5.76MHz clock signal for the microprocessor, the 3.84MHz clock signal for voltage-ratio measurement, and the 8kHz reference signal for the phase-locked loop on the A2 board.

8-43. Figure 8-23 shows the flow diagram for the measurement sequence.

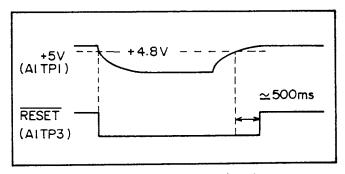


Figure 8-22. Reset Signal.

Table 8-5. I/O Enable Signals

Address Lines			
5	4	3	I/O Enable Signal
0	0	0	IOENO
0	0	1	IOEN1
0	1	0	IOEN2
0	1	1	IOEN3
1	0	0	IOEN4
1	0	1	IOEN5

Table 8-4. ROM Addresses

	Address Lines															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0	Addressed ROM
0	0	0	0	-	-	-	-	-	-	-	-	-	-	-	-	ROMO (A1U5)
0	0	0	1	-	-	-	-	-	-	-	-	-	-	-	-	ROM1 (A1U6)
0	0	1	0	-	-	-	-	-	-	-	-	-		-	-	ROM2 (A1U7)
0	0	1	1	-	-	-	-	-	-	-	-	-	-	-	-	ROM3 (AlU8)
0	1	0	0	-	-	-	-	-	-	-	-	-	-	-	-	ROM4 (A1U9)
0	1	0	1	-	-	-	-	-	-	-	-	-	-	-	_	ROM5 (A1U10)

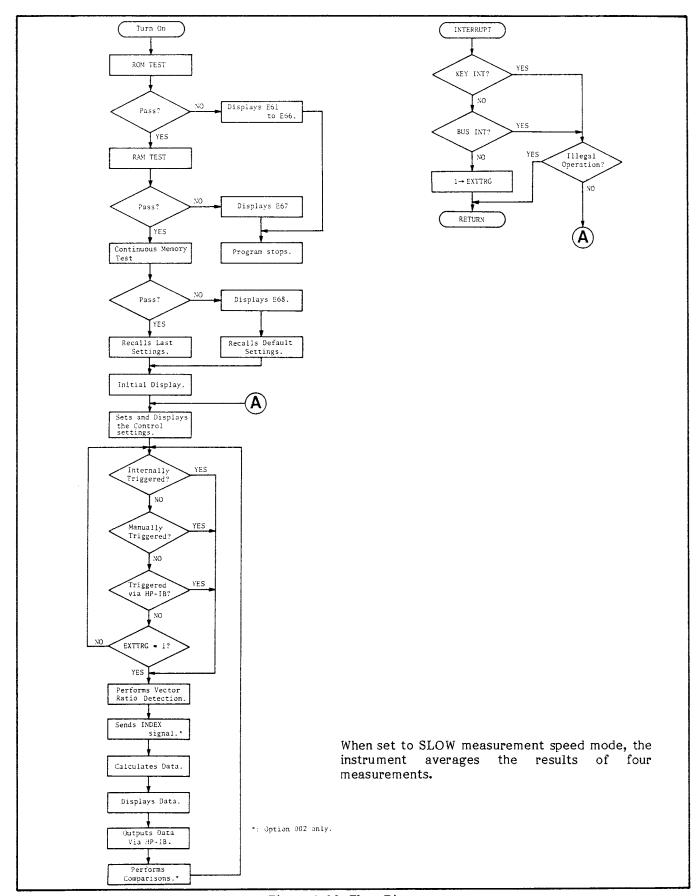


Figure 8-23. Flow Diagram.

#### 8-44. OPTIONS

8-45. The theory of operation for the 4276A's optional circuits is outlined in the following paragraphs.

8-46. OPTION 001 INTERNAL DC BIAS (A22)

8-47. The A22 board primarily contains a DAC (A22U3) and an output amplifier, as shown in Figure 8-24. The DAC outputs a dc voltage whose polarity and magnitude are determined by the reference voltage,  $V_{\text{ref}}$ , and the digital is determined by the digital data sent from the microprocessor and stored in latches Ul and U2. Output voltage is calculated as follows:

Vour = 
$$-Vref \cdot \sum_{n=1}^{10} Bn \cdot 2^{-n}$$
(Bn: 0 or 1)

where V ref is determined as follows:

Internal DC Bias Voltage Range	Vref
40.0V to .00V	-6.3V
01V to -40.0V	+6.3V

The output voltage of DAC is shown in Figure A.

 $V_{\rm ref}$  is switched by an analog switch controlled by Data Bus lines BD6 and BD7 via the latch A22U2. The two latches, A22U1 and U2, are successively enabled by clock signals  $\overline{\rm IOEN_1}$  and  $\overline{\rm IOEN_2}$  to output digital data to the DAC and to control the analog switches. Another analog switch selects the attenuation factor — xl or x1/5 — in accordance with the internal dc bias voltage setting as follows:

Internal DC Bias Voltage Range	Attenuator
±(.00V to 9.99V)	x 1/5
±(10.0V to 40.0V)	x 1

The DAC output voltage shown in Figure A is attenuated by a xl or xl/5 attenuator to obtain the linear characteristic shown in Figure B. This attenuated voltage is amplified by the x8 output amplifier.

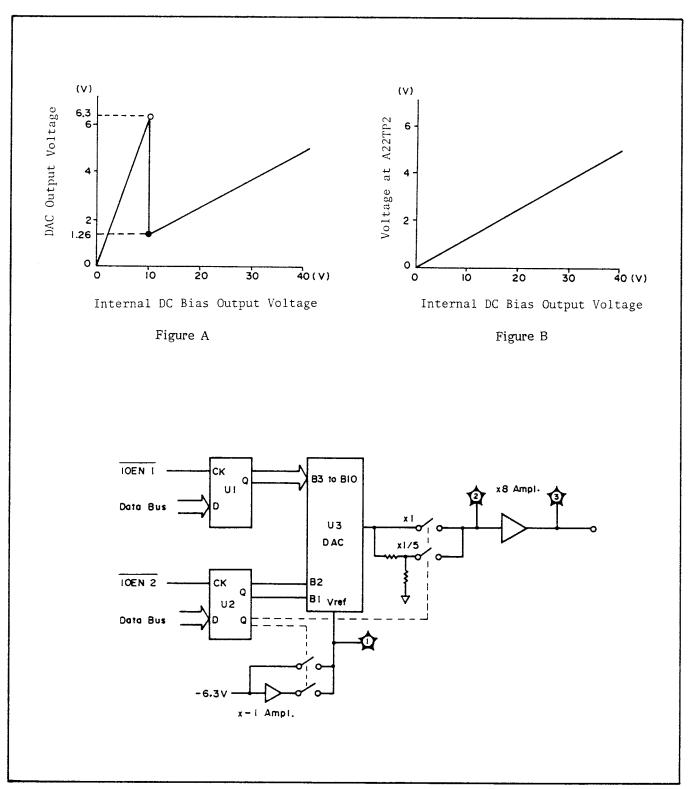


Figure 8-24. A22 Board Block Diagram.

#### 8-48. TIMING DIAGRAM DISCUSSION

8-49. Figure 8-26 shows the timing diagram for the various signals necessary for VRD operation. In the figure, the REFDET signal goes HIGH when the  $\mathbf{e}_{\rm ref}$  signal crosses zero volts in a negative voltage-to-positive voltage direction after the PSH signal goes LOW. The REFDET signal provides the start timing for the 8F , which is the source of the phase reference signals used by the phase detector. Refer to Figure 8-25.

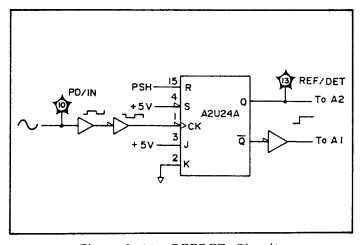


Figure 8-25. REFDET Circuit.

8-50. In each voltage ratio detection period — , , and — the integrator charge time,  $T_0$ , is constant for the selected measurement speed, as listed in Table 8-6.

Table 8-6. Integrator Charge Time

Measurement Speed	То
MED, SLOW	5ms
FAST	1.25ms

The discharge times  $T_\alpha$ ,  $T_\beta$ , and  $T_\gamma$  are measured by a programmable counter, AlU39, which counts the pulses of a 3.84MHz clock signal.

8-51. For accurate dual-slope analog-to-digital conversion, the VRD contains a pre-integrator which converts the phase detector's output signal into a dc voltage for integration by the main integrator. The pre-integrator charge time,  $T_1$ , depends on the test frequency and the measurement speed. Refer to Table 8-7 .

Table 8-7. Pre-Integrator Charge Time

Test Frequency	Measurement Speed			
1000 Troquency	MED, SLOW	FAST		
100Hz to 199Hz	1/f [s]	1/f [s]		
200Hz to 498Hz	5ms	1/1 [3]		
500Hz to 20kHz	Sins	2ms		

f: Test Frequency

#### Note

On the ranges shown in the table below, pre-integrator charge time,  $T_1$ , is 10 times longer than those listed in Table 8-7 .

C Range		Test Fre	quency	
C Kange	100Hz	200Hz	2kHz	20kHz
10mF				
1mF			Ì	
100µF				
10μF			_	
1μF				
100nF				
10nF				
lnF				
100pF				
10pF				

Capacitance Measurement

 $8\text{-}52\,.$  For the reference convenience, abbreviated signal names are listed in Table 8-8 .

Table 8-8. Signal Abbreviations

Abbreviation	Description
PSH	Phase Search
PS1, PS2, PS3	Phase Signals 1 through 3
REFDET	Reference Detection
AZT	Auto Zero Time
PC	Polarity Check
ZST	Zero Shift
IRST	Integrator Reset
PRST	Pre-Integrator Reset
IOFF	Integration Off
IOFS	Integration Off Switch

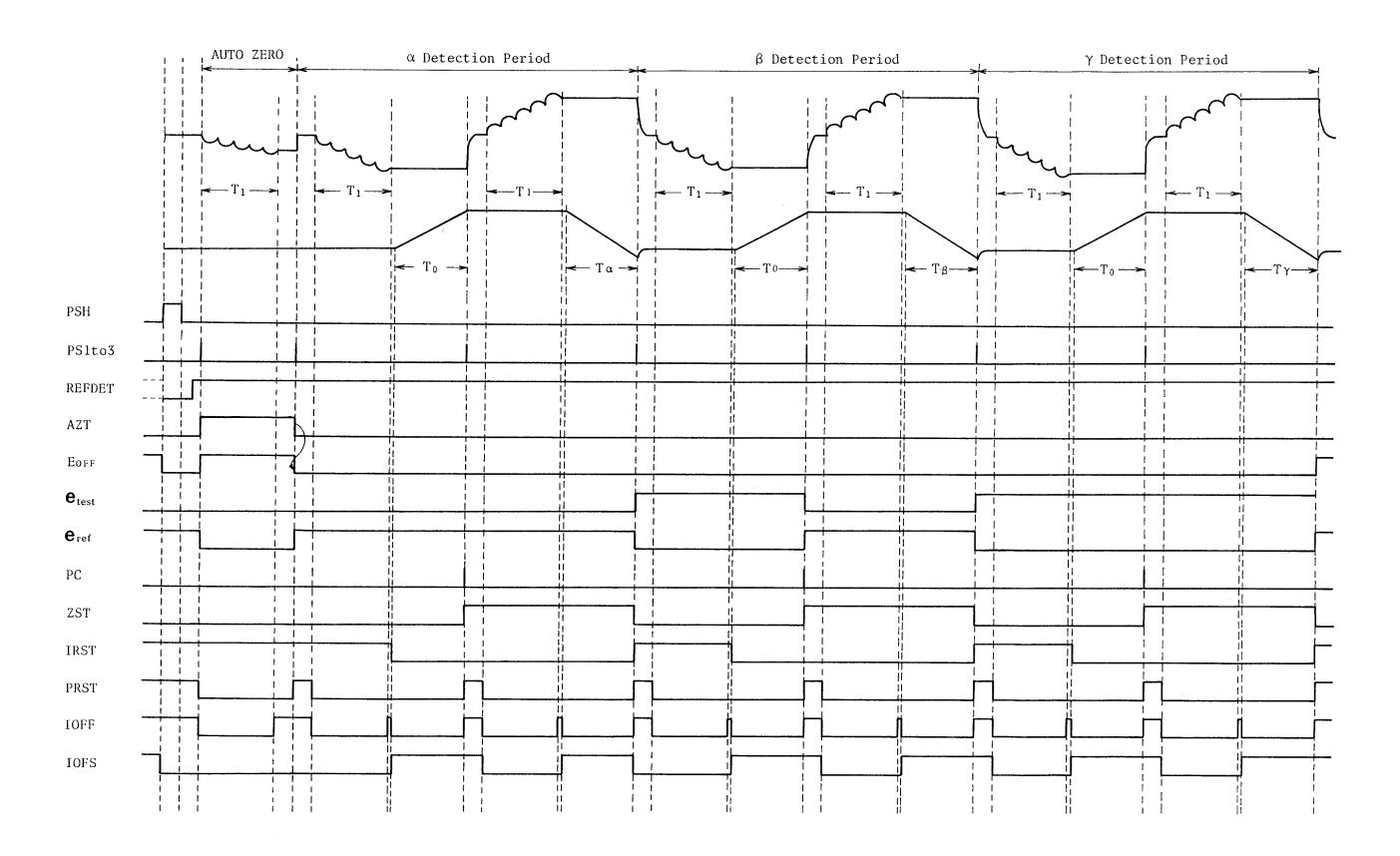


Figure 8-26. Timing Diagram.

8-53. Repair.

8-54. Board assembly locations are shown in Figure 8-27. Graphic symbols and abbreviated signal names used in schematic diagrams are explained in Figure 8-28 and Table 8-9, respectively.

Model 4276A

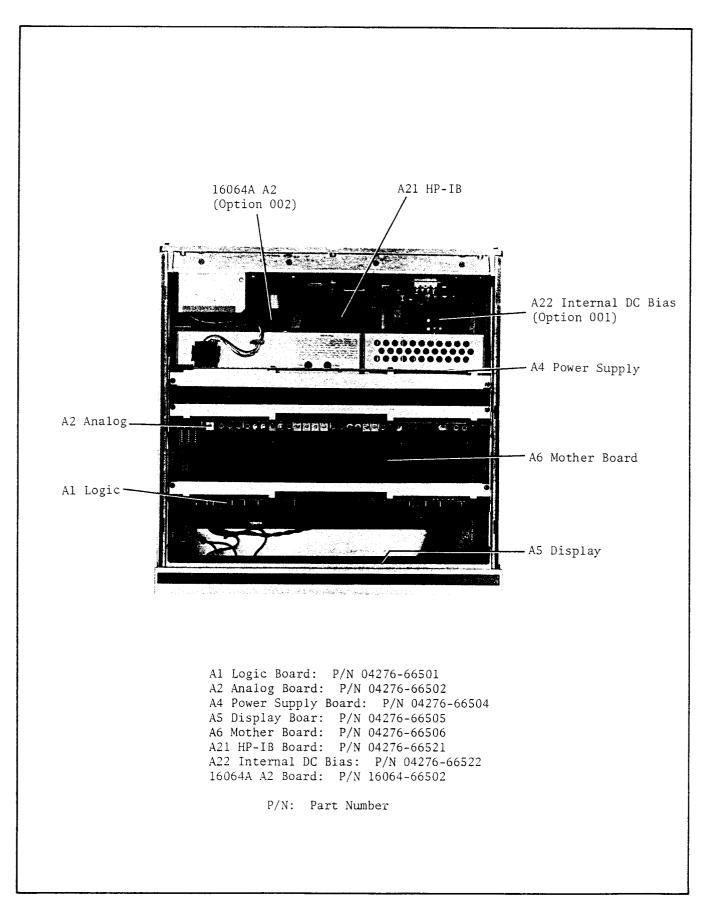


Figure 8-27. Assembly Locations (Top View).

P/0	Part of.
$\circ$	Knob control.
9	Screwdriver adjustment.
	Circuit assembly boarderline.
*	Asterisk denotes a factory selected value. Value shown is typical, part may be omitted.
*	Bead inductance.
_0_	Circuit board pattern inductance.
**************************************	Heavy line indicates main signal path.
	Heavy dashed line indicates main feedback path.
<b> CW A</b>	Wiper moves towards CW with clockwise rotation of control (as viewed from shaft or knob).
7	Numbered test point. Measurement aid provided.
<del>- 947 -</del>	Denotes wire color code. Code used is the same as the resistor color code (e.g., 9.4.7 denotes white/yellow/violet).
	Encloses front panel designations.
[]	Shielded area.
<u>_</u>	Indicates direct conducting connection to earth.
工	Indicates conducting connection to chassis or frame.
$\Rightarrow$	Indicates circuit common connection.

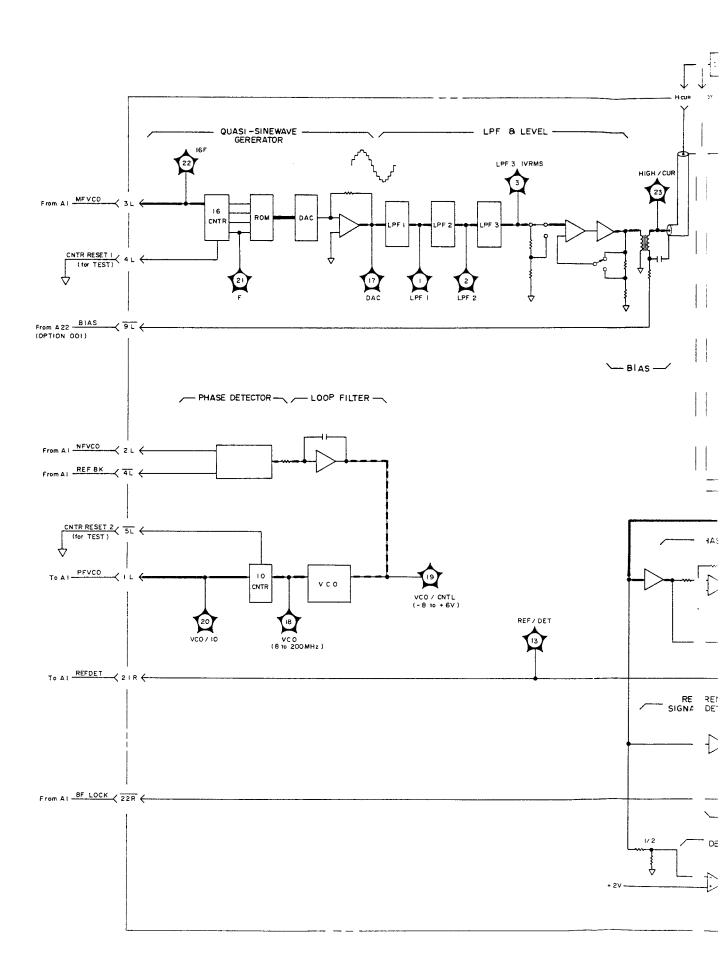
Figure 8-28. Schematic Diagram Notes.

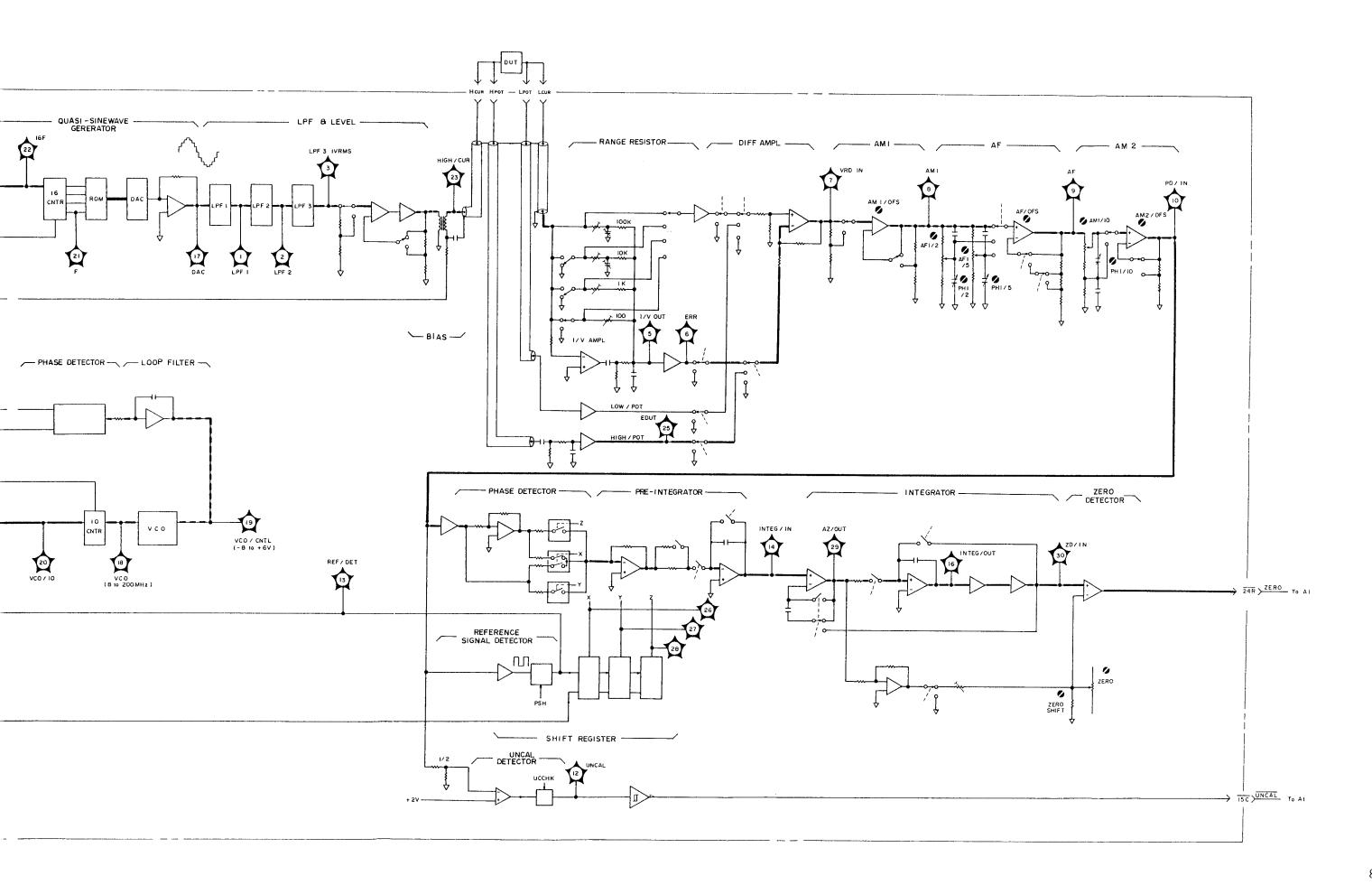
Table 8 9. Mnemonic Information (Sheet 1 of 2)

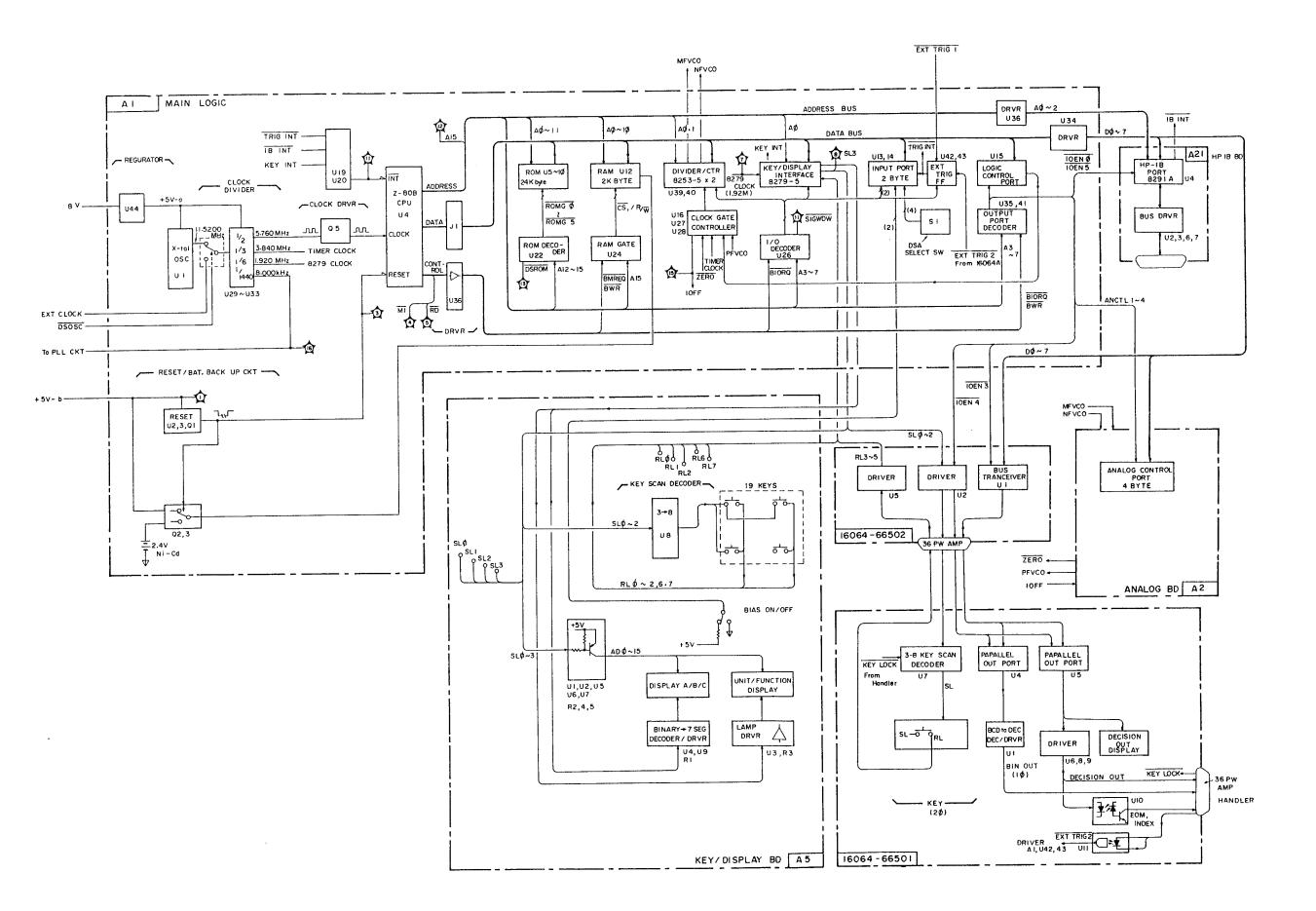
Mnemonic	DESCRIPTION
AF	Amplifier-Frequency
AM1	Amplifier-Magnitude 1
AM2	Amplifier-Magnitude 2
ANACTL 1 - 5	Analog section control lines
BA0-2	Buffered address bus lines to the A21 HP-IB board
BATTERY	Continuous Memory battery supply and charge line
BCLK	Buffered master clock (5.760MHz)
BD0-7	Buffered data bus lines
BEXT TRIG 2	Buffered external trigger signal from the 16064A
BIAS	Output from the internal dc bias source (0 to ±40V)
BIASG	Internal dc bias source common
BIAS	A22 DC BIAS board is installed.
BIAS ON	Front panel DC BIAS switch status
BIN1 - 9	Comparison results output to an external component handler
BIOEN3 - 4	Buffered I/O enable lines from the 16064A A2 board to the Alboard.
BIORD	Buffered I/O read line from the 16064A A2 board to the A1 board.
BIOWR	Buffered I/O write line from the 16064A A2 board to the Alboard.
BLANK	Display blanking signal
BRL3-5	Buffered return line from the 16064A keyboard
BSLO- 2	Buffered send line to the 16064A keyboard
CPWT	Comparator wait signal
DA 0 - 3	Front panel display data lines
DB 0 - 3	Front panel display data lines
DQ HI, IN, LO	Comparison results output to an external component handler.
DUT	Device under test
EOM	End of measurement. Informs an external component handler that the comparison has been made.
EXT DCV1	External DC supply input to HANDLER INTERFACE for open collector outputs.
EXT DCV2	External DC supply. Input from an external component handler.
EXT DCV2 COM	EXT DCV2 return line to an external component handler.
EXT TRIG	External trigger input from an external component handler.
EXT TRIG 1	External trigger signal from the rear panel.
EXT TRIG 2	External trigger signal from 16064A.
Нс	Center conductor of high current terminal

Table 8-9. Mnemonic Information (Sheet 2 of 2)

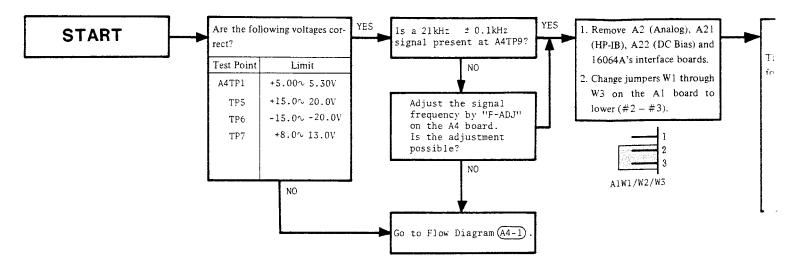
Mnemonic	DESCRIPTION
Нсс	Outer conductor of high current terminal
Нр	Center conductor of high potential terminal
IB INT	HP-IB interrupt
INDEX	Signal which informs an external component handler that the instrument has finished analog measurement.
<u>10EN 0</u> - 5	I/O enable lines
IOFF	Integrator off
IORD	I/O read
TOWR	I/O write
KEY LOCK	16064A keyboard disable. Input from an external component handler.
Lc	Center conductor of low current terminal
<b>L</b> cc	Outer conductor of low current terminal
LC HI, IN, LO	Comparison results output to an external component handler
Lp	Center conductor of low potential terminal
MF VCO	Divided VCO output (16 times the test frequency)
NF VCO	Divided VCO output (8kHz)
OUT OF BIN	Comparison result output to an external component handler
PD	Phase detector
PF VCO	Divided VCO output (fvco ÷ 10)
PWF	Power failure
REF DET	Reference phase detect
REF8K	VCO reference frequency, 8kHz
RESET	Reset signal for A21 HP-IB and A22 DC BIAS boards
RLO - 2, 6-7	Return lines from the A5 DISPLAY board.
SL 0 - 3	Send lines to the A5 DISPLAY board.
UNCAL	Uncalibration detect
ZERO	Integrator zero detect
8FLCK	Signal for phase detector (8 times of test frequency)



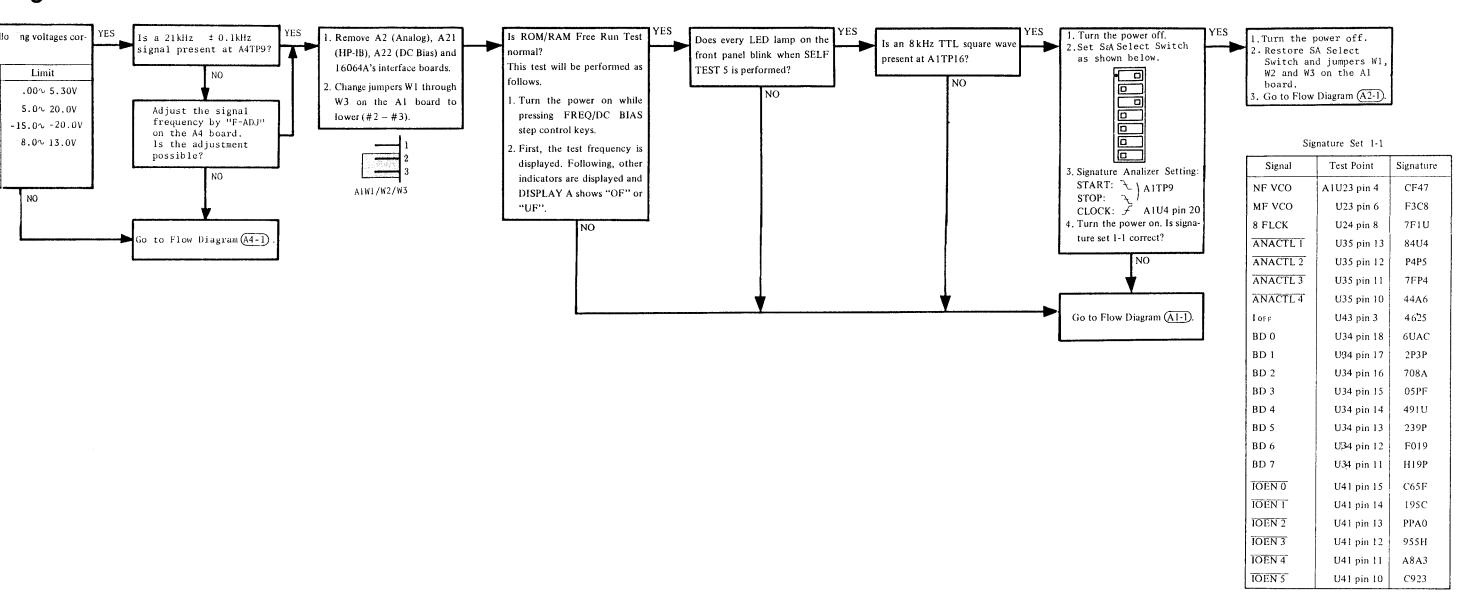


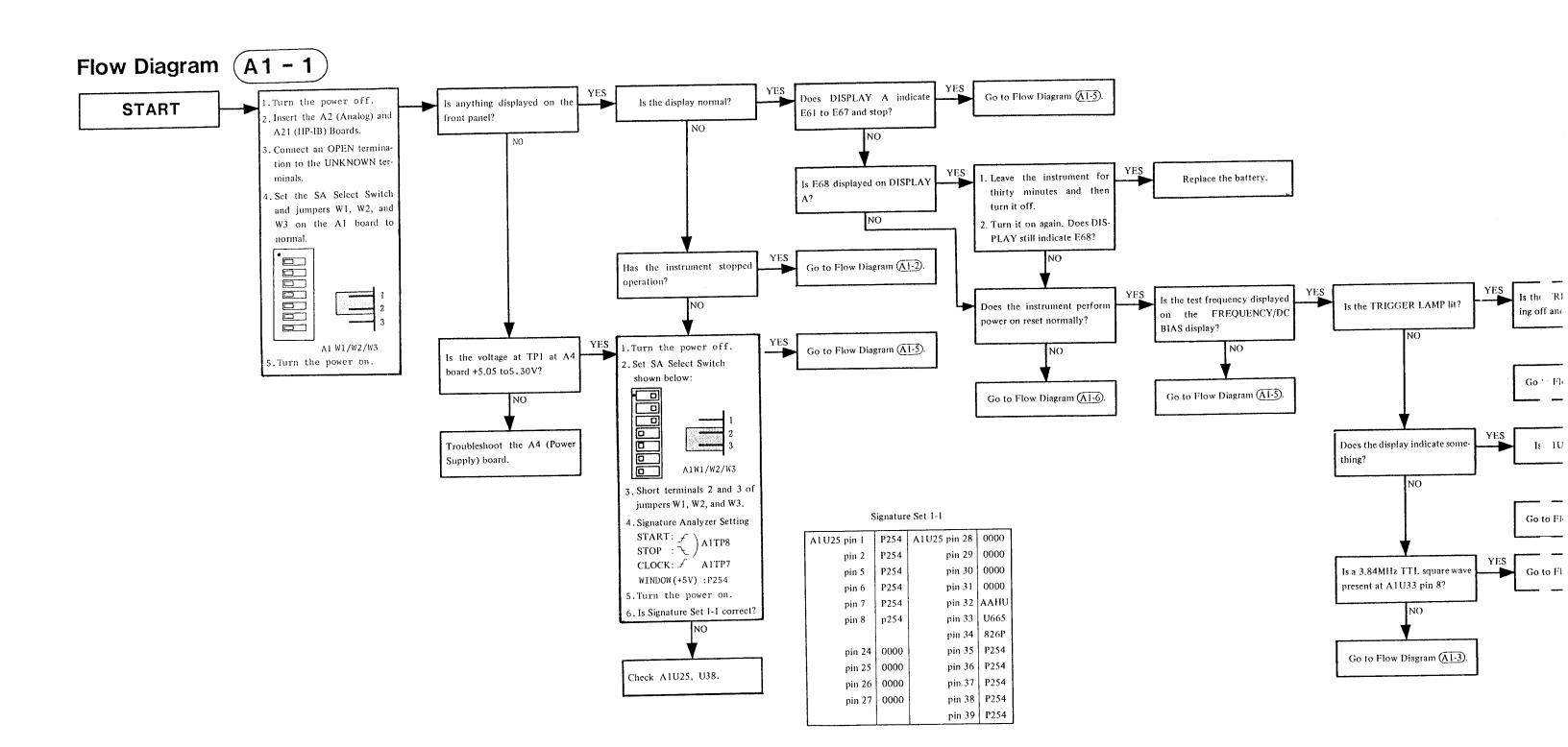


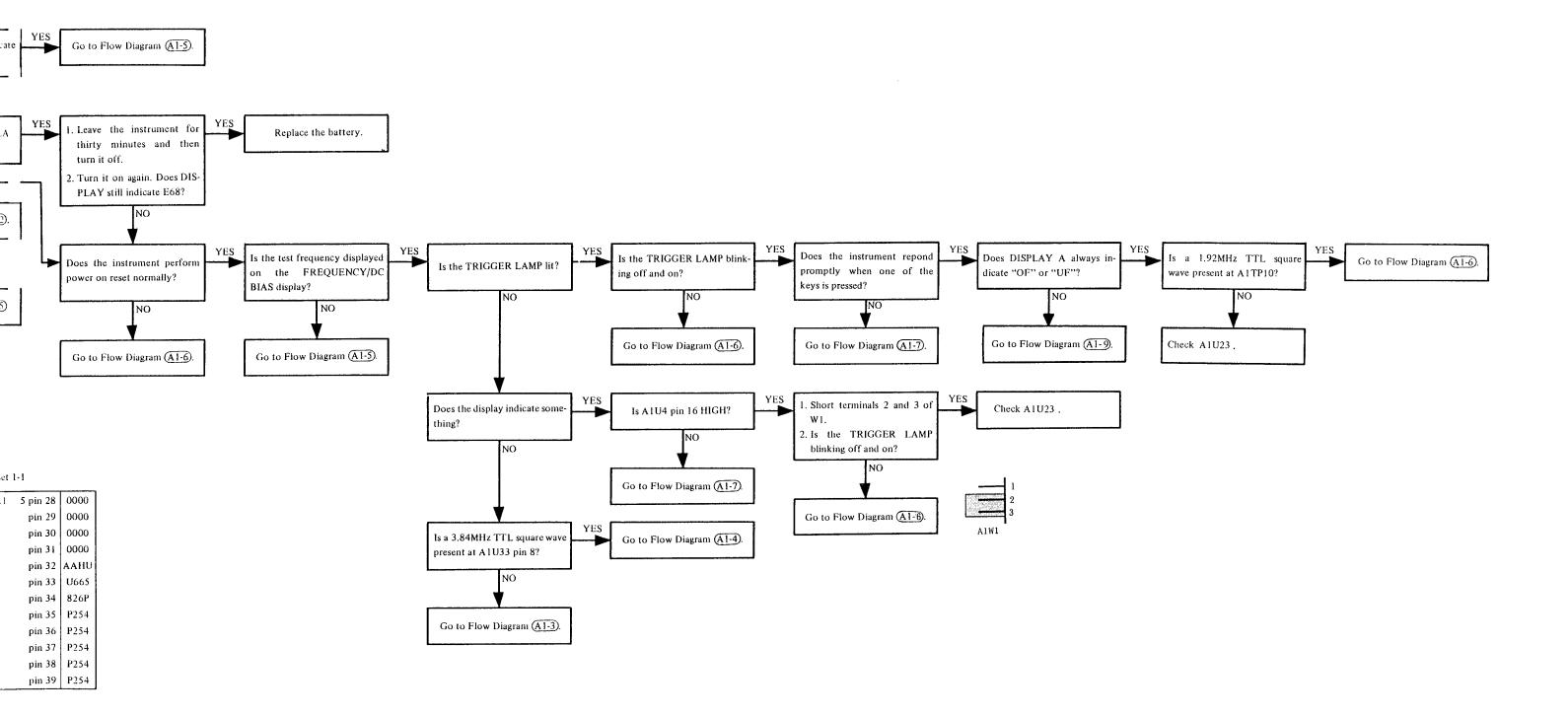
# **Board Isolation Flow Diagram**

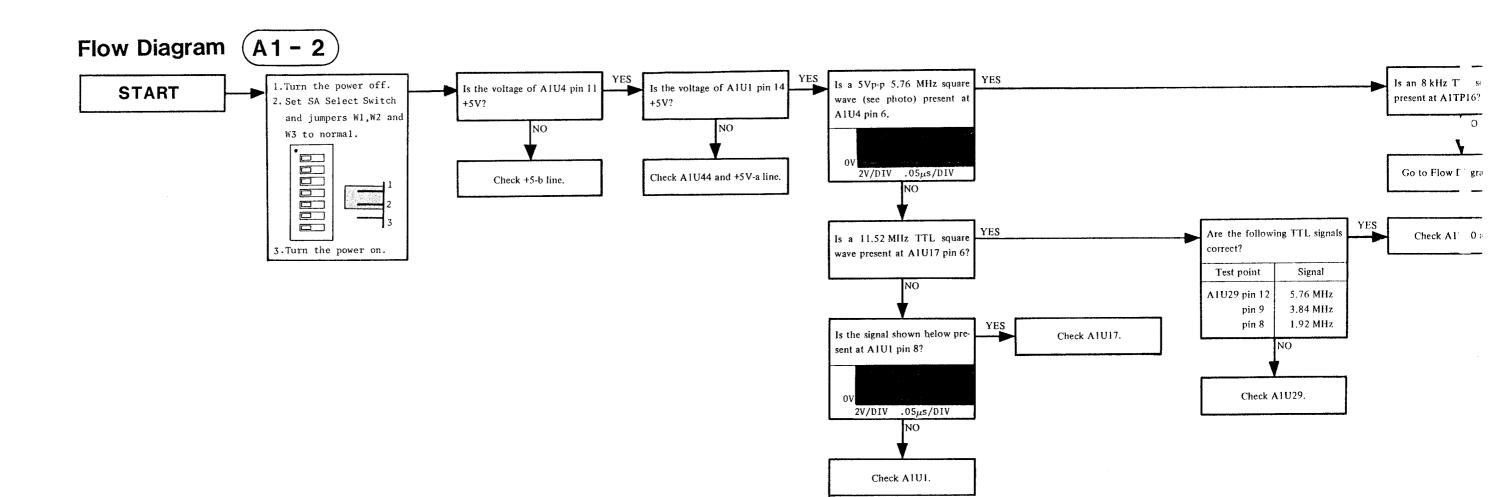


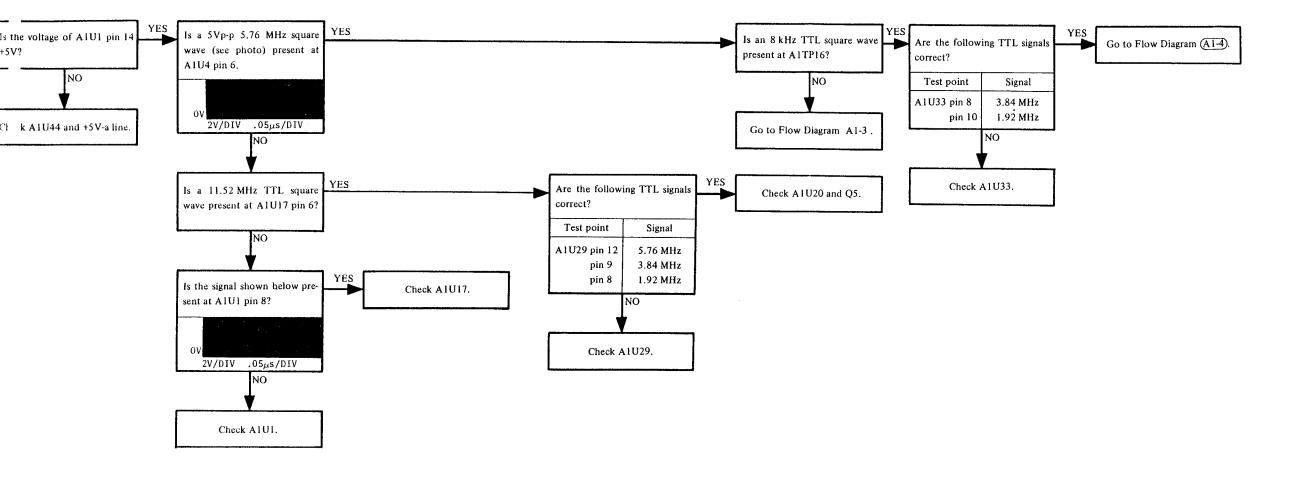
## agram

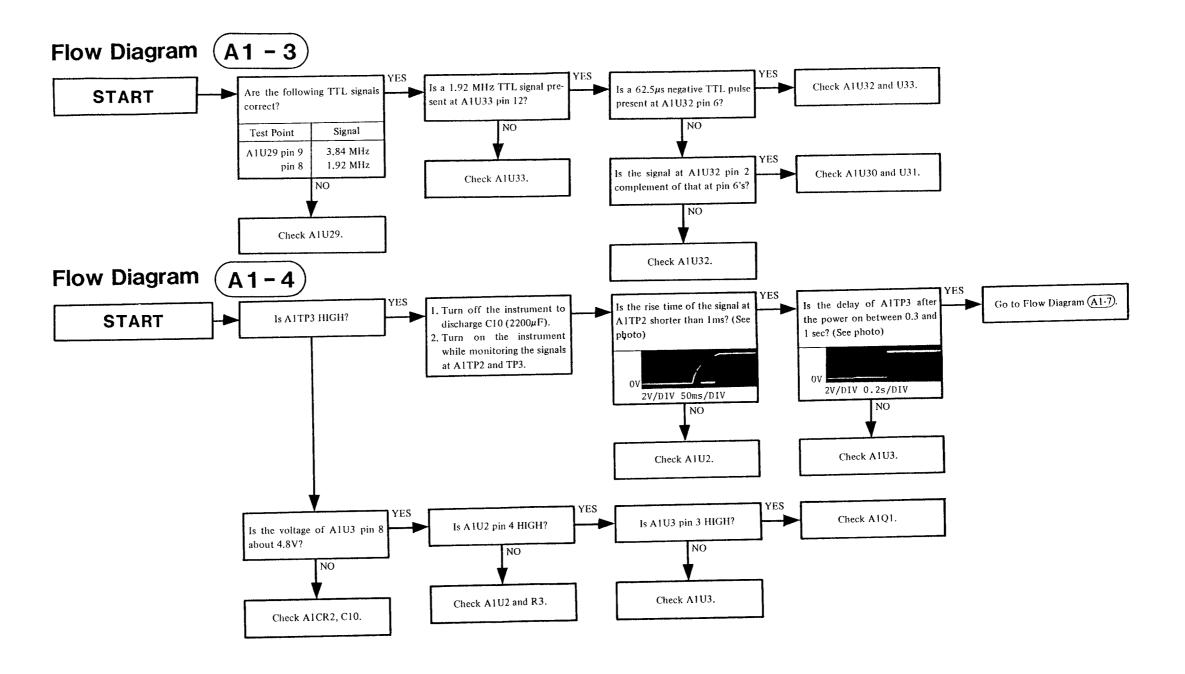


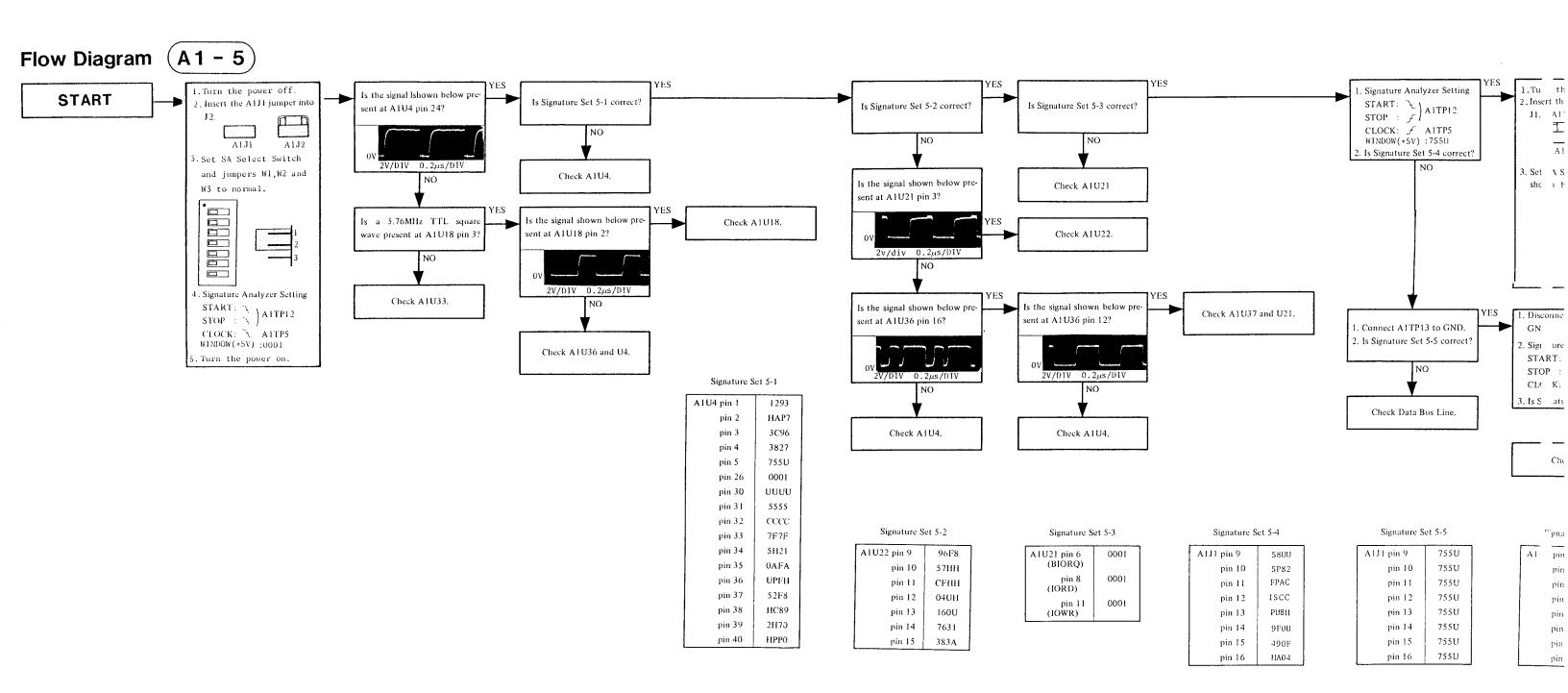


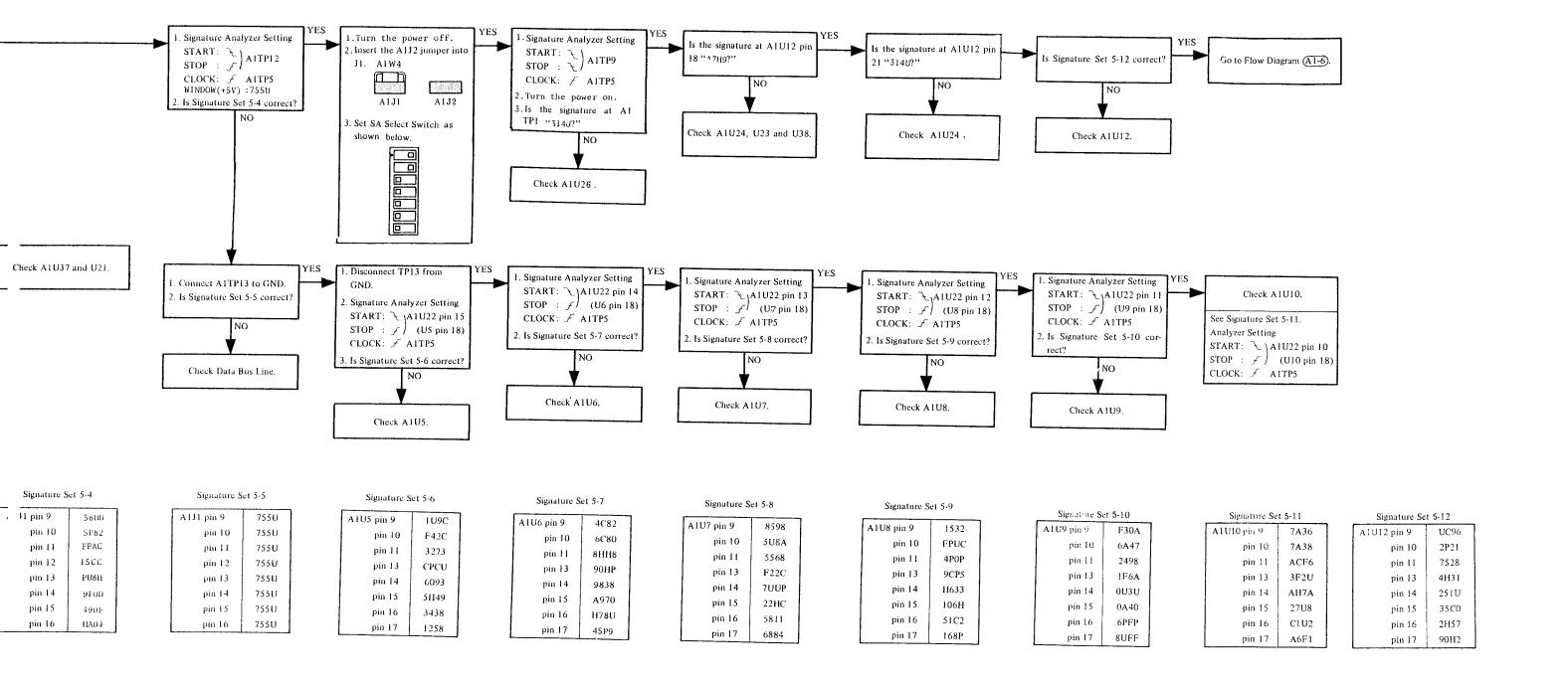


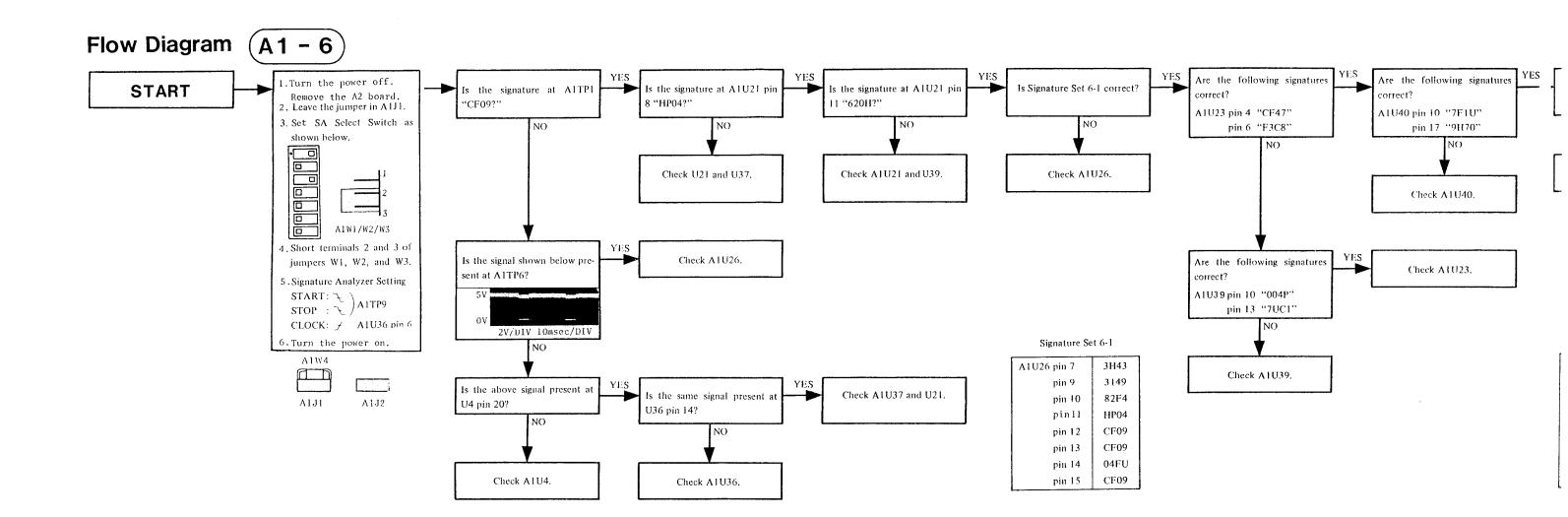


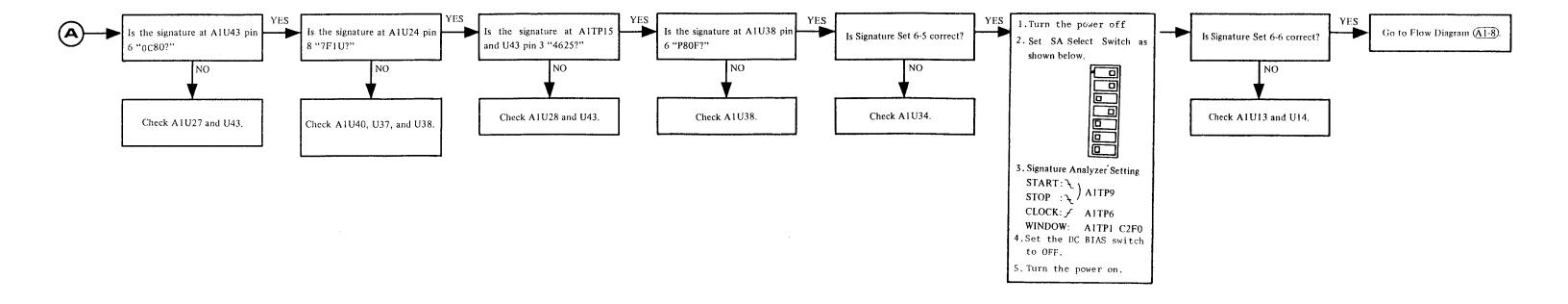


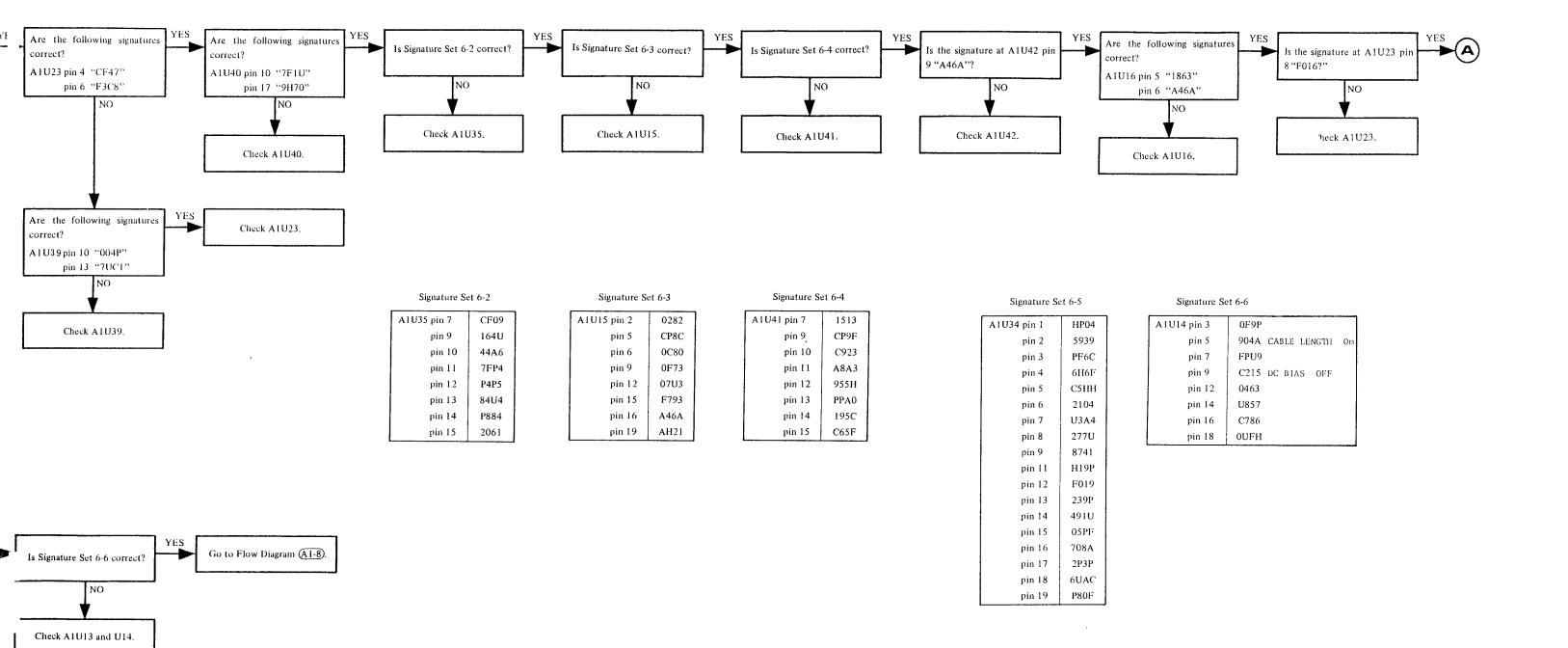


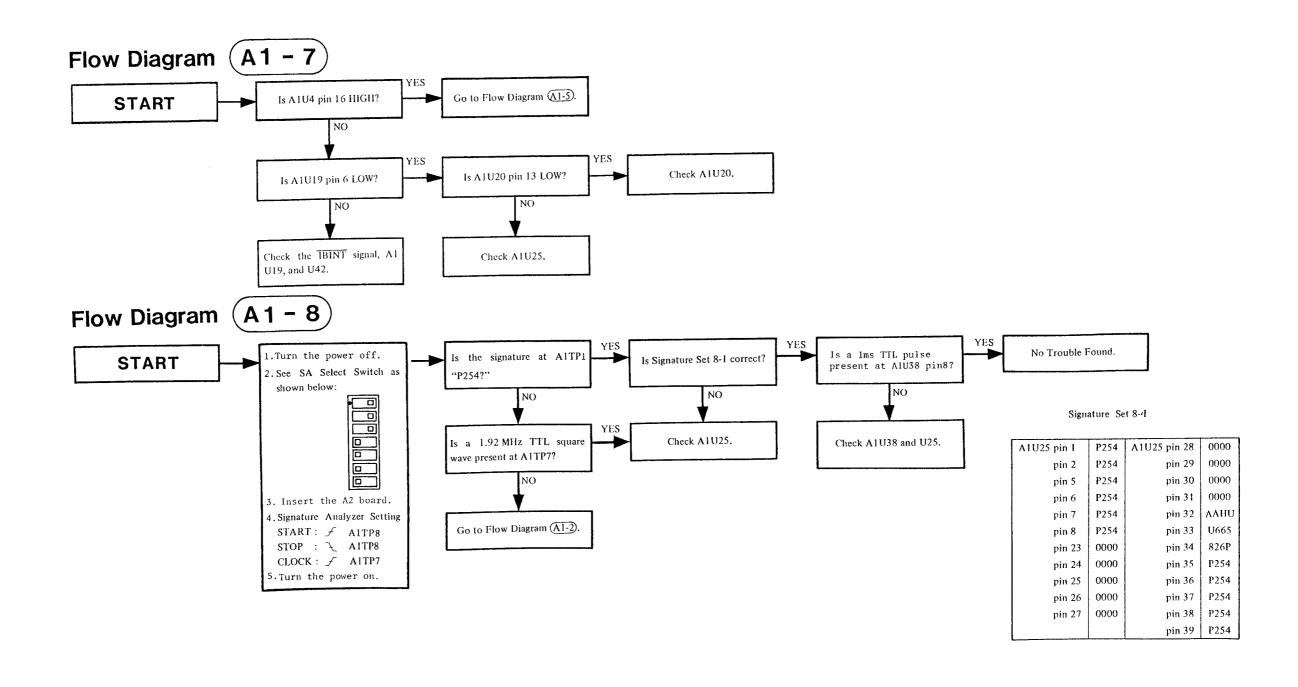


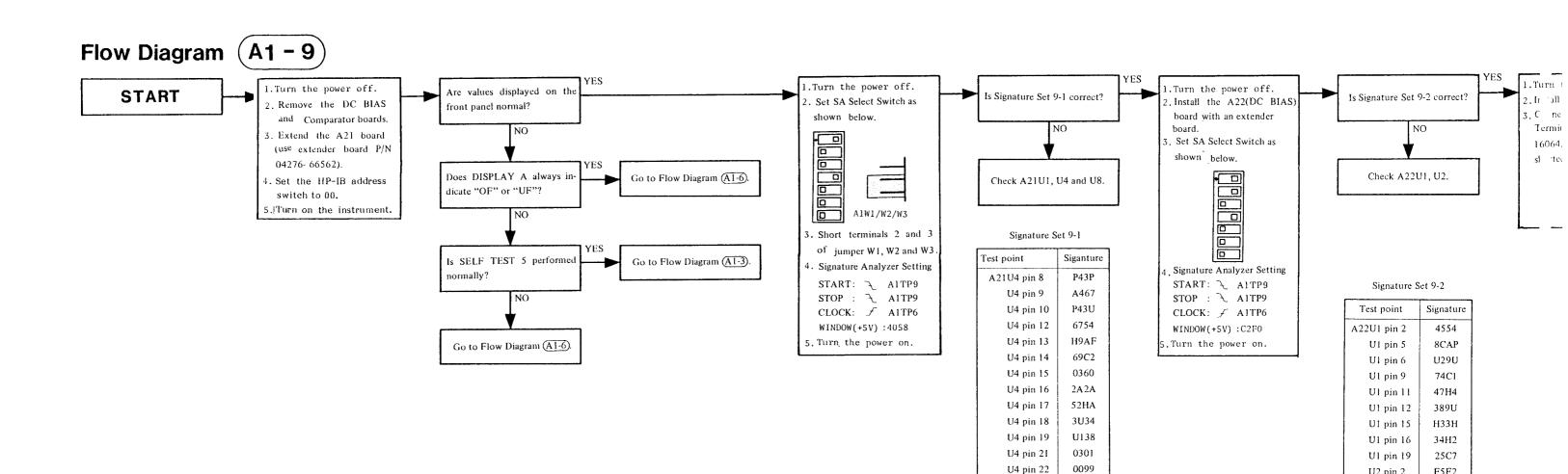












U4 pin 23

023A

F5F2

571F

1PC7

66P4 49C8

870P

2489

U2 pin 2

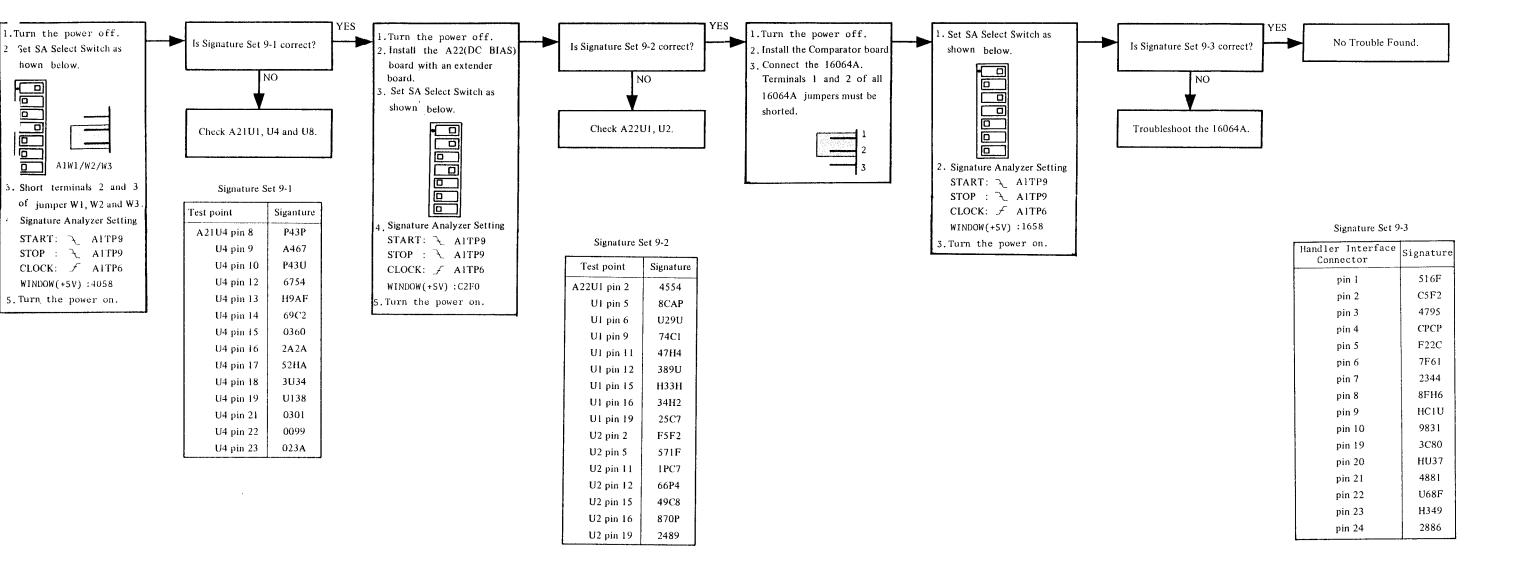
U2 pin 5

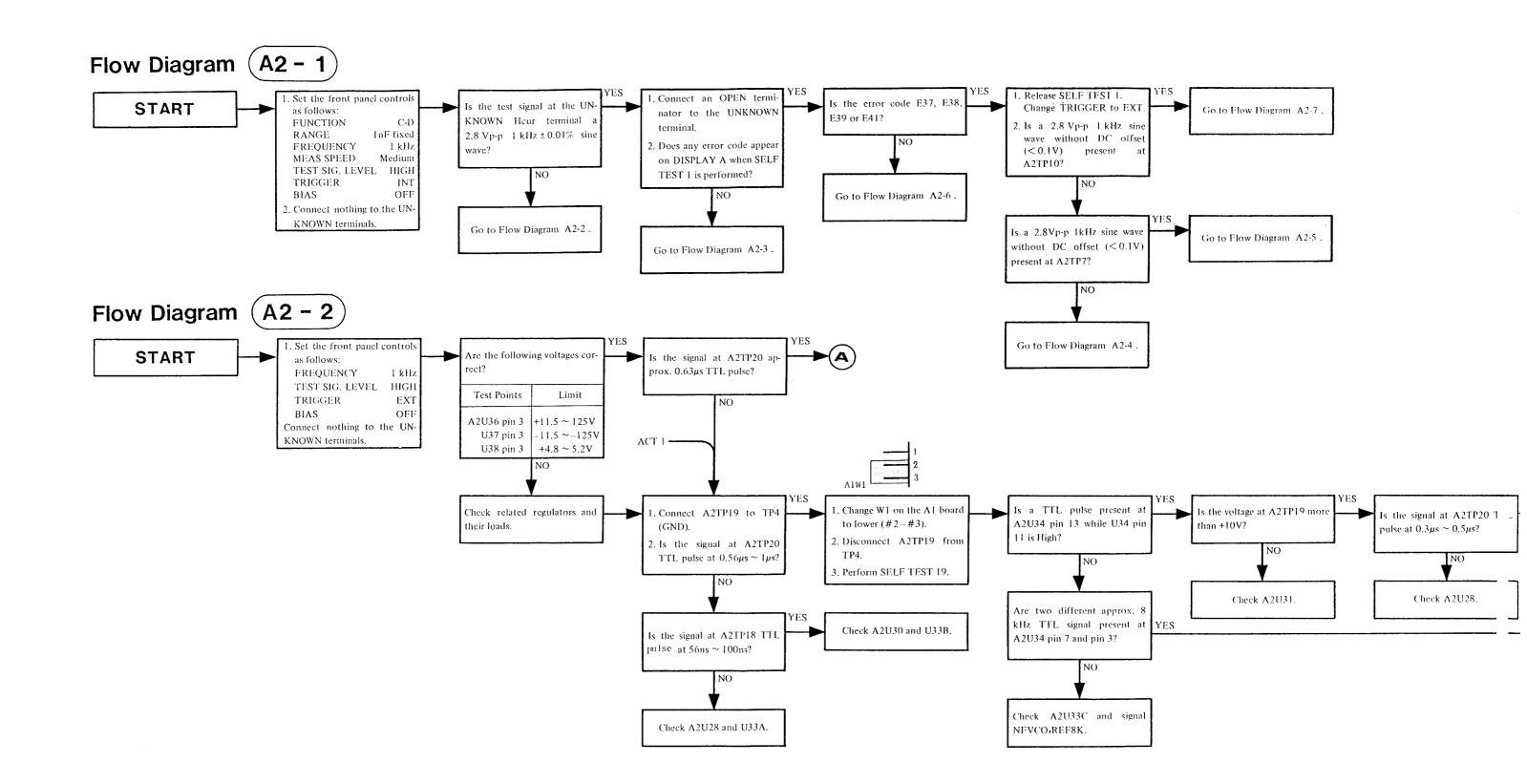
U2 pin 11

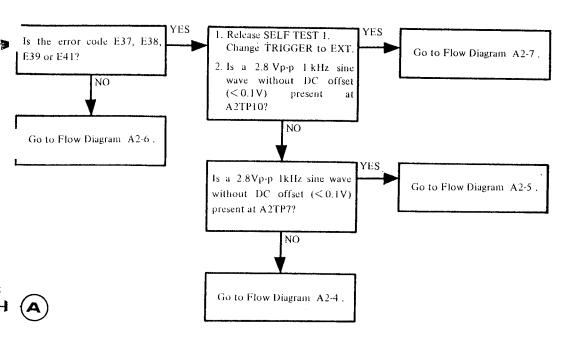
U2 pin 12

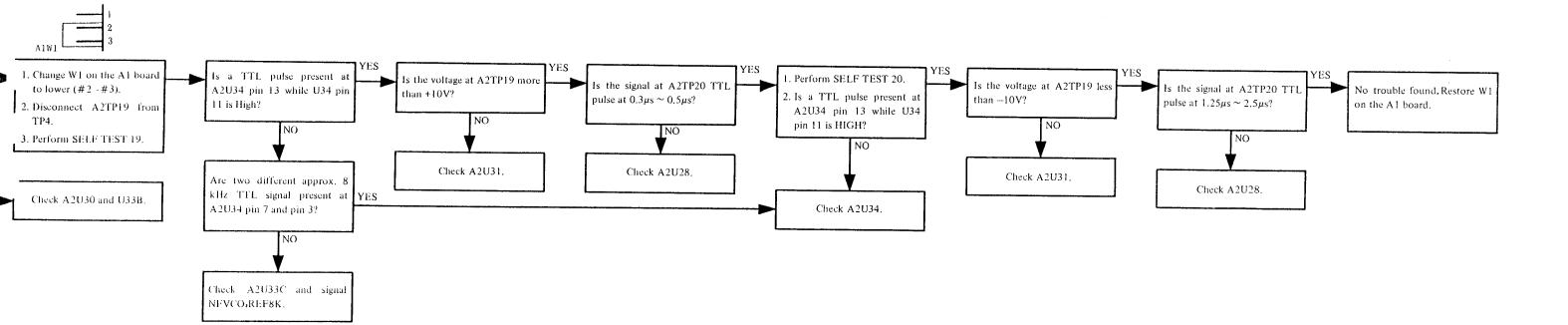
U2 pin 15

U2 pin 16 U2 pin 19

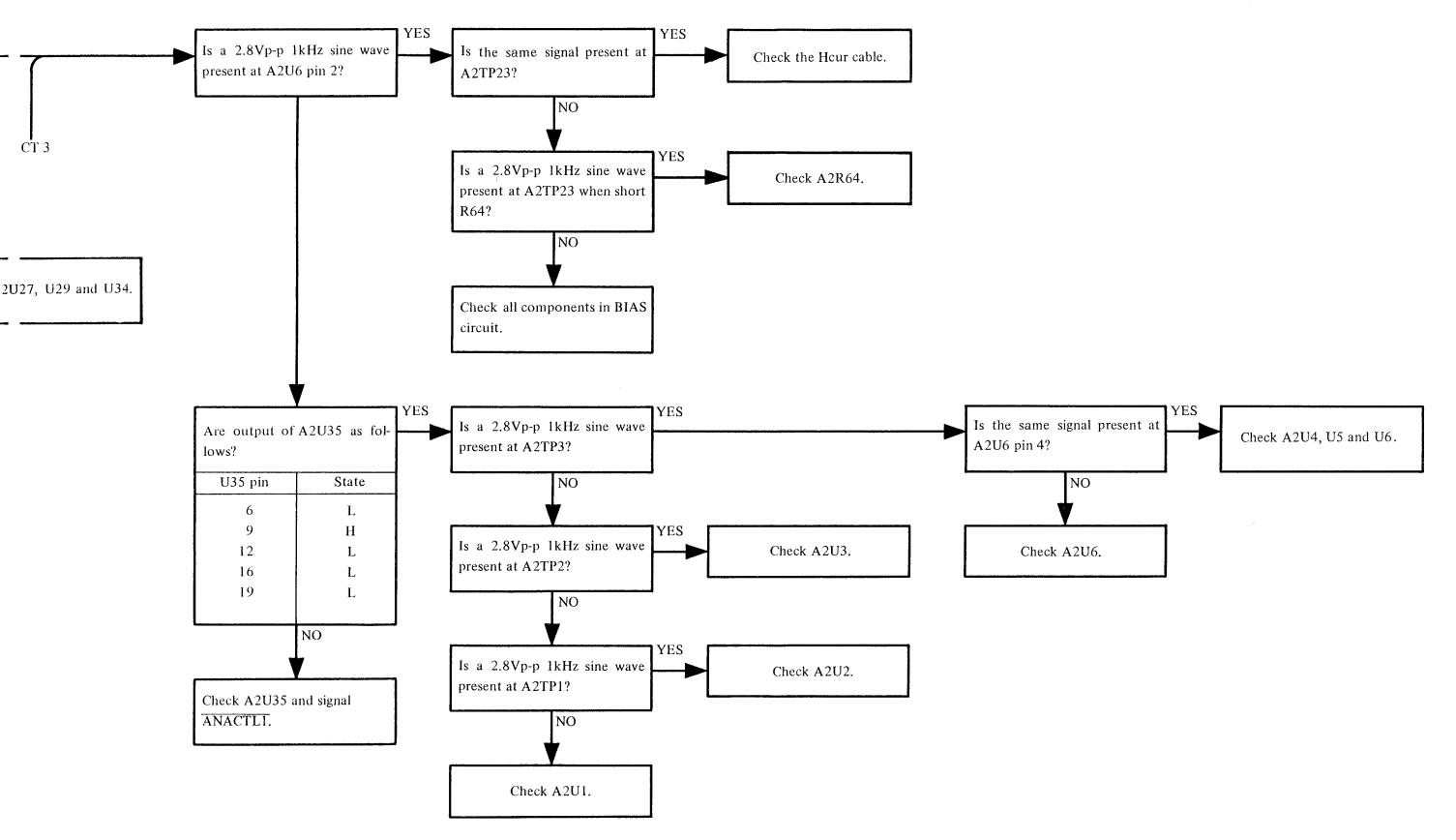


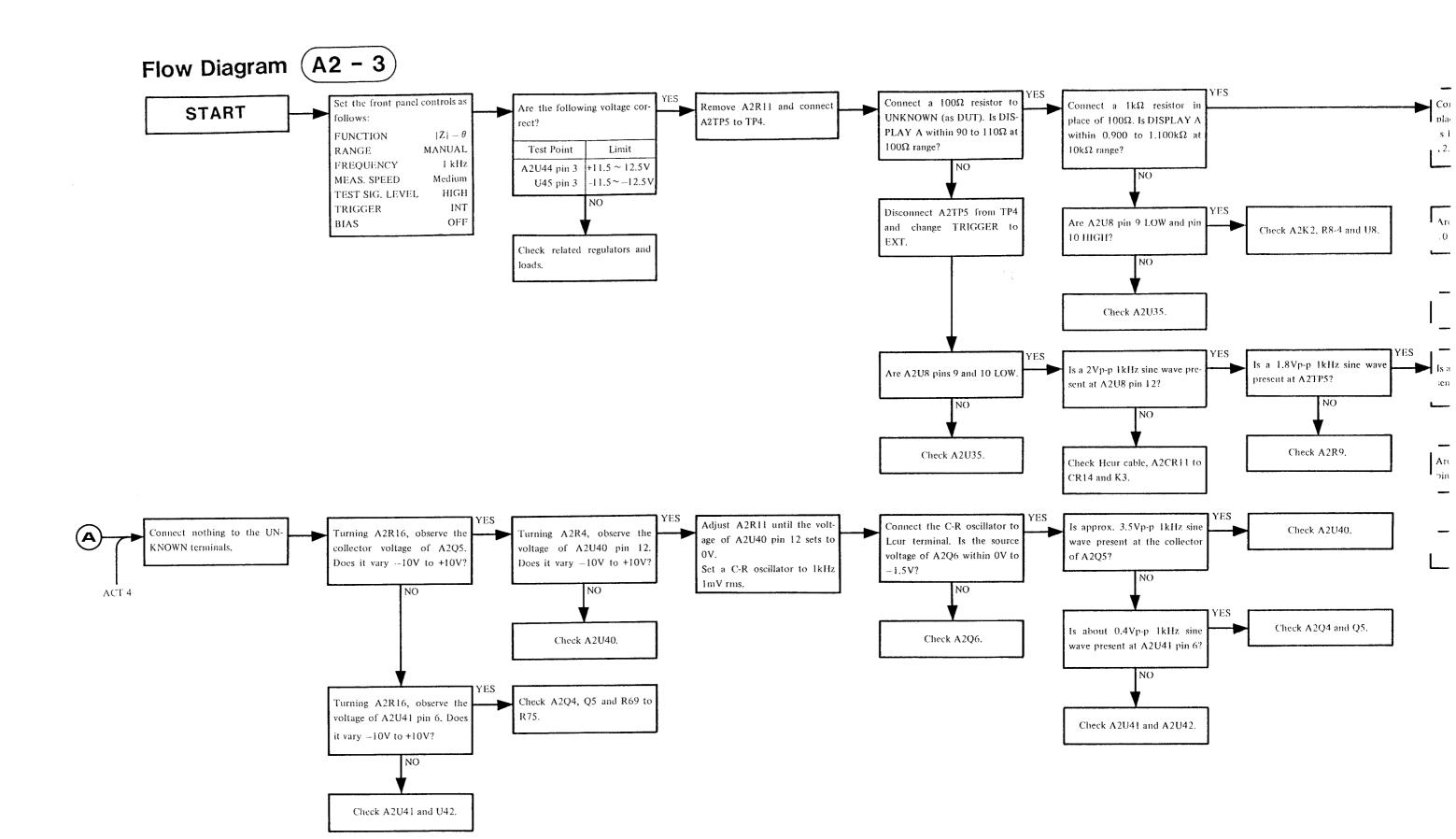


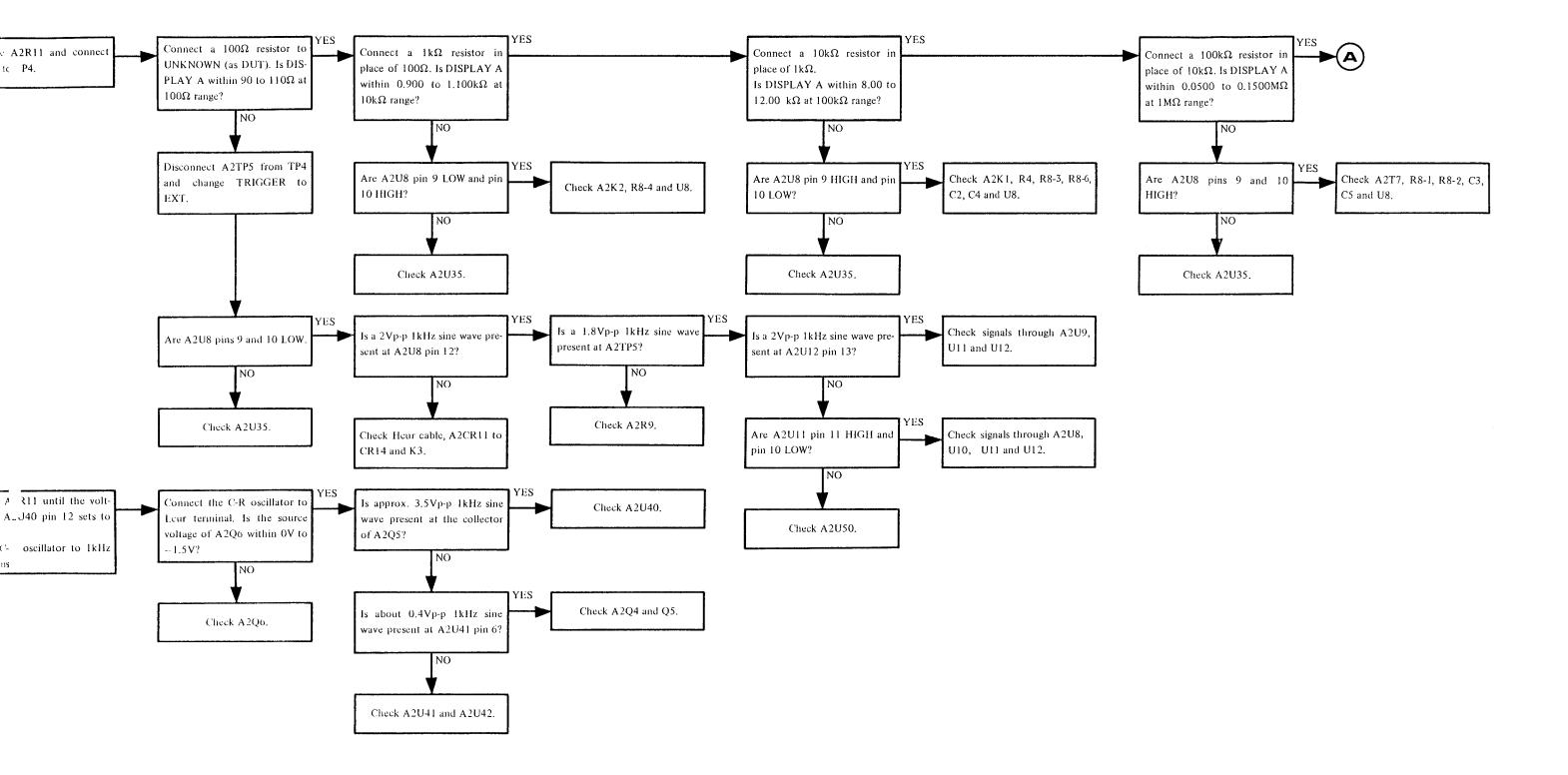


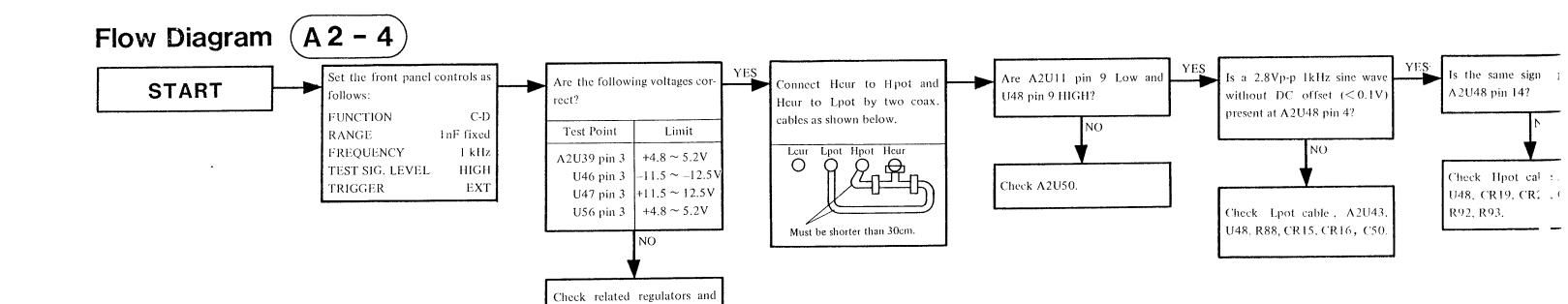


### Flow Diagram (A2 - 2) YES Check the Hour cable. Is the same signal present at Is a 2.8Vp-p 1kHz sine wave Is the signal shown below A2TP23? present at A2U6 pin 2? (Quasi-sine wave) present at A2TP17? ACT 3 Is a 2.8Vp-p 1kHz sine wave Check A2R64. present at A2TP23 when short 1V/DIV 0.1ms/DIV NO NO ACT 2 Is a 1kHz TTL square wave Check A2U27, U29 and U34. Check all components in BIAS present at A2TP21? circuit. NO YES Check A2U32 and U33. Is a 2.8Vp-p 1kHz sine wave Are output of A2U35 as folpresent at A2TP3? State U35 pin L 6 YES H Is a 2.8Vp-p 1kHz sine wave Check A2U3. 12 L present at A2TP2? L 16 L 19 NO NO YES Is a 2.8Vp-p 1kHz sine wave Check A2U2. present at A2TP1? Check A2U35 and signal NO ANACTL1. Check A2U1.

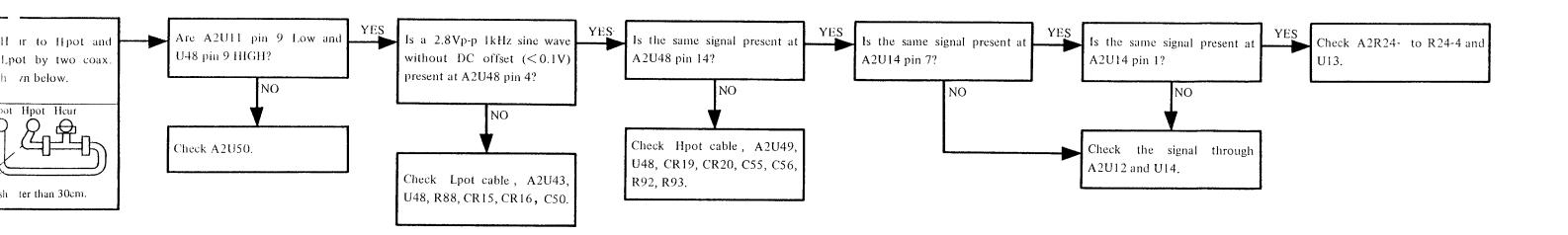


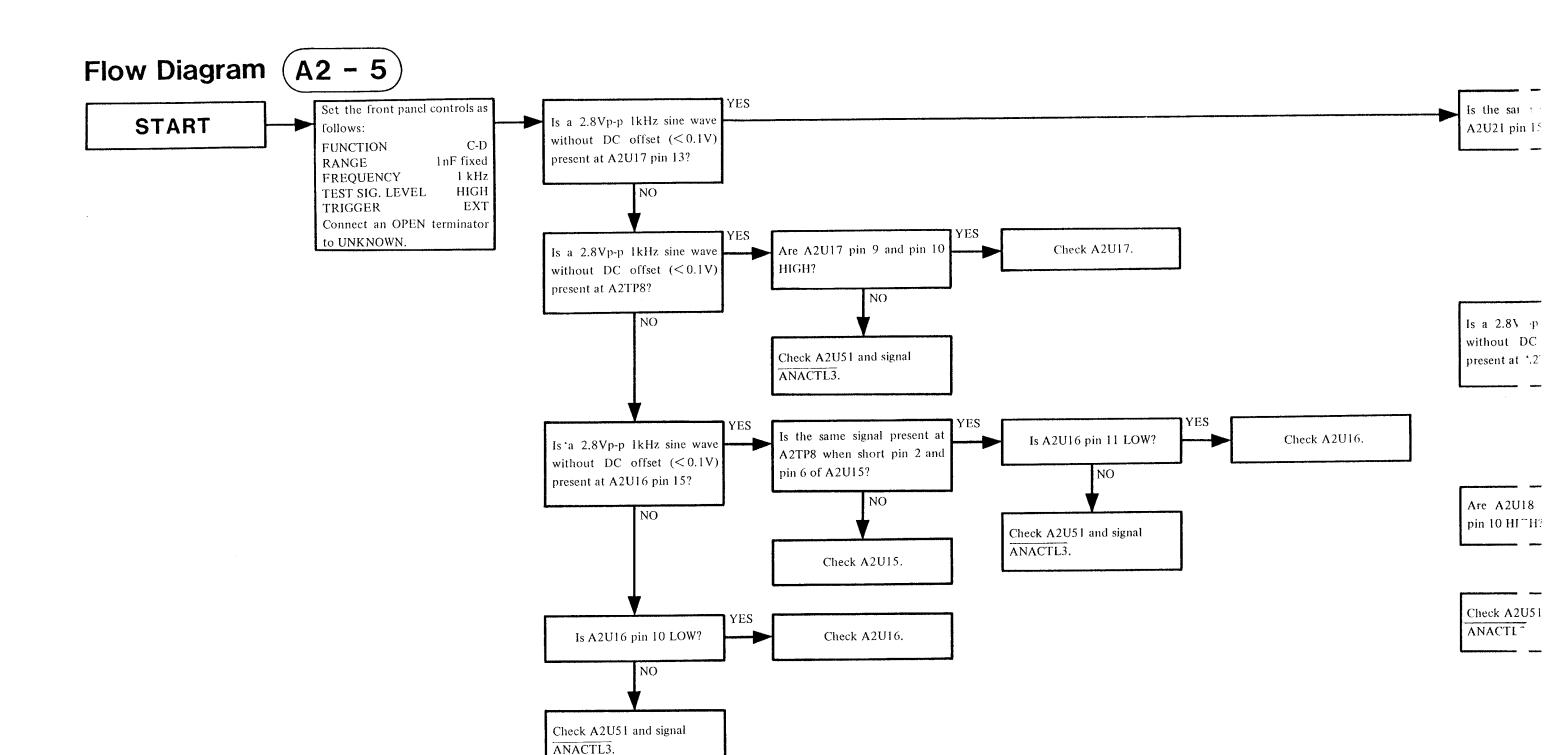


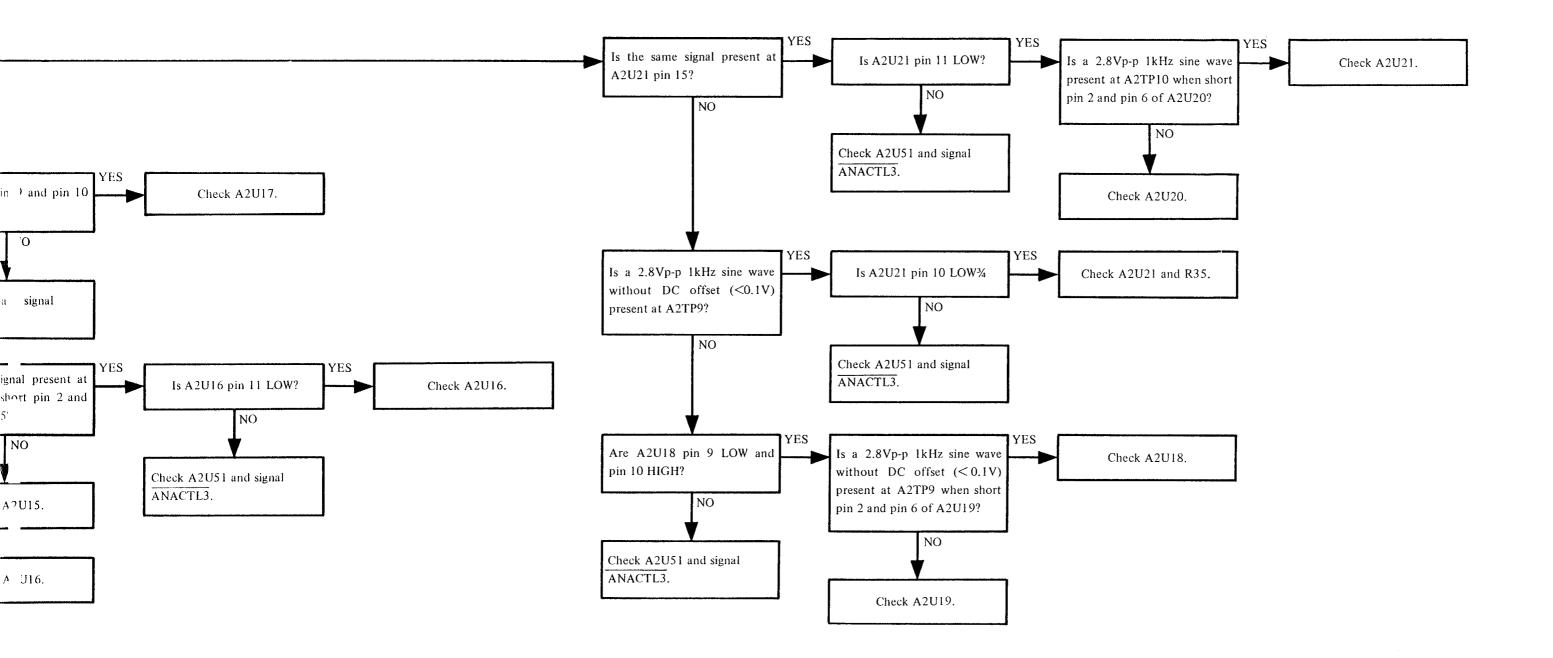


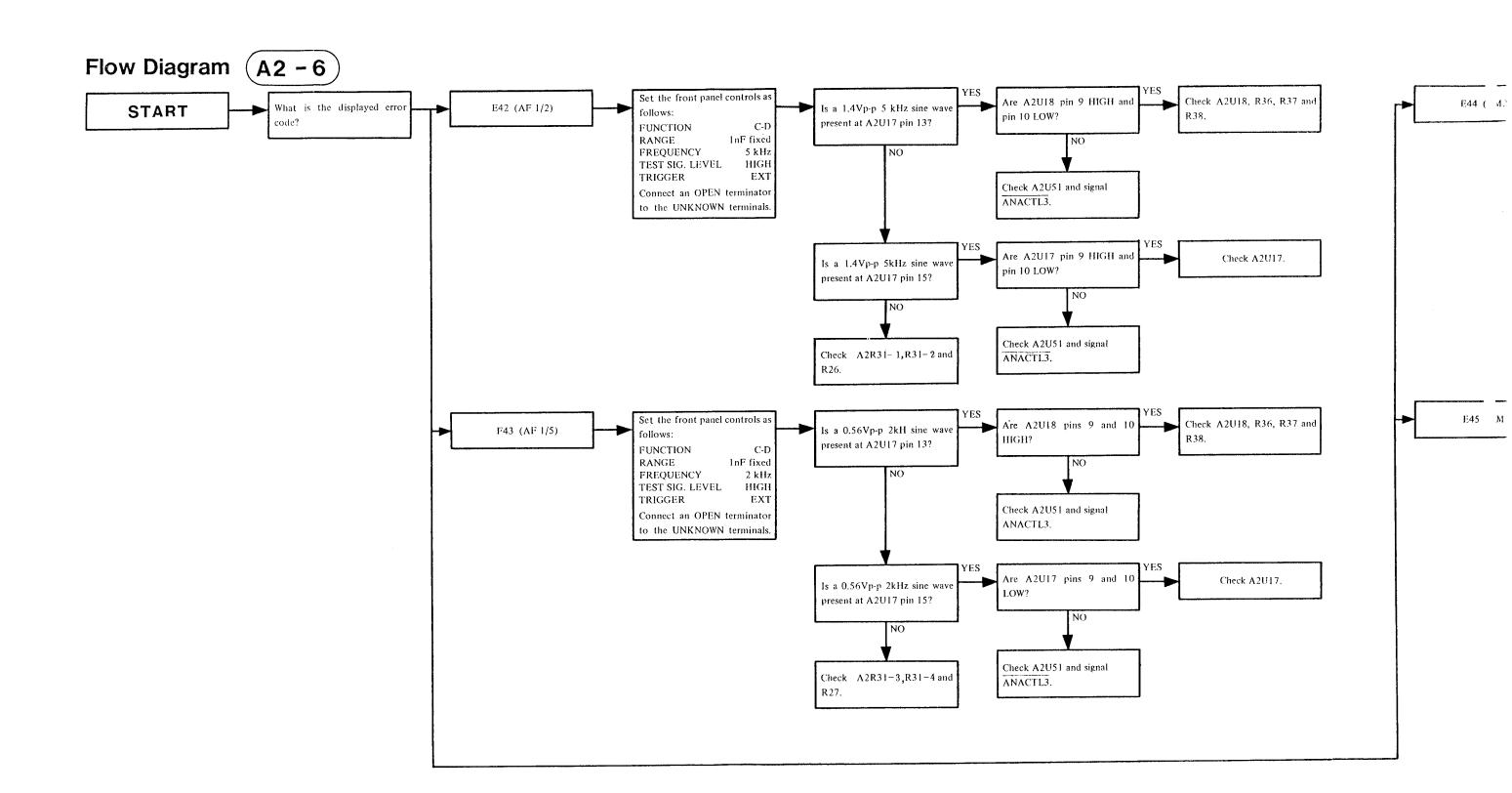


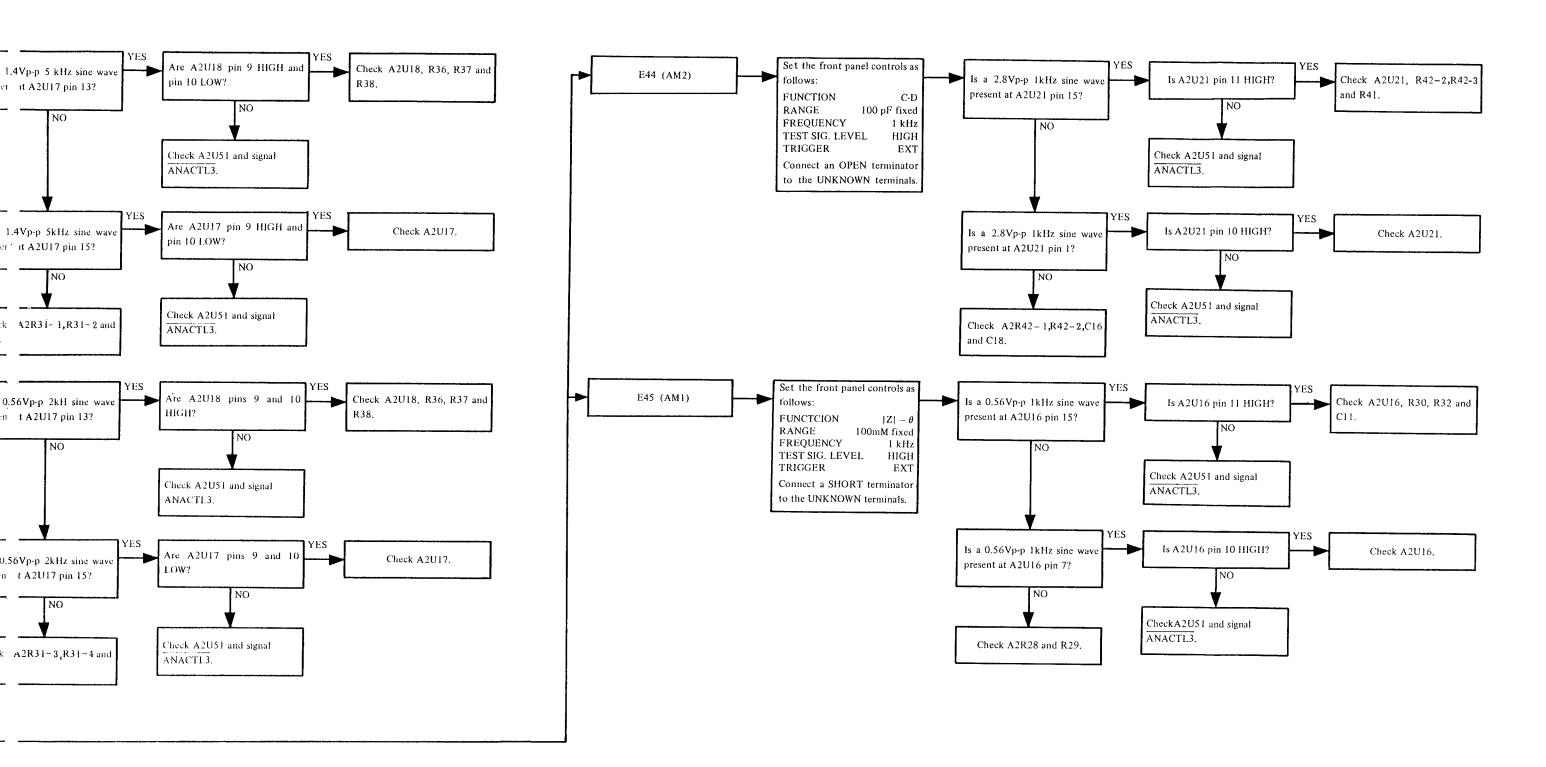
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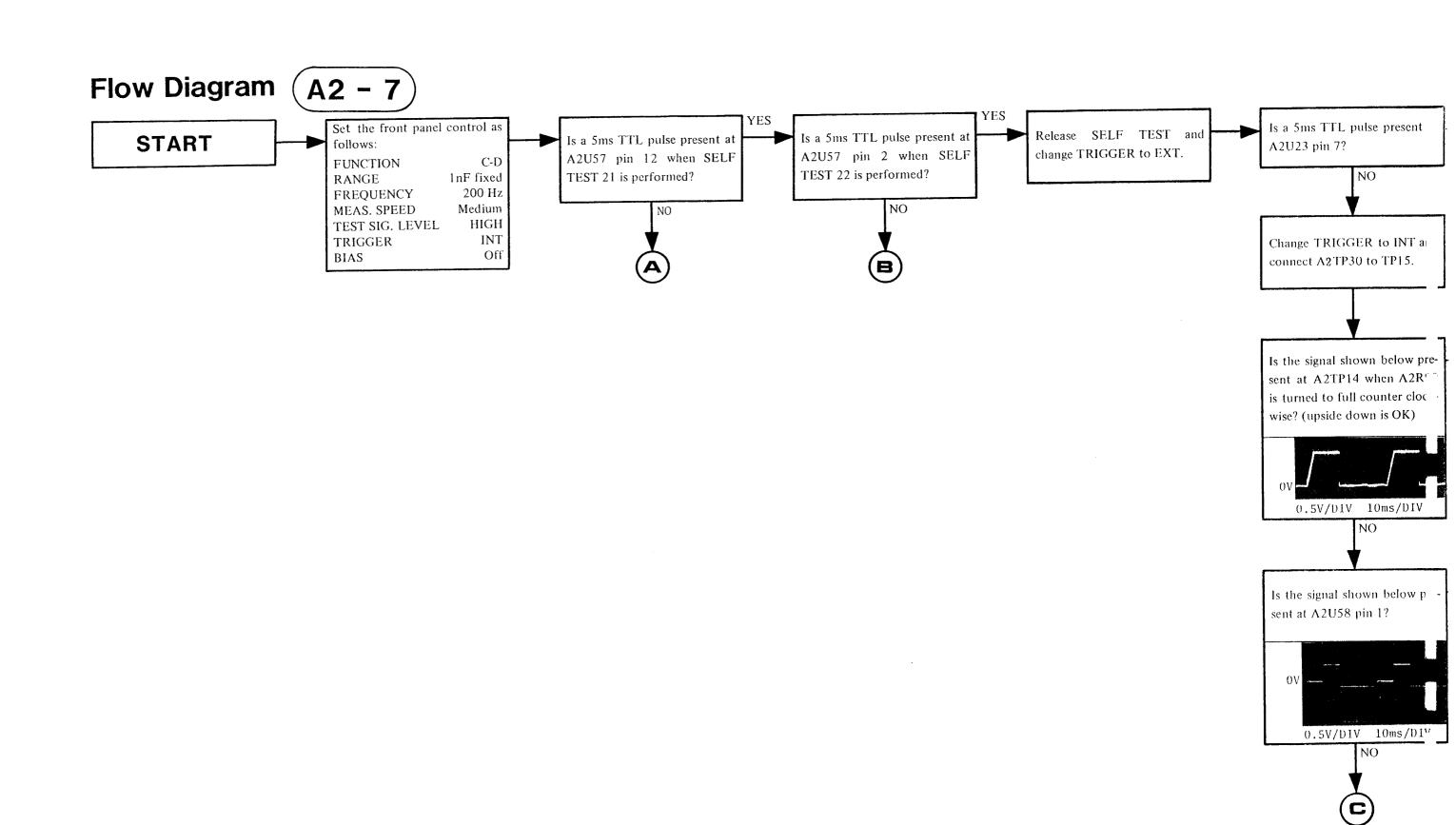


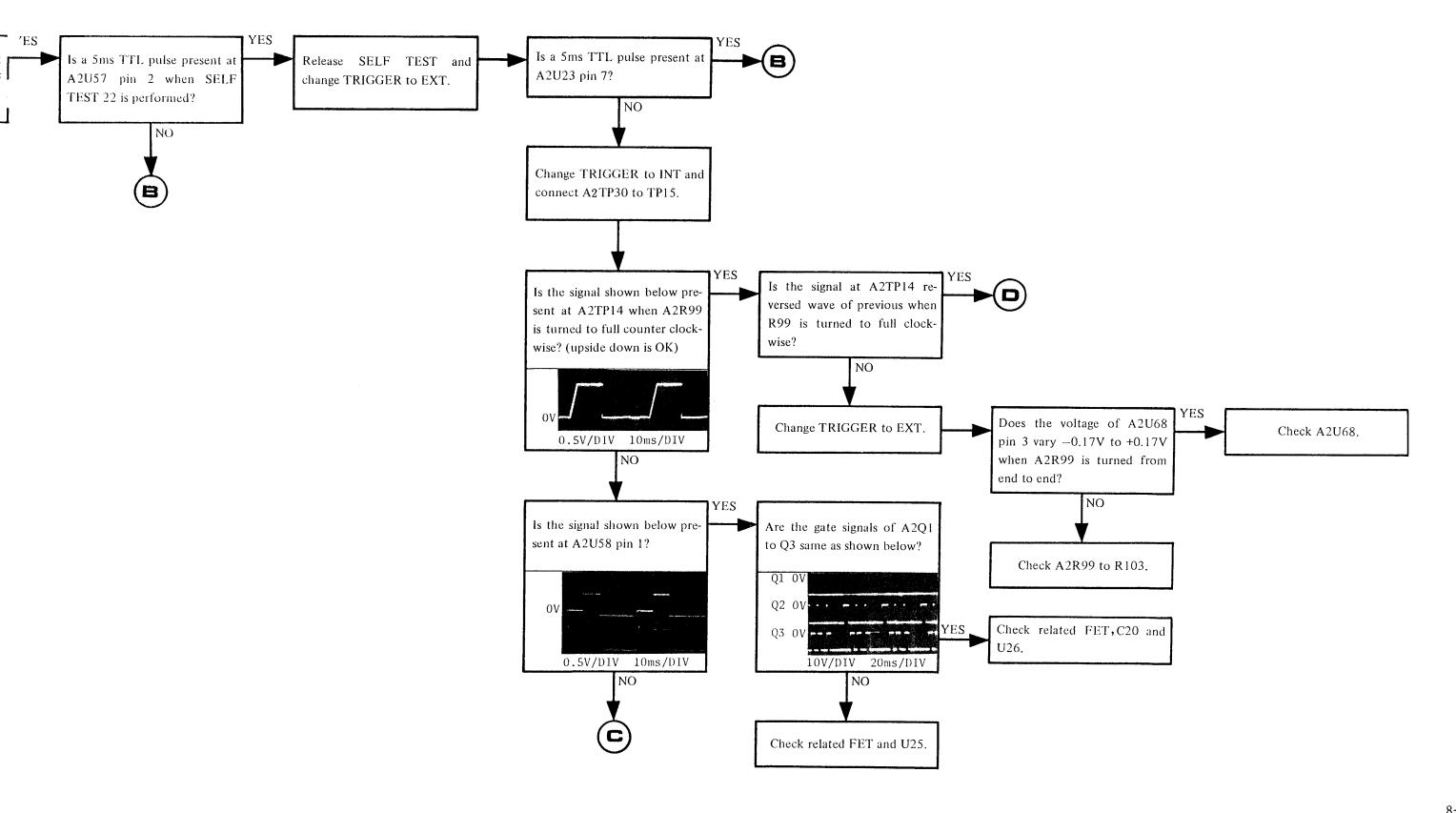




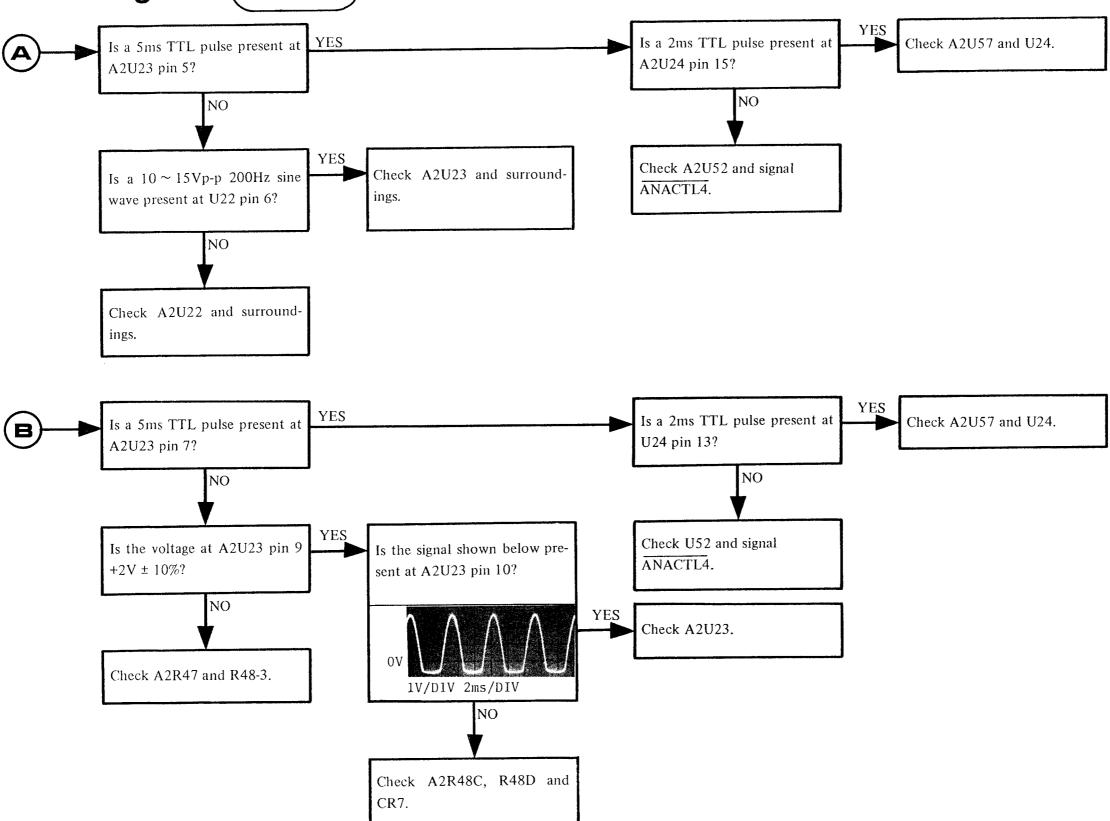


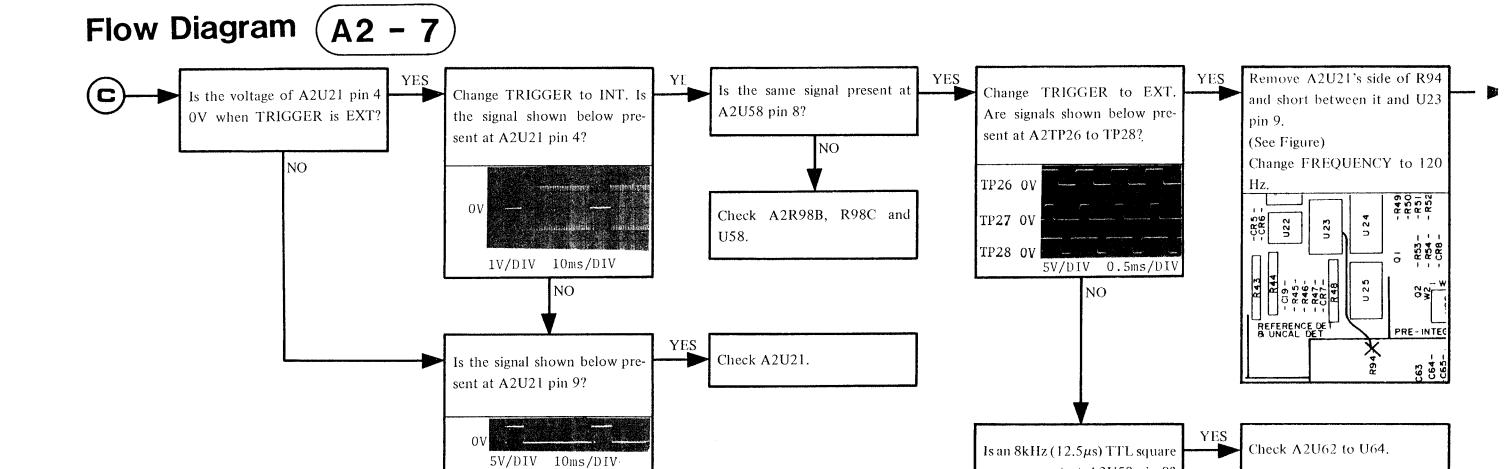






## Flow Diagram (A2 - 7)





NO

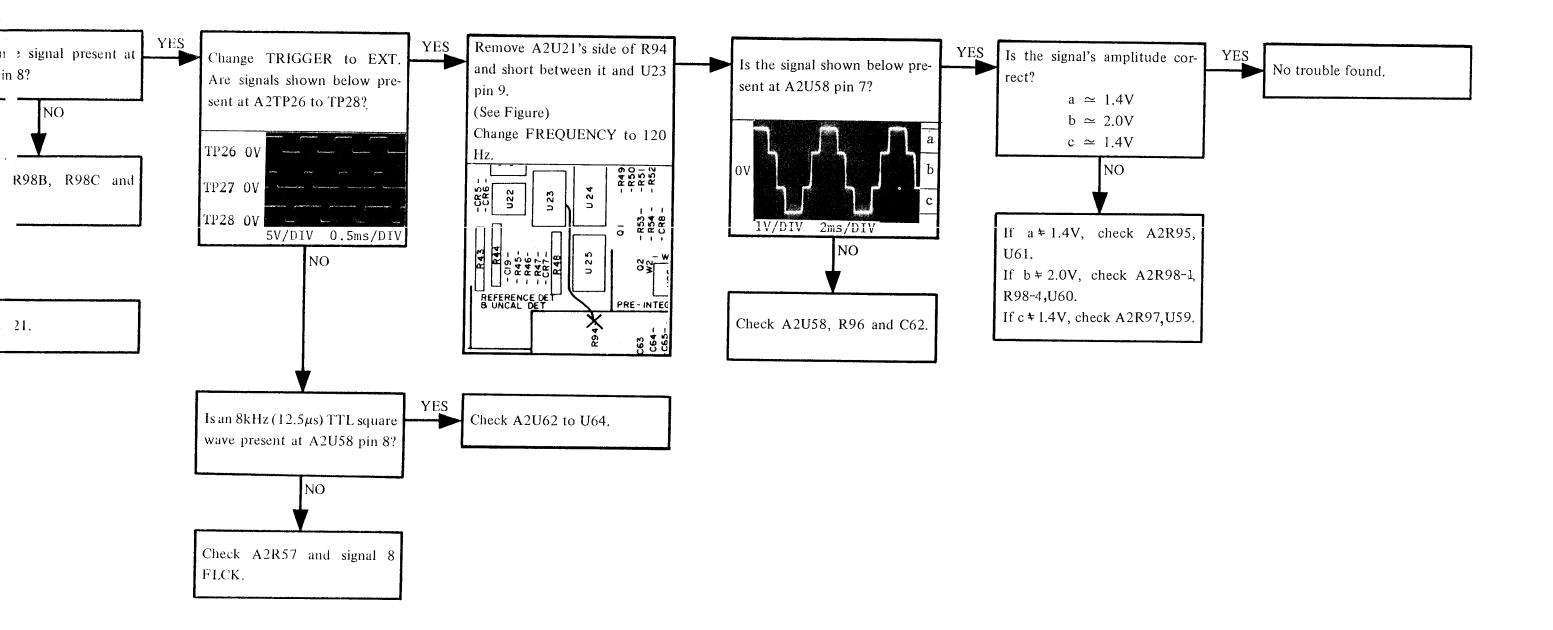
Check A2U50.

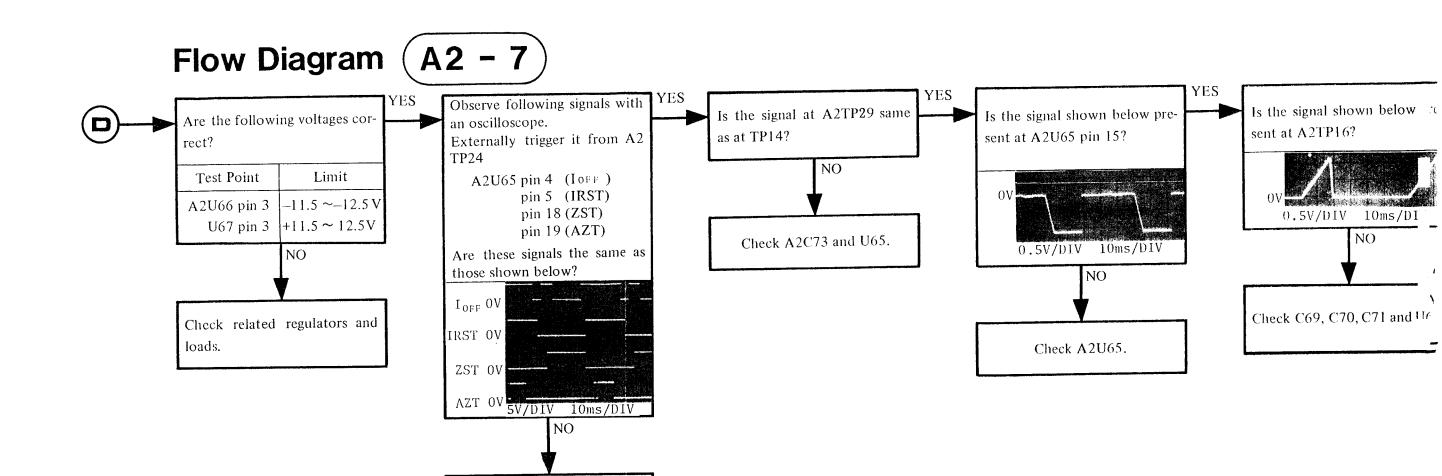
wave present at A2U58 pin 8?

Check A2R57 and signal 8

FLCK.

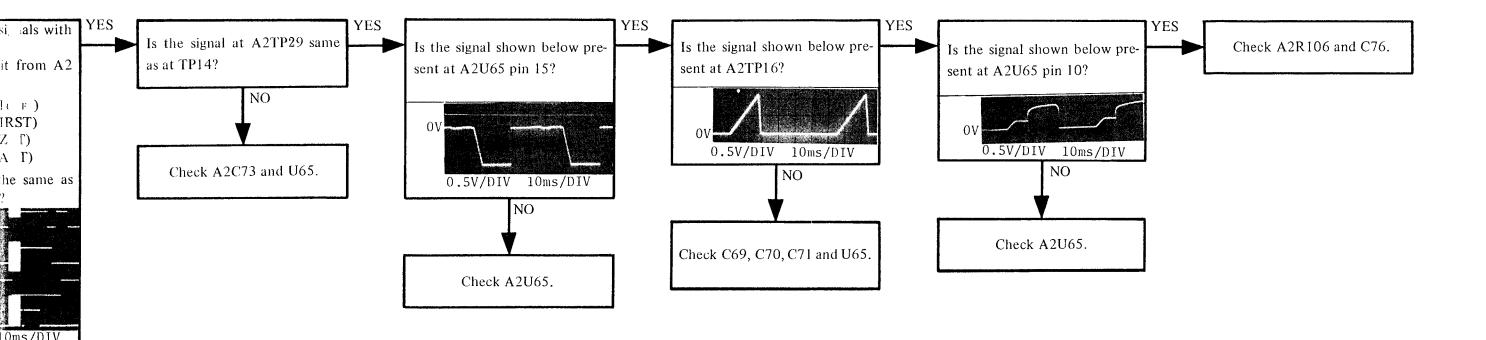
NO





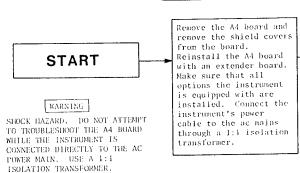
Check A2U52. If I off is ab-

normal, also check U18.



lc. = is ab-

#### Flow Diagram (A4 - 1)



Is the voltage at the

junction of A4F1 and

A4CR21 (reference is

is turned on?

them.

A4TP13) approximately +12V when the instrument

Check A4F1, F2, F3,

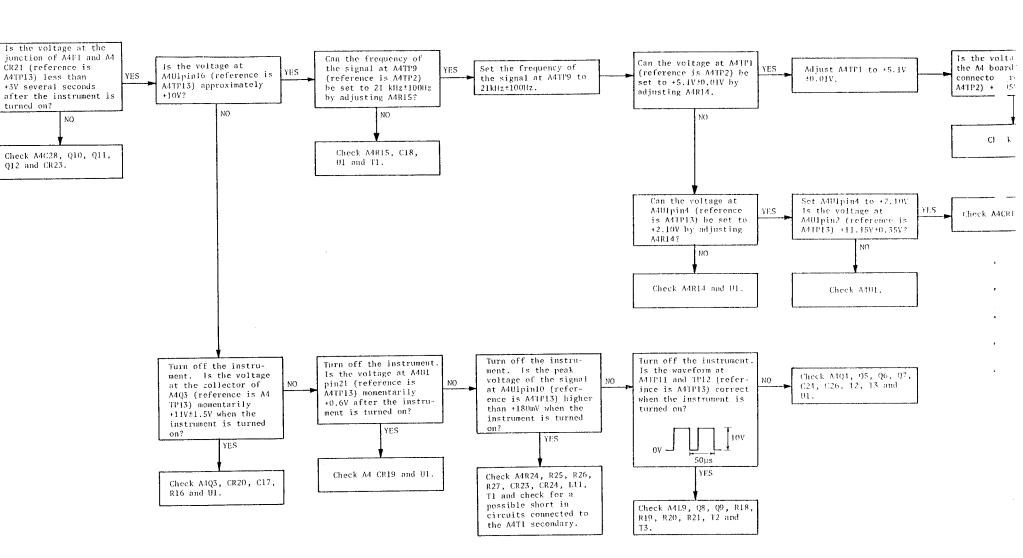
and the parts around

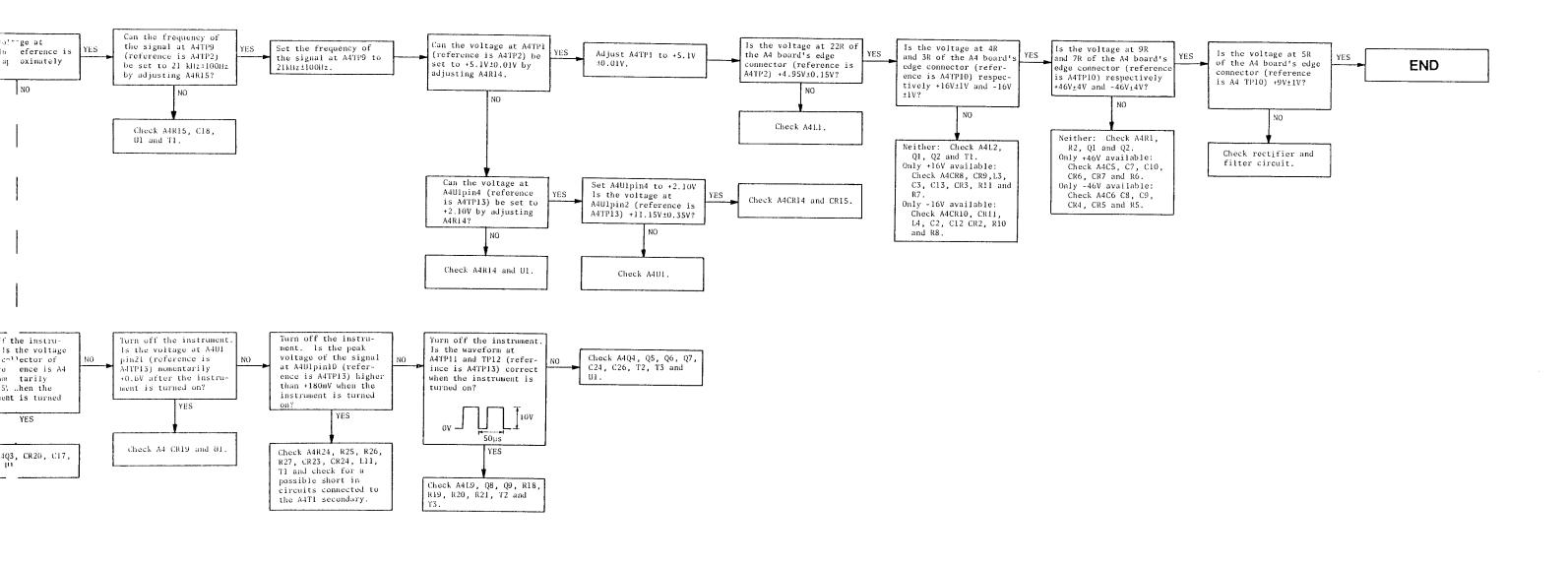
C28, Q10, Q11, Q12

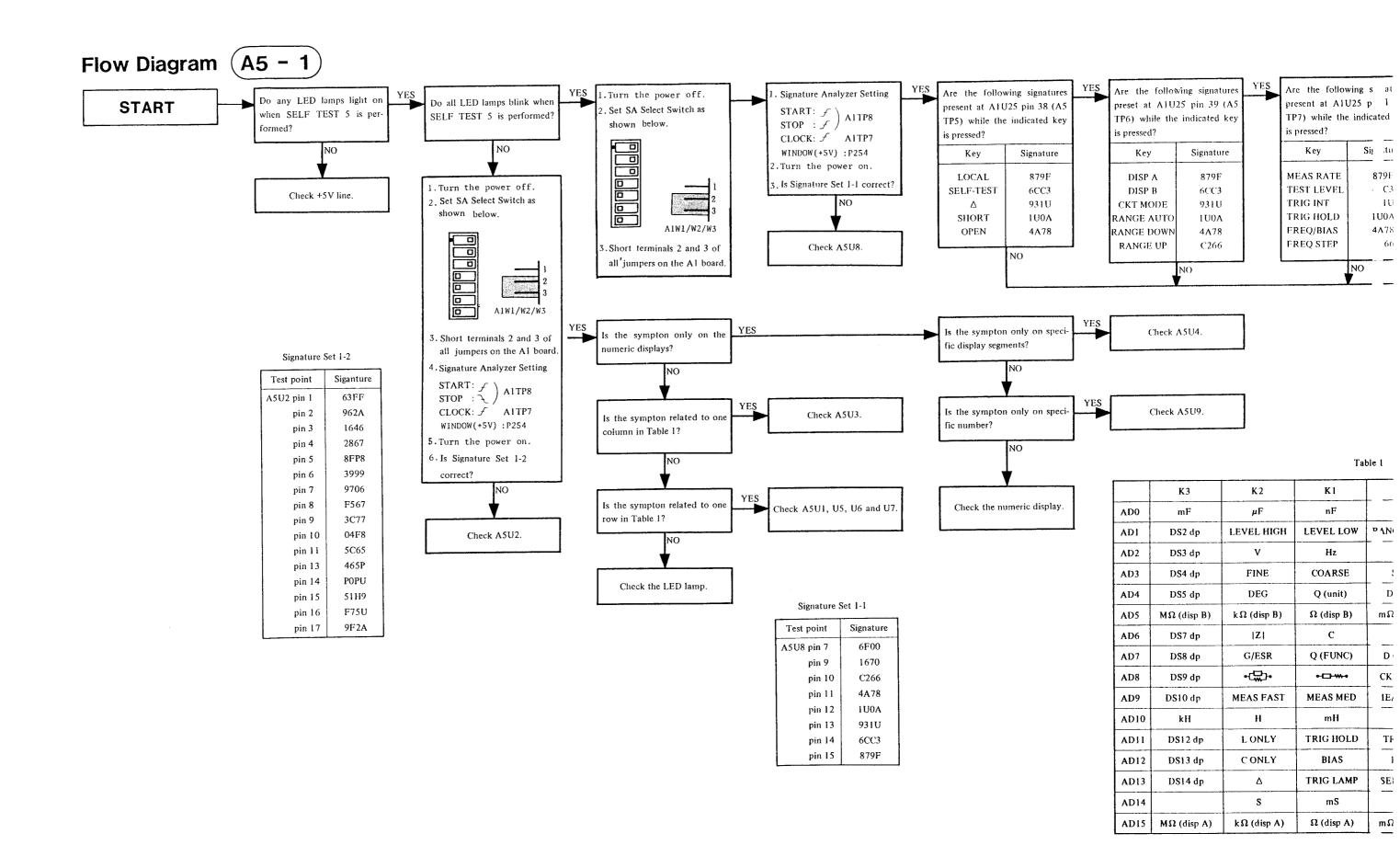
A4TP13) less than

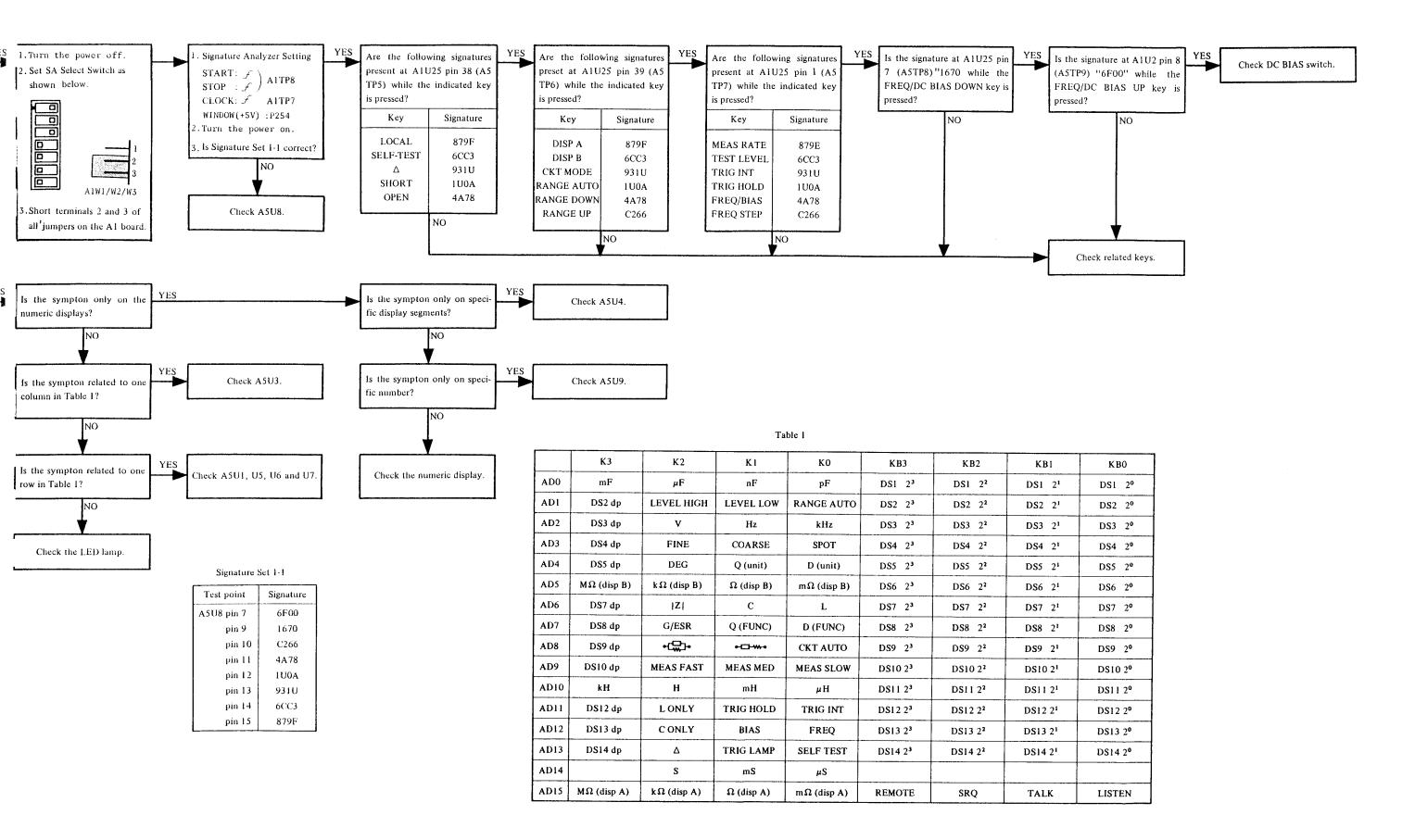
turned on?

Q12 and CR23.

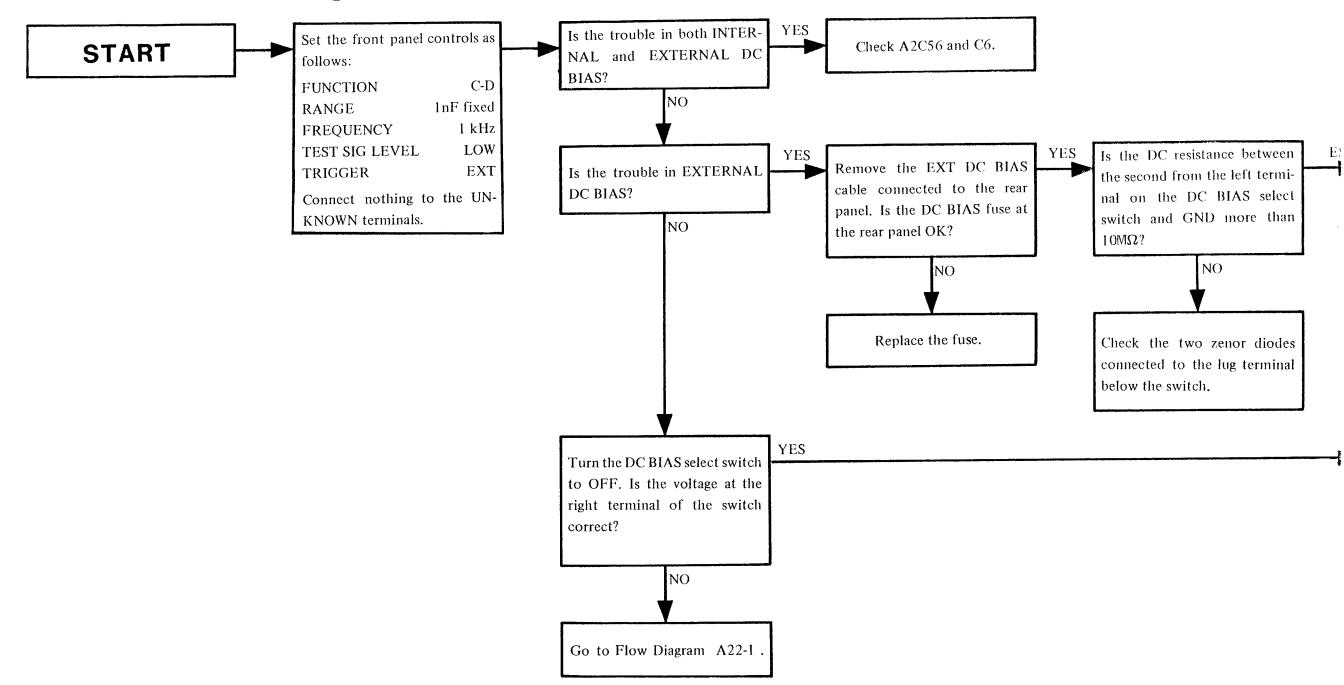




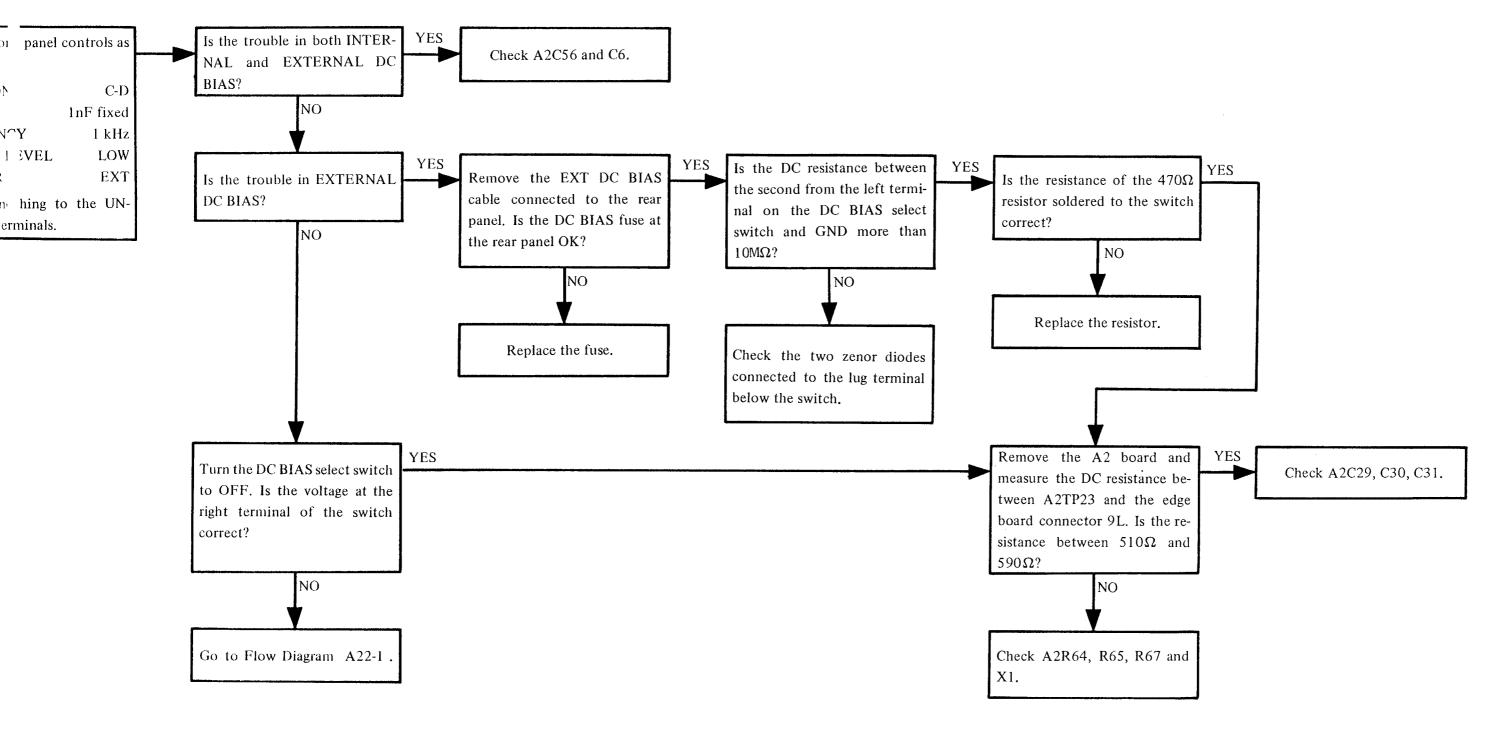


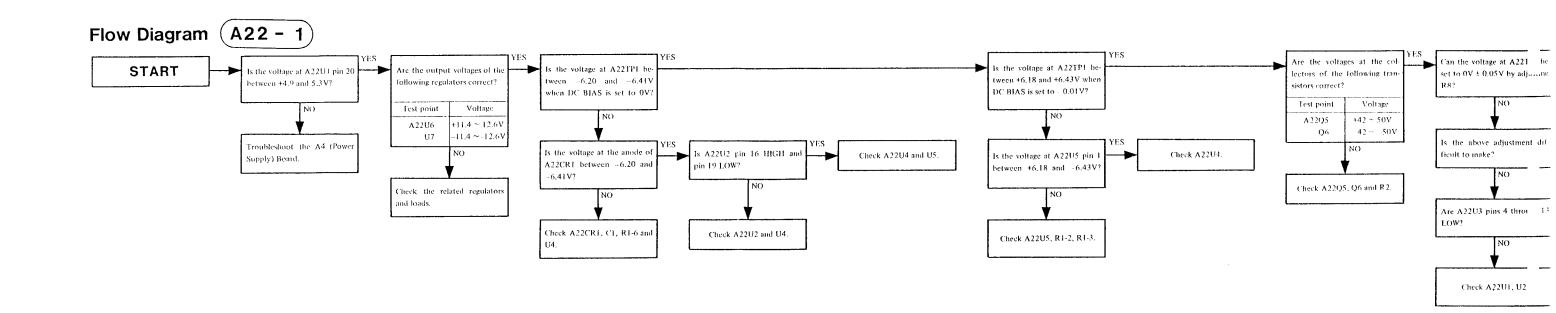


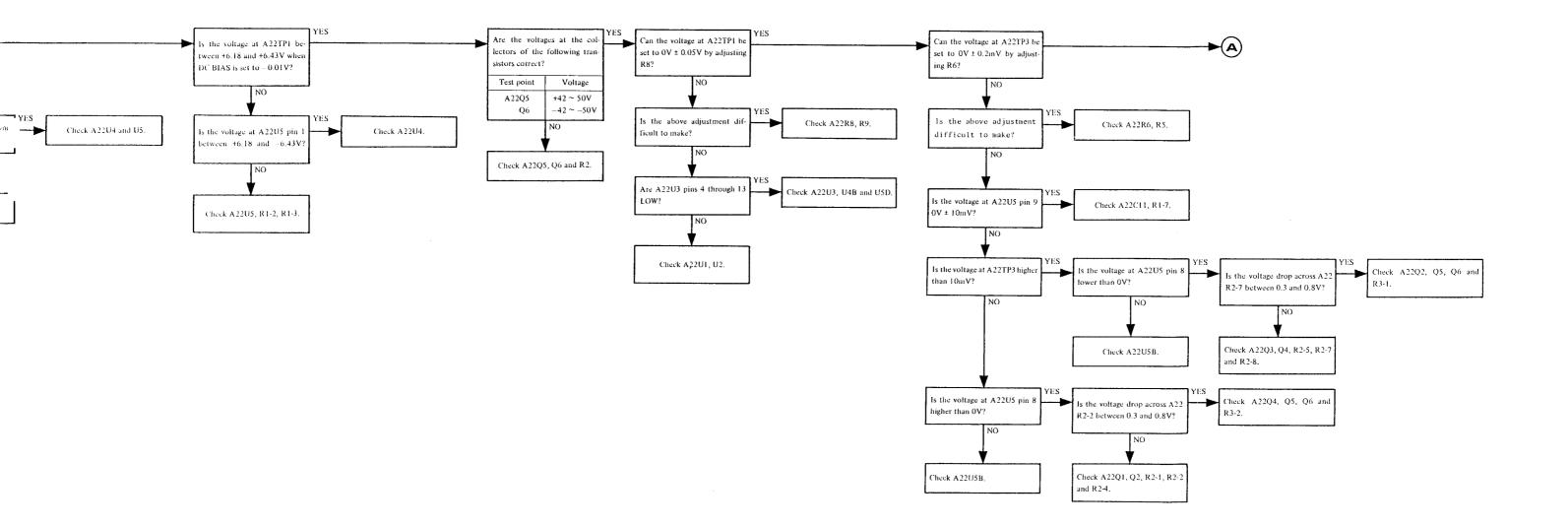
### DC BIAS Flow Diagram

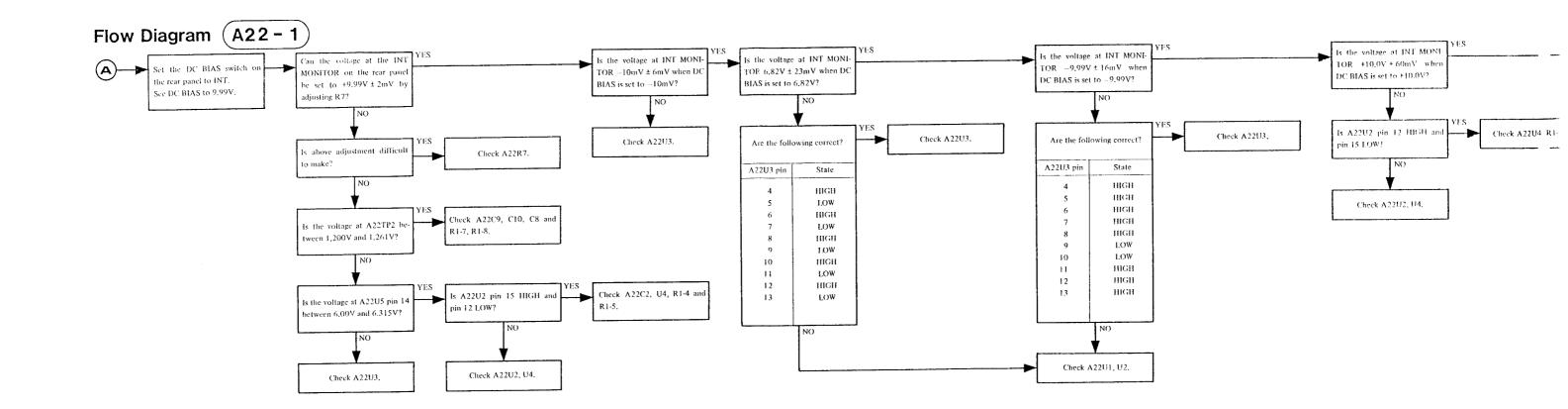


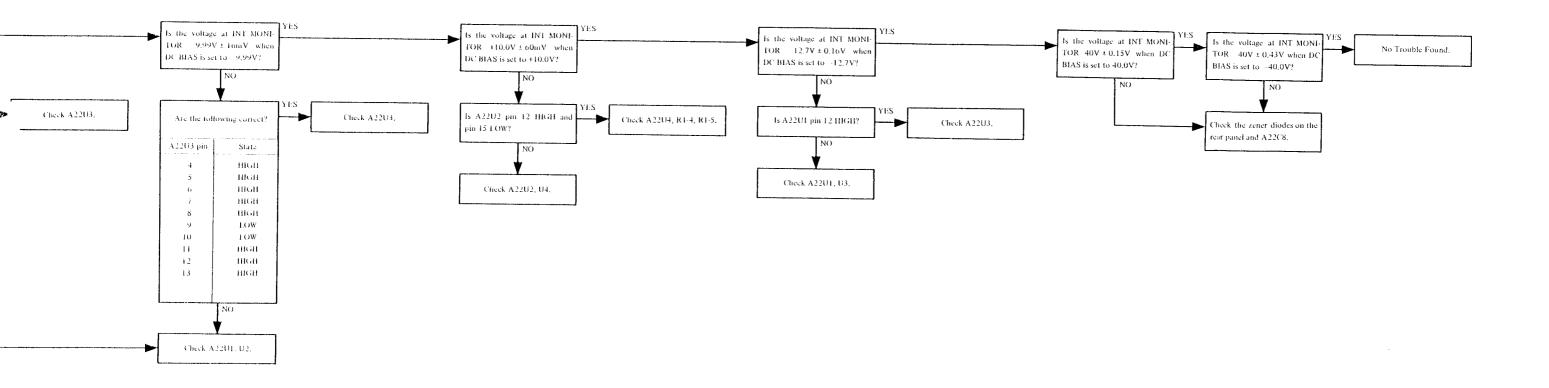
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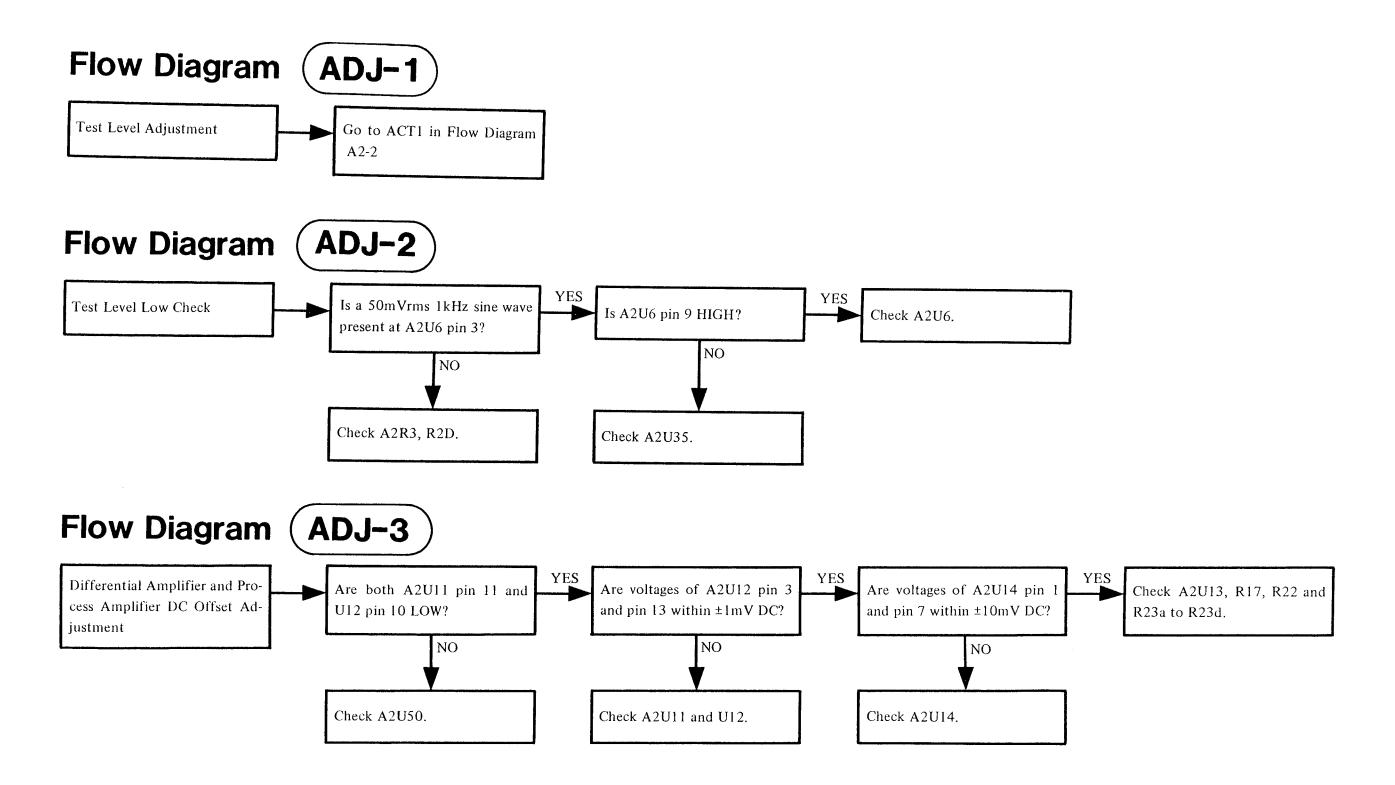


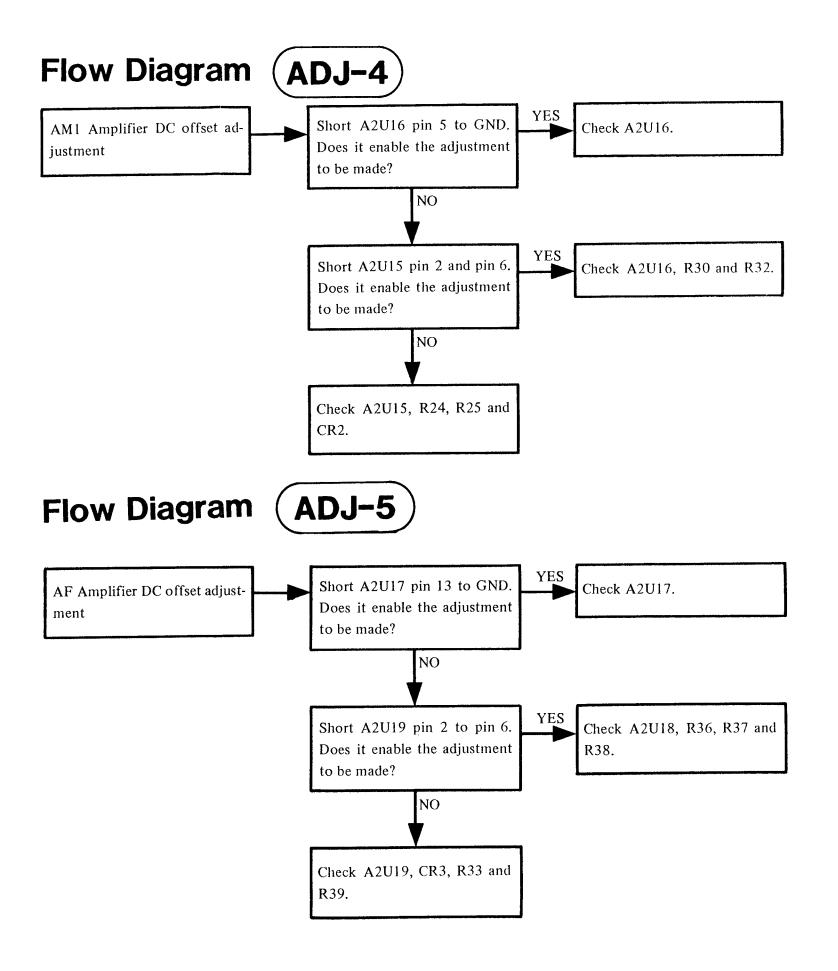


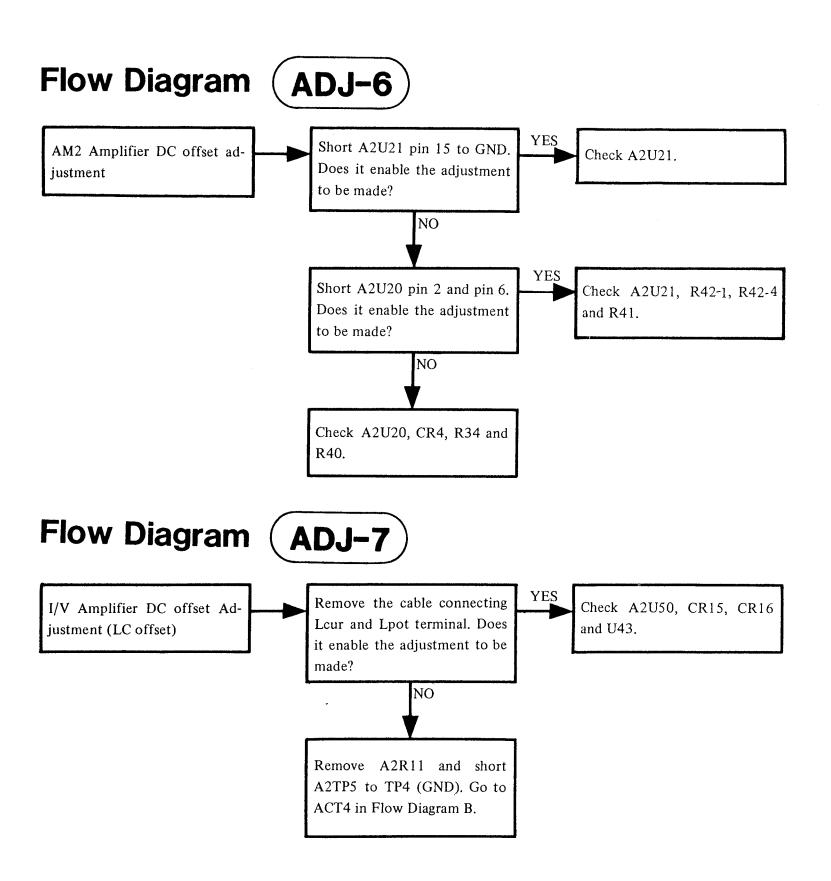


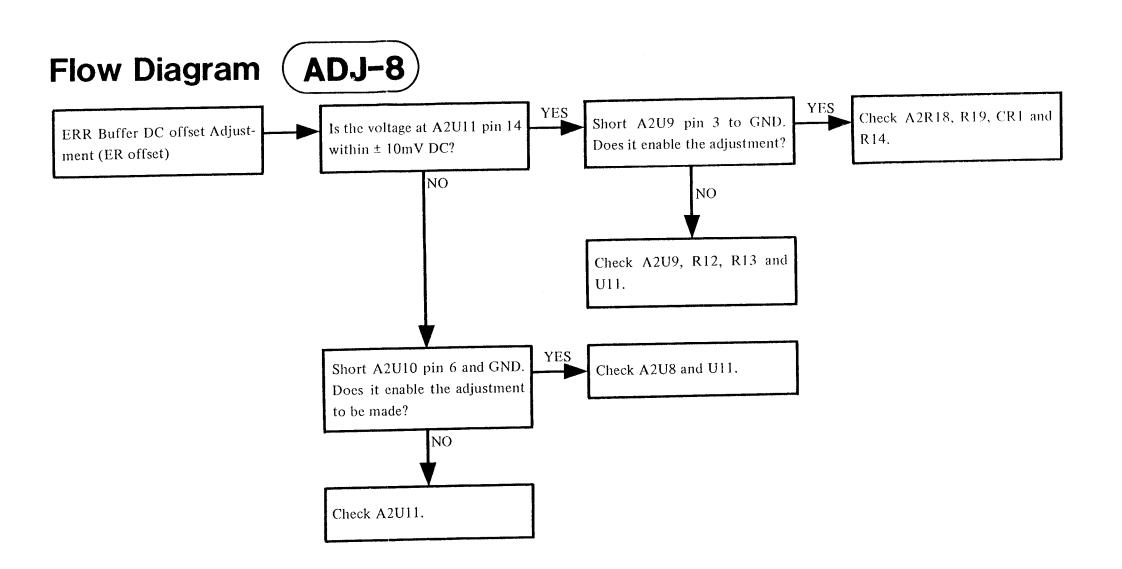




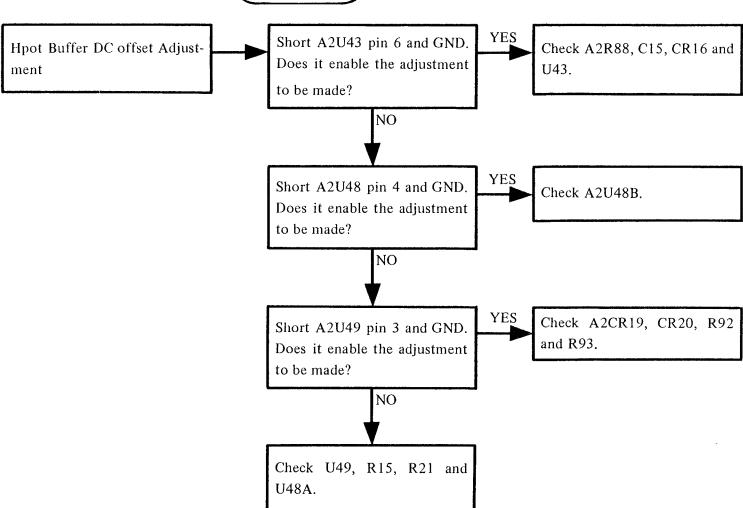


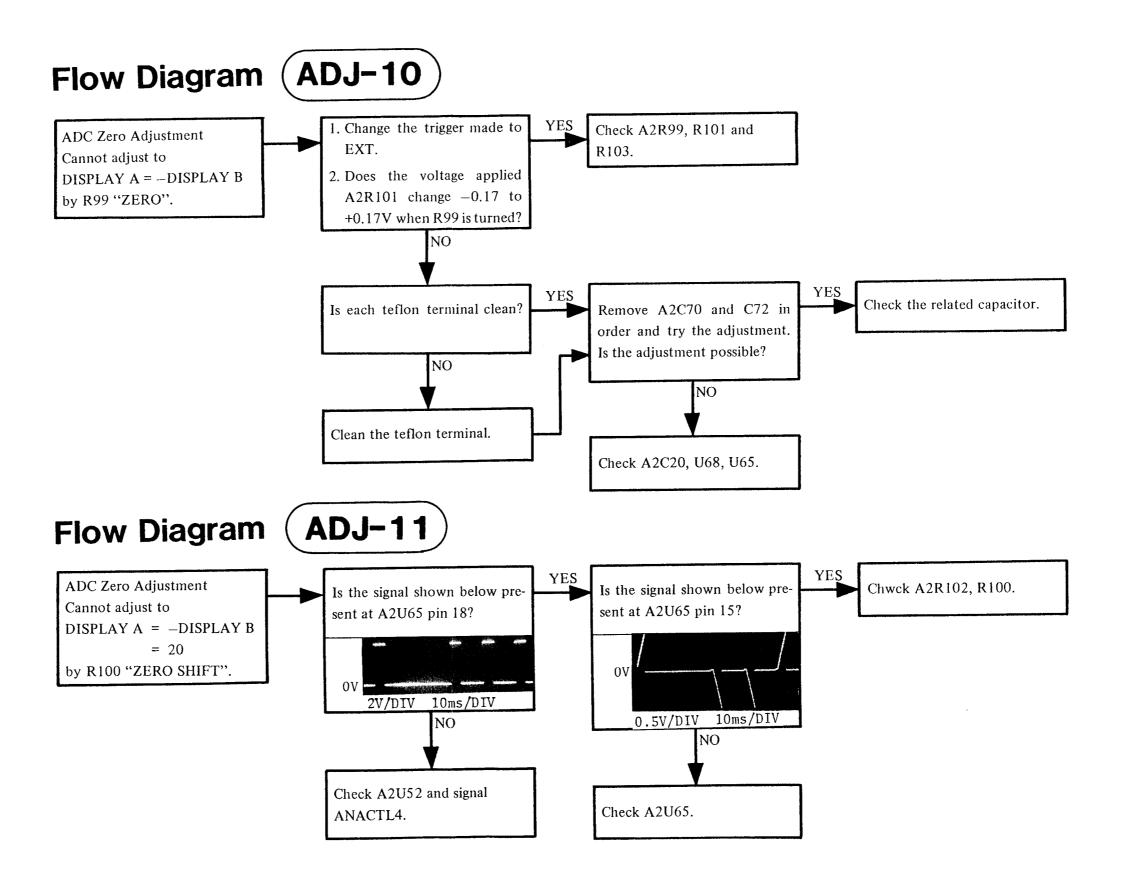


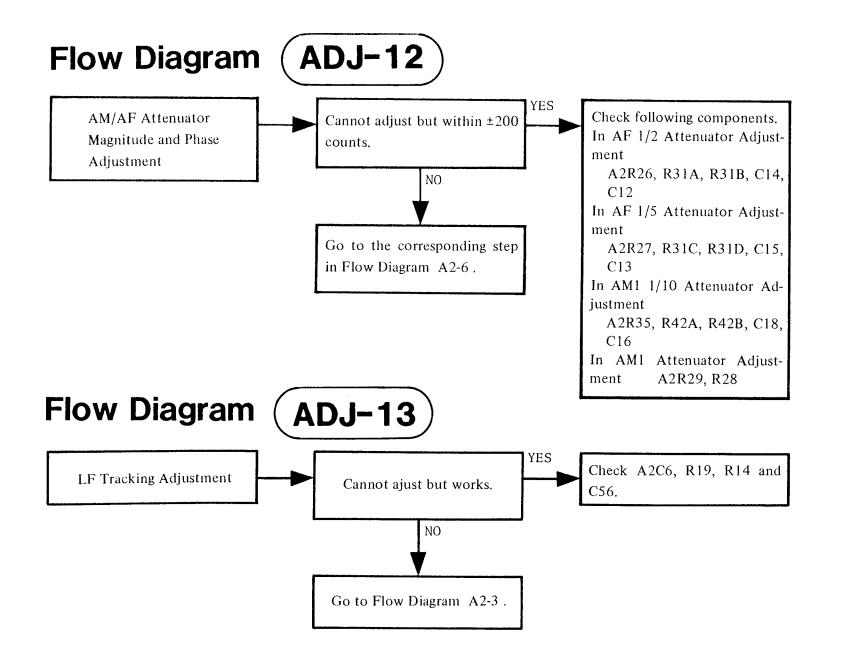


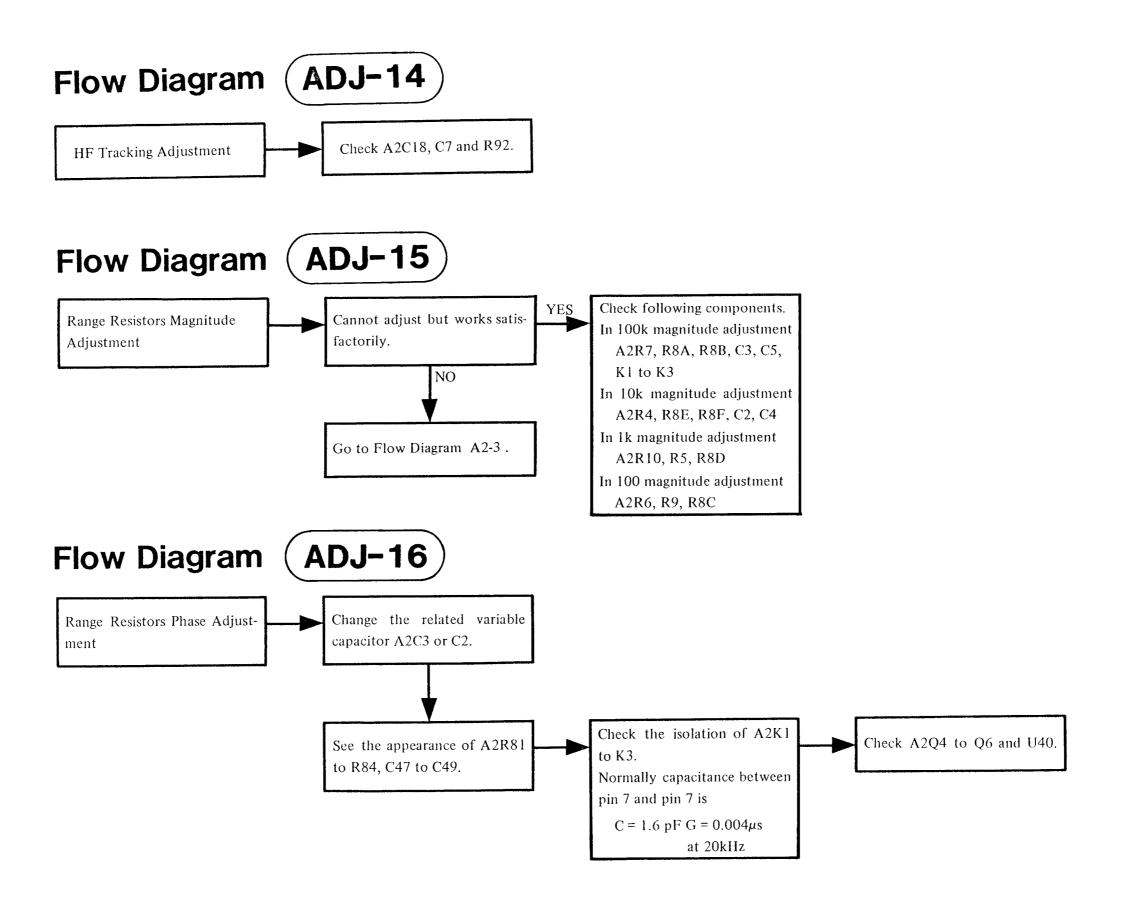


# Flow Diagram ADJ-9









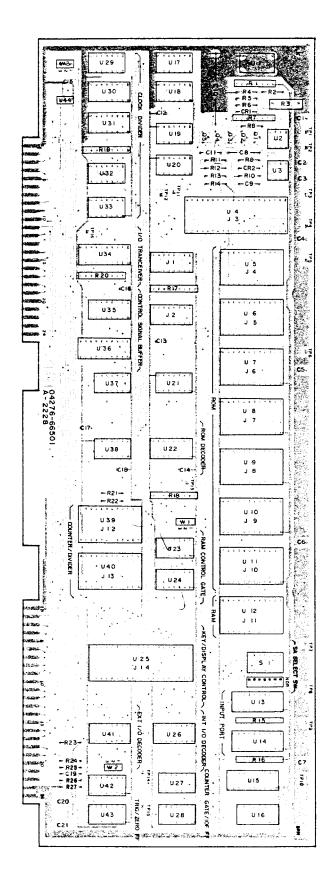
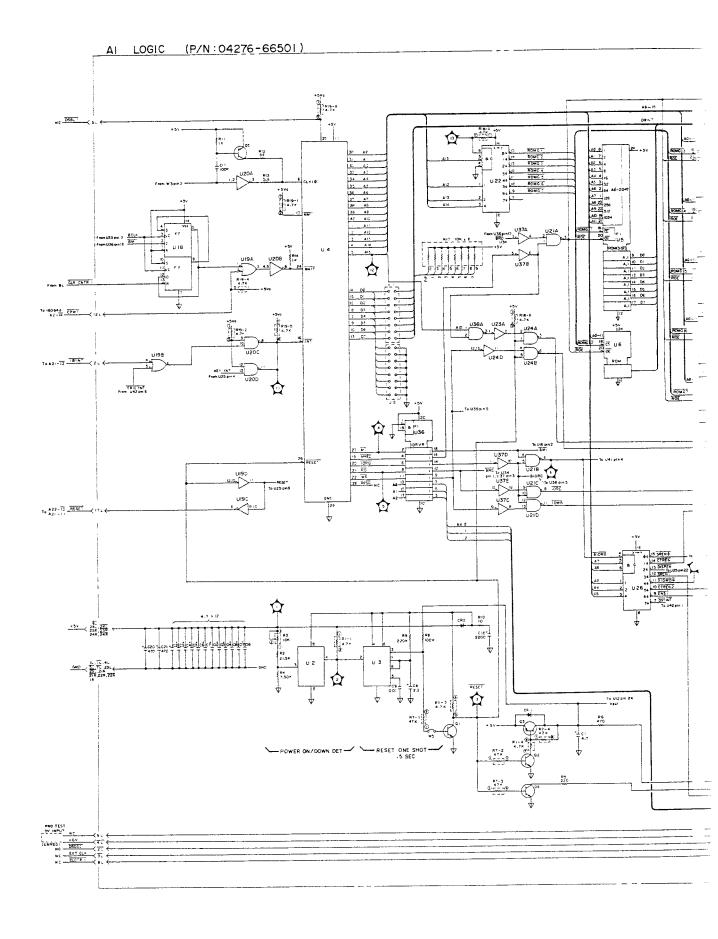
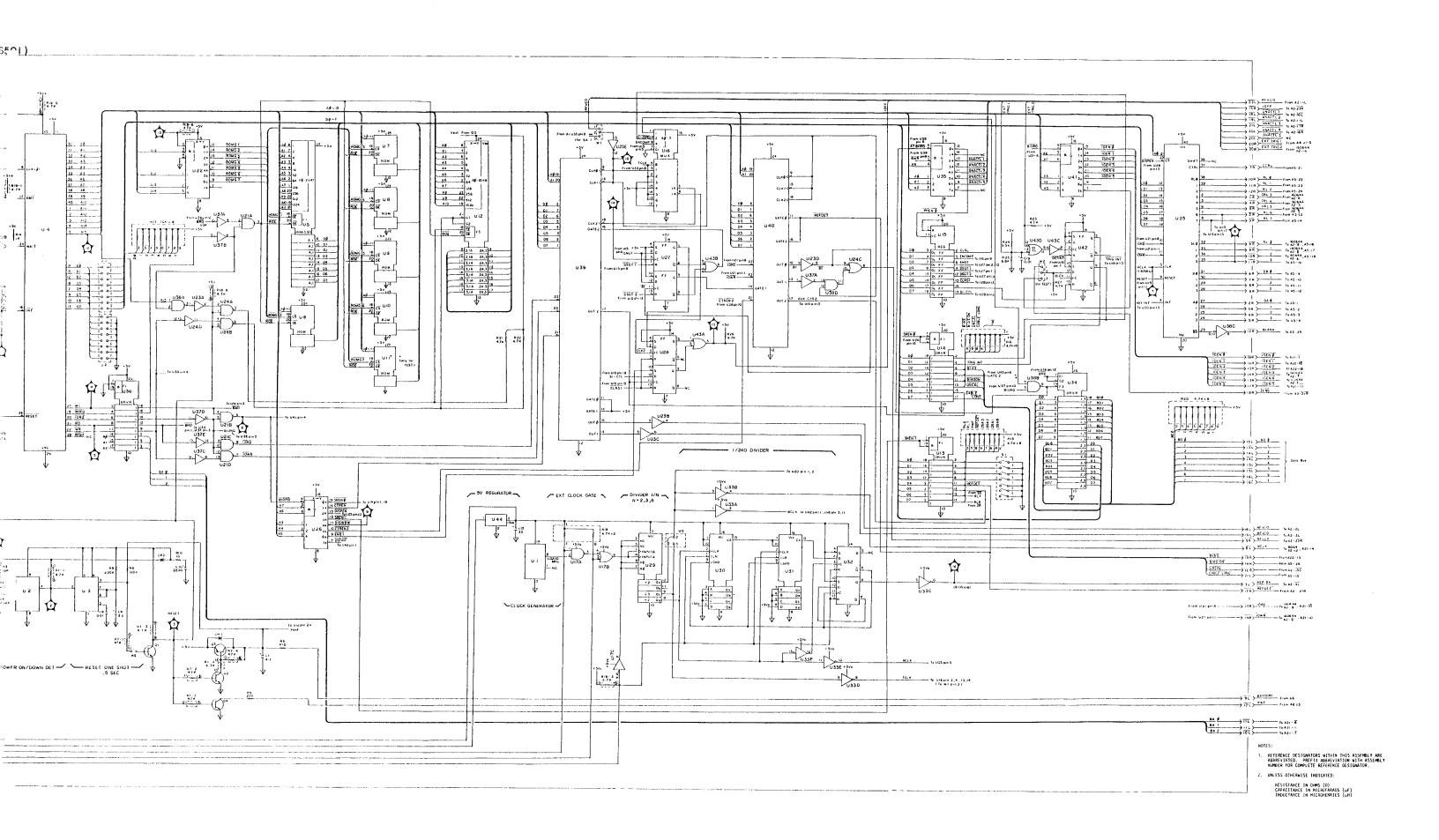


Figure 8-XX. Al LOGIC Board Assembly Component Locations.





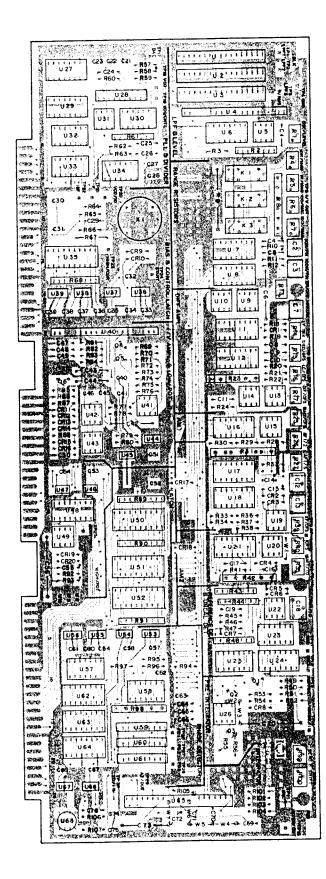
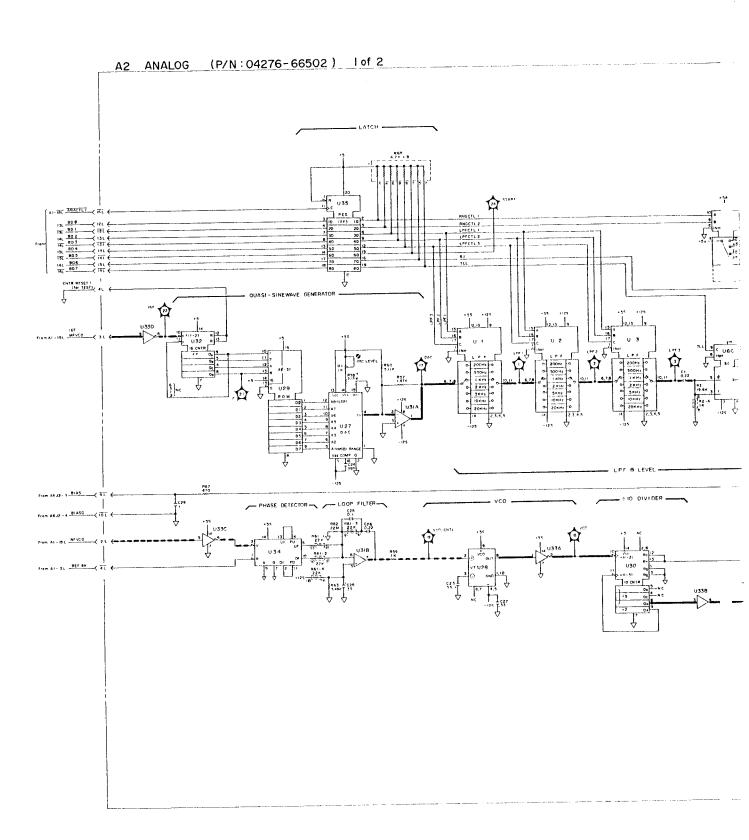
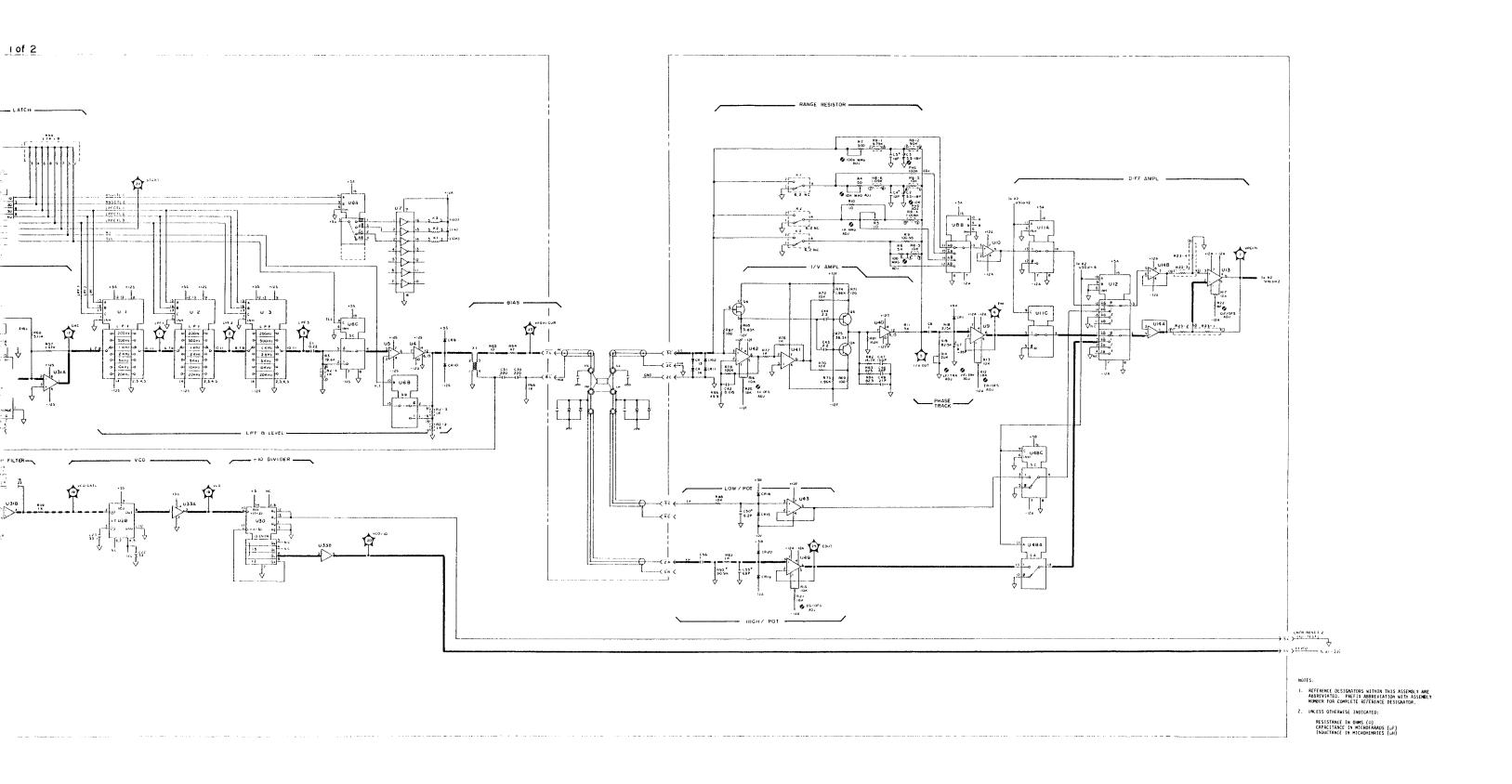


Figure 8-XX. A2 ANALOG Board Assembly Component Locations.





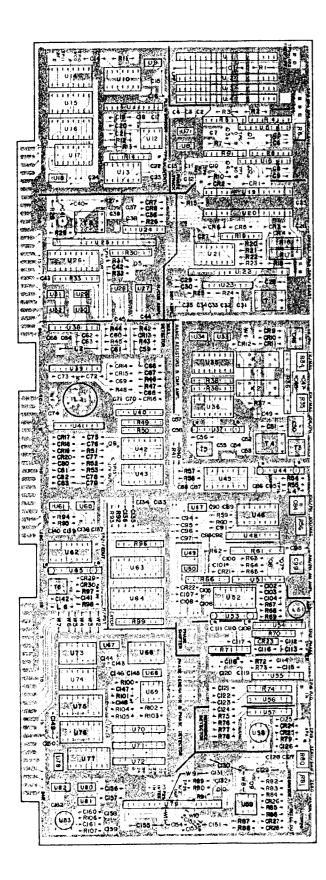
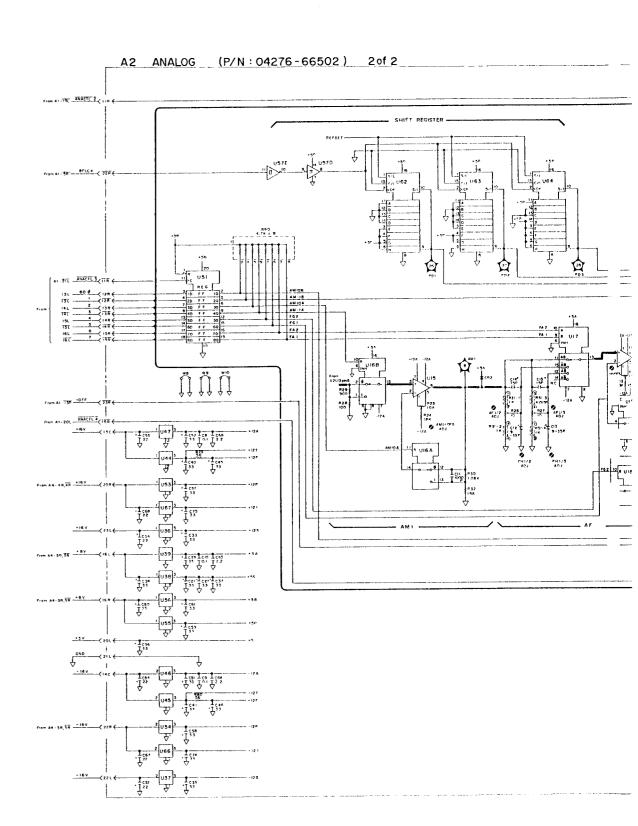
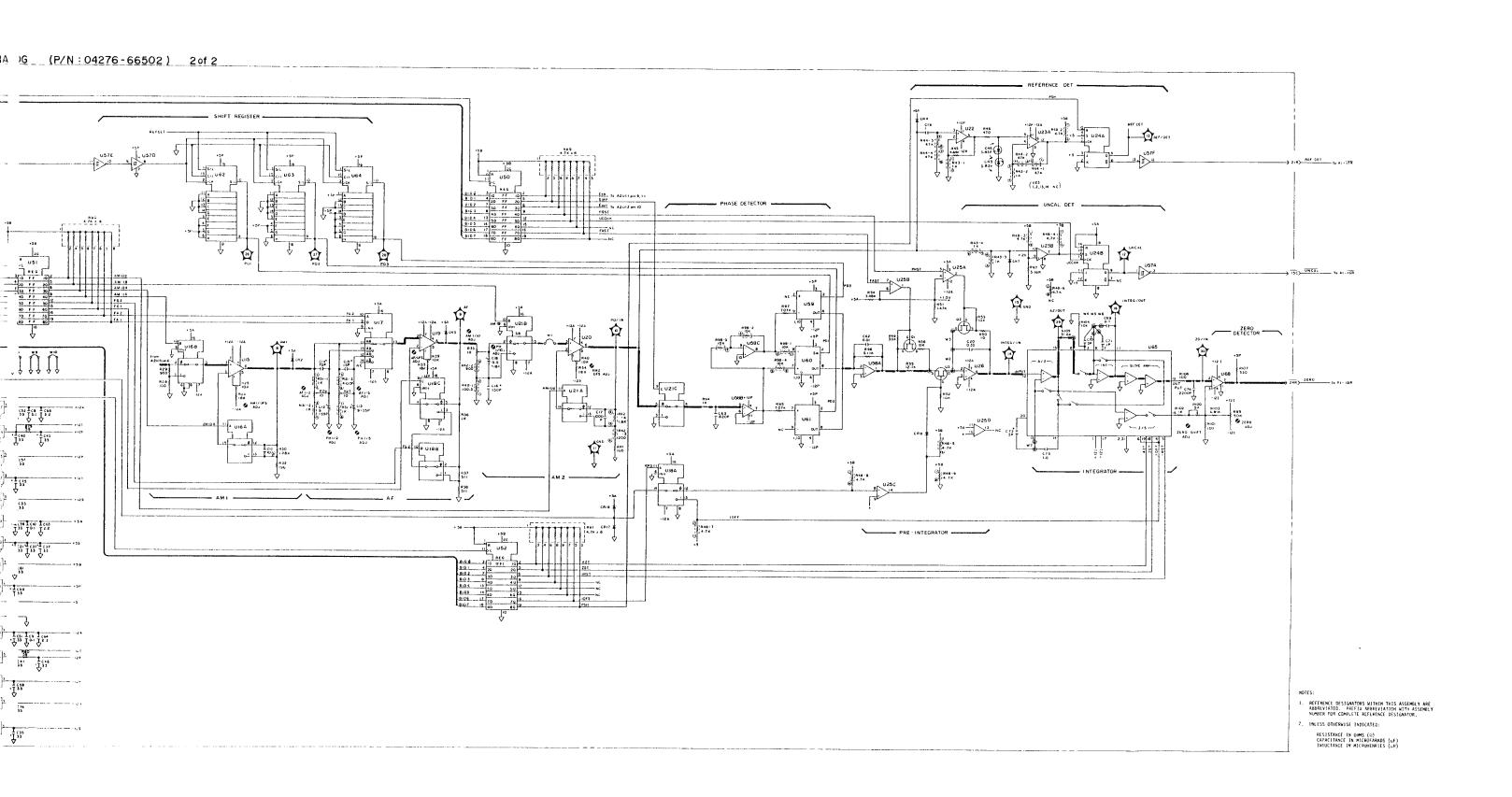


Figure 8-XX. A2 ANALOG Board Assembly Component Locations.





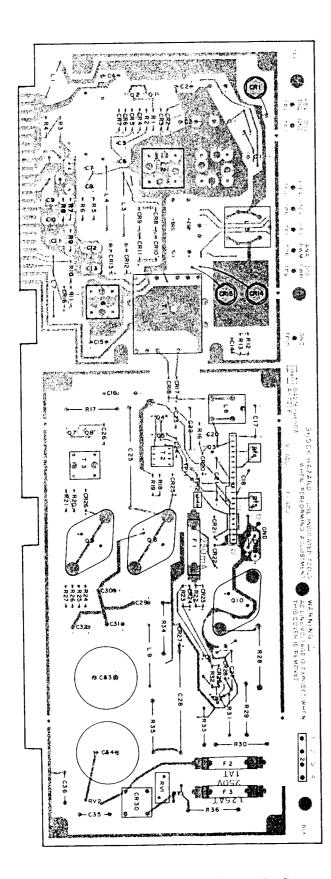
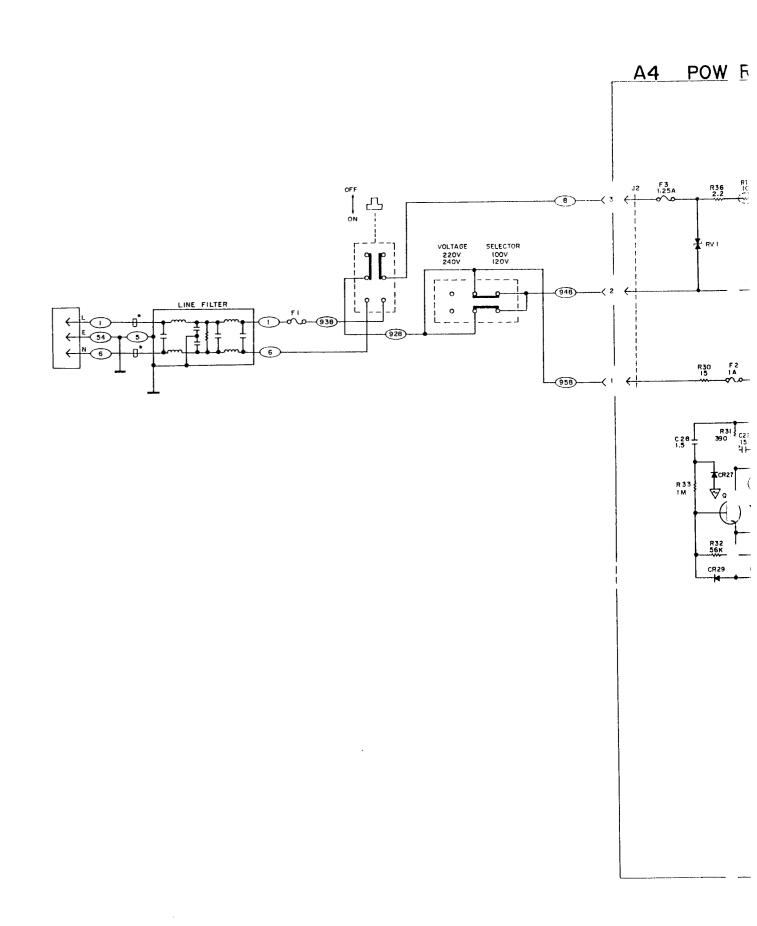
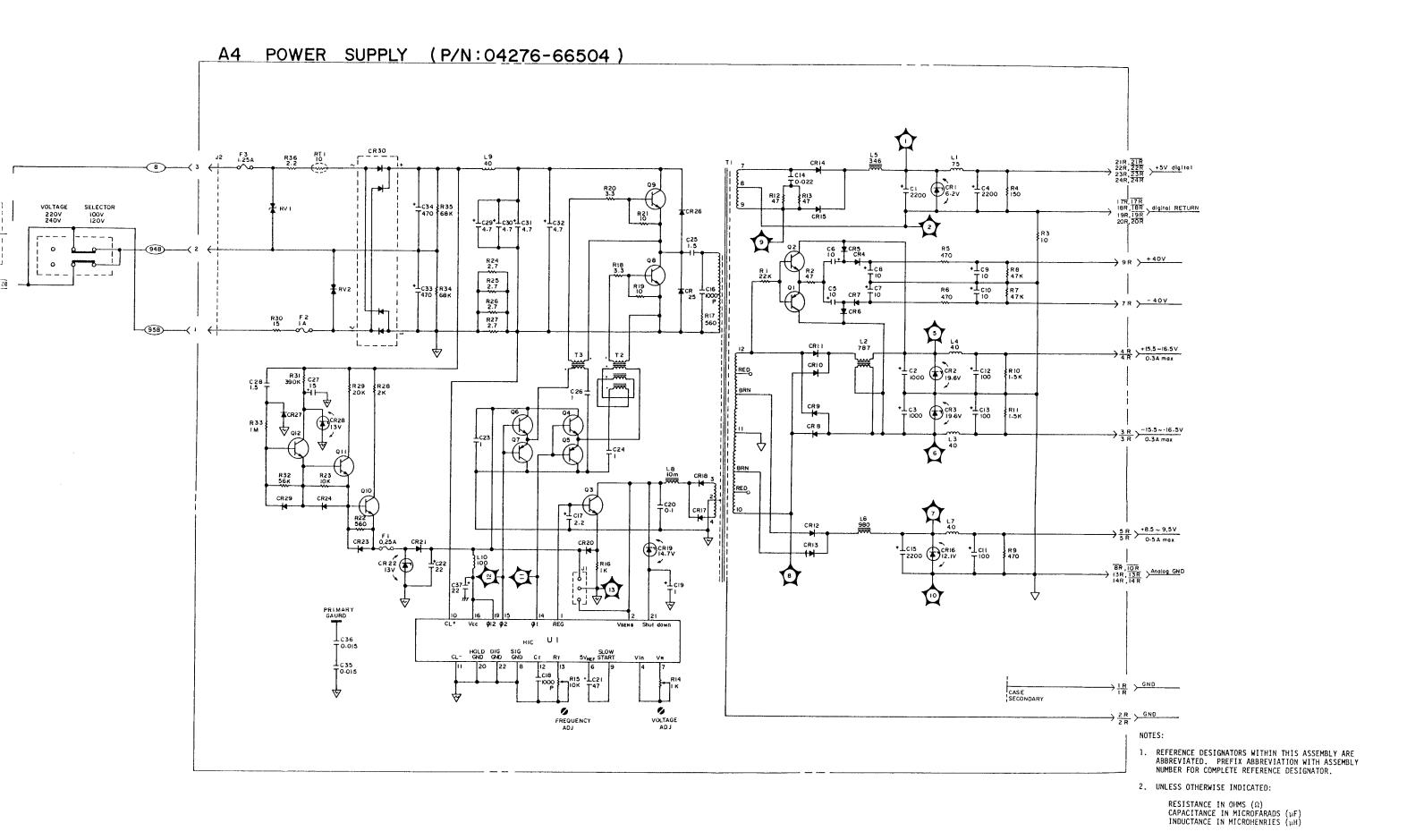


Figure 8-XX. A4 POWER SUPPLY Board Assembly Component Locations.





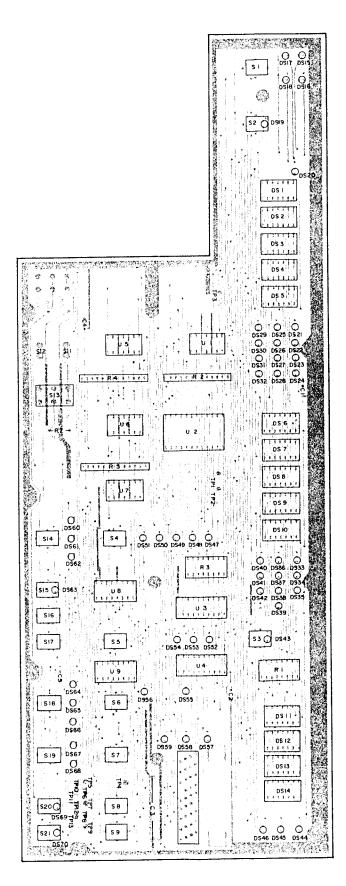
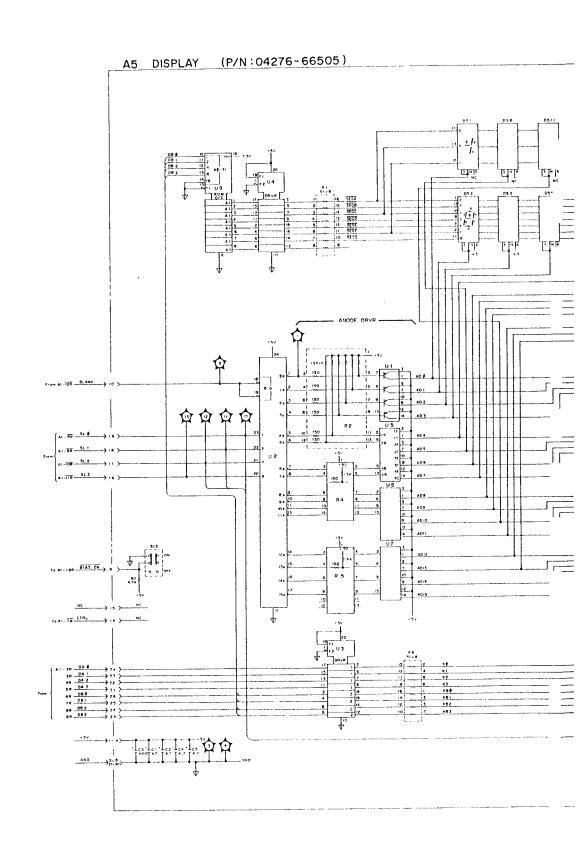
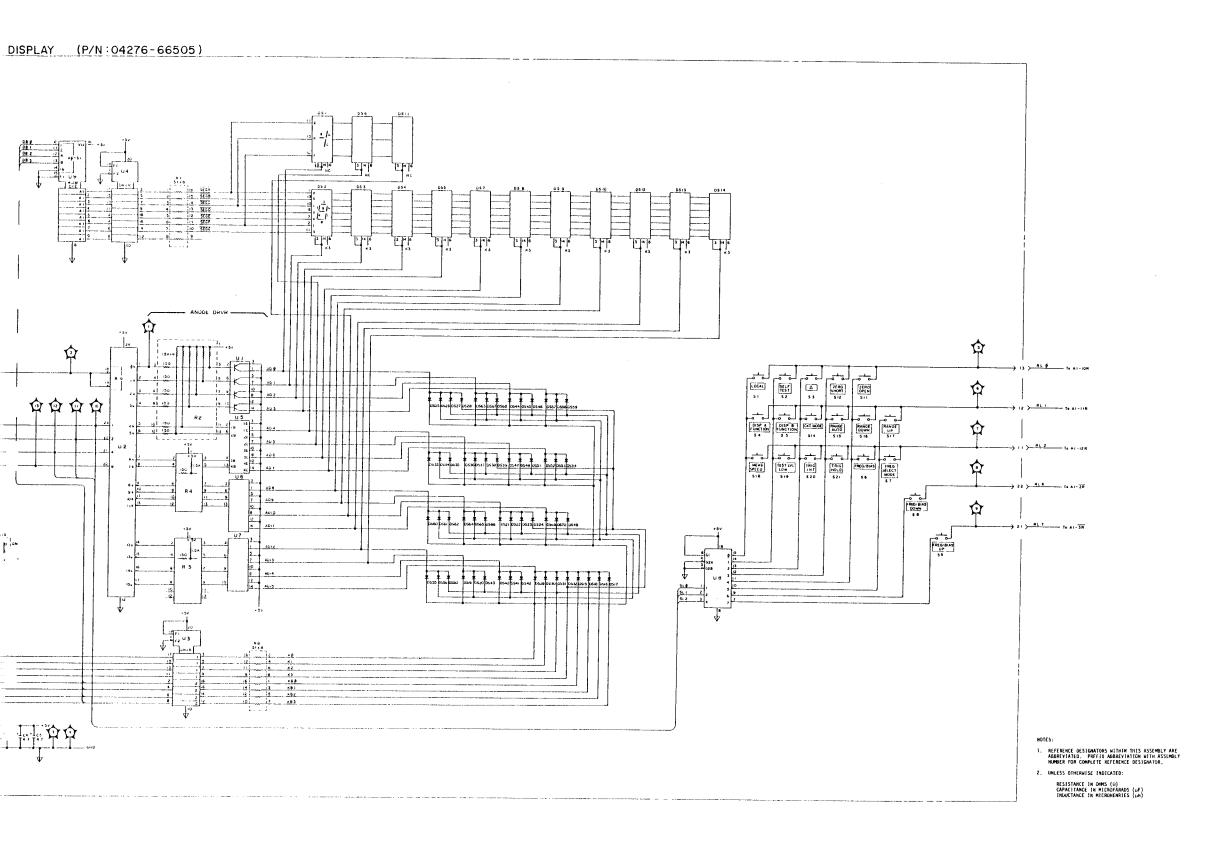
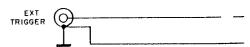
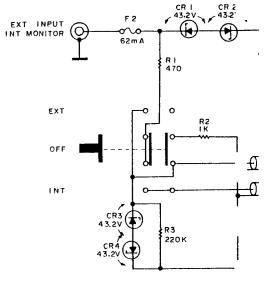


Figure 8-XX. A5 DISPLAY Board Assembly Component Locations.









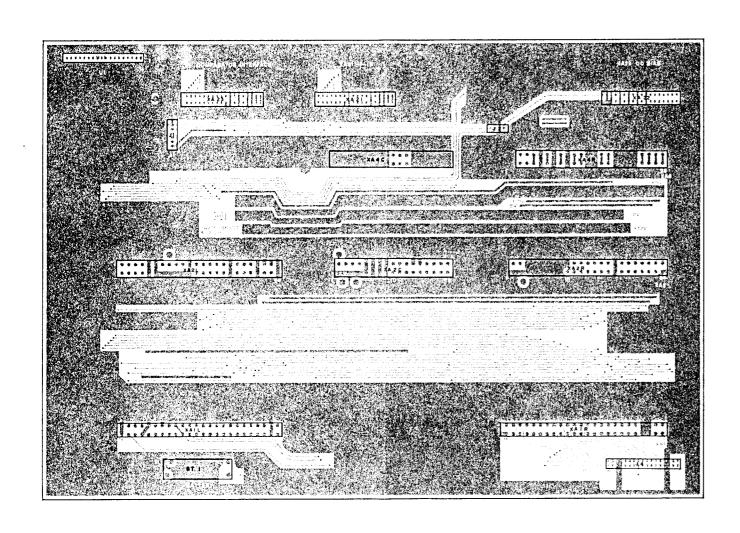
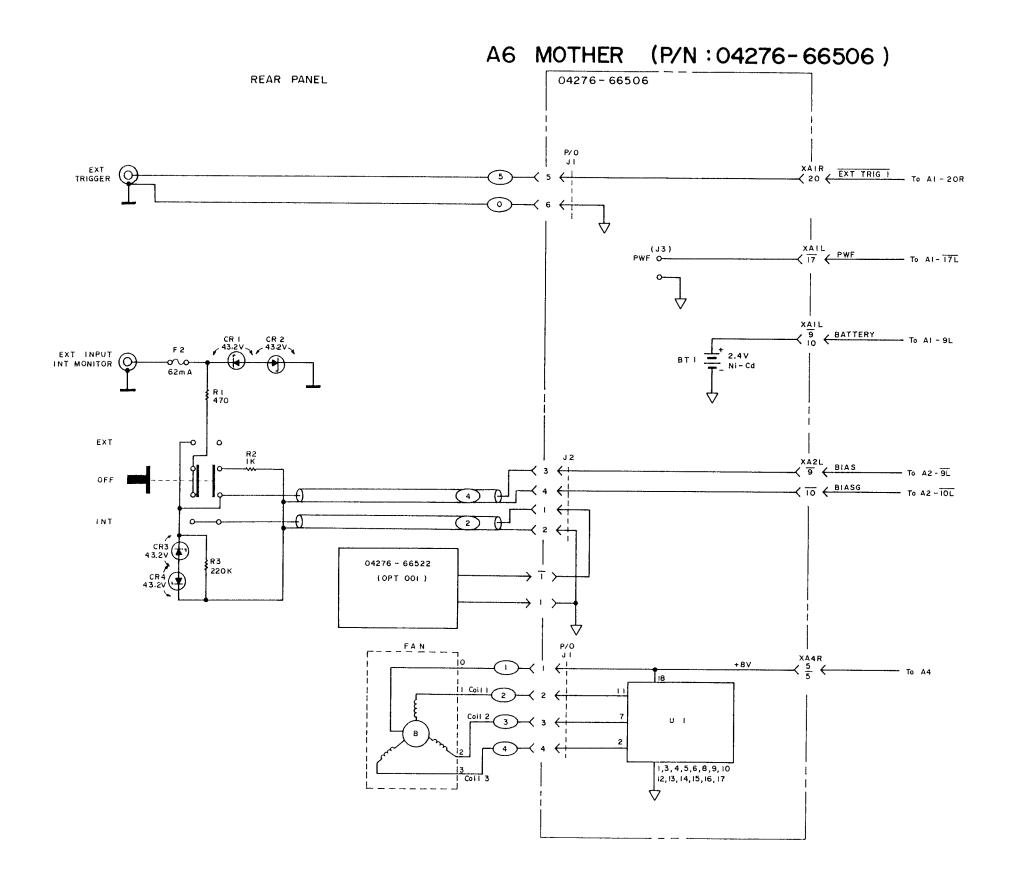


Figure 8-XX. A6 MOTHER Board Assembly Component Locations.



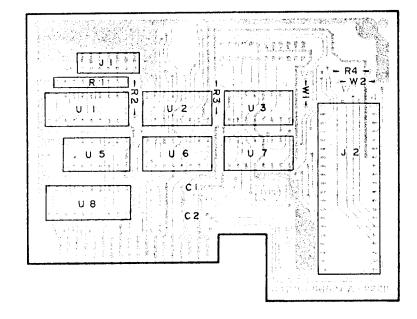
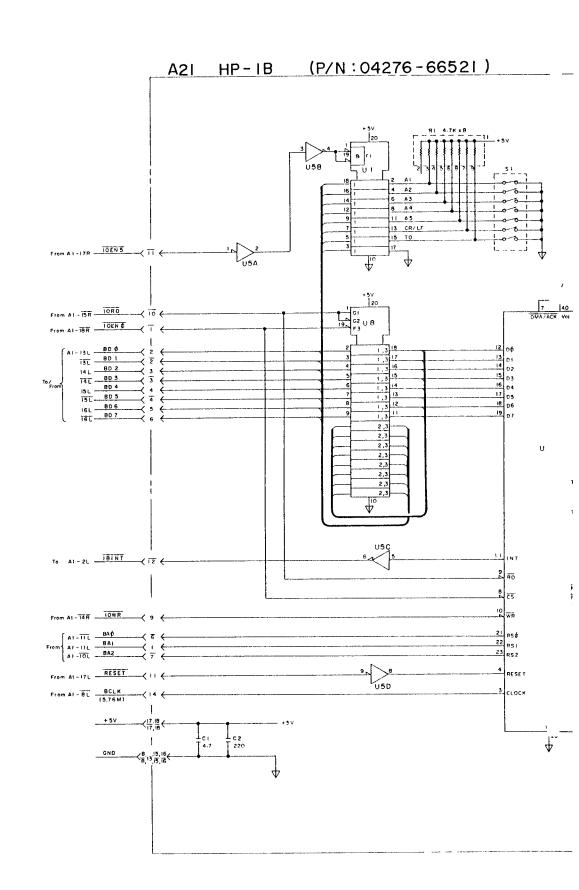
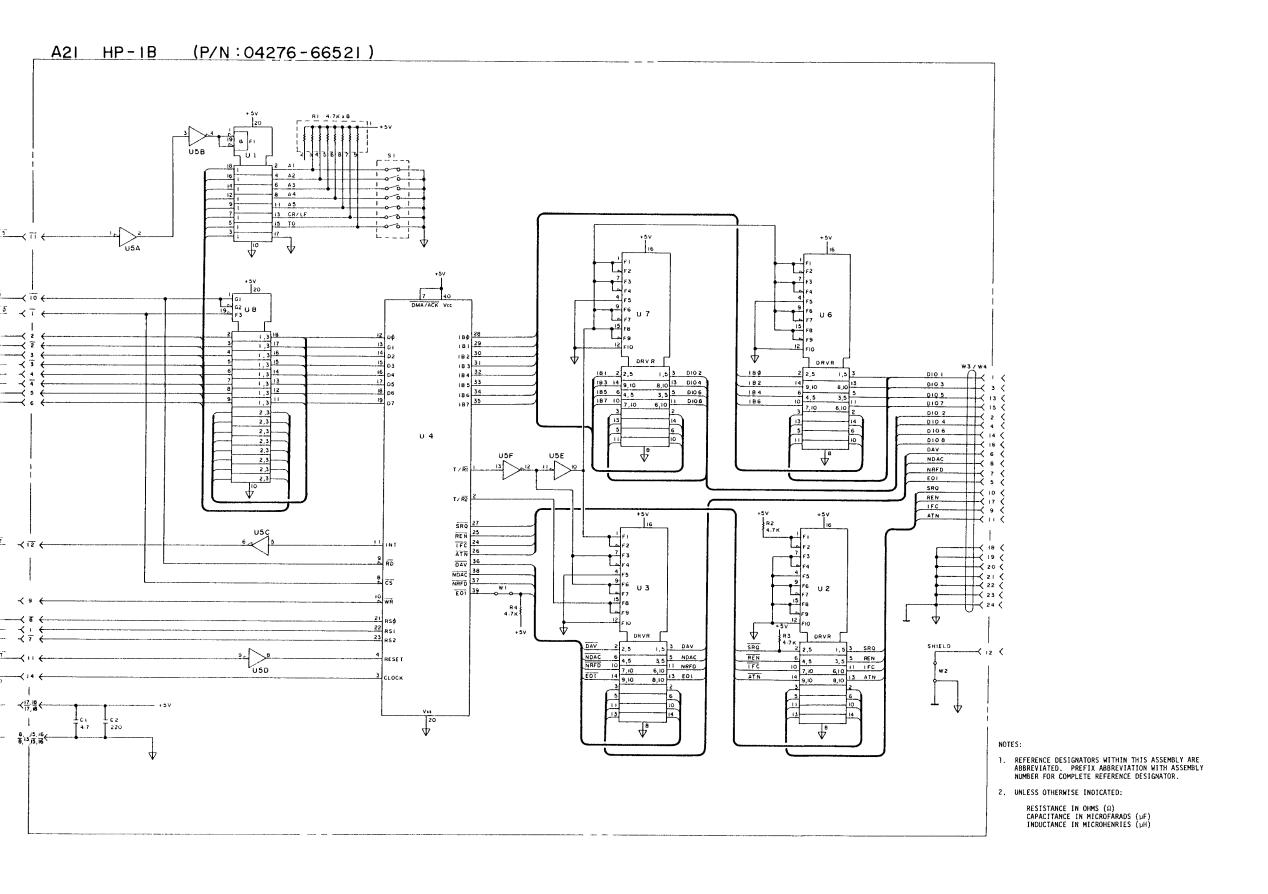


Figure 8-XX. A21 HP-IB Board Assembly Component Locations.





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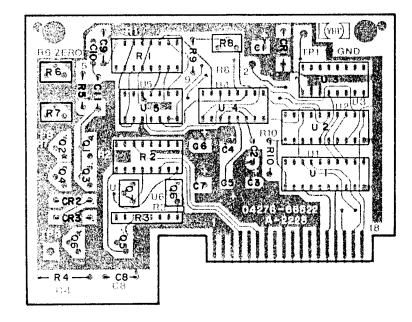
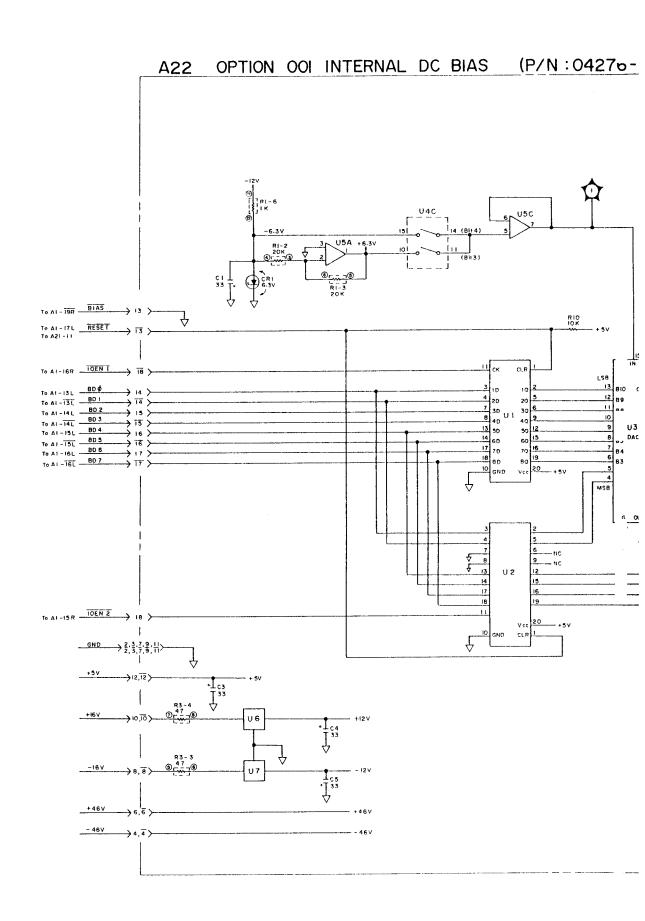
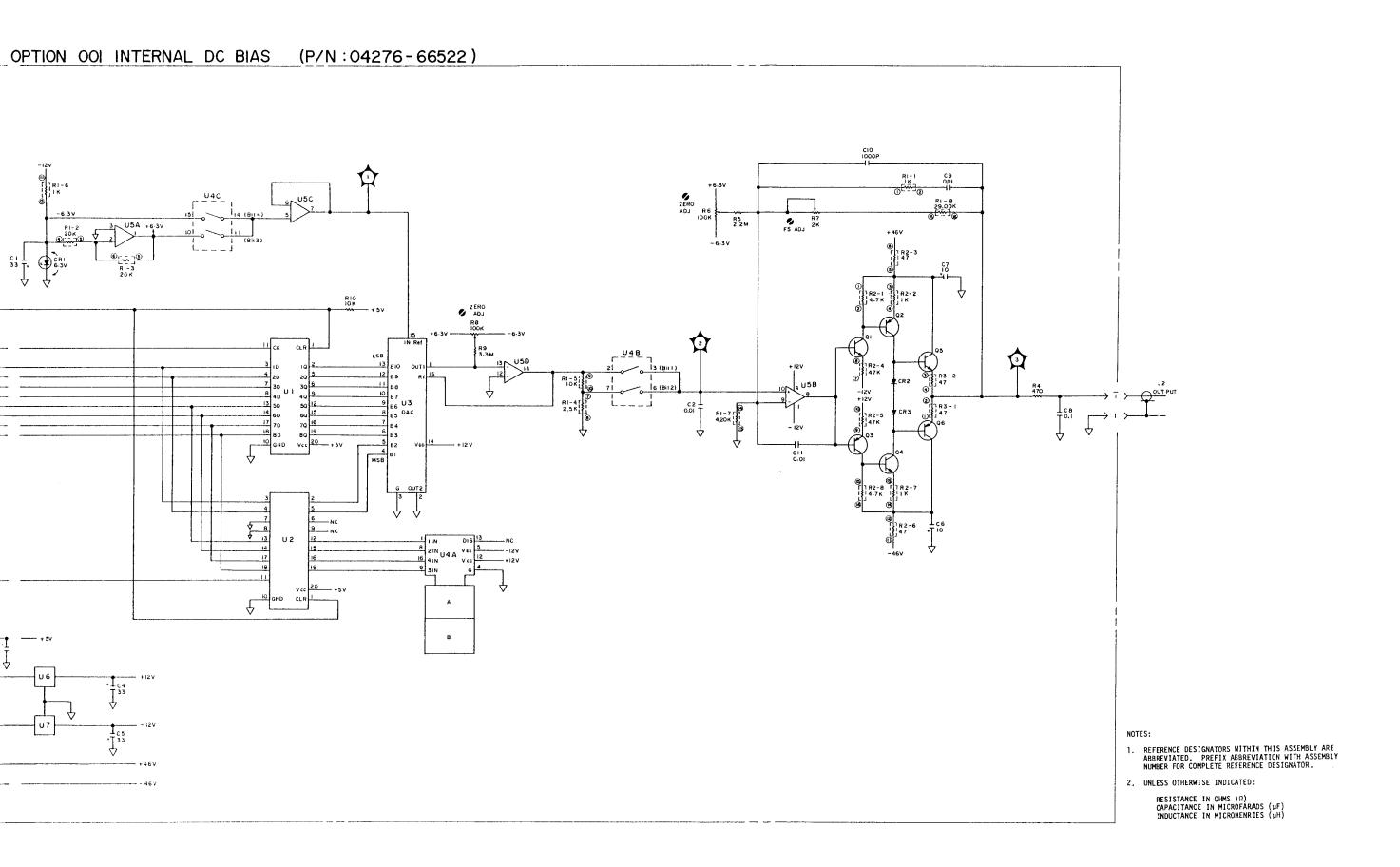


Figure 8-XX. A22 OPTION 001 INTERNAL DC BIAS Board Assembly Component Locations.





### **SALES & SUPPORT OFFICES**

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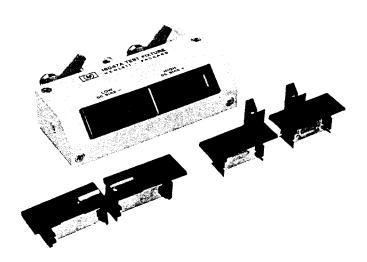
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# **TEST FIXTURE**

16047A



HEWLETT PACKARD

#### WARRANTY AND ASSISTANCE

All Hewlett-Packard products are warranted against defects in materials and workmanship. This warranty applies for one year from the date of delivery, or, in the case of certain major components listed in the operating manual, for the specified period. We will repair or replace products which prove to be defective during the warranty period provided they are returned to Hewlett-Packard. No other warranty is expressed or implied. We are not liable for consequential damages.

For any assistance, contact your nearest Hewlett-Packard Sales and Service Office. Addresses are provided at the back of this manual.

#### 1. INTRODUCTION

This operating note provides complete information on the Hewlett-Packard Model 16047A Test Fixture. The 16047A is shown pictorially on the front-cover, its phyical dimensions are given in Table 1, and typical characteristics related to offset error are given in Table 2. To order additional copies of this operating note, use the part number listsed on the rear cover.

Table 1. Specifications.

For use with Hewlett-Packard Models 4274A, 4275A, 4276A, Function: 4277A, and 4192A. Permits connecting axial and radial lead components to the UNKNOWN terminals (4-terminal configuration) of the 4274A, 4275A, 4276A, 4277A, or 4192A.

Contact Inserts: Three kinds: one for axial lead components, one for radial lead components, and one for radial short lead

components.

Dimensions in mm (inches):  $124(4.9) \times 31(1.2) \times 62(2.4)$ 

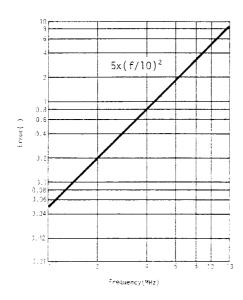
Weight in grams (1bs): 205(0.45)

Table 2. Typical Characteristics.

		Applicable Measurement Ranges		Incremental error at freq. ≥ 1MHz	
	Model	Parameter value	Measurement frequency	Parameter reading error	Offset value for D
	4274A*				
	4275A	Full range	Full range	$\pm 5 \times \left(\frac{f}{10}\right)^2 \%$	$\pm 0.02 \times \left(\frac{f}{10}\right)^2$
Γ	4192A				_0.02 ^ (10 /

f is the measurement frequency in megahertz. The incre-Note: mental errors calculated from the equation in the table for measurements at frequencies above 1MHz are additive.

\*: The maximum frequency of the 4274A is 100kHz.



0.06 0.04  $0.02x(f/10)^2$ 0.02 in D 0.008 0.006 value 0.004 Offset 3.002 0.0008 0.0006 0.0004 0.0001 Frequency (MHz)

Parameter reading error vs frequency. Offset value in D vs frequency.

Table 2. Typical Characteristics (cont'd).

	Applicable Measurement Ranges		Incremental error at 1MHz	
Mode1	Parameter value	Measurement frequency	Parameter reading error	Offset value for D
4276A**	F11	5 11		
4277A	Full range	Full range	±0.05%	±0.0002

\*\*: The maximum frequency of the 4276A is 20kHz.

#### 2. DESCRIPTION

The Model 16047A Test Fixture is designed for use with the following instruments:

Model 4192A LF Impedance Analyzer

Model 4274A Multi-Frequency LCR Meter

Model 4275A Multi-Frequency LCR Meter

Model 4276A LCZ Meter

Model 4277A LCZ Meter

It is a direct attachment, 4-terminal pair configuration type test fixture for measurements on both axial and radial lead components. Three contact inserts — labelled (1), (2), and (3) in Figure 1 — are available: one, (1), for measurements on axial components and two, (2) and (3), for measurements on radial lead components. DC bias levels up to  $\pm 35$ V can be applied to the device under test (DUT) through this test fixture. When used with the 4276A or 4277A, however, the 16047A can handle dc bias voltages up to  $\pm 40$ V. The dimensions of the contact inserts are given in Figure 2.

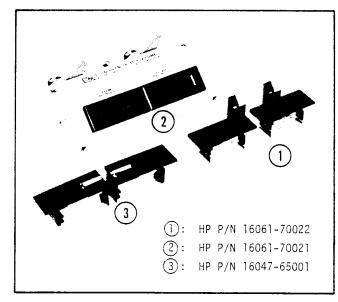


Figure 1. 16047A Test Fixture.

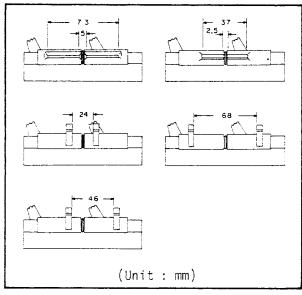


Figure 2. Dimensions of Test Fixture Contacts.

#### 3. ZERO OFFSET ADJUSTMENT

The 16047A has inherent stray capacitance, residual inductance, and residual resistance that affect the accuracy of measured values. To compensate for, or negate, these residuals to minimize measurement error, the instrument's zero offset adjustment procedure should be performed. The procedure is given in Section III of the instrument's operating manual. For SHORT zero offset adjustments a low impedance copper or phosphor bronze shorting bar such as the one shown in Figure 3 is recommended.

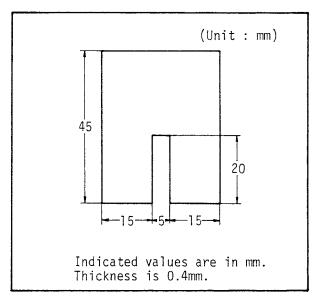


Figure 3. Shorting-bar dimensions.

#### 4. OPERATION

Setup and measurement procedure is as follows:

- a. Set the CABLE LENGTH switch (4275A, 4277A and 4192A only) on the instrument's front-panel to  $\mbox{Om}\,.$
- b. Connect the 16047A directly to the UNKNOWN terminals of the instrument.
- c. Perform ZERO OFFSET ADJUSTMENT as described in Section III of the instrument's operating manual.
- d. Insert the DUT into the test fixture.

#### 5. MAINTENANCE

The internal wiring of the 16047A is shown in Figure 4 and an exploded view — for parts identification — in Figure 5. Do not disassemble any further than shown. Maintenance consists principally of cleaning contacts and replacing worn or damaged parts. Take special care when cleaning contacts. To order parts, use the Hewlett-Packard part numbers listed in Figure 5. If a faulty part is located in an assembly that cannot be disassembled, order the next higher assembly or return the whole device to the nearest Hewlett-Packard Sales/Service Office for repair or replacement.

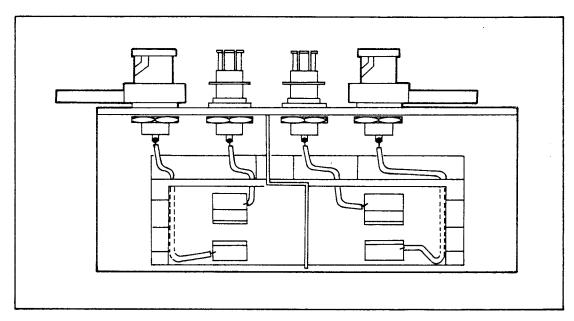


Figure 4. Internal Wiring of 16047A.

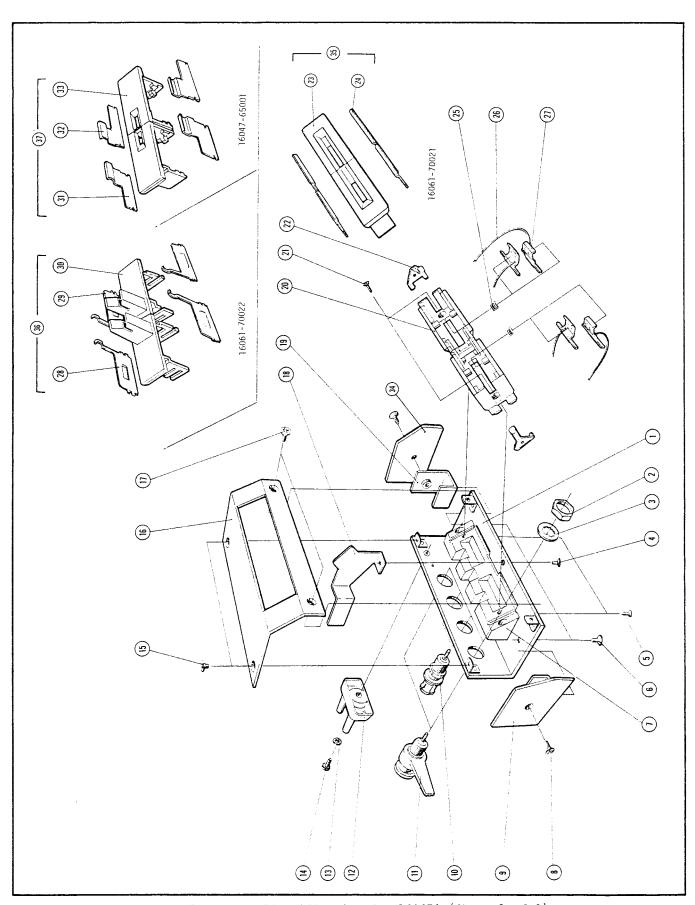


Figure 5. Parts Identification for 16047A (Sheet 1 of 2).

Reference	HP Part No.	Qty	Description
1	16047-04001	1	COVER-BOTTOM NUT WASHER SCREW SCREW
2	2950-0043	4	
3	2190-0016	4	
4	2200-0140	1	
5	0624-0203	2	
6 7 8 9	2360-0192 16061-50022 2360-0192 16047-00603 1250-1798	2 1 2 1 2	SCREW BASE SCREW SIDE PLATE (LEFT) CONNECTOR-BNC
11	16012-7122	2	CONNECTOR-BNC STOPPER WASHER SCREW SCREW
12	16047-40000	1	
13	3050-0229	1	
14	2200-0105	1	
15	2360-0192	2	
16	16047-04000	1	COVER-TOP SCREW SHIELD PLATE ANGLE SOCKET
17	2360-0192	2	
18	16047-00600	1	
19	16047-01200	2	
20	16061-50023	1	
21	0624-0202	2	SCREW SPRING SOCKET-RADIAL CONTACT-RADIAL SPRING
22	16061-10027	2	
23	16061-50031	1	
24	16061-10031	4	
25	1460-0343	2	
26	8150-0447	4	WIRE CONTACT CONTACT-AXIAL CONTACT-AXIAL SOCKET-AXIAL
27	16061-10026	4	
28	16061-10033	2	
29	16061-10032	2	
30	16061-50032	1	
31	16047-00604	2	CONTACT CONTACT SOCKET SIDE PLATE (RIGHT) SOCKET ASSEMBLY ( 23) and 24)
32	16047-00605	2	
33	16047-40001	2	
34	16047-00602	1	
35	16061-70021	2	
36	16061-70022	2	SOCKET ASSEMBLY ( 30 , 39 and 30 ) SOCKET ASSEMBLY ( 31 , 32 and 33 )
37	16047-65001	2	

Figure 5. Parts Identification for 16047A (Sheet 2 of 2).

## MANUAL CHANGES

4276A

LCZ METER

MANUAL IDENTIFICATION -

Model Number: 4276A

Date Printed: JUL. 1983

Part Number: 04276-90000

This supplement contains important information for correcting manual errors and for adapting the manual to instruments containing improvements made after the printing of the manual.

To use this supplement:

Make all ERRATA corrections.

Make all appropriate serial number related changes indicated in the tables below.

١.	SERIAL PREFIX OR NUMBER	MAKE MANUAL CHANGES	SERIAL PREFIX OR NUMBER MAKE MANUAL CHANGES
	2517J01681 and above	1	
Į			

► NEW ITEM

#### ► CHANGE 1

Page 2-5, Figure 2-3;

Change the Figure as shown on the next page.

#### NOTE

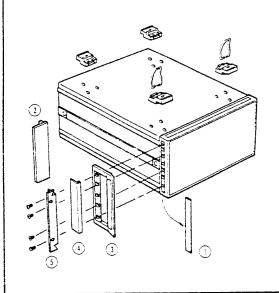
Manual change supplements are revised as often as necessary to keep manuals as current and accurate as possible. Hewlett-Packard recommends that you periodically request the latest edition of this supplement. Free copies are available from all HP offices. When requesting copies quote the manual identification information from your supplement, or the model number and print date from the title page of the manual.

Date/Div: May 17, 1985/33

Page 1 of 2



Option	Description	Kit Part Number
907	Handle Kit	5061-9690
908	Rack Flange Kit	5061-9678
909	Rack Flange & Handle Kit	5061-9684



- Remove adhesive-backed trim strips (1) from side at right and left front of instrument.
- 2. HANDLE INSTALLATION: Attach front handle 3 to sides at right and left front of instrument with screws provided and attach trim 4 to handle.
- 3. RACK MOUNTING: Attach rack mount flange 2 to sides at right and left front of instrument with screws provided.
- 4. HANDLE AND RACK MOUNTING: Attach front handle 3 and rack mount flange 5 together to sides at right and left front of instrument with screws provided.
- When rack mounting (3 and 4 above), remove all four feet (lift bar at inner side of foot, and slide foot toward the bar).

Figure 2-3. Rack Mount Kit.