## 7B90P PROGRAMMABLE TIME BASE

## INSTRUCTION MANUAL

Tektronix, Inc.
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070-2309-00
Product Group 42
$\qquad$

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INSTRUMENT SERIAL NUMBERS

Each instrument has a serial number on a panel insert, tag, or stamped on the chassis. The first number or letter designates the country of manufacture. The last five digits of the serial number are assigned sequentially and are unique to each instrument. Those manufactured in the United States have six unique digits. The country of manufacture is identified as follows:

| B000000 | Tektronix, Inc., Beaverton, Oregon, USA |
| :--- | :--- |
| 100000 | Tektronix Guernsey, Ltd., Channel Islands |
| 200000 | Tektronix United Kingdom, Ltd., London |
| 300000 | Sony/Tektronix, Japan |
| 700000 | Tektronix Holland, NV, Heerenveen, |
|  | The Netherlands |

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## WARNING

> The remaining portion of this table of contents lists the Servicing instructions. These Servicing instructions are for use by qualified personnel only. To avoid electrical shock, do not perform any servicing other than that called out in the Operating instructions unless qualified to do so.

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## SAFETY SUMMARY

This manual contains safety information which the user must follow to ensure safe operation of this instrument. WARNING information is intended to protect the operator; CAUTION information is intended to protec: the instrument. The following are general safety precautions that must be observed during all phases of operation and maintenance.

## WARNING

## Ground the Instrument

To reduce electrical-shock hazard, the mainframe (oscilloscope) chassis must be properly grounded. Refer to the mainframe manual for grounding information.

## Do Not Operate in Explosive Atmosphere

Do not operate this instrument in an area where flammable gases or fumes are present. Such operation could cause an explosion.

## Avoid Live Circuits

Electrical-shock hazards are present in this instrument. The protective instrument covers must not be removed by operating personnel. Component replacement and internal adjustments must be referred to qualified service personnel.

## Do Not Service or Adjust Alone

Do not service or make internal adjustments to this instrument unless another person, capable of giving first aid and resuscitation, is present.

## WARNING

## Warning Statements

Warning statements accompany potentially dangerous procedures in this manual. The following warnings appear in this manual and are listed here for additional emphasis.
ro avoid electrical shock, disconnect the instrument from the power source before soldering.

To avoid electrical shock, disconnect the instrument from the power source before replacing components.

## PREFACE

This manual contains two main divisions: operation and service. These divisions are separated by a heavy divider page. Information following this divider sheet is intended for qualified service personnel only.

The first division (operator's portion) begins with an introductory section that briefly describes the 7B90P Programmable Time Base. Specifications and programable features of the plug-jn are also listed. Section 2 describes each of the front-panel controls and connectors, and gives complete instructions for manual operation of the plug-in. Measuring techniques and applications are also discussed in Section 2. The last section of the operator's portion describes programmable operation of the 7B90P via the IEEE 488 Interface Bus, also known as the General Purpose Interface Bus (GPIB).* A complete list of device mnemonics is provided in this section.

The second division of the manual (service portion) begins with a section on preventive and corrective maintenance. Included in this section are some general troubleshooting procedures. Section 5 then provides a detailed circuit description of both the analog and digital portions of the 7B90P. Section 6 is a calibration procedure; it lists the equipment needed, and the steps required, for calibrating the instrument within specified limits. The remaining three sections provide electrical parts lists, schematic diagrams, and mechanical parts lists.

At the back of the manual is a removable sheet which serves as a pocket programming aid. It lists all of the high-level mnemonic codes and low-level hexadecimal codes for exercising programmable control of the 7B90P Programmable Time Base.
*Detailed information on the IEEE 488 bus is provided in IEEE Standard 488-1975, published by the Institute of Electrical \& Electronics Engineers--345 E. 47 Street, New York, NY 10017.


The 7B90P Programmable Time Base

## SECTION 1

## INTRODUCTION

The TEKTRONIX 7B90P Programmable Time Base provides calibrated sweep rates for TEKTRONIX 7000-Series programmable mainframes. The 7B90P may be inserted in any 7000-Series mainframe without damage, but is recommended for use only in a programmable mainframe.

The 7B90p has calibrated sweep rates ranging from 500 picoseconds/ division to 500 milliseconds/division. Other operating features that can be selected are: normal or automatic triggering, high or low frequency roll-off (attenuation) of the triggering signal, variable trigger holdoff, internal or external triggering, 1 X or 10 X sweep magnification, and repetitive or single-sweep mode.

The 7B90P has readout encoding capability. Thus, horizontal scale factors can be displayed on a programmable mainframe having readout display capability.

All functions of the 7B90P can be remotely programmed except for the Sweep Calibration (SWP CAL) adjustment and the External Trigger Input Termination (TERM) switch. The status of each programmable function can be set or read by commands sent over the IEEE 488 bus. These commands can be sent in either a high-level or low-level language, joth of which are decoded by a microprocessor in the 7B90P.

## Specifications

The following tables describe the electrical, programming, environmental, and physical specifications of the 7B90P. Characteristics whose specifications are checked in the calibration procedure are listed under a column entitled PERFORMANCE REQUIREMENTS. Specifications for all other characteristics are listed under SUPPLEMENTAL INFORMATION.
Information in the SUPPLEMENTAL INFORMATION column is of a general nature and should not be considered as performance requirements. In cases where a particular plug-in specification is mainframe dependent, a note has been included referencing the manual for the mainframe.

TABLE 1-1

## ELECTRTCAL SPECIFTCATIONS

Sweep Generator


TABLE 1-1 (cont.)

| CHARACTERISTIICS | PERFORMANCE REQUIREMENTS | SUPPLEMENTAL |
| :---: | :---: | :---: |
| Sweep Length (unmagnified) | At least 10.2 divisions at all sweep speeds. |  |
| Position Range <br> (at $1 \mathrm{~ms} /$ div) <br> Programmed POS Ø <br> (high-level mnemonic) <br> or POSITION control <br> at midrange. |  | Sweep starts 5 divisions to left of graticule center $\pm 1.0$ divisions. |
| Programmed POS 6.4 or POSITION control fully clockwise. |  | Sweep starts 1.4 divisions to right of graticule center $\pm 1.0$ divisions. |
| Programmed POS -6. 39 or POSITION control fully counterclockwise. |  | Sweep starts 11.39 divisions to left of graticule center $\pm 1.0$ divisions. |
| Programmed Step Size |  | 0.0125 div ( 80 steps/ div) unmagnified; 0.125 div (8 steps/ div) magnified. |
| Trigger Hold off Time $500 \mathrm{ps} / \mathrm{div}$ to $2 \mu \mathrm{~s} / \mathrm{div}$ $5 \mu \mathrm{~s} / \mathrm{div}$ to $20 \mu \mathrm{~s} / \mathrm{div}$ $50 \mu \mathrm{~s} / \mathrm{div}$ to $200 \mu \mathrm{~s} / \mathrm{div}$ $500 \mu \mathrm{~s} / \mathrm{div}$ to $2 \mathrm{~ms} / \mathrm{div}$ $5 \mathrm{~ms} / \mathrm{div}$ to $500 \mathrm{~ms} / \mathrm{div}$ | $\operatorname{Min}(\mathrm{ccw})$ Max (cw) <br> $\leq 3.5 \mu \mathrm{~s}$ $\geq 90 \mu \mathrm{~s}$ <br> $\leq 3.5 \mu \mathrm{~s}$ $\geq 900 \mu \mathrm{~s}$ <br> $\leq 350 \mu \mathrm{~s}$ $\geq 9.0 \mathrm{~ms}$ <br> $\leq 3.5 \mathrm{~ms}$ $\geq 90 \mathrm{~ms}$ <br> $\leq 35 \mathrm{~ms}$ $\geq 900 \mathrm{~ms}$ | When operated in 7912AD, Hold-off is disabled at some sweep speeds. |
| Programmed Range |  | Min. = programmed HOL $\varnothing$ <br> Max. = programmed HOL 63 |
| Programmed Resolution |  | 62 steps between min. and max. |

TABLE 1-1 (cont.)

Triggeriag

| CHARACTERISTICS | PERFORMANCE REQUIREMENTS |  |  | SUPPLEMENTAL |
| :---: | :---: | :---: | :---: | :---: |
| Triggering Sensitivity <br> (from repetitive signal) | Triggering <br> Freq. Range | Min. Tr <br> Signal | ggering <br> equired | External signals are increased 10 times for EXT +10 operation. |
|  |  | Int. | Ext. |  |
| PP AUTO Mode (AC or DC coupling) | $\begin{aligned} & 200 \mathrm{~Hz} \text { to } \\ & 50 \mathrm{MHz} \end{aligned}$ | 0.5 div | 125 mV |  |
|  | $\begin{aligned} & 50 \mathrm{MHz} \text { to } \\ & 400 \mathrm{MHz} \end{aligned}$ | 1.5 div | 375 mV |  |
|  | at least $50 \mathrm{~Hz}$ | 2.0 div | 500 mV |  |
| NORM or SGL SWP Mode |  |  |  |  |
| AC coupling | $\begin{gathered} 30 \mathrm{~Hz} \text { to } \\ 50 \mathrm{MHz} \end{gathered}$ | 0.3 div | 100 mV |  |
|  | $\begin{aligned} & 50 \mathrm{MHz} \text { to } \\ & 400 \mathrm{MHz} \end{aligned}$ | 1.5 div | 250 mV |  |
| AC LF REJ coupling | $\begin{aligned} & 30 \mathrm{kHz} \text { to } \\ & 50 \mathrm{mHz} \end{aligned}$ | 0.3 div | 100 mV | Does not trigger on sine wave at or below 60 Hz with amplitudes less than 8 divisions INT or 3 V EXt. |
|  | $\begin{aligned} & 50 \mathrm{MHz} \text { to } \\ & 400 \mathrm{MHz} \end{aligned}$ | 1.5 div | 250 mV |  |
| AC HF REJ coupling | $\begin{gathered} 30 \mathrm{~Hz} \text { to } \\ 50 \mathrm{KHz} \end{gathered}$ | 0.3 div | 100 mV | Does not trigger on 50 MHz sine wave with amplitude 1.5 divisions or less INT or 0.15 V or less EXT. |
| DC coupling | $\begin{aligned} & \text { DC to } \\ & 50 \mathrm{MHz} \end{aligned}$ | 0.3 div | 100 mV |  |
|  | $\begin{aligned} & 50 \mathrm{MHz} \text { to } \\ & 400 \mathrm{MHz} \end{aligned}$ | 1.5 div | 250 mV |  |

TABLE 1-1 (cont.)


* ccw = LEV -6.4 (high-level code) cw $=$ LEV 6.39 (high-level code)

TABLE 1-1 (cont.)

| CHARACTERISTICS | PERFORMANCE REQUIREMENTS | SUPPLEMENTAL |
| :---: | :---: | :---: |
| Internal Trigger Jitter | 0.1 nanosecond or less at 400 MHz . |  |
| External Trigger <br> Input <br> Maximum Input Voltage 1 megohm |  | 250 volts DC plus peak AC. |
| 50 ohms |  | 1 watt |
| Input R and C 1 megohm |  | 1 megohm in parallel 20 picofarads $+10 \%$. |
| 50 ohms |  | 50 ohms $+5 \%$ |
| VSWR with 50 ohm termination |  | 1.22 maximum at 400 MHz . |

TABLE 1-2

PROGRAMMING SPECIFICATIONS

| CHARACTERISTICS | PERFORMANCE REQUIREMENTS | SUPPLEMENTAL |
| :--- | :--- | :--- |
| Set-up Time (Full <br> front panel) | Front-panel set-up <br> time is: <br> (decode time) + (byte <br> count) X (time/byte) <br> (overhead). |  |
| High-level |  | Decode time $\simeq 14 \mathrm{~ms}$ <br> Byte count $\simeq 75$ <br> Overhead $\simeq 0.4 \mathrm{~ms}$ |
| Low-level |  | Decode time $\simeq 1.5 \mathrm{~ms}$ <br> Byte count $=15$ <br> Overhead $\simeq 0.1 \mathrm{~ms}$ |

TABLE 1-3

## ENVIRONMENTAL SPECIFICATIONS

For temperature, altitude, vibration, shock, and humidity specifications, refer to the manual for the associated mainframe.

TABLE 1-4

PHYSICAL SPECIFICATIONS

| Weight | $1.2 \mathrm{Kg} \cdot(2.6 \mathrm{lbs})$. |
| :--- | :--- |
| Dimensions | Fits 7000-Series mainframe plug-in compartment. |

## Summary of Programmable Features

The TEKTRONIX 7B90P Programmable Time Base can be operated in either Local or Remote mode. In Local mode, it operates like a non-programmable plug-in, such as the TEKTRONIX 7B80. That is, its functions are controllable only from the front panel.

When the 7B90P is set to Remote mode, however, the front panel is rendered inoperative except for two non-programmable functions; the Sweep Calibration adjustment (SWP CAL) and the External Trigger Input Terminator swi.tch (TERM). Under remote control, functions can be controlled only by commands sent over the IEEE 488 bus. However, as control settings are effected remotely, the front panel lights will indicate ancordingly.

The Remote/Local state of the 7B90P is slaved to the Remote/Local state of the mainframe in which it is installed. Thus the mainframe must be set to Remote mode for programmed control of the plug-in, and must be set to Local mode for front-panel control of the plug-in.

## Front-Panel Buttons

Each pushbutton on the 7B90P front panel (except TERM) serves as both a switch and an indicator. A Light Emitting Diode (LED) mounted behind each button indicates the present state of the function controlled by that button, regardless of whether the corresponding function was set manually or under program control. The switches themselves are a momentary-contact type; it is only necessary to tap a particular button to set that function.

The operation of the front-panel buttons generally falls into one of two categories. The first category includes the +SLOPE and MAG buttons. Since there are just two settings for each of these funtions, buttons in this category are essentially toggle switches. The lighting of each button indicates the current state of the function controlled by that button. For example, repeatedly pressing the MAG button alternately illuminates and extinguishes the MAG button. When MAG is lit, the sweep is magnified; when MAG is not lit, the sweep is not magnified. (Incidentally, pressing the MAG button causes the sweep speed to change and this is indicated by the TIME/DIV buttons; thus the TIME/DIV buttons indicate the correct sweep speed, regardless of the setting of the NAG button.)

The second category of buttons includes the following functions:

$$
\begin{aligned}
& \text { TRIGGERING-MODE (PP AUTO, NORM, or SGL SWP/RESET) } \\
& \text { TRIGGERING-COUPLING (AC, AC LF REJ, AC HF REJ, or DC) } \\
& \text { TRIGGERING-SOURCE (INT, LTN, EXT, or EXT* } 10 \text { ) } \\
& \text { TIME/DIV ( } 1,2,5,10,20,50,100,200 \text {, or } 500 \text {; and } \mathrm{ms}, \mu \mathrm{~s}, \mathrm{~ns} \text {, or } \mathrm{ps} \text { ) }
\end{aligned}
$$

Since there are more than two switch settings for each of these functions, pushbuttons in this category are part of a ganged-switch arrangement. Pressing a particular button cancels the setting of the previously pressed button and selects the new setting. For example, if ExT (Triggering Source) has been pressed or remotely set as indicated by the lighting of the EXT button, pressing INT will extinguish the EXT light and switch the triggering source to internal; the INT button will then be lit to indicate the new setting.

## Front-Panel Controls

There are three 7B90P front-panel rotary controls: POSTION, HOLD OFF, and LEVEL (TRIGGERING). Each of these is fully programmable. When the 7B90P is in Local mode, these controls operate as though they were part of a non-programable plug-in. That is, the trace position, hold-off period, and triggering level are controlled by the current position of each respective knob. When the 7B90P is set to Remote mode, however, each control is disabled and the corresponding functions are now set by commands which set a Digital-to-Analog Converter (DAC). There is one DAC for each of the three controls. Since the 7 B 90 P front panel is disablec during remote operation, the current knob settings may not accurately reflect the true trace position, hold-off period, and triggering level. However, when the 7B90P is again set to Local mode, the trace position, hold-off period, and triggering level will again revert to the settings currently indicated by their respective knobs.

Table 1-5 provides a list of all the 7B90P functions and indicates which functions can be set or read (queried) under manual and remote operation. Setting a function manually refers to pressing a given button or turning a control; reading a function manually refers to looking at the illumination of a particular button or noting the position of a control. Setting and reading functions under remote control refers to setting a function or interrogating the status of a function by sending and receiving messages over the IEEE 488 bus. Information on how this is done is contained in Section 3.

For more information on each of the functions and how they are used, refer to Section 2.

TABLE 1-5

## LIST OF 7B90P FUMCTIONS

| FUNCTION | MANUAL OPERATION |  | REMOTE DPERATION |  |
| :---: | :---: | :---: | :---: | :---: |
|  | SET | READ | SET | READ |
| TRIGGERING |  |  |  |  |
| MODE | X | X | X | X |
| COUPLING | X | X | X | X |
| SOURCE | X | X | X | X |
| SLOPE | X | X | X | X |
| LEVEL Control | X | X | X | X |
| POSIIION Control | X | X | X | X |
| TIME/DIV | X | X | X | X |
| MAGnified Sweep | X | X | X | X |
| HOLD OFF Control | X | $X$ | X | X |
| SWP CAL Adjustment | X |  |  |  |
| TERMination (input) | X | X |  |  |
| Plug-in type ${ }^{1}$ |  | X |  | X |
| Trigger Light |  | X |  | X |
| End-of-Sweep SRQ Enable |  |  | X | X |
| Single Sweep Ready Light |  | X |  | X |

## NOTE

An "X" in the above chart denotes that a particular function can be set or read.

1 The plug-in type (7B90P) is inscribed on the front panel and can be read under program control with the ID command.

## Block Diagranim

The following is a simplified block diagram of the 7B90P (Fig. 1-1) and a brief description of the blocks. It is included here to provide background information for the sections that follow.

The trigger generator provides a stable display by starting the sweep generator at a controlled point for each sweep. The trigger signal is taken from the internal or line trigger signals from the mainframe or the EXI TRIG IN connector on the front panel. The sweep logic sets up the control signals for PP AUTO (Peak-to-Peak AUTOmatic), NORM (NORMal) or SGL SWP (SinGLe SWeeP) modes. This circuit also generates the hold-off timing for the sweep generator. The sweep generator provides a precise ramp voltage for the sweep outputs. The sweep time is variable in calibrated steps from 500 picoseconds to 500 milliseconds per division. When TIME/DIV is set for 100 ms or slower, the sweep generator asserts a mainframe control line that limits the intensity to protect the CRT from burns due to high intensity at slow sweep speeds. The horizontal preamp provides differential sweep signals to the mainframe and adds DC current for sweep positioning. If the MAGnifier is selected, the gain of the horizontal preamp is increased to magnify the sweep.

The 7B90P sweep speed, triggering, front panel LED's, and IEEE 488 circuit are controlled by a microprocessor system in the plug-in. The heart of this system is a Motorola M6800 Microprocessing Unit (MPU). A control program, resident in $4 \mathrm{~K}(1 K=1024)$ bytes of Read Only Memory (ROM) directs the MPU activity. The MPU also uses 128 bytes of Random Access Memory (RAM) as a "scratch pad".

Five Peripheral Interface Adapters (PIA's) handle internal communication between the MPU and other circuits in the plug-in. The IEEE 488 interface handles the handshaking and communication on the IEEE 488 bus. The clock circuit generates the 2 -phase clock signals required by the MPU. When power-up occurs, the power-up circuit initializes the MPU and PIA's, and starts the clock circuit.

A detailed circuit description of these blocks is included in Section 5.


Fig. 1-1. Block diagram of the 7B90P.

## SECTION 2

## MANUAL OPERATION

This section contains a description of the 7B90P controls, connector, and indicators, and includes instructions for manual (non-programmable) operation of the plug-in. This is followed by a "functional check" section which provides a quick verification of instrument performance. Also included are instructions for installing the plug-in, general information on triggering, and other subjects that pertain to various measurement applications.

## Installation

The 7B90P is calibrated and ready for use as received. It may be installed in any compartment of a TEKTRONIX 7000-Series programmable mainframe, but is intended primarily for use in a horizontal plug-in compartment. The 7B90P may also be installed in a vertical plug-in compartment to provide a vertical sweep on the CRT. However, when used in this manner, there are no internal triggering or retrace blanking provisions, and the unit may not meet the specifications given in Section 1.

## NOTE

Always turn off the mainframe power before installing or removing plug-ins to prevent damage to circuitry.

To install the plug-in, align the upper and lower rails of the 7B90P with the mainframe tracks and insert the plug-in. The front panel will be flush with the front of the mainframe when the plug-in is fully inserted, and the latch at the bottom left corner will be in place against the front panel.

To remove the 7 B 90 P , pull on the latch (inscribed with the identification "7B90P") and the plug-in will unlatch. Continue pulling on the latch to slide the 7B90P out of the mainframe.

## Controls, Connector, and Indicators

## (1) Level Control

Selects a point on the trigger signal where triggering occurs. When the LEVEL is set in the positive region, triggering occurs on the positive excursion (top half) of the signal; when the LEVEL is set in the negative region, triggering occurs on the negative excursion (bottom half) of the signal.

## (2) SLOPE Switch

Determines whether the sweep is triggered on the negative slope or positive slope of the trigger signal. When lit, the trigger switch is set for a positive slope; when unlit, it is set for a negative slope.

## (3)TRIG'D Indicator

Illuminates when the sweep is triggered.

## (4) READI Indicator

Illuminates when the sweep circuit is armed, and is extinguished when the sweep is completed. (Applies only to single-sweep mode.)

## (5) Triggering-mode Pushbuttons

These three buttons select the triggering mode of the 7B90P according to which button was last pressed:

PP AUTO: Provides a triggered display at any setting of the LEVEL control whenever an adequate trigger signal is applied. The range of the LEVEL control in the Peak-to-Peak Automatic mode is between approximately $10 \%$ and $90 \%$ of the peak-to-peak amplitude of the trigger signal.

NORM: Provides normal triggering. That is, the sweep is triggered only if a point on the triggering signal corresponds to the conditions set by the LEVEL and SLOPE controls.


Fig. 2-1. Front panel of the 7B90P.

SGL-wSP/RESET: Sets the 7B90P to single-sweep mode and resets (arms) the sweep. A single sweep will occur when a point on the trigger signal corresponds to the conditions set by the LEVEL and SLOPE controls. Pressing this button a second time rearms the sweep.

## (6) TRIGGERTNG-COUPLING Pushbuttons

These four buttons select the trigger coupling mode according to which button was last pressed:

AC: Selects AC coupling of the triggering signal, thereby blocking any direct-current component of the signal.

AC LF REJ: Selects a trigger-coupling circuit that passes high-frequency AC signals, but blocks DC and low-frequency AC signals.

AC HF REJ: Selects a trigger-coupling circuit that passes low-frequency AC signals, but blocks $D C$ and high-frequency $A C$ signals.

DC: Selects DC coupling of the triggering signal, thereby passing all frequency components of the signal.

## (7) TRIGGERING-SOURCE Pushbuttons

These four buttons select the source of the trigger signal according to which button was last pressed:

INT: Selects the internal signal supplied by the mainframe vertical system as the triggering signal.

LINE: Selects a sample of the mainframe $A C$ power input as the triggering signal.

EXT: Selects the signal supplied to the EXT TRIG IN connector as the triggering signal.

EXT+10: Selects the signal supplied to the EXT TRIG IN connector as the triggering signal, but attenuates it by a factor of 10 to allow greater input voltage range.

## (8) TIME/DIV Pushbuttons

The sweep rate (time/division) is entered in engineering notation and therefore both a number and a multiplier must be entered via the following switches:
$1,2,5,10,20,50,100,200,500$ : These switches determine the numeric part of the sweep rate.
ms, $\mu \mathbf{s , n s , p s : ~ T h e s e ~ s w i t c h e s ~ d e t e r m i n e ~ t h e ~ m u l t i p l i e r ~ a n d ~ u n i t s ~ o f ~ t h e ~}$ sweep rate in milliseconds (ms), microseconds ( $\mu s$ ), nanoseconds ( $n s$ ), or picoseconds (ps).

MAG: Selects the magnified $10 X$ sweep rate, thereby generating a sweep speed that is 10 times greater. When lit, the switch is set for magnified sweep; when unlit, the sweep is not magnified.

## (9) POSITION Control

Varies the horizontal position of the trace by applying a DC offset voltage.

## (10) Hold off Control

Varies the trigger hold-off period (the time following the end-of-sweep that the sweep generator is inhibited) to improve triggering stability on repetitive signals. Hold-off time is maximum when the control is turned fully clockwise.

## (11) SWP CAL Adjustment

This front panel screwdriver adjustment varies the sweep calibration to achieve the desired trace width. This allows for differences in gain between mainframes. All sweep rates are affected by this adjustment.

## EXT TRIG IN

(12) EXT TRIG IN Connector: Input connector (BNC type) for an external trigger signal.
(13) TERM: Terminates the EXTernal TRIGger INput connector with either 50 ohms or 1 megohm of impedance. In the IN position, it selects 1 megohm; in the OUT position, it selects 50 ohms.


The signal to the EXT TRIG IN connector must not exceed 250 volts, DC plus peak AC. If 50 ohms is selected, the power of the input signal must be limited to 1 watt.

## Functional Check

The following procedure demonstrates the use of the controls, connector, and indicators of the 7B90P, while at the same time providing a means of checking the basic operation of the instrument. Refer to the description of the controls, connector, and indicators while performing this procedure. If performing the functional-check procedure reveals a malfunction or improper adjustment, first check the operation of the associated mainframe. If the mainframe seems to be working properly, have the 7B90P examined by a. qualified service technician.

## Preliminary Setup

With the mainframe power off, install the 7B90P into the horizontal plug-in compartment of a $7000-S e r i e s ~ p r o g r a m m a b l e ~ m a i n f r a m e . ~ I n s t a l l ~ a ~$ 7A-Series amplifier unit into the vertical plug-in compartment.

With the mainframe power on, set the mainframe trigger-source switch (if applicable) to select the appropriate vertical compartment. Set the amplifier unit for $A C$ input coupling. Also, adjust the mainframe intensity knob(s), if necessary. NOTE: On some mainframes, both the beam intensity and graticule intensity must be adjusted.

## Power-up Sequence

When the mainframe is turned on, the 7 B 90 P is set to a predetermined state as indicated by the lighted front-panel buttons. The status of the front-panel buttons should indicate the following:

| TRIGGERING MODE: | PP AUTO |
| :--- | :--- |
| TRIGGERING COUPLING: | AC |
| TRIGGERING SOURCE: | INT |
| TRIGGERING SLOPE: | + SLOPE |
| TIME/DIV: | $1 \mu \mathrm{~s}$ |
| MAG: | unmagnified (1X) |

Also, the outputs of the three front-panel controls (LEVEL, POSITION, and HOLD OFF) are tracked to their current knob settings during the power-up sequence.

## Procedure

## NORMAL OPERATYON

1) After an adequate warm-up period ( 20 minutes or more), adjust the 7B90P POSITION control for midscreen positioning of the trace; adjust the HOLDOFF control to minimum (fully counterclockwise). Readjust the mainframe intensity knob(s), if necessary.
2) Adjust the amplifier POSITION control so that the trace is vertically centered on the CRT monitor.
3) Set the amplifier unit to a deflection factor of $1 \mathrm{~V} / \mathrm{div}$ (or 100 $\mathrm{mV} / \mathrm{div}$ if a 10X probe is being used). The VARIABLE control should be pressed in to the calibrated position.
4) Set the 7B90P sweep speed to $1 \mathrm{~ms} / \mathrm{div}$. Adjust the mainframe intensity knob(s) to the proper levels.
5) Connect a 4 volts (peak-to-peak), 1 kilohertz square wave to the input of the amplifier unit, and (if applicable) select the proper input connector.
6) Check the display for one cycle per division. If necessary, adjust the front-panel SWP CAL screwdriver adjustment for one cyole per division over the center 8 graticule divisions.
7) Readjust the 7B90P POSITION control, if necessary.

## MAGNTFIED SWEEP

8) Press the MAG button and check that the MAG button illuminates. Note that the center portion of the unmagnified display (the area within one-half division of centerline) is expanded to 10 divisions. Also note that the sweep-speed buttons now indicate $100 \mu \mathrm{~s} / \mathrm{div}$. (Adjust the mainframe intensity knob(s), if necessary.)
9) Reset the 7B90P to unmagnified mode by again pressing the MAG button. Note that the MAG button is no longer lit and that the display is no longer magnified. Also note that the sweep-speed buttons now indicate $1 \mathrm{~ms} / \mathrm{div}$. (Adjust the mainframe intensity $\mathrm{knob}(\mathrm{s})$, if necessary.)

## PP-AUTO TRIGGERING MODE

10) Connect an $8 \mathrm{~V} P-\mathrm{P}, 1 \mathrm{kHz}$ sine wave to the input of the amplifier unit.
11) Rotate the LEVEL control from fully counterclockwise to fully clockwise and note that the signal can be triggered over the full range of the LEVEL control in PP AUTO mode.
12) Reduce the input signal amplitude to 1 V P-P. Note that the signal can still be triggered in PP AUTO mode, as indicated by the TRI'G light.
13) Disconnect the input signal and check for a free-running baseline trace. Note that the TRI'G light is no longer lit.

## NORMAL TRIGGERING MODE

14) Press the NORM button. Check that the button is lit and that the baseline trace now disappears in the absence of the trigger signal.
15) Reconnect the $1 \mathrm{~V} P-\mathrm{P}, 1 \mathrm{kHz}$ sine wave to the amplifier input. Note that the waveform is displayed (the 7B90P is triggered) only when the LEVEL control is set near 0 .

## SINGLE-SWEEP TRIGGERING MODE

16) With the 7B90P set for a triggered display, press the SGL SWP/RESET button. Check that the button is lit and that the displayed waveform disappears from the CRT monitor. The 'RRI'G light is also extinguished at this time.
17) Again press the SGL SWP/RESET button. Note that one sweep occurs (the waveform briefly appears on the CRT monitor) each time this button is pressed. Also note that the TRI'G and READY lights flash when SGL SWEEP/RESET is pressed. NOTE: It may be necesaary to increase the mainframe intensity during this step.
18) Disconnect the input signal and press SGL SWP/RESET. The READY light should now be lit. Check that TRI'G briefly lights and that READY is extinguished when the input signal is reconnected.

## LEVEL CONTROL

19) Press the PP AUTO button and set the LEVEL control fully counterclockwise. Note that the waveform is triggered on the negative excursion (bottom half) of the signal.
20) Set the LEVEL control fully clockwise and note that the input signal is now triggered on the positive excursion (top half) of the signal.
21) Set the LEVEL control to 0 and note that triggering occurs midway between the peak and trough of the signal.
+SLOPE
22) Press the +SLOPE switch. Check that the button is no longer lit and that triggering now occurs on the negative slope of the input signal.
23) To again trigger the input signal on its positive slope, press +SLOPE. The button should now be lit, indicating positive-slope triggering.

## HOLD OFF CONTROL

24) Rotate the HOLD OFF control fully clockwise. Note that the trigger hold-off period increases. (On a CRT monitor, this is evidenced by increased flickering of the input signal.)
25) Rotate the HOLD OFF control fully counterclockwise to restore minimum hold-off time.

## TRIGGERING COUPLING

26) Press the $A C, A C H F R E J$, and $D C$ coupling buttons. Note that in each case the respective button is lit and that a stable display is achieved.
27) Press the AC LF REJ button and check that it lights. Note that a stable display is not achieved due to attenuation of the 1 kHz input signal.

## EXIERNAL TRIGGERING

28) By means of a BNC $T$ connector, connect the input signal to the amplifier input and al so to the EXT TRIG IN connector of the 7B90P. Press the EXT source button. Note that the button lights and that stable triggering now occurs if TERM is pressed in. NOTE: with the TERM switch in the extended (50 ohm) position, the displayed signal amplitude is one-half that displayed when TERM is in the unextended (1 megohm) position.

## General Operating Information

## Triggering Switch Logic

The MODE, COUPLING, and SOURCE pushbuttons of the TRIGGERING switches are arranged in a sequence that places the most-often used position at the top of each series of pushbuttons. With this arrangement, a stable display can usually be obtained by pressing the top three pushbuttons: PP AUTO, $A C$, and INT.

When an adequate trigger signal is applied and the LEVEL control is correctly set, the unit is triggered as indicated by the illuminated TRIG'D light. If the TRIG'D light is not on, the LEVEL control is either at a setting outside the range of the trigger signal applied to the 7B90P from the vertical unit, the trigger signal amplitude is inadequate (less than 0.3 divisions), or its frequency is below the lower frequency limit of the AC COUPLING switch on PP AUTO range. If the desired display is not obtained with these buttons lit, other selections must be made. Refer to the following discussions or the instruction manuals for the associated mainframe and amplifier unit for more information.

## Triggering Modes

The MODE pushbutton switches select the mode in which the sweep is triggered. The following discussion describes the purpose of each mode:

PP AUTO: The PP AUTO MODE provides a triggered display at any setting of the LEVEL control whenever an adequate trigger sigral is applied. The range of the LEVEL control in the PP AUTO MODE is between approximately $10 \%$ and $90 \%$ of the peak-to-peak amplitude of the trigger signal. The LEVEL and SLOPE controls can be set so that the displayed waveform starts at any point within this range on either slope. The trigger circuits automatically compensate for a change in trigger-signal amplitude. Therefore, if the LEVEL control is set to start the waveform display at a certain percentage point on the leading edge of a lowmamplitude signal, it triggers at the same percentage point on the leading edge of a high-amplitude signal if the LEVEL control is not changed. When the trigger repetition rate is outside the parameter given in the Specifications (Section 1), or when the trigger signal is inadequate, the sweep free runs at the rate indicated by the TIME/DIV buttons to produce a bright base-line reference trace (TRIG'D light off). When an adequate trigger signal is again applied, the free-running condition ends and a triggered display is presented.

The PP AUTO MODE is particularly useful when observing a series of waveforms, since it is not necessary to reset the LEVEL control for each observation. The PP AUTO MODE is used for most applications because of the ease of obtaining a triggered display. The NORM and SINGLE-SWP MODE settings may be used for special applications.

NORMAL: The NORM MODE provides a triggered display with the correct setting of the LEVEL control whenever an adequate trigger signal is applied. The TRIG'D light indicates when the display is triggered. When the TRIG'D light is off, no trace is displayed.

The normal trigger mode must be used to produce triggered displays when the trigger repetition rate is below about 30 hertz.

SINGLE SWEEP: When the signal to be displayed is not repetitive, or varies in amplitude, waveshape, or repetition rate, a conventional repetitive-type display mode may produce an unstable presentation. Under these circumstances, a stable display can often be obtained by using the single-sweep mode. The single-sweep mode is also useful in photographing non-repetitive or unstable displays.

To obtain a single-sweep display of a repetitive signal, first obtain the best possible display in the NORM mode. Then, without changing the other TRIGGERING controls, press the SGL SWP button. A single trace is presented each time this button is pressed. Further sweeps cannot be presented until the SGL SWP button is pressed again. If the displayed signal is a complex waveform composed of varying amplitude pulses, successive single-sweep displays may not start at the same point on the waveform. To avoid confusion due to the crt persistence, allow the display to disappear before pressing the SGL SWP button again. At fast sweep rates, it may be difficult to view the single-sweep display. The apparent trace intensity can by increased by reading the ambient light level or by using a viewing hood.

## Trigger Coupling

The TRIGGERING COUPLING pushbuttons select the method in which the trigger signal is connected to the trigger circuits. Each position permits selection or rejection of some frequency components of the signal which triggers the sweep.

AC: AC COUPLING blocks the DC component of the trigger signal. Signals with low-frequency components below about 30 hertz are attenuated. In general, AC COUPLING can be used for most applications. However, if the signal contains unwanted frequency components or if the sweep is to be triggered at a low repetition rate or DC level, one of the other COUPLING switch positions will provide a better display.

AC LF REJ: AC LF REJ COUPLING rejects $D C$, and attenuates low-frequency trigger signals below about 30 kilohertz. Therefore, the sweep is triggered only by the higher-frequency components of the trigger signal. This position is particularly useful for providing stable triggering if the trigger signal contains line-frequency components. Also, the AC LF REJ position provides the best alternate-mode vertical displays at fast sweep rates when comparing two or more unrelated signals.

AC HF REJ: AC HF REJ COUPLING passes all low-frequency signals between about 30 hertz and 50 kilohertz. DC is rejected and signals outside the above range are attenuated. When triggering from complex waveforms, this position can provide a stable display of the low-frequency components.

DC: DC COUPLING can be used to provide stable triggering from lowfrequency signals which would be attenuated in the other COUPLING switch
positions. DC COUPLING can be used to trigger the sweep when the trigger signal reaches a DC level set by the LEVEL control. When using internal triggering, the setting of the vertical unit position control affects the triggering point.

## Trigger Source

The TRIGGERING SOURCE pushbuttons select the source of the trigger signal that is connected to the trigger circuits.

INTERNAL: The INT button connects the trigger signal from the vertical plug-in unit. Further selection of the internal trigger signal may be provided by the vertical plug-in unit or by the mainframe; see the instruction manuals for these instruments for more information. For most applications, the internal source can be used. However, some applications require special triggering which cannot be obtained in the INT position. In such cases, the LINE or EXT buttons must be used.

LINE: The LINE button connects a sample of the power-line voitage from the mainframe to the trigger circuit. Line triggering is useful when the input signal is time-related (multiple or submultiple) to the line frequency. It is also useful for providing a stable display of a linefrequency component in a complex waveform.

EXTERNAL: The EXT button connects the signal from the EXT TRIG IN connector to the trigger circuit. The external signal must be time-related to the displayed waveform for a stable display. An external trigger signal can be used to provide a triggered display when the internal signal is either too low in amplitude for correct triggering or contains signal components on which triggering is not desired. It is also useful when signal tracing in amplifiers, phase-shift networks, wave-shaping circuits, etc. The signal from a single point in the circuit can be connected to the EXT TRIG IN connector through a probe or cable. The sweep is then triggered by the same signal at all times and allows amplitude, time relationship, or waveshape changes of signals at various points in the circuit to be examined without resetting the TRIGGERING controls.

The EXT+10 pushbutton attenuates the external trigger signal by a factor of 10. Attenuation of high amplitude external trigger signals is desirable to increase the effective range of the LEVEL control.

## Trigger Slope

The SLOPE button determines whether the trigger circiit responds on the positive or negative slope of the trigger signal. When the SLOPE switch is in the (+) position (as indicated by the lighting of the SLOPE button), the display starts on the positive slope of the waveform (see Fig. 2-2). When several cycles of a signal appear in the display, the setting of the SLOPE switch is often unimportant. However, if only a certain portion of a cycle is to be displayed, correct setting of the SLOPE switch is important to provide a display that starts on the desired slope of the input signal.

## Trigger Level

The LEVEL control determines the voltage level on the trigger signal at which the sweep is triggered. When the LEVEL control is set in the (+) region, the trigger circuit responds at a more positive point on the trigger signal. When the LEVEL control is set in the (-) region, the trigger circuit responds at a more negative point on the trigger signal. Figure 2-2 illustrates this effect with different settings of the SLOPE switch.

To set the LEVEL control, first select the TRIGGERING MODE, COUPLING, SOURCE, and SLOPE. Then set the LEVEL control fully counterclockwise and rotate it clockwise until the display starts at the desired point.

## Horizontal Sweep Rates

The TIME/DIV buttons provide calibrated sweep rates from 500 milliseconds/division to 500 picoseconds/division in a 1-2-5 sequence. In most cases, selecting a sweep rate requires pressing two buttons: a numbered button ( $1,2,5,10,20,50,100,200$, or 500 ), and a bution representing the multiplier and units (ms, $\mu s, n s$, and $p s$ ). In some cases, the MAG button is also involved.

When an attempt is made to enter a sweep rate faster than 500 picoseconds/division, the 7B90P internal logic corrects the mistake as follows:

1) If a numbered button other than 500 is pressed while ps (picoseconds) is selected, the multiplier button is automatically switched to ns (nanoseconds).


Fig. 2-2. Effect of the LEVEL control and SLOPE switch on CRT display.
2) If the ps button is pressed while any numbered button other than 500 is selected, the number is automatically switched to 500 .

In either case, the button just pressed is accepted. This allows the operator to enter a valid sweep rate without regard to the order in which the buttons are pressed. For example, to $\varepsilon$ form 500 picoseconds/division to 20 microseconds/division, the operator can press $\mu s$ finst (intermediate state $=500 \mu s$ ) and then 20 . Or, 20 can be pressed first (intermediate state $=20 \mathrm{~ns}$ ) and then $\mu \mathrm{s}$.

The correct sweep speed is always indicated by which buttons are lit. It is not necessary to multiply the indicated sweep speed by a factor of 10 when selecting the MAGnified mode. When the MAG button is pressed, the indicated sweep speed automatically increases by a factor of 10 . The correct sweep speed is also indicated by the digital readout displayed on the CRT monitor or other display device.

## Sweep Magnification

On a TIME/DIV setting of 10 ns or slower, the MAG button can be used to expand the display by a factor of 10. (When the MAG bu'ton is lit, the sweep is magnified.) The center portion of the display is the part that is magnified (see Fig. 2-3). The equivalent length of the magnified sweep is more than 100 divisions; any 10 division portion can be viewed by adjusting the 7B90P POSITION control to bring the desired portion into view.

Magnified sweep mode is automatically selected when a TIME/DIV of 5 , 2 , or 1 nanoseconds/division or 500 picoseconds/division is selected. When in the magnified mode at a sweep rate of 1 nanoseconds/division or slower, pressing MAG returns the sweep length to 10 divisions and TIME/DIV to a rate 10 times slower. When in the magnified mode at a sweep rate of 500 picoseconds/division, the sweep length is also returned to 10 divisions when MAG is pressed, but TIME/DIV is switched to 10 nanoseconds/division, the fastest unmagnified sweep rate.

## Time Measurement

When making time measurements from the graticule, the area between the second and tenth vertical lines of the graticule provides the most linear time measurements (see Fig. 2-4). Position the start of the timing


Fig. 2-3. Operation of the sweep magnifier.
area to the second vertical line and adjust the TIME/DIV switch so the end of the timing area falls between the second and tenth vertical lines.

## Variable Hold Off

The HOLD OFF control improves triggering stability on repetitive complex waveforms by effectively changing the repetition rate of the horizontal sweep signal. The HOLD OFF control should normally be set to its minimum setting. When a stable display cannot be obtained with the TRIGGERING LEVEL control, the HOLD OFF control can be varied for an improved display. If a stable display cannot be obtained at any setting of the LEVEL and HOLD OFF controls, check the TRIGGERING COUPLING and SOURCE switch settings. (NOTE: When the 7B90P is used in a TEKTRONIX 7912AD Programmable Digitizer, the HOLD OFF control is overridden by the mainframe at certain sweep speeds.


Fig. 2-4. Area of the graticule providing the
most accurate tiame measurement.

## Applications

The following information describes procedures and techniques for making basic time measurements with the 7B90P and an associated TEKTRONIX programmable mainframe and amplifier. (To some extent, this information is also applicable to measurements made under program control.) These procedures provide enough detail to enable you to adapt them to other related time measurements. Contact your Tektronix Field Office or representative for assistance in making specific measurements.

Since time is a function of the sweep rate and the horizontal distance (in divisions) that the sweep travels, the time interval between any two points on a waveform can be accurately measured. The following procedures provide methods to measure some of the more common time-related characteristics of a wəveform such as period, frequency, rise time, fall time, and pulse width. The procedure for each of these measurements is essentially the same, except for the points between which the measurements are made. The time interval between any two selected points on a displayed waveform can be measured with basically the same technique.

## Period and Frequency Measurements

Perform the following procedure to measure the period and determine the frequency of a displayed waveform:

1) Connect the signal to be measured to the selected input of the amplifier unit.
2) Set the TRIGGERING switches and LEVEL control for a stable display (see General Operating Information for selecting proper triggering).
3) Set the vertical deflection factor and amplifier POSITION control for about a 5-division display, vertically centered on the graticule.
4) Set the TIME/DIV switches and 7B90P POSITION control for 1 complete cycle displayed within the center 8 graticule divisions as shown in Fig. 2-5.
5) Measure the horizontal distance in divisions over 1 complete cycle of the displayed waveform (see Fig. 2-5).
6) Multiply the horizontal distance measured in Step 5 by the TIME/DIV setting.

Example: Assume that the horizontal distance over 1 complete cycle is 7 divisions, and the TIME/DIV setting is 0.1 ms (see Fig. 2-5).

Using the formula:

$$
\text { Period }=\text { Horizontal distance } X \text { TIME/DIV setting }
$$

Substituting values:

```
Period = 7 X 0.1 ms = 0.7 millisecond
```

7) Determine the frequency of the displayed waveform obtained in steps 1 through 6 by taking the reciprocal of the period of 1 cycle.

Example: Assume that the period of the displayed waveform is 0.7 millisecond.


Fig. 2-5. Measuring the period and determiniag the
frequency of a displayed waveform.

Using the formula:

$$
\text { Frequency }=\frac{1}{\text { Period }}
$$

Substituting values:

$$
\text { Frequency }=\frac{1}{0.7 \mathrm{~ms}}=1.43 \mathrm{kilohertz}
$$

## Rise-Time and Fall-Time Measurements

Perform the following procedure to measure the rise time and fall time of a displayed waveform:

1) Connect the signal to be measured to the selected input of the amplifier unit. Obtain a triggered display as previously described.
2) Set the vertical deflection factor and amplifier POSITION control for about a 5-division display, vertically centered on the graticule.
3) Set the TIME/DIV switches and 7B90P POSITION control to display the rising or falling portion of the waveform within the center 8 graticule divisions as shown in Fig. 2-6 (see General Operating Information in this section for a discussion of timing measurement accuracy).
4) Determine rise time or fall time by measuring the horizontal distance in divisions between the point on the rising or falling portion of the waveform that is $10 \%$ and the point that is $90 \%$ of the total display amplitude (see Fig. 2-6)
5) Multiply the horizontal distance measured in step 4 by the TIME/DIV setting.

Example: Assume that the horizontal distance from the $10 \%$ to $90 \%$ points is 2.5 divisions and the TIME/DIV setting is $0.1 \mu s$ (see Fig. 2-6).

Using the formula:


Substituting values:

$$
\text { Rise Time }=2.5 \mathrm{X} 0.1 \mu \mathrm{~s}=0.25 \text { microsecond }
$$

## Pulse-Width Measurements

Perform the following procedure to measure the pulse width of a displayed waveform:

1) Connect the signal to be measured to the selected input of the amplifier unit. Obtain a triggered display as previously described.
2) Set the vertical deflection factor and amplifier POSITION control for about a 5-division display vertically centered on the graticule.


Fig. 2-6. Measuring the rise time and fall. time of a displayed waveform.
3) Set the TIME/DIV switches and 7B90P POSITTON control for 1 complete pulse displayed within the center 8 graticule divisions as shown in Fig. 2-7.


Fig. 2-7. Measuring the pulse width of a displayed waveforil.
4) Measure the horizontal distance in divisions between the $50 \%$ amplitude points of the displayed pulse (see Fig. 2-7).
5) Multiply the horizontal distance measured in step 4 by the TIME/DIV setting.

Example: Assume that the horizontal distance between the $50 \%$ amplitude points is 3 divisions, and the TIME/DIV setting is 0.1 ms (see Fig. 2-7).

Using the formula:

$$
\text { Pulse Width }=\begin{gathered}
\text { Horizontal } \\
\text { distance }
\end{gathered} \quad \times \begin{array}{r}
\text { TIME/DIV } \\
\text { setting }
\end{array}
$$

Substituting values:

```
Pulse Width = 3 X 0.1 ms = 0.3 millisecond
```


## SECTION 3

## PROGRAMMIMG

The 7 B 90 P can be operated by remote control over a versatile instrument bus known as the IEEE 488 bus. A detailed description of the bus is given in IEEE Standard 488-1975 as well as ANSI Standard MC 1.1-1975. A brief introduction to the IEEE 488 standard is included in this section as background information.

The 7B90P is interfaced to the IEEE 488 bus through the mainframe in which it is installed. The mainframe provides a transparent interface between the IEEE 488 bus and the internal 7000-Series bus used to program the plug-ins.

The Remote/Local state of the 7B90P is slaved to the Remote/Local state of the programable mainframe. When the mainframe is set to Remote mode, the 7 B 90 P can be operated by remote control over the IEEE 488 bus. Either the controller-in-charge or other designated talker and listeners can then set or read any of the 7B90P programmable functions. In Remote mode, the front panel is disabled except for the settings of the Sweep Calibration adjustment (SWP CAL) and the External Trigger Input Terminator switch (TERM); these are the only non-programmable funetions on the 7B90P.

After the $7890 P$ has been set to Remote state, it can be remotely controlled by messages sent over the bus. This remote programming can be accomplished by either of two types of device-dependent messages: a high-level language (ASCII character strings) or a low-level language (hexadecimal codes). The advantage of the highwlevel language is that messages can be sent with simple, easy-to-remember mnenonics. Since the plug-in itself decodes these high-level commands, it is not necessary to incorporate this decoding capability into special driver software. The main advantage of using the low-level code is that fewer keystrokes and less bus time is required than when programming with the high-level language. When access time is more important than ease of progranming, the low-level language of the $7 B 90 P$ should be used. Both methods of programing the 7B90P are explained following a brief introduction to the IEEE 488 bus.

## Introduction to the IEEE 488 Bus

The IEEE 488 bus is a versatile instrument bus designed to provide an effective communications link for data and instructions. The bus itself is entirely passive. The active components of the interface are contained within each device. Instruments designed to operate according to this universal standard can be connected directly to the bus and operated by a controller with appropriate programming. The instructions and data generated by instruments can be coded in either ASCII or binary. The IEEE standard specifies only the mechanical, electrical, and functional aspects of the interface. The operational, or device dependent, aspects of the system are purposely not specified to allow greater flexibility as to the types of devices that can be interconnected.

## A Typical System

The IEEE 488 bus uses eight data lines and eight control lines. Information is transferred bit-parallel, byte-serial by an asynchronous handshake. The handshake signals guarantee that each data byte has been transferred properly before allowing another byte to be transferred across the bus. This allows instruments with different data transfer rates to operate together if they conform to the handshake state diagrams defined in the IEEE 488 standard.

Types of Instruments. Instruments connected to the bus can be classified as either controllers, talkers, or listeners. A controller designates which devices are to talk or listen and exercises other bus management functions; at any given time, there can be only one controller. A talker is a device capable of transmitting data and instructions on the Data lines; there can be only one talker at a time to avoid confusion in message and data transfer. A listener is a device capable of responding to data or instructions received on the Data lines; there can be more than one listener at a time since no confusion results.

A device need not be a talker or listener or controller at all times. It may be idle part of the time. Other devices (such as a digital multimeter) may alternately function as a talker or listener depending on whether they are listening to instructions or generating data.

A typical system is diagrammed in Fig. 3-1. It includes a controller (such as a TEKTRONIX CP4165 Controller), a talker (such as a counter or
digital multimeter), and a listener (such as a line printer or signal generator). Also included is a TEKTRONIX 7912AD Programable Digitizer which may either talk or listen.


Fig. 3-1. A typical system based on the IEEE 488 Bus.

Types of Messages. Messages on the bus are either interface messages or device-dependent messages. Interface messages are used to manage the interface functions of the instruments. They designate talkers and listeners, determine local or remote operation of devices, indicate service requests, and communicate other important interface conditions. Device-dependent messages, by contrast, are not used to change the state or configuration of the interface, but are used to control the operating modes or device functions of designated instruments. Device-dependent messages can also
be data, such as waveform data generated by the TEKTRONIX 7912AD Programmable Digitizer.

Maximum Number of Devices. Up to 15 devices can be connected on the IEEE 488 bus. More than 15 devices can be interfaced if they are not directly connected to the bus but are interfaced through another device. Such a scheme is used for the 7B90P plug-in housed in a 7000 -Series programmable mainframe such as the 7912AD; the mainframe provides a transparent interface between the IEEE 488 bus and the plug-ins. Secondary addresses are used for the plug-ins. More than half of the main devices connected at any time must be powered-up for the system to be operational.

Maximu Cable Length. The maximum length of cable that can be used to connect a group of devices on the bus is:

1) 2 meters times the number of devices
2) Or 20 meters, whichever is less

Cables may be connected in either a star or linear configuration, or in a combination of the two methods.

Electrical Specifications. The relationship between the binary logic states of the bus and the voltages present on the signal lines is as follows:

Logical 1 corresponds to a low voltage level (+0.8 V or less) and the signal is said to be "asserted".

Logical 0 corresponds to a high voltage level (at least +2.0 V) and the signal is said to be "unasserted".

The electrical states are based on standard TTL (Transistor-Transistor Logic) levels where the power source does not exceed +5.25 Volts DC referenced to logic ground.

## Bus Signal Lines

The IEEE 488 bus is functionally divided into three component busses: an eight-line Data Bus, a three-line Transfer (or handshake) Bus, and a five-line Management Bus. This bus structure is diagrammed in Fig. 3-1.

The eight lines of the Data Bus (DI01 through DI08) are bidirectional active-low lines used to convey data or device-dependent messages. Device addresses and universal commands are also transferred over these lines when ATN is asserted. One byte of information is transferred over the bus at a time. DIO1 represents the least significant bit in the byte; DI08 represents the most significant bit. Data is transferred in bytemserial, bit-parallel fashion. Data bytes can be formatted in ASCII with or without parity, or they can be formatted in machine-dependent binary code. The term "machine-dependent binary code" refers to an internal binary format used by a device to store certain programs and data.

The Transfer Bus is used to communicate a handshake sequence that is executed between the talker and all designated listeners each time a byte is transferred over the Data Bus. This handshake sequence prevents the talker from placing a new byte on the bus until the slowest listener has captured the previous byte. Thus the talker can not transmit at a rate faster than can be received by the slowest listener. The three active-low lines of the Transfer Bus are NRFD, DAV, and NDAC. (See Fig. 3-2 for a basic timing relationship between these signals). Their functions are:

NRFD (Not Ready For Data) - This signal line is asserted until all assigned listeners are ready to receive the next data byte. When all of the assigned listeners have released NRFD, the NRFD signal is unasserted, thereby allowing the talker to place the next byte on the Data lines.

DAV (Data Valid) - The DAV signal line is asserted by the talker shortly after placing a valid byte on the Data lines. This tells each listener to capture the byte presently on the Data lines. DAV can not be asserted until NRFD has been unasserted.

NDAC (Not Data Accepted) -- This signal line is asserted by each listener until all have captured the byte currently on the Data lines. When the slowest listener has captured the data byte, NDAC is unasserted thereby allowing the talker to remove the byte from the Data lines. At that point, the DAV line is unasserted and the entire handshake cycle is repeated.


Fig. 3-2. A typical hamshake sequence.

The Management Bus is a group of signal lines used to control data transfers over the Data Bus. These lines communicate important interface messages such as detecting an interrupt from a device, setting a device to remote control, and denoting the end of a message. The five active-low signal lines are ATN, IFC, SRQ, REN, and EOI; their functions are:

ATN (Attention) -- Asserted by the controller-in-charge to
specify how information on the Data Bus is to be interpreted. When ATN is not asserted, the information on the Data Bus is interpreted as device-dependent messages and data. When ATN is asserted, the Data Bus conveys universal commands, addressed commands, talk addresses (MTA), listen addresses (MLA), or secondary addresses (MSA). Just which addresses and commands are sent depends upon the byte currently on the Data Bus. The codes corresponding to various commands and addresses are defined in Appendix E of the IEEE 488 standard.

IFC (Interface Clear) -w Asserted by the system controller to initialize the interface functions of all instruments to a quiescent state and return control to the system. The IFC function does the same thing as UNListen, UNTalk, Serial Poll Disable, and resets all devices except the system controller to the idle state.

SRQ (Service Request) -- Asserted by an instrument to request service from the controller-in-charge. The controller usually interrupts its current task and conducts a serial poll to determine which device interrupted. The controller can then branch to an interrupt service routine where appropriate action is taken. After the interrupt has been processed, the controller may resume execution of the previous task.

EOI (End Or Identify) -- Asserted by a talker to indicate the last byte of its message. When EOI is asserted with ATN, the controller conducts a parallel poll of the devices connected to the bus.

REN (Remote Enable) -- Asserted by the controller-in-charge to allow devices on the bus to go to Remote mode, thereby allowing remote control of their programmable functions. When in Remote mode, the front panels of the instruments are disabled except for any non-programmable functions.

## Bus Messages

As previously noted, messages on the Data Bus are either interface messages or device-dependent messages. When the ATN line is asserted by the controller, all devices "pay attention" since interface messages are to be transferred over the Data lines. (By "pay attention" it is meant that all devices handshake and process all bytes transferred on the bus.) Interface messages can generally be classified as follows:

1) talk addresses
2) listen addresses
3) secondary addresses
4) universal commands
5) addressed commands

The first three categories refer to how a device is to be addressed. That is, they designate whether a device is to be a talker or a listener. To designate a device as a talker, the controller asserts ATN and places the device's talk address on the Data lines. Similarly, the controller designates a listener by asserting ATN and placing the address of the listener on the Data lines. In cases where secondary addressing is designed into a particular device, it is necessary to transmit the device's secondary address following the primary talk or listen address.

The fourth category listed above (universal commands) consists of those interface commands which affect all instruments connected to the bus, regardless of whether they are currently addressed as talker or listeners. Examples of universal commands are LLO (Local Lockout) and DCL (Device Clear).

The fifth category in the preceding list (addressed commands) consists of those interface commands which affect all instruments currently addressed as listeners. Examples of addressed commands are GTL (Go To Local) and GET (Group Execute Trigger). A complete list of universal and addressed commands is provided in Appendix E of the IEEE 488 standard.

In contrast to interface messages, device-dependent messages are sent with ATN unasserted and are transmitted only between a designated talker and one or more designated listeners. A device-dependent message can be either an instruction (e.g., set the input polarity to normal) or data (e.g., 3.456 volts). The format of instructions and data js entirely up to the device designer. Instructions and data are normally coded in ASCII or binary, but this is not required by the IEEE standard.

This has only been a brief introduction to the IEEE 288 interface. Further information can be found in IEEE Standard 488-1975, IEEE
Standard Digital Interface for Programmable Instrumentation. A detailed description of the actual handshake timing sequence is covered in Appendix $B$ of the standard.

## 7B90P Interface Function Subsetis

The IEEE 488 standard is designed in such a way that not all devices on the bus need to have the same capability to comply with the standard. The instrument designer can choose from a "menu" of device functions, and implement only those capabilities (known as "functional subsets") that are appropriate to a particular device. The functional subsets are described in detail in the standard. The degree to which the 7B90P implements each of the ten interface functions is described below.

1) Source Handshake Function: SH 1

The SH function provides a device with the ability to initiate and terminate transfer of multiline messages on the Data Bus. The 7B90P, in conjunction with the mainframe, conforms to subset SH1, meaning it has full capability with no states omitted.
2) Acceptor Handshake Function: AH1

The AH function provides a device with the capability to guarantee proper reception of messages on the Data Bus as well as the capability of delaying initiation or termination of such messages. The 7B90P conforms to subset AH1, meaning it has full capability with no states omitted.
3) Talker Function: TE6

The $T$ function enables a device to send device-dependent data (including status information) over the bus to other devices. The 7B90P conforms to subset TE6, meaning it is an extended talker honoring secondary addresses. It has full capability except that it does not have a Talk-Only mode.
4) Listener Function: LE 4

The L function allows a device to receive device-dependent data over the bus from other devices. This capability exists only when the device is addressed to listen. The TB90P conforms to subset LE4, meaning it is an extended listener honoring secondary addresses. It has full capability except that it does not have a Listen-Only mode.
5) Service Request Function: SR1

The SR function enables a device to asynchronously request service from the controller-in-charge of the interface. The 7B90P conforms to subset SR1, meaning it has full capability.
6) Remote/Local Function: RL1

The RL function provides a device with the capability to select between two sources of information: remote (programmed control) or local (front-panel control). The 7B90P, in conjunction with the mainframe, conforms to subset RL1, meaning it has full capability.
7) Parallel Poll Function: PP $\emptyset$

The PP function allows a device to present one bit of status to the controller-in-charge without being previously addressed to talk. The 7B90P conforms to subset PPø, meaning it has no capability for responding to a parallel poll.
8) Device Clear Function: DC1

The DC function allows a device to be cleared (initialized) either by itself or as a group of devices. The 7B90P conforms to subset DC1, meaning it has full capability. When the 7B90P receives a DCL (Device Clear) interface message, its internal buffers are cleared and the front panel is set to the powerup state. However, DCL does not affect plug-in settings in Local mode.
9) Device Trigger Function: DT 1

The DT function allows the operation of a device to be triggered (initiated) either individually or as part of a group. The 7B90P conforms to subset DT1, meaning it has full capability. The 7B90P does a Single--Sweep Reset when it receives a GET (Group Execute Trigger) command.
10) Controller Function: C $\varnothing$

The C function provides a device with the capability for sending device addresses, universal commands, and addressed commands over the bus. The 7B90P conforms to subset $C \varnothing$, meaning it cannot serve as controller.

## Addressing the 7B90P

It was noted previously that two types of communication occur on the IEEE 488 bus: interface messages and device-dependent messages. The first type of communication occurs when the system controller asserts ATN and begins sending messages to all devices on the bus. The second type of communication occurs when ATN is unasserted; in this case, communication occurs only between a designated talker and one or more designated listeners.

Interface messages can occur without the assignment of talkers and listeners since, by definition, the controller has an active source handshake and all other devices have an active acceptor handshake. On the other hand, before a device-dependent message can be sent, a talker must be assigned and at least one listener must be assigned. The process of assigning talkers and listeners is referred to as "addressing".

## Addressing Schemes

There are two types of addressing schemes allowed by the IEEE 488 standard: primary addressing and secondary addressing. It is up to the device designer to decide which type of addressing scheme is to be implemented on a particular device. That is, some devices use only primary addressing while others require secondary addressing. The 7B90P and its mainframe use a secondary addressing scheme.

On a device that uses primary addressing, up to 31 possible talk or listen addresses can be assigned to a particular device. Once a talk or listen address has been assigned to a device, this address becomes the "name" by which the controller refers to that device. However, this "name" can usually be changed since the address is generally determined by a bank of switches on the back panel of the instrument. (In some cases, the address is determined by straps on a circuit board inside the instrument.)

On a device that uses secondary addressing, a slightly different scheme is used. In this case, there are still 31 possible talk or listen addresses available. However, for each primary address there can be 31 possible secondary addresses corresponding to the primary address. This allows a total of 961 possible addresses for a device incorporating secondary addressing. Secondary addressing is sometimes used on instruments incorporating plug-in modules. An example is the TEKTRONIX 7912AD Programmable Digitizer and its two programmable plug-ins: the 7A16P and the 7B90P.

Primary and secondary addresses are sent over the Data Bus with ATN asserted. Data lines DIO1 through DIO5 are used to convey the actual device address. DI01 corresponds to the least significant bit and DIO5 corresponds to the most significant bit. Data lines DIO6 and DIO7 determine whether this address is to be interpreted as a primary talk address (abbreviated MTA for My Talk Address), a primary listen address (abbreviated MLA for My Listen Address), or as a secondary address (abbreviated MSA for My Secondary Address). Incidentally, if DIO6 and DIO7 are $\varnothing \varnothing$, lines DI01-5 correspond to a universal or addressed comand rather than a device address. This scheme is diagrammed in Fig. 3-3.


Fig. 3-3. Interpretation of the Data Bus (DI01-8) when ATN is asserted.

## Determining Device Addresses

The 7B9OP does not have any provision for changing its device addresses since these addresses are a function of the mainframe in which the plug-in is installed. In the case of the TEKTRONIX 7912AD Programmable Digitizer, the MSA of the plug-in installed in the vertical compartment (such as the 7A16P) is interpreted to be the MSA of the 7912AD plus 1. Similarly, the MSA of the plug-in installed in the horizontal compartment (such as the "(B90P) is interpreted to be the MSA of the 7912AD plus 2. However, this rule is not necessarily true for all mainframes. Refer to the manual for your particular mainframe if you need more information on this point.

So far, we have talked about addressing in general terms but have not discussed the actual procedure for addressing the 7B90P for a talk or listen operation. Perhaps an example using the 7912AD will best illustrate the method. Suppose that the 7912 AD is set for a primary address of $\emptyset \emptyset \emptyset \emptyset \emptyset$ and a secondary address of $\varnothing \varnothing \varnothing \emptyset 1$. According to the scheme previously mentioned, the MLA, MTA, and MSA of the mainframe are:

$$
\begin{aligned}
& M L A=\emptyset 1 \emptyset \emptyset \emptyset \emptyset \emptyset=40_{8}=32_{10}=20_{16} \\
& M T A=1 \emptyset \emptyset \emptyset \emptyset \emptyset \emptyset=100_{8}=64_{10}=40_{16} \\
& M S A=11 \emptyset \emptyset \emptyset \emptyset 1=141_{8}=97_{10}=61_{16}
\end{aligned}
$$

To address the 7B90P as a listener, the controller asserts ATN and sends the MLA of the mainframe $\left(32_{10}\right)$ on the Data Bus. With ATN still asserted, the controller sends the MSA of the 7B90P $\left(99_{10}\right)$ on the Data Bus. Recall that the MSA of the time base is that of the 7912AD plus 2. ATN can now be unasserted so that the talker can begin sending devicedependent messages to the 7B90P.

To address the 7B90P as a talker, the controller asserts ATN and places the MTA of the mainframe $(6410)$ on the Data Bus. With ATN still asserted, the controller places the MSA of the 7B90P (9910) on the Data Bus. ATN can now be unasserted so that the listener(s) can begin receiving device-dependent messages from the 7B90P.

## High-Level Messages

7B90P high-level messages are device-dependent and therefore are not specified in the IEEE 488 standard. However, these high-level messages do conform to Tektronix standards intended to enhance Tektronix compatibility with other bus-compatible instruments. To accomplish this, codes and syntax are designed to be unambiguous, to correspond to those used by other Tektronix devices, and to be as simple and obvious as possible. This minimizes the cost and time required to program the 7B90P by making it easier for the programmer to write and understand the needed device-dependent code.

The 7B90P responds to device-dependent messages that contain one or both of two types of commands: Set and Query. During a Set conmand, the plug-in acts as a listener and uses the incoming information to effect certain operating modes or front-panel settings. During a Query command, the plug-in initially acts as a listener until receiving an UNListen command then the plug-in is made a talker and returns the status of a specified function or operating parameter. The syntax of a Set and Query cormand is explained later.

A device-dependent message begins when the plug-in is addressed as a talker or listener, with ATN asserted. Then ATN is unasserted and the actual device-dependent code is transmitted from talker to listener. The message is terminated when EOI is asserted. EOI is sent concurrent with the last byte of the message, whether it be a data byte, a delimiter, or a format character.

When special driver software is being used, the user need not concern himself with all the details of initiating and terminatirg each message, since this is done automatically by the software. Most of this section thus deals with message syntax rather than message transmission. However, for those who are not using special-driver software, a brief discussion is included later as to the methods of initiating and terminating a message.

## Explanation of Command Syntax

In describing the syntax of high-level commands, a modified form of the Backus waur format is used. According to this format, descriptive words such as "header", "format", or "argument" are enclosed in angle brackets (<>). Optional items are enclosed in square brackets ([]). An ellipsis
（．．．）indicates that the preceding argument may be repeated one or more times．Punctuation marks such as the semicolon and question mark are listed exactly as they are typed；hence they are not enclosed in angle brackets．

The syntax allows format characters at several pcints in each command． Where they are allowed，the word＂format＂is spelled cut．Format characters are always optional，so the word＂format＂is always shown in brackets．A format item can be a space，carriage return，line feed，or any combination of these characters．

## Set Comands

Unless specifically designated as query（read）only，the headers and arguments listed in Table 3－1 can be used to form Set commands．Set commands are used to set the states of the various $7 B 90 P$ programmable functions．A Set command for the 7B90P has the following syntax：
［＜format＞］＜header＞＜header delimiter＞［＜format＞］＜argument＞

As previously noted，a format character is optional and may be a carriage return，line feed，space，or any combination thereof．The header delimiter is a space and is always required．The following are all examples of legal Set comands，where＜cr＞denotes a carriage return and 〈lf〉 denotes a line feed：

1） CPL DC
2）$\langle$ If $\rangle$ CPL $D C$
3）$\langle c r>C P L D C$
4）〈cr＞＜lf〉 CPL DC＜cr＞
5）T／D． 005
6）〈cr＞T／D．$\varnothing 05$

Examples 1 through 4 set the trigger CouPLing to DC and examples 5 and 6 set the Time／Division to 5 milliseconds．

One or more Set commands can be included in the same message as explained later under the heading＂Messages＂．

## Query Commands

All of the headers in Table 3-1 can be used to form Query commands. Query commands are used to determine the states of the various 7B90P programable functions. A Query command has the following syntax:
[<format>]<header>?

Again, <format> refers to an optional carriage return, line feed, space, or any combination thereof. The question mark (?) following the header argument must be typed without a preceding space or other format character.

Examples of Query commands are:

1) CPL?
2) $\langle c r\rangle \mathrm{CPL}$ ?
3) 〈cr><lf〉 CPL?
4) $T / D$ ?
5) $\langle 1 \mathrm{f}\rangle \mathrm{T} / \mathrm{D}$ ?

Examples 1 through 3 query the state of the CouPLing mode and examples 4 and 5 query the setting of the Time/Division switches.

One or more Query commands can be included in the same message as explained next.

TABLE 3-1

7B90P HIGH-LEVEL (ASCII) COMMAND SET

| HEADER | ARGUMENT | DESCRIPTION |
| :---: | :---: | :---: |
| MOD | $\begin{aligned} & \text { PPA } \\ & \text { NOR } \\ & \text { SSW } \end{aligned}$ | Peak-to-Peak Automatic triggering mode is selected. Normal triggering mode is selected. <br> Single-sweep triggering mode is selected. |
| CPL | AC <br> DC <br> LFR <br> HFR | Trigger signal is AC coupled. <br> Trigger signal is DC coupled. <br> Trigger signal is $A C$ coupled with low frequency rolloff. <br> Trigger signal is $A C$ coupled with high frequency rolloff. |
| SRC | $\begin{aligned} & \text { INT } \\ & \text { LIN } \\ & \text { EXT } \\ & \text { E1 } \end{aligned}$ | Trigger source is internal. <br> Trigger source is the line voltage. <br> Trigger source is external input. <br> Trigger source is external attenuated 10X. |
| SLO | $\begin{aligned} & \text { POS } \\ & \text { NEG } \end{aligned}$ | Positive trigger slope is selected. <br> Negative trigger slope is selected. |
| LEV | xxxx | Trigger level is set to xxxx; range is -6.4 to +6.35 in .05 steps. Query returns <nr2>. 1 |
| T/D | Xxxx | Time/Division is set to xxxx; range is $5 \mathrm{E}-10$ to $5 \mathrm{E}-1$ in a 1-2-5 sequence. Query returns 〈nr3>. 1 |
| MAG | $\begin{aligned} & \text { ON } \\ & \text { OFF } \end{aligned}$ | Sweep magnifier is turned on (10X). <br> Sweep magnifier is turned off (1X). |
| POS | Xxxx | Horizontal position of sweep is set to xxxx; range is -6.4 to +6.39 in 0.0125 steps ( 80 steps/division). Query returns <nr2>. 1 |
| HOL | Xxxx | Trigger holdoff period is xxxx; range is 0 to 63 uncalibrated. |
| EOS | ON | End-of-Sweep SRQ signal is enabled. |

(cont. next page)

| TRI | ON <br> OFF | Trigger light is on（read only）． <br> Trigger light is off（read only）． |
| :--- | :--- | :--- |
| SSW | ARM | Single Sweep is armed．A GET（Group Execute Trigger <br> universal command）has the same effect． <br> Single Sweep is disarmed（read only）． |
| ID |  | Returns the plug－in type：TEK／7B90P，V77．1，LLL <br> This means Tektronix 7B90P with language version 77．1 <br> and low－level language．2 |
| SET | Returns the setting of all of the above functions with <br> the exception of the read－only ones． |  |

FOOTNOTES：
1．〈nr1〉，＜nr2＞，and＜nr3＞are numerical formats defined in ANSI Standard X3．42．

2．The 77.1 nomenclature refers to the version of protocol implemented；it may be updated at a future date．

## Suxamary of ANSI 83.42 Muserical Formats

＜nr1＞Signed or unsigned integers（no decimal point）preceded optionally by spaces（＜sp＞）．The＜nr1＞representation of value＂zero＂must not contain a minus sign．

Examples：＋1245，〈sp＞－328，〈sp＞＜sp＞475，＋øøøø
＜nr2＞Signed or unsigned numbers with a decimal point preceded optionally by spaces（＜sp＞）．The＜nr2＞representation of value＂zero＂must not contain a minus sign．

Examples： $0.123,\langle\mathrm{sp}\rangle+5.41,-6.42 \emptyset,\langle\mathrm{sp}\rangle\langle s p\rangle \varnothing . \varnothing \varnothing \varnothing$
＜nr3＞Floating－point numbers expressed in modified scientific notation．The mantissa always includes a decimal point and is preceded by sign（ + ，－，or 〈sp＞）．The exrad following the mantissa begins with the character E followed by a plus or minus sign and then one or more digits for the exponent． The＜nr3＞representation of value＂zero＂must contain an＜nr2＞ zero followed by an exrad with plus sign and all zero digits．

```
Examples: -1.5E+\varnothing3, <sp>2.E-1, +5.\emptysetE-2, +\varnothing. ØE + व\varnothing
```


## Messages

A Set or Query command is referred to as a "message unit". One or more message units can be concatenated to form a message, if each unit is delimited by a semicolon (;). Using the preceding terminology, the syntax of a message may be described as:
<message unit>[; [<format>]<message unit>]...[;][<format>]

Notice that format characters may be included but are not required. Also, the ellipsis indicates that one or more message units, and accompanying optional format characters, may be included if they are delimited by a semicolon.

The simplest message consists of a single Set or Query command. For example, the following Set message sets the Time/Division to 5 milliseconds:

T/D. . 005

Similarly, the following Query message reads the setting of the Time/Division switches:

T/D?

When the $7 \mathrm{B90P}$ is made a talker, it responds by sending the current Time/Division setting. For the case of the preceding example, the following query response would be obtained:

T/D 5.E-3

This indicates that the time/division setting is $5 \times 10^{m} 3$ seconds ( 5 milliseconds).

Messages with Multiple Sets. More than one Set command can be sent in the same message by concatenating them with semicolons. For example:

SLO NEG; MAG OFF

This message sets the triggering SLOpe to the NEGative portion and then turns the sweep MAGnifier OFF.

Here are some other example of messages with multiple Set commands:

1) MOD PPA;CPL AC; MAG OFF;T/D. $\emptyset \emptyset \emptyset 5$


In each of the preceding two examples, there are four Set commands that accomplish the same settings, namely: triggering MODe is set to Peak-to-Peak Auto, AC CouPLing of the trigger signal is used, the sweep MAGnifier is turned OFF, and the Time/Division is set to 500 microseconds.

Messages with Multiple Queries. In a similar fashion, more than one Query command can be executed in the same message. With the 7B90P addressed as a listener, each query is received and the results are queued internally until the plug-in is made a talker. When the 7B90P is made a talker, the results of the query are output in the same order they were received. The response to the query is valid at the time of the response, rather than at the time the query is received. The following is an example of a multiple-query message and its resulting output:


Notice that the results of the query are executed in sequence and that a semicolon, carriage return, and line feed are generated after all but the last response.

When more than one Query command of the same type is included in the same message, only the last occurrence of the Query is executed. For example:

MAG?; EOS?; MAG?

Executing the above message would cause an output such as:

EOS ON;
MAG OFF

Since two MAG queries were executed in the same message, the first one was ignored.

The SET? Query. A special type of Query command is the SET? query. It interogates the status of all the programable functions of the 7B90P. The SET? query is sent just like any other query message. That is, the $7 \mathrm{B90P}$ is made a listener and then SET? is sent over the Data Bus. When the plug-in is made a talker, an output such as the following is obtained:

```
T/D 1.E-6;
POS -0.25;
HOL 16;
MAG OFF;
MOD PPA;
CPL AC;
LEV -6.4;
EOS OFF;
SLO POS;
SRC INT
```

Normally, a SET? query will be the only query in a message. When a SET? query follows some other query in the same message, all queries preceding the SET? query are ignored. Also, any queries following the SET? query cause the SET? query to be ignored.

Messages with Sets and Queries. When Set and Query commands are included in the same message, the following action is taken: First, all Set commands are executed in the order they are received. This is true even if there are more than one Set command for the same function. After all Set commands are executed, the plug-in is made a talker and the results of the Query comands are executed in sequence. If more than one query for the same programmable function is received, all but the last occurence of that query is ignored.

To demonstrate the effect of multiple Sets and Queries in the same message, consider the following:

MOD PPA; MOD?; MOD NOR; MOD?

Executing the above message first sets the triggering mode to Peak-to-Peak Auto mode momentarily. Then, without further delay, the triggering mode is returned to normal as the next Set command is executed. With both Set commands executed, the queries are then executed. Since the MOD? query occurs twice, only the second occurence is honored. When the plug-in is made a talker, the following output is seen:

MOD NOR

Query Responses. The 7B90P responds to a query by sending the status of the queried function at the time of the response rather than at the time the query is received. A query remains valid (if the plug-in is made a talker, it will return the requested status) until one of the following occurs:

1) the plug-in is made a talker and sends the requested status. 2) a later message unit contains the same query (in this case the old query is cleared and replaced by the new one).
2) a DCL (Device Clear interface message) is received.

This allows you to mix Set and Query commands without regard to where the query is in the message. For example, executing:

SLO POS; SLO?; SLO NEG
causes a response of:

SLO NEG

Similarly, executing:

MAG ON ; SET?; MAG OFF
results in an output such as:

T/D 1.E-6;
POS - $-7.2 \varnothing$;
HOL 17;
MAG OFF;
MOD NOR;
CPL AC;
LEV -6.4;
EOS OFF;
SLO NEG;
SRC INT

Notice that in each case, the result of the query is valid as though it was included last in the message.

## Low-Level Messages

All of the programmable functions of the 7B90P can be set or queried by a low-level language that is completely redundant to the high-level language. The low-level code merely provides an alternative method of commication that requires less bus time for programming the plug-in because data is moved in fewer bytes. Also, the language is decoded faster.

Unlike the high-level language which allows Set and Query commands to be combined in the same message, the low-level language requires that Sets and Queries be entered as separate messages. However, it is possible to set or query more than one function in the same message, provided certain rules of syntax are followed.

The following conventions are used in discussing the 7B90P low-level code. The term "hexadecimal" or "hex" refers to a base-16 number. Syntax items shown inside square brackets ([]) are optional and can therefore be omitted. Angle brackets (<>) are used to show that the enclosed argument is a descriptive term.

## NOTE

The TEKTRONIX 7912AD Programmable Digitizer has an internal strap that allows it to generate an EOI when a linefeed is detected. If the 7912AD is strapped in this way and a 7B90P is installed, the 7B90P low-level code will not be functional. However, the high-level language will be functional if all messages to the plug-in are terminated by at least 17 spaces.

## Set Messages

Both Set and Query messages are implemented in low-level code. The format of a Set message is as follows, where each line is a separate 8-bi.t byte representing a hexadecimal number:

```
        15
        <address>
        <data>
        <data>
    .
    .
        <checksum>
```

The hex 15 (ASCII NAK character) identifies this message as a low-level Set message. The <address> byte specifies the starting address in 7B90P memory for storing the <data> byte(s) which follow(s). The <address> byte corresponds to a particular programmable function and the first <data> byte corresponds to the setting of that function. (See Table 3-2 for the "hexadecimal" codes corresponding to various functions and their settings). The allowable range for the starting address is $\varnothing \varnothing$ to hex $\varnothing$ C. This address is automatically incremented between <data> bytes if more than one <data> byte is sent. If the address is automatically incremented beyond $\varnothing \mathrm{C}$ by sending too many bytes, the extra bytes are ignored. If arguments other than the ones specified in Table 3-2 are sent, the 7B90P may be set to an illegal operating mode, but no damage will result.

The checksum is the 2 's complement of the modulo-256 sum of all the preceding bytes of the message (including the 15 and address byte). Therefore the modulo-256 sum of all bytes in the message, including the checksum, is zero (for a correctly transmitted block).
table 3-2a

7B90P LOW-LEVEL (HEXADECIMAL) CODES

| ADDRESS | DATA | DESCRIPTION |
| :---: | :---: | :---: |
| $0 \varnothing$ | 90 | Plug-in Type. (read only) <br> Any data other than hex $9 \varnothing$ written to this address causes a command error. |
| ø1 | xx | Hold-Off <br> Range $\varnothing \varnothing$ to FC , low two bits must be zero |
| $\square 2$ $\boxed{y}$ | xX xx | ```Position High bits Position Low bits. Position range moves center of sweep -6.39 divisions ( \(\varnothing \varnothing \emptyset \emptyset\) ) to +6.4 divisions ( \(\varnothing 3 \mathrm{FF}\) ). Center screen \(=\varnothing 2 \emptyset \emptyset\). Step size \(=0.0125\) divisions.``` |
| 04 | $\begin{aligned} & \emptyset \emptyset \\ & \emptyset 8 \end{aligned}$ | Sweep Magnifier ${ }^{1}$ Magnifier Off Magnifier On |
| $\square 5$ $\varnothing 6$ | xx xx | Sweep Rate <br> High bits, see Table 3-2b <br> Sweep Rate <br> Low bits, see Table 3-2b |
| $\boxed{\square}$ | xx | ```Trigger Level Range = -6.4 divisions ( }\varnothing\mathrm{ ) to +6.35 divisions (FF), with step size of 0.05 div.``` |

(cont. next page)

| ADDRESS | Data | DESCRIPTION |
| :---: | :---: | :---: |
| $\emptyset 8$ | $\begin{aligned} & \emptyset 2 \\ & \varnothing E \\ & \emptyset B \\ & \emptyset 8 \end{aligned}$ | Trigger Coupling AC LFR HFR DC |
| $\emptyset 9$ | $\begin{aligned} & 1 \varnothing \\ & 2 \emptyset \\ & C \emptyset \\ & 4 \varnothing \end{aligned}$ | Trigger Source <br> Internal <br> Line <br> External <br> External:10 |
| $\emptyset A$ | $\begin{aligned} & \varnothing \varnothing \\ & \emptyset 8 \end{aligned}$ | Trigger Slope Negative Positive |
| $\emptyset \mathrm{B}$ | $\begin{aligned} & \varnothing \varnothing \\ & 4 \varnothing \\ & 80 \end{aligned}$ | Trigger Mode $\quad$ Normal PPA Single Sweep |
| $\emptyset C$ | x $x$ | Single Sweep Arm, End-of-Sweep SRQ Enable See Table 3-2c |

## FOOTMOTE

1 To set read-out and front panel, sweep rate low bits must be changed also. If not changed, sweep is still magnified, but T/D? will return T/D error.

TABLE 3-2b

TIME/DIV CODE

| TIME/DIV | MAG OFF | MAG ON |
| :---: | :---: | :---: |
| 500 ps | - - - | 0118 |
| 1 ns | - - - | Ø1 11 |
| 2 ns | - - - | 00 11 |
| 5 ns | - - - | Q7 11 |
| 10 ns | 0115 | 1115 |
| 20 ns | Ø0 15 | 1015 |
| 50 ns | $\emptyset 715$ | 1715 |
| 100 ns | 1119 | 1519 |
| 200 ns | 1019 | 1419 |
| 500 ns | 1719 | 1 B 19 |
| 1 s | 1512 | 1912 |
| 2 s | 1412 | 1812 |
| 5 s | 1B 12 | 2712 |
| 10 s | 1916 | 2516 |
| 20 s | 1816 | 2416 |
| 50 s | 2716 | 2B 16 |
| 100 s | 25 1A | 291 A |
| 200 s | 24 1A | 281 A |
| 500 s | 2B 1A | 47 1A |
| 1 ms | 2913 | 4513 |
| 2 ms | 2813 | 4413 |
| 5 ms | 4713 | 4 B 13 |
| 10 ms | 4517 | 4917 |
| 20 ms | 4417 | 4817 |
| 50 ms | 4 B 17 | 4 F 17 |
| 100 ms | 49 øВ | - - - |
| 200 ms | 480 B | - - - |
| 500 ms | $4 F \square B$ | - - |

TABLE 3-2c

SINGLE-SWEEP, EOS CODE

| CODE | DESCRIPTION |
| :---: | :---: |
| xxyø xxxx <br> Øxxx xxxx <br> 1xxx xxxx | Single Sweep Arm (write only). <br> Enable End-of-Sweep SRQ (write only). <br> Disable End-of-Sweep SRQ (write only). |
| $\begin{aligned} & \varnothing ø \mathrm{xx} \\ & 1 \mathrm{xxx}_{\mathrm{xx}} \emptyset \varnothing \varnothing \end{aligned}$ | End-of-Sweep SRQ Enabled (read only). <br> End-of-Sweep SRQ Disabled (read only). |
| $\begin{aligned} & \mathrm{x} \varnothing \emptyset \mathrm{x} \emptyset \emptyset \emptyset \emptyset \\ & \mathrm{x} \varnothing 1 \mathrm{x} \emptyset \varnothing \varnothing \varnothing \end{aligned}$ | Sweep Triggered (read only). <br> Sweep Not Triggered (read only). |
| $\mathrm{x} \varnothing \mathrm{x} \varnothing \varnothing \varnothing \varnothing$ <br> $x \not x \times 1$ øøøø | Single Sweep Armed (read only). <br> Single Sweep Disarmed (read only). |

## Computing the Checksum

To compute the checksum, required as the last byte of a low-level Set or Query message, do the following:

1) Find the hexadecimal (base-16) sum of the preceding bytes of the message.
2) If the sum found in step 1 is greater than hex FF ( $256_{10}$ ), convert it to modulo-256 by successively subtracting FF until the remainder is less than FF but greater than 0.
3) Convert the remainder from step 2 to binary representation.
4) Find the 2's complement by complementing all bits and adding 1 .
5) Convert the resulting number back to hexadecimal representation, if desired.

To determine what <data> and <address> bytes are required to set a particular operating parameter, refer to Table 3-2. For example, let's suppose you wanted to set the triggering slope of the 7B90P to negative. By looking at the table, you note that the address of the trigger slope switch is $\varnothing A$ and the value corresponding to negative slope is $\emptyset \emptyset$. Thus the following hexadecimal numbers should be entered in sequence to effect negative-slope triggering:

15
$\emptyset A$
$\emptyset \emptyset$
E 1

The last byte sent (E1) is the 2's complement of the modulo-256 sum of the preceding three bytes (hex 1F).

Consider another example. Suppose that we want to set the following trigger parameters via the low-level code:

```
trigger level: -6.4 divisions
trigger coupling: DC
trigger source: internal
trigger slope: positive
trigger mode: Peak-to-Peak Automatic
```

To effect these settings, the following hexadecimal numbers must be sent over the Data lines in the sequence listed:

15
07
$\square \varnothing$
$\varnothing 8$
10
$\emptyset 8$
40
84

The explanation for the above numerical sequence is as follows: First a 15 is sent to indicate the beginning of a low-level Set message. Next, a $\emptyset 7$ is sent which is the address corresponding to the 7B90P trigger level. The $\emptyset \emptyset$ which follows sets the trigger level to -6.4 divisions as shown in Table $3-2 a$. The rest of the numbers in the sequence ( $08,10,08$, and 40 )
correspond to DC trigger coupling, internal trigger source, positive trigger slope, and PP AUTO triggering mode, respectively. The last number in the sequence ( 84 ) is the checksum.

When setting more than one programmable function with the same low-level Set message, it must be remembered that the address counter of the plug-in is incremented automatically if more than one data byte is sent. Thus the data bytes must be entered in sequence such that they correspond to the incremented address counter. In other words, successive Set codes can be entered only if they correspond to the normal sequence of the address counter as listed in Table 3-2a. If they do not correspond, then more than one low-level Set message must be used to set more than one programmable function.

## Query Messages

Using the same syntax notation as just used for Set messages, the format of a low-level Query message can be defined as:

```
        11
```

$\left[\begin{array}{l}\text { <address }\rangle \\ {[\langle\text { count }\rangle]}\end{array}\right]$
<checksum>

The hex 11 (ASCII DCI character) identifies this message as a low-level Query message. The <address> byte specifies the first address in 7B90P memory to be read. The <count> byte specifies the number of addresses to be read. The <address> byte corresponds to a particular programmable function and the <count> byte designates how many successive functions are to be queried (see Table 3-2). If 〈address> and <count> are both omitted, all addresses are read beginning with address zero. If just <count> is omitted, only the address specified is read (count defaults to 1).

If the address specified is within range ( $\varnothing \emptyset$ to hex $\emptyset C$ ), but <count> is too large, no error is reported; however, the output line includes data only up to and including $\varnothing \mathrm{C}$.

The checksum is always required and is computed in the same manner as for the Set message. That is, the checksum is the 2's complement of the modulo-256 sum of all the preceding bytes in the message.

To determine what <address> byte should be used in a particular Query message, refer to Table 3-2. If no <count> byte is included, the specified <address> byte will determine the programmable function to be queried. If a <count> byte is included, the <address> byte determines the first function to be queried, and as the address counter is incremented, additional functions following the first one queried are also queried.

As an example, suppose you wanted to query one function $-\cdots$ the setting of the sweep magnifier switch. In this case, no <count> byte is necessary and thus the following sequence of hexadecimal numbers would work:

11
04
EB

The 11 identifies this message as a low-level query. The number following (4) is the address that contains the current setting of the sweep magnifier switch. Finally, the $E B$ is the $2^{\prime}$ 's complement of the modulo- 256 sum of the preceding two bytes (hex 15). When the plug-in is made a talker, the setting of the sweep magnifier switch will be returned.

The response to a low-level Query message looks identical to a corresponding low-level Set message. For instance, if the setting of the sweep magnifier is queried as in the preceding example, a response like the following hexadecimal numbers could be obtained:

## 15

04
$\varnothing 8$
DF

The 15 designates a low-level query response. The 64 is the <address> byte corresponding to the sweep magnifier, and the 08 is the <data> byte corresponding to the OFF position. DF is the checksum.

Consider another example. This time let's query the state of the trigger source, trigger slope, and trigger mode switches. Since these switches correspond to consecutive addresses in Table 3-2, we can query all three settings in a single low-level Query message consisting of the following hexadecimal numbers:

As in the previous example, the 11 identifies this message as a low-level query. The $\emptyset 9$ which follows is the starting address in 7B90P memory that will be queried; it corresponds to the source of the trigger signal (i.e. internal, line, external, or external $\div 10$ ). The $\varnothing 3$ indicates that the address counter is to be incremented twice so that two more functions can be queried. These two successive functions are the trigger slope and trigger mode switches, corresponding to addresses $\emptyset_{A}$ and $\emptyset_{B}$ respectively. The message ends with the checksum E3.

The response to the above multiple-query message might be the following hexadecimal numbers:

## 15

09
10
$\emptyset 8$
40
8A

Again, the 15 indicates a response to a low-level query, and the $\varnothing 9$ indicates the starting <address> byte -- which corresponds to the trigger source. The next three bytes ( $10, \emptyset 8,4 \varnothing$ ) indicate that the trigger source is internal, the trigger slope is positive, and the trigger mode is Peak-to-Peak Automatic. The 8 A is the checksum.

As a final example, consider the following low-level query:

11
EF

This query contains only the query designator (11) and the checksum (EF). Since the <address> and <count> bytes were both omitted, the state of all functions will be queried (addresses $\emptyset \emptyset$ through $\varnothing C$ inclusive). This is analagous to high level SET? query which indicates all states of the plug-in when the plug-in is made a talker.

## Transmitting Messages

The discussion so far has centered around the syntax of the high-level and low-level messages, rather than the process by which these messages are transferred from talker to listener and vice versa. When special driver software is being used, it is sufficient to know only the message syntax since most of the details of message transfer are handled automatically by the software. However, when using controllers for which driver software is not available, it is necessary to know something about the message protocol discussed in the IEEE 488 standard. The following information will provide these details by showing the states of the interface signal lines for typical message transfers.

## Transwitting Set Messages

First, consider the case where a talker is transmitting a message to effect a certain front-panel setting on the 7B90P. Here is a summary of the steps involved:

1) Clear the bus. Before initiating any message transfer, it is a good idea to clear the bus of any previously assigned talker and/or listener (s). (In some cases, such as immediately following power-up, this step won't be necessary since no talker or listener will have been assigned.) To remove any talker or listener, the controller asserts ATN and sends the UNT and UNL interface messages on the Data lines. (According to Appendix E of the IEEE 488 standard, UNT corresponds to hexadecimal 5 F and UNL corresponds to hex 3 F. )
2) Assign the talker and listener(s). Before communication of devicedependent messages can occur, the controller must designate which device is to talk and which device(s) are to listen. This is done with ATN still asserted. To designate the talker, the controller sends the primary talk address (MTA) of that device. This is followed by the talker's secondary address (MSA) if secondary addressing is employed. Similarly, the controller designates the listener (the 7B90P in this case) by placing its primary listen address (MLA) on the Data lines. This must also be followed by the secondary address (MSA) of the listener since secondary addressing is employed in the 7B90P. The process is repeated for each listener on the bus, although some listeners may not have secondary addresses.

If the ATN signal line is continuously asserted during the period when the talker and listener(s) are assigned, the order in which the talker and listener(s) are assigned is unimportant.

In the case where the controller is to be the talker, it must have addressed itself as such and any other talkers must be idle. (Its talk address does not need to be sent over the bus.)
3) Send the device-dependent message. With the talker and listener(s) assigned, the ATN line is now unasserted to initiate transfer of the device-dependent message.

In the case of a high-level message, each ASCII character of the message is transferred in sequence according to the previously explained syntax. Only lines DI01 through DI07 of the Data Bus are used. DI01 corresponds to the least-significant bit of each 7-bit ASCII character and DI07 corresponds to the most-significant bit. Line DI08 of the Data Bus is a "don't care" bit. As each character is transferred over the Data Bus, the entire handshake cycle is executed. This guarantees proper reception of each character before permitting a new character to be transmitted.

In the case of a low-level message, the process is similar except that all eight Data lines (DI01 through DI08) are used for transferring the message. Lines DI01-4 transfer the low-order hexadecimal digit and lines DI05-8 transfer the high-order hexadecimal digit. The handshake sequence is essentially the same as for the high-level message.
4) Terminate the message. As the last character or digit of the message is sent, the talker asserts the EOI signal line along with the data byte. This is to inform the controller that the talker is finished so that other operations can begin. (All Tektronix instruments functioning as talkers assert EOI automatically when outputting the last data byte of the message, but this is not necessarily true of instruments from other manufacturers -- particularly those designed prior to the formal release of IEEE 488-1975.)
5) Clear the bus. Though step 4 actually terminates the message transfer, it is often desirable to clear the bus of talker and listener(s) so that other operations can proceed immediately. The procedure for doing this is identical to step 1.

As an example of the above steps, assume that the controller is to effect two front-panel settings on a 7B90P: the trigger slope will be set
to positive, and the trigger mode will be set to Peak-to-Peak Automatic. (The plug-in is installed in a 7912AD mainframe set for a primary address of $\emptyset \varnothing \emptyset \emptyset \emptyset$ and a secondary address of $\varnothing \emptyset \emptyset 1 \varnothing$.) The following steps would accomplish this:

1) The controller asserts ATN, then transmits hexadecimal 5F (UNT) and hex 3F (UNL) over Data lines DIO1 through DIO7.
2) With ATN still asserted, the controller sends hex 20 (the MLA of the mainframe) followed by hex 64 (the MSA of the 7B90P) on the Data lines. Recall from the subsection on Addressing the 7B90P that the MSA of the 7B90P is that of the mainframe plus 2. Since the controller is to be the talker, no talk address is sent.
3) a) In the case of a high-level message, the talker (controller) transmits each of the following ASCII characters over the Data lines:

SLO POS ; MOD PPA
b) In the case of a low-level message, the following hexadecimal bytes are transferred over the Data lines:

15
øA
$\emptyset 8$
40
99
4) As the last character or hex byte is transmitted, the talker asserts EOI indicating the end of the message.
5) The controller again clears the bus by transmitting 5 F (UNI) and $3 F$ (UNL) over the Data lines.

Figure 3-4 illustrates the above sequence more graphically.

Figure 3-4 does not indicate the status of the IFC, REN, and SRQ signal lines because they are relatively unimportant to the point being made. However, it should be understood that the REN (Remote Enable) line is continuously asserted whenever the plug-in is being operated under remote (program) control, as opposed to local (front-panel) control. The


In the above diagram, ${ }_{F}^{5}$ and ${ }_{F}^{3}$ are the UNTalk and UNListen commands; ${ }_{\emptyset}^{2}$ and ${ }_{4}^{6}$ are the Listen Address and Secondary Address, respectively. All of the message is coded in hexadecimal.

Fig. 3-4. Status of the ATN, EOI, and DIO signal lines during a high-level (a) and low-level (b) Set message.

IFC line will normally be unasserted during the entire seçuence. The SRQ line can be asserted whenever a device requests service. However, it is not relevant to the process of data transmission diagrammed in Fig. 3-4 and is not shown.

While the handshake lines (DAV, NRFD, and NDAC) are important in the transmission of messages and data, their status changes several times during the transmission of each character in the message. Appendix $B$ of the IEEE 488 standard gives a detailed description of this handshake cycle.

## Transmitting Query Messages

The previous sequence describes the process for transmitting a Set message - - - in which case the plugnin is always a listener. A similar sequence occurs when a Query message is sent to the plug-in. However, in this case the plug-in is first made a listener so that it can receive the Query message; then the plug-in is made a talker so that it can transmit its response to the query. Here is a summary of the steps involved:

1) Clear the bus. As previously noted, this step is not always necessary but is good practice. The controller asserts ATN and sends UNT (hex 5F) and UNL (hex 3F) over the Data lines.
2) Assign the talker and listener(s). The controller designates which device is to send the query and which device or devices are to receive the Query message. This is done with ATN still. asserted. To designate the talker, the controller sends the primary talk address (MTA), followed by the talker's secondary address (MSA) if secondary addressing is used for that device. Similarly, the controller designates the listener (the 7B90P in this case) by sending its primary listen address (MLA) and secondary address (MSA). This process is repeated for each listener on the bus.

When the controller is to be the talker, it must have addressed itself as such and any other talkers must be idle. (Its talk address does not need to be sent over the bus.)
3) Send the device-dependent query. With the talker and listener(s) assigned, the ATN line is now unasserted to initiate transfer of the device-dependent query. Each ASCII character or hex byte of the message is transferred in sequence over the Data lines according to the previously explained syntax.
4) Terminate the query. As the last character or digjt of the high-level query or series of queries is sent, the talker asserts EOI along with the message byte. This informs the controller that the talker is finished so that the query response can proceed.
5) Clear the bus. At this point i.t is necessary to clear the bus so that the plug-in can be addressed as a talker. Again, this is done by the controller asserting ATN and by sending the UNT and UNL interface commands over the Data lines.
6) Assign the talker and listener(s). With ATN still asserted, the controller now assigns the plugmin as a talker and assigns any additional listeners. If the controller is to be the only listener, it does not need to address itself as a listener.
7) Send the device-dependent query response. The ATN line is now unasserted by the controller to initiate transfer of the query response from the talker (plug-in) to any designated listener (s). Each ASCII character or hex byte of the query response is transmitted over the Data lines according to the previously described syntax for a query response.
8) Terminate the query response. As the last character or digit of the query response is sent, the talker (plug-in) asserts EOI along with the data byte, thereby indicating the end of the response.
9) Clear the bus. Though not required for this operation, the controller can now clear the bus of talker and listener ( $s$ ) so that further operations can proceed.

As an example of the above steps, assume that the controller is to query and receive the status of two front-panel functions on the 7B90P: the sweep magnifier and the time/division settings. (The plug-in is installed in a 7912AD mainframe set for a primary and secondary address of $\emptyset \emptyset \emptyset 11$. ) The following steps describe the operations involved:

1) The controller asserts ATN, then transmits hexadecimal 5F (UNT) and $3 F$ (UNL) over the Data lines, thereby clearing the bus.
2) With ATN still asserted, the controller transmits hexadecimal 23 (the MLA of the mainframe) followed by hex 65 (the MSA of the 7B90P) over the Data lines. This assigns the 7B90P as a listener.
3) a) In the case of a high-level message, the talker (controller) transmits each of the following ASCII characters over the Data lines:

MAG?; T/D?
b) In the case of a low-level message, the following hex bytes are transferred over the Data lines:

11
04
03
E8
4) As the last character of the query is transmitted, the talker (controller) asserts EOI.
5) The controller clears the bus of talkers and listeners by asserting A'IN and sending hex 5 F and $3 F$ on the Data lines.
6) With ATN still asserted, the controller transmits hexadecimal 43 (the MTA of the mainframe) followed by hex 65 (the MSA of the 7B90P) over the Data lines. This assigns the 7B90P as a talker.
7) The talker (7B90P) now sends its query response to the listener (controller).
8) As the last character of the query response is transmitted, the talker (7B90P) asserts EOI.
9) The controller again clears the bus by asserting ATN and sending hex 5 F and 3 F .

Figure 3-5 illustrates the state of the ATN EOI, and Data lines for the above sequence and a typical query response.


Fig. 3-5. Status of the ATK, EOI, and DIO signal lines during a highlevel (a) and low-level (b) Query message and response.

## Serial-Poll Responses

The 7B90P reports any of the following errors or special conditions when polled by the controller:

1) Power on
2) Command error
3) Execution error
4) End-of-sweep
5) Busy

If any of the above conditions (except Busy) occurs, the 7B90P asserts the SRQ line to initiate a serial poll. The controller, if programmed to do so, can then process the interrupt and provide an appropriate response. The following is a detailed description of each of the five conditions:

1) Power on: Occurs whenever the plug-in goes through its initialize sequence, either at the time that power is first applied, or when it is reapplied following momentarily interruption.

The serial-poll response byte is $\varnothing 1 \varnothing x \emptyset \varnothing \varnothing 1$, where x is a 1 if the $7 \mathrm{B9OP}$ is busy, and a $\varnothing$ otherwise.
2) Conand error: Occurs whenever the plug-in cannot understand or implement the data it receives.

Command errors detected for low-level messages are: (a) the message is more than 16 bytes long, (b) the modulo- 256 sum of all bytes in the message is not zero, (c) the starting address is too large, (d) no address is supplied with a Set message, or (e) an attempt was made to set address $\emptyset$ (plug in type) to the wrong value.

Command errors detected for high-level messages are: (a) syntax is incorrect, (b) a Set command refers to a readmonly function, or (c) message units are not recognized by the 7B90P.

If an error is detected in a low-level message, the 7B90P ignores the entire command line. If the error is a low-level query address out of range, the output queue is cleared (i.e. if there are any query-responses pending, they are cleared).

If an error is detected in a high-level message, any message unit preceding the one with the error is processed normally. The one with the error and any following are ignored.

The serial-poll response byte for a command error is $\$ 11 \times \varnothing \varnothing \emptyset 1$, where x signifies the busy status.

Occasionally, the 7B90P may be addressed as a talker, but have nothing to say. When this occurs, the plug-in sends a data byte with all Data lines set to a logical 1 (hex FF ) and EOI asserted. Also, the plug-in reports a command error.
3) Execution error: Occurs when the numerical argument for a high-level Set command is invalid or out of the allowable range. The command for that function is ignored, as are any subsequent commands in the same message.

The serial-poll response byte for this type of programming error is $\emptyset 11 \mathrm{x} \varnothing \varnothing 1 \varnothing$, where x signifies the busy status.
4) End of sweep: Occurs at the end of sweep in the single-sweep mode if the End-of-Sweep $S R Q$ has been enabled.

The serial-poll response byte is $11 \varnothing \mathrm{x} \varnothing \varnothing \varnothing 1$, where x signifies the busy status.
5) Busy: Occurs when a serial poll is performed while the plug-in is decoding a command; the response is xxxixxxx. When not busy, the response is $\operatorname{xxx} \varnothing$ xxxx ( $x$ is a "don't care" bit). The plug-in will not accept data on the bus while reporting busy.

The 7B90P queues power-up, command error, and execution error reports. If two or three conditions are waiting to be reported, $S R Q$ is reasserted after reporting one error to indicate that there is more to report.

If an error is found in a high level command string, $S R Q$ is not asserted and the error is not reported until the message has been completely received (i.e. a byte with EOI asserted is received).

## WARNING

THE FOLLOWING SERVICNG WSTRUCTIONS ARE FOR USE BY QUALIFIED PERSONNEL ONLY TO AVOID PERSONAL INURIY, DO NOT PERFORM ANY SERVICNG OTHER THAN THAT CONTANED IN OPERATNG ANSTRUCTIONS UNLESS YOU ARE OUALFIED TO DO SO REFER TO OPERATORS SAFETY SUMMARY ANO SERVICE SARETY SUMMARY PFIOR TO PERFORNING ANY SERVICE:

## SECTION 4

## MAINTENANCE

This section contains maintenance information for use in preventive maintenance, corrective maintenance, and troubleshooting of the 7B90P.

Further maintenance information relating to general maintenance can be found in the instruction manuals for the 7000 -series mainframes.

## Preventive Maintenance

Preventive maintenance, consisting of cleaning, visual inspection, etc., performed on a regular basis, will improve the reliability of this plug in unit. Periodic checks of the semiconductor devices used in the unit are not recommended as a preventive maintenance measure. See semiconductor-checking information given under Troubleshooting.

The severity of the environment to which this instrument is subjected determines the frequency of maintenance. A convenient time to perform preventive maintenance is preceding adjustment of the instrument.

## Cleaning

$$
\begin{aligned}
& \text { Avoid the use of chemical cleaning agents } \\
& \text { which might damage the plastics used in } \\
& \text { this unit. In particular, avoid chemicals } \\
& \text { that contain benzene, toluene, xylene, } \\
& \text { acetone, chlorothane, or similar chemicals. }
\end{aligned}
$$

Exterior. Loose dust may be removed with a soft cloth or a dry brush. Dirt that remains can be removed with a soft cloth dampened with a mild detergent and water solution. Abrasive cleaners should not be used.

Interior. Dust on the interior of the unit should be removed occasionally due to its electrical conductivity under high humidity conditions. Cleaning the interior of the unit should precede calibration, since the cleaning
process could alter the settings of the calibration adjustments. Use lowvelocity compressed air to blow off the accumulated dust. Hardened dirt can be removed with a soft, dry brush, cotton-tipped swab, or cloth dampened with a mild detergent and water solution.

## Visual Inspection

This instrument should be inspected occasionally for such defects as broken connections, improperly seated semiconductors, damaged circuit boards, and heat-damaged parts.

The corrective procedure for most visible defects is obvious. However, particular care must be taken if heat-damaged components are found. Overheating usually indicates other trouble in the instrument; therefore, it is important that the cause of overheating be corrected to prevent recurrence of the damage.

## Lubrication

There are no components in this instrument that require a regular lubrication program during the life of the instrument.

## Troubleshooting

The following is provided to augment information contained in other sections of this manual when troubleshooting the 7B90F. The schematic diagrams, circuit description, and calibration sections should be used to full advantage. Section 5 gives detailed information on circuit behavior.

## Troubleshooting Aids

Schematic Diagrams. Schematic diagrams are provided on foldout pages in Section 8. The circuit number and electrical value of each component are shown on the diagrams. Power supply voltages are also shown. Components that are mounted on circuit boards are outlined on the diagrams with a heavy black line.

Circuit-Board Illustrations. Illustrations of circuit boards are shown opposite the schematic diagrams. Each board-mounted electrical component is identified by its circuit number.

Component-Locator Grids. Schematic diagrams and circuit-board ill ustrations are bounded by component-locator grids. When used with the associated lookup tables, these grids allow you to quickly locate a component on either the schematic or the circuit board.

Component and Wiring Color Code. Colored stripes or dots on resistors and capacitors signify electrical values, tolerances, etc., according to the EIA standard color code. Components not color coded usually have the value printed on the body.

The insulated wires used for interconnection in the 7B90P are color coded to facilitate tracing wires from one point to another in the unit.

Semiconductor Lead Configuration. The lead configurations of the semiconductor devices used in this instrument are shown in Fig. 4-1.

Rear Interface Connector Pin Locations. The 7B90P SWEEP board couples the plug-in to the associated mainframe. Figure 4-2 identifies the pins on the interface connector as shown on the schematic diagrams.


Fig. 4-15. Seaiconductor lead configurations.


Fig. 4-2. Location of pin numbers on interface connector.

## Troubleshooting Equipent

The following equipment is useful for troubleshooting the 7 B 90 P .

1) Semiconductor Tester -- Some means of testing the transistors, operational amplifiers, comparitors, diodes, and FET's used in this instrument is helpful. A transistor-curve tracer such as the TEKTRONIX Type 577-177 or 577-178 will give the most complete information.
2) Multimeter -- A voltmeter is required for checking voltages within the circuits, and an ohmmeter for checking resistors and diodes. The voltmeter should have an input impedance of at least 10 megohm, a range of at least 0 to 50 volts $D C$, and an accuracy of $0.01 \%$. The ohmeter should have a range of 0 to 20 megohms.
3) Test Oscilloscope - - A test oscilloscope (with DC-100 MHz bandwidth) is required to view waveforms at different points in the circuits. A

TEKTRONIX 7000-series Oscilloscope equipped with a readout system, 7D13 Digital Multimeter unit, 7B-series Time-Base unit, and a 7A-series Amplifier unit with a 10 X probe will meet the needs of both items 2 and 3.
4) Plug-in Extender -- This fixture permits operation of the 7B90P outside of the plug-in comparment for better accessibility during troubleshooting. Order TEKTRONIX Part Number 067-0589-00.

## Troubleshooting Procedure

This troubleshooting procedure is arranged in an order which checks the simple trouble sources before proceeding with extensive troubleshooting. The first few checks ensure proper connection, operation, and adjustinent. If the trouble is not located by these checks, the remaining steps aid in locating the component. When the defective component is located, it should be replaced using the replacement procedure given under Corrective Maintenance.

1) Check Control Settings. An incorrect setting of the 7B90P controls can indicate a problem that does not exist. If there is any question about the correct function or operation of a control or front-panel connector, refer to the operating instructions provided in Section 2.
2) Check Associated Equipwent. Before proceeding with troubleshooting of the 7B90P, check that the equipment used with this instrument is operating correctly. If possible, substitute a time-base unit known to be operating correctly into the mainframe and see if the problem persists. Check that any interconnecting cables are not defective.
3) Visual Check. Visually check the portion of the instrument in which the trouble is suspected. Many problems can be located by visual indications such as unsoldered connections, broken wires, damaged circuit boards, damaged components, or components bent over and touching other parts or circuit board runs.
4) Check Instrument Adjustment. Check the adjustment of the unit or the affected circuit by performing the Performance Check in Section 6. The apparent trouble may only be a result of mis-adjustment and may be corrected by calibration. Complete calibration instructions are given in Section 6 .
5) Isolate Trouble to a Circuit. To isolate trouble to a circuit, note the trouble symptom. The symptom often identifies the circuit in which the trouble is located. When trouble symptoms appear in more than one circuit, check the affected circuits by taking voltage readings. Incorrect operation of all circuits sometimes indicates trouble in the power supply.

After the defective circuit has been located, proceed with step 6 to locate the defective components.
6) Check Individual Components. The following methods are provided for checking the individual components. Components which are soldered in place are best checked by disconnecting one end to isolate the measurement from the effects of surrounding circuitry.


To avoid component damage, disconnect the power source before removing or replacing semiconductors.

## NOTE

To locate intermittent or temperature sensitive components mounted on the circuit boards, Quik Freeze (Miller Stephenson, MS-240, TEKTRONIX Part Number 006-0173-01) is recommended. Dry ice or dichloro-difluoromethane (Freon 12, Dupont or Can-0-Gas) may also be used.
A) TRANSISTORS. The best check of transistor operation is actual performance under operating conditions. Transistors that are soldered to the circuit board should first be checked in-circuit using a dynamic transistor tester; then a replacement can be substituted to further verify that the old transistor is bad. Socketed transistors can be checked immediately by substituting a known good component; however, be sure that circuit conditions are not such that a replacement might also be damaged. If substitute transistors are not available, check the old transistor out-of-circut using a dynamic tester (such as the TEKTRONIX Type 577-177). Static--type testers may be used, but since they do not check operation under simulated operating conditions, some defects may go unnoticed. Be sure the power is off before attempting to remove or replace any transistor.
B) INTEGRATED CIRCUITS. Analog IC's such as comparitors and operational amplifiers can usually be checked in-circuit with a voltmeter or test oscilloscope. An understanding of the device and circuit operation is essential for this type of troubleshooting. (For example, an op amp can be tested by measuring the input and output circuit voltages and comparing this ratio to the ratio of input and feedback resistors.) Analog IC's that are socketed can also be checked out-of-circuit using a dymamic tester such as the TEKTRONIX 577-178.

Digital. IC's are best checked in-circuit using a logic probe or voltmeter. Use care when checking voltages and waveforms around DIP (Dual-In-line-Package) IC's so that adjacent leads are not shorted together. A convenient means of connecting a test probe to 14 - and 16 - pins IC's is with an IC test clip. This device also doubles as an extraction tool.
C) DIODES. A diode can be checked for an open or shorted condition by measuring the resistance between terminals with an ohmeter set to the $\mathrm{R} X 1 \mathrm{k}$ scale. The diode resistance should be very high in one direction and very low when the meter leads are reversed. A diode can also be tested with a dynamic tester (transistor curver tracer).


Do not use an ohmmeter scale that has a high internal current. High currents may damage the diodes under test.
D) RESISTORS. Check resistors with an ohmmeter. Resistor tolerance is given in the Electrical Parts List. Resistors normally do not need to be replaced unless the measured value varies widely from the specified value.
E) CAPACITORS. A leaky or shorted capacitor can be detected by checking resistance with an ohmmeter on the highest scale. Use an ohmmeter that will not exceed the voltage rating of the capacitor. (Be careful to observe correct polarity when checking electrolytic capacitors.) The resistance reading should be high after initial charge of the capacitor. An open capacitor can best be detected with a capacitance meter, or by checking whether the capacitor passes AC signals.
7) Repair and Readjust the Circuit. Special techniques required to replace components in this unit are given under Cowinonent Replacement. Be
sure to check the performance of any circuit that has been repaired or that has had any electrical components replaced. Recalibration of the affected circuit may be necessary.

## Specific Troubleshooting Information

The following information provides provides an aid to troubleshooting parts of the analog circuitry and the microprocessor system. This information is provided as a guide to troubleshooting the instrument; it is not intended to be a complete troubleshooting procedure. Before troubleshooting the instrument, an understanding of the circuit operation is required. Refer to the Circuit Description for a discussion of the circuit operation.

The following information is provided as an aid to troubleshooting only; it is not intended as a complete troubleshooting procedure.

Analog Circuits. Some of the transistors in the 7 B 90 P are mounted in sockets to allow feedback loops to be opened as an aid to trouble-shooting. For example, when Q406 (part of the sweep-start current switch -- see schematic 2) is removed, the sweep ramp goes high (11-14 volts, measured at the collector of Q406) and remains there. If Q406 is replaced and Q400 removed, the sweep ramp stays low. The hold-off start logic, gate generator and sweep-start circuits can also be checked using this technique.

Trigger Circuits. If the trigger circuits are malfunctioning, the following steps may help to isolate the problem:

1) Measure the voltages around the trigger I.C. (U220). Pin 10 should be near D.C. ground, with about 20 millivolts of A.C. signal.
2) One of pins 2, 4, 6, or 8 should be near ground potential. All other pins should be held at about -2 volts. The SOURCE selected determines which pin is grounded as follows:

## SOURCE GROUNDED PIN

| LINE | 2 |
| :--- | :--- |
| EXT | 4 |
| EXT +10 | 6 |
| INT | 8 |

DS 348 and DS 440 provide a visual indication of the selected source and check the operation of the associated biasing circuit. The LED's should be lit as follows:

## SOURCE LEDS ON

| LINE | DS 348, DS 440 |
| :--- | :--- |
| EXT | DS 440 |
| EXT +10 | DS 348 |
| INT | DS 348, DS 440 |

3) Check the levels on pin 1 of U240. When + SLOPE is selected, pin 1 should be at about 0.5 volts. When - SLOPE is selected, the pin should be at about 4 volts.

Sweep Generator. Trouble in the sweep generator may be isolated using the following steps:

1) For the sweep to run, a GATE input to Q262 or AUTO SENSE to Q264 must be present. If either of these signals are present and there is no sweep, pull Q406 from its socket and check the voltage at TP422. The voltage should be about +5 volts. If not, check Q512, U110, and the associated components.
2) Remove Q400 and measure the voltage at the junction of R512 and R514. The voltage should be near ground.
3) DS001 and DSO08 provide a visual indication of the state of the $1,2,5$ voltage reference (U110). For example, when a multiple of 2 sweep speed (e.g. 2 milliseconds/division, 20 microseconds/division, etc.) is selected, DSOO1 and DSOO8 are both on. The LED states for 1, 2, and 5 sweep speed steps are shown below.

## SWEEP SPEED LEDs LIT STEP

| 1 | DS 008 |
| :--- | :--- |
| 2 | DSOO1, DS OO8 |
| 5 | NONE |

Readout Circuits. 41000 accepts digital signals from the microprocessor and the timewslot counter and generates the analog row and column current that is sent to the mainframe. The levels on pins 4 and 5 of U1000 select the sweep speed symbol (milliseconds, microseconds, nanoseconds, or picoseconds) as shown below:


Pins 8 and 9 select the number of zeros in the readout. For example, a readout with two zeros $(100,200,500)$ pin 8 is low and pin 9 is high. The levels on pins 8 and 9 are shown in the following table:

## U1000 PIN NO.

## 89

| $1,2,5$ | 1 | 1 |
| :--- | :--- | :--- |
| $10,20,50$ | 1 | 0 |
| $100,200,500$ | 0 | 1 |

Pins 11 and 13 select the 1,2 , or 5 multiplier for the readout. The levels are summarized in the table below:

## U1000 PIN NO.

## 1113

| 1 | 1 | 1 |
| :--- | :--- | :--- |
| 2 | 1 | 0 |
| 5 | 0 | 1 |

Pins 14, 15, 16, and 17 are the time-slot count inputs from U1005. U1005 counts the time-slots in a straight binary sequence and provides time-slot input to U1000.

Microprocessor System. If the 7B90P fails to power-up in the correct state, the following procedure provides an aid to locating the fault without special microprocessor service equipment.

1) Measure all the supply voltages. If they are out of tolerance, the power-up circuit in the plug-in will not initialize the micro-processor system correctly. Check for faults in the plug-in. If none are found, check the mainframe power supplies.
2) Temporarily remove the $\overline{\mathrm{STOP}}$ strap, P1526, on the HORIZONTAL PROGRAMMING board. Press any front panel button (except TERM) to cause an interrupt and start the MPU clock. Be sure to replace the STOP strap when you are through troubleshooting the clock circuit.
3) Check for valid two-phase clock signals at TP132 (Phase 1) and TP130 (Phase 2). Refer to the Calibration Procedure for a discussion of the MPU clock adjustment. If the clock signals are present, go to step eight.
4) Check that the clock jumper strap, P130 is correctly placed on the corresponding square pins.
5) Check the output (pin 3) of U010A. This pin should be high $(+2.4$ to 5.0 volts). If it is not, check the flip-flop formed by U310B and U1630B. The STOP line (pin 5 of U 1630 B ) should be high.
6) Check pin 12 of U310B. If this pin is low, there is a fault in the power-up circuit (assuming the supply voltages are correct).
7) If no faults are found in steps four through six, there is a fault in the clock circuit.
8) Check pin 40 of U 1220 (MPURES) and pin 34 of U510 (RESET). If either of these lines is low, there is a fault in the power-up circuit.
9) Momentarily short pin 34 of $U 510$ to ground. If the plug-in powers-up correctly, there is a fault in the power-up circuit.
10) Check that pin 2 of $U 1220$ is high and pin 39 is Jow.
11) Check that pins 36 and 37 of $U 1220$ are tied together.
12) If these tests fail to identify the fault, troubleshooting with a microprocessor debugging aid (such as a Tektronix 8000 meries microprocessor lab) may be necessary.

Self Test Addresses. If the 7B90P fails to power-up and the above steps fail to identify the problem, the MPU may have detected a fault while performing the self-test routines. If a fault is detected the MPU "hangs" at a specified address. The fault can be determined by measuring the state of address bus lines ( $A \emptyset-A 11$ ) with a DVM or logic analyzer (such as the Tektronix 7Dø1). Table $4-1$ shows the addresses and the fault(s) detected, by the MPU for each address. If a fault is indicated in the ROM, RAM or PIA's, replace the involved components and see if the problem is cured.

## TABLE 4-1

## 7B90P Self Test Hang Addresses

790A16 - The RAM was filled with ones. Data was read back and complemented, and a non-zero byte was found. The RAM is left filled with zeros.

791516 - Any ones left from the previous test hangs the 6800 here.
$793 \varnothing_{16}$ - The RAM failed a pattern-sensitivity test.

795816 - The PIA control registers were filled with zeroes. The registers were read back and a one was found.

797816 - The PIA registers failed a read/write test.

799216 - The high-level ROM (U830) failed a checksum test.

799E16 - The low-level ROM (U630) failed a checksum test.

Control Line State Tables. The microprocessor controls the trigger and sweep generator circuits through several control lines. The following tables show the states of these lines for each sweep speed and triggering mode.

Table 4-2 shows the state of the sweep timing control lines (ST0-ST7) and the Intensity Limit (IL) control line for each sweep speed. The selected
timing capacitor, reference voltage, decade resistor, and hold-off timing components are also shown for each sweep speed. (Refer to the Circuit Description for a discussion of these parameters.) For example, when the 7B90P TIME/DIV setting is one microsecond per division, sweep timing lines ST2, ST3 and ST5 are high. The remaining lines are low. At this sweep speed, C412 is selected as the timing capacitor (in parallel with C510 refer to schematic 2) and R203 is selected as the decade resistor. The timing reference voltage is 2 volts (measured at the collectors of U110) and the timing current through Q512 is 500 microamps. The holdoff time is 3.5 microseconds, set by C644. The information in this table provides an aid to troubleshooting the ramp generator and its control circuitry.

Table 4-3 shows the state of the triggering control lines when each combination of TRIGGERING MODE, SOURCE and COUPLING is selected. For example, in PP AUTO MODE with INTernal SOURCE and AC COUPLING, the top row in the table indicates the state of each control line. A complete discussion of each of the control lines is given in the circuit description. The information in this table provides an aid to trouble-shooting the trigger generator and its control circuitry.

TABLE 4-2

Sweep Timing Information

*Times shown in this column are minimum values. (HOLDOFF control fully counterclockwise).

## - <br> Digitally signed by http://www.aa4df.com

Trigger Control Line States


## Corrective Maintenance

Corrective maintenance consists of component replacement and instrument repair. Special techniques required to replace components in this instrument are given here.

## Obtaining Replacement Parts

Standard Parts. All electrical and mechanical part replacements for the 7B90P can be obtained through your local Tektronix Field Office or representative. However, many of the electronic components can be obtained locally in less time than is required to order them from Tektronix, Inc. Before purchasing or ordering replacement parts, check the parts list for value, tolerance, rating and description. The vendor's or manufacturer's part number and address are also provided in the parts list.

## NOTE

When selecting replacement parts, it is important to remember that the physical size and shape of a component may affect the performance of the instrument, particularly at high frequencies. All replacement parts should be direct replacements unless it is known that a different component will not adversely affect instrument performance.

Order all special parts directly from your local Tektronix Field Office or representative.

Ordering Parts. When ordering replacement parts from Tektronix, Inc., include the following information:

1) Instrument Type.
2) Instrument Serial Number.
3) A description of the part (if electrical, include circuit number found in the Replaceable Electrical Parts list - Section 7).

CKI NUMBERING EXAMPLE:

4) TEKTRONIX Part Number.

## Soldering Techniques


#### Abstract

WARNING

Disconnect the instrument from the power source before soldering.


Circuit Boards. The components mounted on the circuit boards in the 7B90P can be replaced using normal circuit-board soldering techniques. Keep the following points in mind when soldering on the circuit boards:

1) Use a pencil-type soldering iron with a wattage rating from 15 to 50 watts.
2) Apply heat from the soldering iron to the junction between the component and the circuit board.
3) Heat-shunt the lead to the component by using a pair of long-nose pliers.
4) Avoid excessive heating of the junction with the circuit board, as this could separate the circuit board wiring from the base material.
5) Use only 60-40 rosin core, electronic grade solder.
6) Clip off any excess lead length extending beyond the circuit board. Clean off any residual flux with a flux-removing solvent.

Metal Terminals. When soldering metal terminals (TEFM switch and EXT TRIG IN connector) use $60-40$ tin-lead solder and a 15 to 50 watt soldering iron. Observe the following precautions when soldering metal terminals:

1) Apply only enough heat to make the solder flow freely.
2) Apply only enough solder to form a solid connection. Excess solder may impair the function of the part.
3) If a wire extends beyond the solder joint, clip off the excess.
4) Clean the flux from the solder joint with a flux-removing solvent.

## Component Replacement

## WARNING

Disconnect the equiprent from the power source before replacing components.

Semiconductor Replacement. Some transistors in the 7B90P are socketed. These should not be replaced unless actually defective. If removed from their sockets during routine maintenance, return them to their original sockets. Unnecessary replacement of transistors may effect the calibration of this instrument. When transistors are replaced, check the performance of any part of the instrument that may be affected.


#### Abstract

WARNING

CMOS parts are very susceptable to damage due to static discharge. The operator and instrument should be grounded when parts are being replaced. Place CMOS parts on static-free foam during storage.


Replacement semiconductors should be of the original type or a direct replacement. The schematic diagram shows the lead configurations of the semiconductors used in this instrument. If the replacement semiconductor is not of the original type, check the manufacturer's basing diagram for proper basing.

Front-Panel LED's. When replacing any front-panel LED's (Light Emitting Diodes), be sure to place the new one(s) in the same exact alignment with the other LED's to facillitate reassembly of the front panel. When soldering LED's, use the minimum heat required to do the job. (For board removal, see A) Removal of LED and SWITCH Boards.)

Free-Standing Components. When replacing components that are freestanding (not directly mounted to circuit boards), be sure to place the new components in the same physical location and position as the old components. If this is not done, the high frequency characteristics of the plug-in may be altered; also, there may be a possibility of components touching and causing a short circuit.

## Circuit Board Removal

The 7B90P contains the following circuit boards:

1) LED board
2) SWITCH board
3) TRIGGER board
4) SWEEP board
5) SHIELD board
6) HORIZONTAL PROGRAMMING board

The LED and SWITCH boards are attached to the front-panel casting. The TRIGGER, SWEEP, SHIELD, and HORIZONTAL PROGRAMMING boards are all connected together via mounting studs and multi-pin (interboard) connectors. These four boards are held to the plug-in frame via special fastening nuts that hold the SWEEP board to the top and bottom plug-in rails. (The SHIELD board simply insulates and shields the SWEEP board from the HORTZONTAL PROGRAMMING board; there are no components mounted to it, and it is electrically grounded to the SWEEP board.)

In general, the HORIZONTAL PROGRAMMING board will rarely need to be removed during troubleshooting since most of its components are readily accessable. The components on the TRIGGER board are also readily accessable, but it may be necessary to remove the TRIGGER board to expose certain components on the SWEEP board. The LED and SWITCH boards must be removed when replacing front-panel components such as the Light Emitting Diodes (LEDs).

The following procedure explains how to remove each of these boards from the rest of the plug-in. Peforming all of these steps in the order listed will result in a complete disassembly of the 7B90P. To reassemble the plug in, reverse the process.

## A) Rewoval of LED and SWITCH Boards

1) Remove the side panels from the plug-in.
2) Remove the front-panel LEVEL, POSITION, and HOLD OFF knobs by loosening their Allen set-screws.
3) Remove the tension spring from the end of the plug-in pull latch (inscribed with the word "7B90P").
4) Dismount the front-panel cover plate by gently prying on the bottom with a regular screwdriver.
5) Remove the jumper wires connecting the LED board to connectors P550 and J436 on the TRIGGER board.
6) Remove the four Phillips screws (one from each corner) of the front-panel casting.
7) Gently rock the front-mpanel casting to remove the front-panel assembly from the edge connector on the SWEEP board.
8) With the front-panel assembly removed, unsolder capacitor C402 and resistor R400 at the points where they connect to the LED board; also unsolder the ground lead (from the LED board) at the point where it connects to the TERM switch.

Note: The position and length of these components should be noted so that they can be replaced as originally installed. Otherwise, the VSWR may be adversely affected.
9) To dismount the LED board, remove all five Phillips screws; then carefully separate the LED board from the SWITCH board.

## B) Removal of TRIGGER Board

1) Remove side panel from left side of plug-in, if not previously removed.
2) Disconnect the four coaxial end-leads that connect to J002, J212, J264, and J436 on the TRIGGER board. Also, pull the jumper wire end-lead that connects to P550 on the TRIGGER board.
3) Loosen the four Phillips screws from the studs connecting the TRIGGER board to the SWEEP board. (These do not remove completely; they are captive screws.)
4) Carefully lift the TRIGGER board from the SWEEP board. Use a gentle prying motion to separate the multi-pin (interboard) connectors. (When later replacing these boards, engage the top multi-pins first, then the lower ones. Be careful not to bend the pins.)

## C) Removal of hORIZONTAL PROGRAMMING and SHIELD Boards

1) Remove side panel from right side of plug-in, if not previously removed.
2) Loosen the six Phillips screws from the mounting studs that hold the HORIZONTAL PROGRAMMING board to the SWEEP board. (These do not remove completely; they are captive screws.)
3) Carefully lift the HORIZONTAL PROGRAMMING board from the SHIELD and SWEEP boards. Use a gentle prying motion to separate the multi-pin (interboard) connectors.
4) To separate the SHIELD board from the SWEEP board, remove the spring retainers from the mounting studs on the SWEEP board.

## NOTE

After removing the TRIGGER, HORIZONTAL PROGRAMMING, and SHIELD boards, it should not be necessary to remove the SWEEP board to access its components. However, the following steps are provided in the event that the entire circuit board must be replaced.
D) Removal of SWEEP Board

1) If the TRIGGER, HORIZONTAL PROGRAMMING, and SHIELD boards have not been removed, perform all steps of procedures $B$ and $C$, above.
2) If the front panel has not been removed, remove the 7B90P Rear Interface panel by unscrewing the four Phillips screws at each corner. (The SWEEP board can be removed out the front or rear of the plug-in.)
3) Remove the screws from the top of the three fasteners that hold the SWEEP board to the top and bottom plug-in rails.
4) Slide the SWEEP board assembly out from the plug-in rails. The plastic clip on the top rail should slide with the circuit board.

## Recalibration After Repair

After any electrical component has been replaced, the calibration of that particular circuit should be checked, as well as the calibration of other closely related circuits. Refer to Section 6 for these procedures.

## Repackaging For Shipment

If the Tektronix instrument is to be shipped to a Tektronix Service Center for service or repair, attach a tag showing: owner (with address) and the name of an individual at your firm that can be contacted. Include complete instrument serial number and a description of the service required.

Save and re-use the package in which your instrument was shipped. If the original packaging is unfit for use or not available, repackage the instrument as follows:

Surround the instrument with polyethylene sheeting to protect the finish of the instrument. Obtain a carton of corrugated cardboard of the correct carton strength and having inside dimensions of no less than six inches more than the instrument dimensions. Cushion the instrument by tightly packing three inches of dunnage or urethane foam between carton and instrument, on all sides. Seal carton with shipping tape or industrial stapler.

The carton test strength for your instrument is 200 pounds.

## SECTION 5

## CTRCUIT DESCRIPTION

## Introduction

This section describes the 7B90P block diagram and circuit operation in detail. Frequent reference is made to the schematic diagrams at the rear of the manual. Each fold-out schematic is tabbed and numbered for ease of use. As you go through the following descriptions, unfold the appropriate schematics and refer to them as needed.

## Block Diagram

Before we begin the detailed circuit descriptions, we will discuss the block diagram shown in Fig. 5-1.

The trigger generator provides a stable display by starting the sweep generator at a controlled point for each sweep. The trigger signal is taken from the INTernal or LINE trjgger signals from the mainframe or the EXT TRIG IN connector on the front panel. The sweep logic sets up the control signals for PP AUTO (Peak-to-Peak Auto), NORM (NORMal) or SGL SWP (SinGLe SWeeP) modes. This circuit also generates the hold-off timing for the sweep generator. The sweep generator provides a precise ramp voltage for the sweep outputs. The sweep time is calibrated from 500 picoseconds to 500 milliseconds per division. The horizontal preamp provides differential sweep signals to the mainframe and adds DC current for sweep positioning. If the MAGnifier is selected, the gain of the horizontal preamp is increased by a factor of ten (or 20 for $500 \mathrm{ps} / \mathrm{DIV}$ ) to magnify the sweep.

The 7B90P sweep speed, triggering, front panel LED's (Light-emitting Diodes) and IEEE 488-1975 interface are controlled by a microprocessor system in the plug-in. The heart of this system is a Motorola M6800 Microprocessing Unit (MPU). A control program, resident in 4 K ( $1 \mathrm{~K}=1024$ ) bytes of Read Only Memory (ROM) directs the MPU activity. The MPU also uses 128 bytes of Random Access Memory (RAM) as a "scratch pad".

Five Peripheral Interface Adapters (PIA's) handle internal communication between the MPU and other circuits in the plug-in. The IEEE 488 (also


Fig. 5-1. 7B90P Block diagram.
called GPIB - General Purpose Interface Bus) interface controls the handshaking and comrounication on the IEEE 488 bus. The clock circuit generates the 2-phase clock signals required by the MPU. When power-up occurs, the power-up circuit initializes the MPU and PIA's and starts the clock circuit.

## Sweep Generator

The sweep generator is divided into three main parts - the timing current source, the ramp generator and the auxiliary sweep preamp. Fig. 5-2 shows a block diagram of the sweep generator.


Fig. 5-2. Sweep generator block diagram.

Timing Current Source. The timing current source generates a precise source current for the ramp generator. The amount of current is controlled by signals from the microprocessor. The reference voltage generator and the source current generator comprise the timing current source (see Fig. 5-2).

U110 (top-center of schematic 2) generates a reference voltage for the source current generator that is selectable in $0.5,1,2$ volt steps. U110E and R103 set the reference voltage for the base of all the transistors in the package. These transistors are well matched, so their base-to-emitter voltages and emitter currents are essentially equal. Twice as much current flows through R102 and the pair of U110A and B, but the current divides
equally between U110A and B. As a result, the current through each individual transistor is the same as the other transistors in the package, providing good temperature tracking and current matching. DS001 and DS008 are current-steering diodes that allow control lines ST1 and ST2 (Sweep Timing 1 and 2) to switch the current through U110C and U110A and B. For example, when ST 1 is high, DS001 is on and the emitter of U110C goes to about +3.8 volts. This reverse biases the transistor and cuts off its collector current. If ST 1 is low, DS001 is off and U110C is on. LED's (Light-Emitting Diodes) are used as current steering diodes because they have a forward bias voltage drop of about 1.2 volts, providing a higher threshold voltage for ST1 1 and ST2.

The sum of the currents through all the forward biased transistors in U110 flows through R104, R105 and R312. (R312 is located at the bottom-center of schematic 5.) R 312 sets the level of current through R105, allowing precise calibration of the timing current and as a result, the sweep speed. The current through these resistors produces a reference voltage at the collectors of $\mathbf{U 1 1 0}$ that is proportional to the sum of the currents through U110. This voltage is fed to the non-inverting (+) terminal of U116.

U212, U116 and the associated components (left-menter of schematic 2) comprise the source current generator. This circuit generates a precise current for the ramp generator. The amount of current is a function of the reference voltage from U110 and the decade range selected by sweep timing lines ST3 and ST4 from the microprocessor.

U212 is a dual four-channel analog multiplexer. ST3 and ST4 drive the A and B inputs through buffers U320D and U320E. These inputs are decoded to select one of the four analog inputs ( $0,1,2,3$ ). Current is supplied to the analog inputs through the decade resistors, R200, R202, R203 and R108. The values of these resistors are selected to provide current to the analog inputs in decade steps (X1, X10, X100 and X1000). The selected analog input is connected to XCOM and YCOM outputs through separate analog switches in U212 (see Fig. 5-3). As a result, ST3 and ST4 select the level of current at the XCOM output.

The analog switches in $U 212$ have some series resistance ( $R_{o n}$ ) when they are on that changes with temperature. Changes in the value of $\mathrm{R}_{\mathrm{on}}$ would affect the current at XCOM, so the YCOM output drives a feedback amplifier, U116, that compensates for these changes. The ron-inverting (+) terminal of U116 is at the reference voltage set by U110. Since U116 is


## Fig. 5-3. Partial block diagram of the analog multiplexer, $\mathbf{U} 212$.

connected as a feedback amplifier, the inverting terminal will also be at this voltage (the feedback path for the amplifier is through the baseemitter junction of Q512, through U212 and back to the inverting terminal of U116). The input impedance of U116 is very high, so almost no current flows through $R_{o n}(Y)$ and the analog input terminal of $U 212$ is at the reference voltage. The current into the analog inputs of $U 212$ is set by the voltage drop across the selected decade resistor ( $\mathrm{R} 200, \mathrm{R} 202, \mathrm{R} 203$ and R108). As a result, the current in the analog input and through XCOM is a function of the selected analog input to $U 212$ and the reference voltage from U110, but is independant of changes in $R_{o n}$.

The output timing current from the XCOM output of U 212 flows through Q512 to the ramp generator. LR512 compensates for the base-to-emitter capacitance of $Q 512$.

Ramp Generator. The ramp generator uses the constant current from the timing current source to charge capacitors, generating a linear ramp voltage. The sweep starts when the trigger circuit asserts GATE (We'll discuss the trigger circuit later). This turns Q400 on (bottom-center of schematic 2) and turns Q406 off, so CR406 becomes reverse-biased.

The sweep timing capacitors are selected by ST5, ST6 and ST7. These lines are buffered by $U 320 A, U 320 B$ and $U 320 C$. The outputs of the buffers drive Q410, Q314, and Q312. When the slower sweep rates are selected ( $100 \mathrm{~ns} / \mathrm{DIV}$ and slower), these transistors switch C412, C314 or C410 in parallel with $C 500$ and C 510 . The constant current charges the selected capacitors and produces a linear ramp voltage.

Q516B and Q620 at the right-center of schematic 2 form a voltage follower. When the ramp at the emitter of $Q 620$ reaches +5 volts, the end of-sweep comparator, U530, switches. R536 and R534 set the refernce voltage for the comparator. When the comparator switches, C530 couples the negative-going output pulse at pin 7 back to the $A$ input to improve the switching time and provide hysteresis on the input. The positive-going output pulse on pin 5 of the comparator goes to the sweep logic and the GATE or AUTO SENSE line (depending on the trigger mode selected) goes to -15 volts. This turns Q400 off and Q406 on. Q406 sinks the timing current and discharges the timing capacitors.

If Q406 continued to sink current through the timing capacitors after they are discharged, the ramp voltage would drop below zero volts. The baseline stabilizer circuit prevents the voltage from falling below zero by sourcing current through CR406. Follow the line from the emitter of Q620 at the right-center of the schematic, back to the base of Q300A. When the ramp voltage falls to zero as the timing capacitors discharge, Q300A turns on the current through Q300B decreases. The base of Q604 goes to zero volts and its emitter current decreases. CR 406 becomes forward-biased and current flows through the diode and Q406. The current through CR406 keeps the baseline of the ramp from dropping below zero volts while the sweep is off. The baseline is held stable at zero volts until the next sweep starts.

Auxiliary Sweep Preamp. The sweep signal from the voltage follower also drives the auxiliary sweep preamp formed by CR732, Q732 and Q748. The differential stage of CR732 and $Q 732$ provides a high voltage-gain while Q748 drives the output. R746 provides negative feedback to set the stage gain. The non-inverting input to the amplifer at the anode of CR732 is grounded to provide a reference. The output of the amplifier drives the ramp output that is available on some mainframes.

Power Supplies. The circuitry in the upper right corner of schematic 2 distributes, filters and decouples the power supplies from the mainframe. A 5-volt, three-terminal regulator, $U 020$, generates the -5 volt supply required by the holdoff and trigger circuits and some of the logic circuits.

## Horizontal Preamplifier

The horizontal preamplifier is shown on schematic 4. This circuit takes the sweep signal from the ramp generator, converts it to a differential drive signal and adds DC positioning current. The circuit is divided into four main parts -- the input differential amplifier, the current follower, the output amplifier and the positioning circuit.

Input Differential Amplifier. The input differential amplifier consists of two differential amplifiers - one for X1 operation (MAG off) and one for X10 MAG operation. Q320, Q220, Q324 and the associated components comprise a constant-current switching circuit. When MAG is off, MAG (center of schematic 4) is high and Q320 is off. Q220 is off and Q324 is on, supplying current to Q224 and Q236 (see Fig. 5-4a). The sweep signal from the ramp genertor drives the base of Q224. DC positioning current is fed to the base of Q236. R236, R237 and R330 set the stage gain. The differential output signal from the collectors of Q226 and Q324 drive the emitters of Q120 and Q242 in a cascode configuration. Since Q220 is off, Q222 and Q234 are alsc off.

When MAG is on, MAG is asserted (low), so Q320 and Q220 are on and Q324 is off. Current is supplied through Q220 to the X10 MAG amplifier, Q222 and Q234 (see Fig. 5-4b). Since Q324 is off, Q224 and Q236 are also off. R228, R234, R235 and R232 set the gain of the stage. R232 allows the gain to be matched to ten times the X1 gain. R233 and C233 increase the gain of the circuit when the ramp starts at high sweep rates to improve linearity. The differential output signal is fed to Q120 and Q242.

Current Follower. Q120 and Q242 form a differential common-base current follower. This circuit improves the thermal stability and high-frequency characteristics of the input differential amplifiers by holding the base-to-collector voltage of the amplifier transistors near zero volts. To see how this circuit works, let's follow the ramp signal through the X10 input amplifer and current follower (The X1 circuit is identical).

Assume that we are at the beginning of a sweep ramp, so the input voltage is zero. The current through R221 forward biases CR222. The base of Q222 is at zero volts and the base of Q120 is one diode-drop (about 0.6 volts) below the base of Q222. The base-emitter juncticn of $Q 120$ produces about the same voltage drop, so its emitter is at zero volts. R128 decouples the base-tomcollector capacitance of Q222. Since R128 has a low value, it drops very little voltage. As a result, the collector of 0222 is very near


Fig. 5-4a. Simplified schematic of the X1 horizontal preaplifier.


Pig. 5-4b. Simplified schewatic of the $X 10$ horizontal preazplifier.
zero volts and its base-to-collector voltage is very low. The low voltage drop causes the device to dissipate very little power, providing good thermal stability. The low base-to-collector voltage also minimizes the effects of the junction capacity.

R220 and C220 improve the high frequency linearity of the circuit.

Output Amplifier. The output amplifier provides a differential drive of 50 millivolts per division (into 50 ohms ) to the mainframe and provides a 2 X magnification for the 500 picoseconds/division sweep speed. The output
of the current follower drives the bases of Q130 and Q132. The network of R122, R238 and R241 provides DC balance for the amplifer.

When the 500 picoseconds/division sweep speed is selected, STOA goes high and Q142 and Q146 saturate. The current through Q146 raises the gate voltage of Q134, forward biasing the gate-to-drain and gate-to-source junctions. This shunts R124 and R133 with a low impedance, changing the emitter current in Q130 and Q132 and doubling the gain of the amplifier. R240 adjusts the 500 picosecond sweep speed by setting the current through Q146 and Q134 which changes the emitter current in Q130 and Q132. When Q134 is off no current flows in the FEP (Field-Effect Transistor) so R124 and R133 set the stage gain. CR138 protects the base-emitter junction of Q146 against reverse breakdown when Q142 and Q146 are off.

Positioning Circuit. The DC positioning current comes from the 10-bit Digital-to-Analog Converter (DAC), U330, in the upper-left corner of schematic 4. The output current of the DAC is proporational* to the value of the 10 -bit input word. The value of this word can be set from the front panel POSITION control or under program control. We'll discuss the circuit that sets the value of the word in more detail in the MPU Control Example section of this circuit description. R342 sets the gain of the DAC and the output of the DAC drives U340. the gain of this amplifier is set by R343 and R344. When the 2 X magnifier is off (500 picoseconds/division not selected), STO is low and the analog switch is in the position shown on the schematic. When the TIME/DIV is set for 500 picoseconds/division, STO goes high and the switch closes, shunting $R 344$ and decreasing the amplifier gain. This causes the POSITION control to have the same resolution at 500 picoseconds/division as it does in other TIME/DIV settings.

When the 2 X magnifier is on at $500 \mathrm{ps} / \mathrm{DIV}$, the sweep positioning is shifted to correct for positioning offsets in the instrument. When STO is asserted as just discussed, U430E (bottom-left of schematic 4) is off (switch open). This shifts the positioning by connecting R446 in series with R644.

## Trigger Generator

The trigger generator (schematic 1) provides for a stable display by starting the sweep generator at a selected point on the input waveform. The trigger point can be set with the + SLOPE and LEVEL controls, or with the equivalent commands under program control. The triggering source can
be selected from the signal being displayed, a sample of the line voltage, or an external signal applied to the EXT TRIG IN connector. Fig. 5-5 shows a block diagram of the trigger generator.

Line Source. When LINE triggering is selected, LT (at the left edge of schematic 1) is asserted. U410A switches, connecting the LINE input (pin 1) to pin 15. A sample of the line voltage from the mainframe is fed through the analog switch (U410A) to pin 2 of the trigger source selector and amplifer (part of U220). If LINE triggering is not selected, LT is low, so pins 2 and 15 of U410A are connected together, applying -2.5 volts to the LINE input (pin 2) through R414 and R410. This voltage reversebiases the input transistor in U220, disabling the LINE input.

Internal Source. A differential copy of the displayed signal from the mainframe is fed to the INTernal trigger inputs (upper-left corner of schematic 1). This signal is split into two paths. Frequencies above about 16 KHz are coupled through C220 and R214 to U220. The low frequency path is through U112A, U350 and U358 to U220. The high and low frequency signal components are recombined in U220 to reproduce the original trigger signal.

When INTernal triggering is selected, IT goes high and U410B opens. Frequencies above 16 KHz are coupled through C220 to the INTernal input of U220. R214 and C220 set the roll-off characteristics of the high frequency path. R213, R212 and R214 set the input impedance of this line to about 50 ohms. If internal triggering is not selected, $U 410 B$ closes, connecting the -5 volt supply to pin 12. R320 and R214 divide the supply voltage down and apply -2.5 volts to the INT input of U220, disabling the input.

The differential internal trigger signal also drives U112A. Low frequencies (below about 16 KHz ) are amplified by U112A and passed to analog switches, U310B and U310C. R120, C110, R114 and C114 set the roll-off characteristics of U112A. If DC coupling is selected, DCI goes low and U310B closes, bypassing C312. When AC coupling is selected, $\overline{D C I}$ goes high, and $\overline{A C I}$ is asserted, closing U310C. This connects C312 in series with the low frequency path. In either case, the output signal is fed to pin 2 of U350 (upper-right of schematic 1 ).

U350, R350 and R218 form an inverting buffer. U358 mixes the internal signal from U350 and the external signal. The amplifier provides a gain of 2 from the non-inverting terminal (EXTernal signal) and unity gain from the inverting terminal (INTernal signal) to the output. Since only one of these signals is fed to the inputs of U358 at any time, the output has the

Fig. 5-5. Block diagram of the trigger generator.
low frequency components of the selected trigger signal. The output of this amplifier drives the LOW FREQ INPUT of U220. R116 keeps the offset voltage low by matching the input impedances on U220. C116 and R128 improve the circuit's stability. R134 provides for balancing the trigger internal source low frequency path in NORMal mode.

External Source. The external trigger signal is fed to J001, the EXT TRIG INput on the front panel. When the TERMination switch (S001) is out, R414 and R410 provide 50 -ohm termination for the line. C412, E412 (Ferrite Bead Inductor) and R412 keep the VSWR (Voltage Standing-Wave Ratio) on the input low. L412 compensates for the input capacitance to improve VSWR. If 1 megohm input impedance is selected (TERMination pressed in), S001 is open and R400 sets the input impedance. C412, C402, C436, C440 and the stray and cable capacitance comprise the 20 picofarad input capacitance.

The external trigger signal is split into two paths, as we discussed for the internal source. Frequencies below 16 KHz are coupled through R400 to the input amplifier, U452. CR464 and CR466 protect the input of U452 from overdrive. When EXT (X1) triggering is selected ET 10 (bottom-left corner of schematic 1) is high and U510C is open. In this mode R448 and C448 comprise the feedback for the amplifier. If EXT X10 zriggering is selected, ET10 goes low and U510C closes. R550 and C540 are connected in parallel with R 448 and C448, reducing the gain of the stage by a factor of ten. C448 and C540 set the frequency response of the stage.

The output of the amplifier is fed to the coupling switches (U510D, U510A and U510B). U510D opens the low frequency external trigger path when LINE or INTernal triggering is selected. U510A and B are open when AC LF REJ COUPLING is selected with EXTernal and INTernal source. When EXTernal and DC coupling are selected, $\overline{D C E}$ is asserted, closing U510A and bypassing C422. $\overline{A C E}$ goes low and $\overline{D C E}$ goes high for AC coupling, connecting C422 in series with the signal path. C422 and R454 set the input's AC coupling bandpass. The output of the coupling switches drives the non-inverting input of the mixer stage (U358). The output of the stage drives the LOW FREQuency INPUT of U220 as previously discussed.

Now refer to the left-center of schematic 1. The high frequency components of the external trigger signal pass through C402 and drive the high-frequency divider. C402, C436 and C440 form a capacitive divider that reduces the effects of cable capacitance and divides the signal down for the X1 and X10 EXTernal trigger modes. As a result, the signal at the gate of the top FET (Field-Effect Transistor), Q327, is ten tines as large as
the signal at the gate of Q440. R434 and C436 set the roll-off characteristics of the X 1 attenuator input signal. R 444 and C 440 perform the same function for the X10 divider. R436 and R442 prevent ringing in the divider.

Signals from the microprocessor control the divider. When EXTernal (X1) triggering is selected, ET1 goes low and Q530 saturates. R536 and R537 set the bias level for Q530. R522 and R524 set the current when the transistor is on and provide for DC offset correction for the EXT X1 circuit. C424 and R424 decouple the AC signal from the control logic signals. When Q530 saturates, its collector voltage rises, biasing, the source follower, Q327, into its active region. C422 keeps the DC from the capacitive divider. At high frequencies Q327 begins to rectify the signal and produces a DC offset voltage, so R420 and C326 feed a portion of the signal around the FET to compensate for this effect. Q326 supplies a. constant current to the source follower for good thermal stability.

The output of this source follower drives the EXT X1 input of U220. The EXT X1 input and the LOW FREQ INPUT form a differential input pair. The impedance of these inputs must be equal to minimize the offset voltages, so the on resistance of Q326 (about 90 ohms) in series with R324 matches the 1 K -ohm impedance on the LOW FREQ INPUT. CR334 and CR342 limit the positive polarity signal amplitude. CR336, DS 348 and the associated components limit the negative polarity $D C$ level and the $A C$ amplitude. When the EXT X1 trigger mode is unselected, Q327 is biased off and the output goes to about -2 volts, reverse biasing the EXT X1 input of U220.

The EXT X10 control circuit is identical to the EXT X1 circuit.

Trigger Level Circuit. The trigger level is adjusted by adding a DC level to the LOW FREQ INPUT of U220. The DC level comes from an 8 -bit DAC (U106), similar to the one used in the positioning circuit. The output current of the DAC is proportional to the value of the 8 -bit input word. The input word can be set from the front panel LEVEL control or under program control. We'll discuss the operation of the circuit in more detail in the MPU Control Example section. U112B and R103 convert the output current of the DAC to a voltage level for the input of U220.

Trigger Source Selector and Amplifier. We have discussed U220 only briefly. Now refer to schematic 1 and the block diagram in Fig 5-6 as we discuss the function of $U 220$ in more detail. The five high-frequency inputs go to the input switching circuit (four of the inputs are used). Unused or disabled inputs are tied to -2.5 volts, reverse biasing the corresponding
input transistor. The output of the switching circuit drives the left amplifier of the pair near the center of the block diagram. The LOW FREQuency INPUT also drives the amplifiers. The low- and high-frequency components are recombined in the amplifier and fed to the variable astenuator through the summing junction.

When High Frequency Reject (HF REJ) is selected the $\overline{H F R}$ control line at the right edge of schematic 1 goes low, reverse-biasing Q128. The HF REJect Select and Biasing circuit at the top of Fig. 5-6 enables the right amplifier of the pair and the high-frequency signals are fed to its noninverting terminal. C137 sets the rollwoff characteristics of the amplifier to reject high frequencies. The output of the amplifier goes through the summing junction to the variable attenuator. In all modes except PP AUTO (Peak-to-Peak Auto), this attenuator is set for X1 attenuation (We'll discuss PP Auto Mode later).

Peak-to-Peak Auto Circuit. $U 220$ also contains the Peak-to-Peak Auto and AGC circuits. The Peak-to-Peak Auto function can be divided into three major blocks. First, the peak detector determines the sigral size and DC positioning. Second, a DC centering loop centers the peak-detected output regardless of the $D C$ input and offset voltages. Third, the peak-tompeak level range output voltage is automatically adjusted until the trigger output reaches the clamp level set by the automatic gain control (AGC) to achieve full level range. The peak detector outputs, pin 14 and 15 of U220 (bottom-right of schematic 1), provide a differential rectified copy of the input trigger signal. C146 and C148 filter this output signal. U150B, CR140, U150A and CR050 form a second rectifier and filter stage. We'll discuss $U 150 B$ and $C R 140$ as an example. The other circuit is identical.

On the positive slope of the signal, the output of U150B follows the input in the positive direction, forward biasing CR140 and charging C140. Since the non-inverting terminal is tied to the anode of CR140, the output voltage follows the PEAK DET OUT signal very closely, without the forward-bias voltage drop of CR140. When the signal begins to go negative, CR 140 is reverse-biased and the feedback path for $U 150 B$ is opened. As a result, the output of the amplifier goes low (about +1 volts) very quickly. C152 slows the transition down to suppress noise. C140 remains charged at the peak output level because it's discharge path is very high impedance. The output of this stage drives the DC centering circuit and the peak-tompeak level range stage.


Fig. 5-6. Block diagram of U220.

Operational amplifier UO3OB adjusts the DC level of the LOW FREQ INPUT of U220 by adding a DC level to the signal. When PP AUYO mode is selected, PPA is asserted, closing U310D and applying the level at the output of UO30B to the LOW FREQ INPUT. This level nulls the DC input voltage and any accumulated DC offsets, allowing the trigger outputs at pins 16 and 17 of U220 to balance when the triggering LEVEL is set to zero. DC balance adjustment R243 provides centering for offset voltages resulting from circuitry external to U220 (e.g., U240).

The peak-to-peak level range amplifies the peak detector DC level to provide constant-amplitude trigger signals and to determine the range of the front panel LEVEL control and LEVEL program command.

The peak detector output level is coupled to pin 3 of U030 through R044 and R052. The gain of this feedback amplifier (consisting of U030A and AGC section of U220) increases as the peak detector output decreases. As a result, the trigger output signal level at pins 16 arid 17 of U220 is constant.

The range of the front panel LEVEL control and the LEVEL program control is zero at minimum triggering signal amplitude. The range increases as the triggering signal amplitude increases, until the maximum range is reached at the AGC threshold. Q134 is used as a switch. When PPA is high (unasserted) Q134 is off, and the peak-to-peak level range circuit is disabled.

The automatic gain control (AGC) stage limits the trigger signal amplitude to approximately 450 millivolts peak-tompeak (at pins 16 and 17 of U220) regardless of the trigger input signal amplitude. Current into the I AGC input (pin 3) of U220 sets the AGC threshold.

Slope Selector and Trigger Generator. U240 converts the differential trigger signal from the outputs of U 220 to a differential gate waveform for use by the gate generator stage. The +SLOPE button (or the SLO command under program control) determines whether the display is triggered on the positive-going or negative-going slope. When the SLOPE is set for positive (+SLOPE button lit or SLO POS set under program control), TL8 goes low. A positive-going signal on pin 13 of $\mathbf{U} 240$ produces a differential output pulse on GATE and GATE. If the SLOPE is set for negative, TL8 is high and a negative-going signal on pin 13 produces a differential GATE pulse. The $+/-$ SLOPE BALance adjustment provides optimum balance between the +SLOPE and -SLOPE modes.

The delay mode control signal into U240 (pin 16) is used only when the unit is operating as a delayed sweep unit in the E Horizontal compartment of a mainframe with two horizontal compartments. When the 7B90P is operating in the B starts after delay mode (as determined by the delaying-sweep time base in the A horizontal compartment), a high on DELAY MODE (pin 16 of U240) forces a GATE pulse, unless SWEEP DISABLE is asserted.

At the end of each sweep, the sweep logic generates a SWEEP DISABLE pulse. The high on SWEEP DISABLE (pins 6 and 10 of U240) disables the trigger generator long enough to allow the sweep generator to retrace and stabilize before another pulse starts the next sweep.

Gate Generator. The gate generator (bottom-left of schematic 2) provides an AUTO ENABLE GATE to the sweep logic and a SWEEP START GATE and Z AXIS to the sweep generator.

When U240 (right-center of schematic 1) asserts GATE and GATE, Q260 turns off and Q262 turns on. The current through Q262 causes SWEEP START GATE to go high, starting the sweep ramp. Q266 and the associated components speed up the transitions on GATE and drive the AUTO ENABLE GATE. At the end of the sweep, GATE goes low and Q262 turns off, ending the sweep ramp as previously discussed. If the time base is in PP AUTO mode and no trigger is received, AUTO SENSE goes high to initiate a sweep. The high on AUTO SENSE turns Q264 on, asserting SWEEP START GATE.

## Sweep Logic

The sweep logic (schematic 3) controls the sweep modes and the associated time base functions. The logic also generates control signals for the mainframe. Fig 5-7 shows the block diagram of the sweep logic.

PP Auto Mode. The PP Auto mode provides a free-running trace in the absence of a trigger signal. Since there is no trigger, GATE (bottom-left of schematic 2) and AUTO ENABLE GATE are low. PPA, pin 19 of U448 (top-center of schematic 3) is asserted. The low on PPA tells $U 448$ that the time base is in AUTO mode. AUTO SENSE goes low and turns Q264 on (bottom-left of schematic 2). This generates a SWEEP START GATE, initiating a sweep. At the end of the sweep, the end-of-sweep comparator switches, pulsing HO START (Hold-Off Start - pin 16 of U448). AU'O SENSE and SWEEP DISABLE go high and stay high until the hold-off circuit times out (we will discuss the hold-off circuit in the MPU control example section.) If no trigger

Fig. 5-7. Sweep logic block diagram.
is received, AUTO SENSE goes low after the hold-off tine and the process is repeated.

If a trigger is received, the time base functions as if it were in NORMal mode except that the trigger point can be set at any point between $10 \%$ and $90 \%$ of the peak-to-peak amplitude of the trigger signal. Pin 7 of U448 goes low and the TRIG'D (Triggered) Light comes on. R650 and C650 keep the trace from free-running between triggers by providing a delay. AUTO TIMING, pin 6 of U448, goes low with each trigger pulse. After the pulse, C650 begins charging through R650. If no triggers are received within about 63 milliseconds, the sweep AUTO TIMING goes high and the sweep begins to free-run.

Normal Mode. When NORMal mode is selected, a sweap is initiated only on a valid trigger. In this mode, SS (Single Sweep-pin 12 of U44) is low and AUTO MODE, pin 19, is unasserted (high). When a trigger is received, GATE goes high to initiate a sweep and the TRIG'D light comes on. At the end of the sweep, HO START is asserted and SWEEP DISABEE goes high, disabling the trigger generator. When the hold-off circuit times out, SWEEP DISABLE goes low and the next valid trigger repeats the process.

Single Sweep Mode. When single sweep is selected, SS (pin 12 of U448) is high. If no trigger is received, GATE is low and the READY light is on. When a valid trigger is received, AUTO ENABLE GATE goes high, the TRIG'D (Triggered) light comes on and the sweep starts. At the end of the sweep, HO START is pulsed high, resetting the single-sweep latch in U448 and the READY light goes out. SWEEP DISABLE goes high and stays high until the single sweep latch is reset. When the hold-off circuit times out, the TRIG'D light goes out.

If RESET is pressed or the plug-in receives a SSW ARM command on the IEEE 488 bus, $\overline{S S R}$ goes low (the single sweep may also be reset by a control line from some mainframes). The low on pin 15 of 4448 initiates an internal pulse that resets the single sweep latch. C448 sets the width of this pulse. The SWEEP DISABLE line goes low and the READY light comes on. The time-base waits for the next valid trigger and repeats the process.

Holdoff Circuit. The hold-off circuit disables the sweep to allow the ramp generator to retrace and stabilize for the next sweep. Since larger timing capacitors are used for the slower sweep speeds, the discharge (retrace) time is longer, so the hold-off time must also be longer. U630, at the bottom-left of schematic 3, decodes ST4, ST6 and ST7 and switches
timing capacitors into the hold-off timing circuit to change the hold-off time as the sweep speed changes.

When HO START (pin 16 of 4448 ) is pulsed, HO OUT (Hold-Off Output) goes high, turning Q252 on. 0250 drives the mainframe HOLDOFF control line. The collector of Q252 goes low, turning Q740 on and Q648 off. A constant current, generated by Q646 and the associated components, charges the selected holdoff timing capacitors. When the voltage on the capacitors exceeds the reference voltage from U654, the hold-off comparator U644 switches and HO TIMING goes high, terminating the hold-of: cycle.

The reference voltage is set by a 6-bit DAC that is similar to the one used in the positioning and trigger level circuits. The value of the input word can be set from the front-panel HOLD OFF control or under program control. This allows the hold-off time to be varied within a fixed range, determined by the sweep speed setting. U654 converts the DAC output current to a reference voltage for the hold-off comparator.

## NOTE

Some mainframes disable the holdoff circuit to generate a display graticule.

Lockout Buffer Amplifier. The SWEEP LOCKOU'T allows the sweep to be disabled for special mainframe functions or when the 7B90P is operated as a delayed time base in a mainframe with two time-base compartments.

Q552 and Q546 form a differential amplifier that provides good thermal stability for the lockout circuit. When SWEEP LOCKOUT is asserted, Q552 is biased off and CR550 turns on, clamping the base of $Q 544$ to about +0.7 volts. Q544 turns off and its collector goes high. The positive pulse turns Q558 on for a time set by R554, R556 and C544. The low on the collector of $Q 558$ initiates a hold-off sequence, stopping the sweep and inhibiting triggers. At the same time, the high on LO IN (LockOut INput) disables the sweep logic until the SWEEP LOCKOUT line is released.

## Microprocessor System

We mentioned in the block diagram discussion that the 7B90P is controlled by a microprocessor system in the plug-in. The major components of the system are shown on schematic 6. Full information on the components that comprise the system is given in the Motorola M6800 Microcomputer System Design Data Manual. It will be helpful to have a copy of this manual for reference as you go through the description.

MPU and Memory. The Motorola M6800 (U920) is an 8-bit Microprocessing Unit (MPU). The MPU has a 16 -bit address bus ( 13 bits are used in this system) that allows it to selectively communicate with the memory and PIA's. A separate 8 -bit bi-directional bus carries data to and from the MPU .

If an external device asserts the MPU's IRQ (Interrupt ReQuest) or NMI (Non-Maskable Interrupt) inputs, the normal progran execution is interrupted. The processor saves its current status, then determines what caused the interrupt and executes an appropriate service routine. These service routines include such tasks as scanning the front panel buttons or getting a byte from the IEEE 488 bus. While servicing the interrupt, the processor can mask its $\operatorname{IRQ}$ input so that further interrupts will not be recognized until the current interrupt has been serviced. The NMI input cannot be masked. We'll discuss the interrupt sequence in more detail later.

The program that controls the MPU's activity resides in two 2048 x 8 -bit ROM's (U830 and U1030). The ROM's are accessed by asserting an address in the range $7000_{16-7 F F F}^{16}$. The MPU also uses a 128 x 3-bit RAM (U1130) as a "scratch pad" memory. The RAM occupies addresses from 0 to 7 F 16.

PIA's. The processor communicates with other parts of the plug-in and the IEEE 488 bus through five M6820 PIA's (U510, U710, U810, U1110, and U1420). Complete information on these chips is contained in the M6800 Microcomputer System Design Data Manual. We'll discuss only the relevant details here.

Figure 5-8 shows a block diagram of the M6820. Each PIA contains two separate Input/Output (I/O) ports and control registers. Each line of the I/O ports can be programmed as an input or output by setting bits in the data direction registers.


Fig. 5-8. Block diagram of the M6820 PIA.

Two interrupt outputs ( $\overline{\operatorname{IRQA}}$ and $\overline{\operatorname{IRQB}}$ ) signal the processor that the PIA has data or needs service. When an interrupt occurs, a flag is also set in the appropriate control register. Since the interrupt lines are OR-tied to the MPU's IRQ input, the processor must read each PIA's control registers to determine which one generated the interrupt.

Power-up Circuit. When power is applied to the plug-in, the power-up circuit monitors the power supplies and generates initialize signals for the MPU, clock circuit, and PIA's. The power-up circuit is located in the bottom-center of schematic 7 .

Comparators U430A, U430B and U430C monitor the power supplies. The outputs of the comparators are OR-tied so that all the supplies must be present before the outputs go high. When they do, C528 begins charging through R430 and R530. When the voltage at the top of R530 reaches about 4.1 volts (about 2 milliseconds after the comparator outputs go high), U430D switches and its output goes low. As a result, POR (Power-On Reset) and the input of $U 530 B$ go low. The output of $U 530 B$ and $P O R$ go high. The POR and POR lines initialize the PIA's and assert the TSC (Three-State

Control) input on the MPU (Refer to the M6800 manual for more information on this input).

U1630B and U310B form an R-S flip-flop that controls the clock circuit. Since the output of U 530 B stays low for 2 milliseconds after the power supplies are stable, this flip-flop powers-up with the output of U310B high. The flip flop will stay in this state until $\overline{S T O F}$ is asserted. We'll discuss the STOP line later.

Since pin 8 of $U 310 B$ is high, when the output of $U 530 B$ goes high, the output of U010A also goes high. The lownto-high transition fires U1530A and the $Q$ output goes low for about 13 microseconds. Fin 8 of U010C is high so MPURES (MPU RESet) is asserted until U1530A times out. This negative-going pulse forces the MPU into a reset sequence. (Refer to the M6800 manual for specific information on the MPU reset sequence.)

Clock Circuit. U030A, U030B and the associated components in the upper-right corner of schematic 4 generate the 2 -phase MPU clock signal.

Recall that the output of U010A makes a low-to-high transition shortly after power-up. This transition fires U030A, so its $\bar{Q}$ output goes low. About 500 nanoseconds later, U030A times out and the low-to-high transition on the $Q$ output fires U030B. U030B times out about 500 nanoseconds later and the negative transition on the $Q$ output fires U030A, repeating the process.

The $\bar{Q}$ outputs of U030A and U030B feed open-collector drivers (U230B and U230A). The outputs of these gates drive phase 1 and phase 2 clock lines. The MPU clock inputs require a shorter rise time than the opencollector drivers can produce with simple pull-up resistors. Q126, Q135 and the associated components form active pull-ups that improve the rise time of the driver outputs. We'll discuss Q126 and the associated circuit as an example.

When the $\bar{Q}$ output of $U 030 B$ makes a high-to-low transition, $C 024$ couples the negative going pulse to the base of Q126. As a result, Q126 turns on for a short period and pulls the output of U230A up very quickly. The positive-going transition does not affect the circuit since the base of Q126 must go below about 4.5 volts to turn on.

The $R-S$ flip-flop of $U 1630 B$ and $U 310 B$ controls the clock generator as previously discussed. When the MPU is idle it asserts 80016 on its
address bus for about 500 nanoseconds and VMA (Valid Memory Address) goes high (upper left of schematic 6). U630 and U1310 decode the address and assert $\overline{S T O P}$ while the address is on the bus. When STOP goes low, the flip-flop is reset and the output of U010A goes low, disabling the clock. The clock remains off until an interrupt occurs ( $\overline{I R Q}$ is asserted) or an Interface Clear message is received on the IEEE 488 bus (IFC is asserted). Since the MPU is a dynamic device, data is lost if the clock is stopped for more than about 10 microseconds. As a result, when the clock is restarted, a power-on sequence is executed as previously ciscussed.

Front Panel Buttons. The front panel buttons provide for local control of the 7B90P. Refer to schematic 5 and notice that the front panel buttons are organized in a $4 \times 8$ matrix (excluding the TERMination button). The microprocessor sets the row lines (R0-R7) low so when a button is pressed, one of the column lines (CO-C3) goes low (pull-ups in the PIA hold the lines high when they are open). If any of the column lines go low, the output of U 310 A (top-right of schematic 6) goes low, pulsing the CA1 input. This generates an interrupt to the MPU. When the MPU services the interrupt, it takes all the row lines high, then it sets the row lines low one at a time and reads the column lines to determine which button was pressed. When the button is found, the MPU executes an appropriate service routine.

Front Panel Lanp Decoders. U630, U110 (bottommerght of schematic 8) and the associated gates decode the sweep timing lines, and drive the front panel TIME/DIV LED's. Schematic 5 shows the front panel LED's and drivers.

## GPIB Interface

The GPIB Interface handles all communication between the MPU and the IEEE 488 bus. This description assumes a basic understanding of the GPIB protocols. Refer to IEEE STD 488-1975 for a complete discussion of the bus.

Before we discuss the GPIB Interface in detail a review of the handshake process is necessary. The IEEE 488 bus uses a three-wire interlocked handshake to transfer all data bytes. When all. the addressed listeners on the bus are ready to receive data, $\overline{N R F D}$ (Not Ready for Da־a) goes high (not-asserted). The talker asserts data on DIO $0-D I O 7$ and, when the data is stable, asserts DAV (Data Valid). When all the addressed listeners have accepted the data, NDAC (Not Data Accepted) goes high, and the talker releases DAV. This process is repeated for every byte tha* is transferred
on the bus. Refer to Appendix B of the IEEE Standard for a complete discussion and timing diagrams of the handshake process.

IEEE Data Bus. The GPIB data bus (DIOØ-DIO7) uses time-slot lines 2 through 9 (TS2-TS9) on the plug-in edge connector as a data bus. These lines are used in non-programmable mainframes for the readout time-slot lines. In a programmable mainframe, IFC (upper-left corner of schematic 7) is tied to a line that is clamped to about +3 volts maximum, turning Q1412 on. U1500 expects the modified time-slot information on TS 1 and TS 10 while TS2-TS9 are enabled as the GPIB data bus. We'll discuss the readout system later.

Acceptor Handshake. The 7B90P must "listen" to the data bus whenever it has been addressed or ATN (Attention) is asserted. When either of these cases are true, the output of U1630D is high, enabling U1630A, U1720A and U1720D. The MPU asserts BUSY left-center of schematic 7) while it is processing a byte from the bus. When BUSY and ATN are not asserted, the output of U530A is high. The output of U530D is low when DATVAL (DATa VALid) is unasserted and the output of U530A is high. NRFD is high (unasserted), indicating that the interface is ready for data, when the output of U1630D is high and the output of U530A is lcw. NRFD is asserted (low) if BUSY or DATVAL are asserted, indicating that the MPU is processing a byte or the interface is receiving a byte. When $\overline{A T N}$ is asserted the plug-in must listen, so $\overline{N R F D}$ goes high.

With NRFD high (ready for data) the interface is ready for the next step in the handshake process. The talker places data on the bus and, after a settling delay, asserts DAV. When DAV goes low the output of U1520D goes high and the clear input of $U 1730 B$ is released. The MPU pulses DATACC (DATa ACCepted) when it has received the byte. On the rising edge of DATACC, U1730B is clocked and the $Q$ output goes low. The low on the input of U1720A causes NDAC to go high. This tells the talker that the data is accepted. Finally, the talker relases $D A V$ and the output of $U 1520 D$ goes low, clearing U 1730 B and asserting $\overline{\mathrm{NDAC}}$. If BUSY is asserted (indicating that the input buffer is full or EOI has been received), it is released when the MPU is ready for more data. This completes the acceptor handshake process.

Source Handshake. The source handshake is not implemented as defined by the IEEE 488 Standard. Instead, the plug in talks through the mainframe using a non-interlocked handshake. The mainframe then generates the standard three-wire GPIB source handshake and transfers the data to the IEEE 488 bus.

When the 7B90P becomes a talker, TALK (left-center of schematic 7) is asserted. The MPU places its data on the bus and pulses SEND low. If $\overline{\text { ATN }}$ is not asserted, the negative transition on SEND fires one-shot U1530B. The $Q$ output of $U 1530 B$ is asserted for about 1 microsecond and the output of U530C (SND) goes low. When U1530B times out, SND goes high and the mainframe latches the data from the plug-in. When the mainframe has completed the transfer, $\overline{D A V}$ makes a low to-high transition (recall the discussion of the GPIB handshake process). This transition clocks U1730A and the $Q$ output goes to the current state of $\overline{\text { ATN }}$. If $\overline{\text { ATN }}$ is high (unasserted), SENT goes high, telling the microprocessor that the data transfer is complete. If $\overline{\text { ATN }}$ is asserted during the data transfer, U 1530 B cleared and BADTLK is asserted, indicating that the transfer was interrupted.

When the 7B90P is listening, U1500 sources current onto the bus. The mainframe can sink this current and control the bus, but other programmable plug-ins cannot. As a result, when one plug-in is talking, all others must be off the bus ( $U 1500$ disabled). U1620C gates SND with the $Q$ output of U1530B. If $\overline{S N D}$ is asserted and the $Q$ output of $U 1530 B$ is low, another plug-in is talking, so the output of U1620C goes low. The output of U1520E and ENABLE go high, disabling U1500.

## Readout Circuit

The readout circuit encodes the sweep speed settings and sends the information to the mainframe for display with the sweep signal. Before we begin the detailed circuit description, a review of the basic readout system is necessary. Refer to the mainframe manual for a complete description of the readout system.

Basic Readout Operation. In a non-programmable mainframe, the plug-in receives time-slot pulses on the time-slot lines (TS1-TS 10). These lines are asserted (Pulled to -15 volts) sequentially so that only one line is asserted at any time. During these pulses, data is sent to the mainframe in the form of analog current on the row current and column current lines. These currents range from zero to 1 milliamp in 100 microamp steps. The current level on these lines defines a particular character in the character selection matrix shown in the mainframe manual.

The data encoded on the row and column lines is assigned a particular meaning during each time slot (see Table 5-1). For example, data on the lines during time slots 9 and 10 define the units of measurement for the plug-in (seconds in this case).

Table 5-1

Standard Readout Format

| Time Slot Nuwber | Description |
| :---: | :---: |
| TS-1 | Determines the decimal magnitude (number of zeroes displayed or prefix change information) or the IDENTIFY function. (No display during this time-slot.) |
| TS-2 | Indicates normal or inverted input (no display for normal input). |
| TS-3 | Indicates calibrated or uncalibrated condition of plug-in variable control. (No display for calibrated condition.) |
| TS-4 | 1-2-5 scaling |
| $\begin{aligned} & \mathrm{TS}-5 \\ & \mathrm{TS}-6 \\ & \mathrm{TS}-7 \end{aligned}$ | Not encoded by plug-in. Left blank to allow addition of zeroes by Readout System. |
| TS -8 | Defines the prefix which modifies the units of measurement. |
| $\begin{aligned} & T S-9 \\ & T S-10 \end{aligned}$ | Defines the units of measurement of the plugin unit. May be standard units of measurement (V,A,S, etc) or special units selected from the character selection matrix. |

Suming The Tiwe Slots. Programmable plug-ins and mainframes use time-slot lines TS2-TS9 as a data bus, as previously discussed. The time-slot pulses are transmitted to the plug-in as a stream of pulses on the TS10 line. The summation of the time slot pulses appears at pin 10 of U1500. Q1404 and R1515 shift the level of the -15 volt pulses to TTL-compatible levels. C1400 makes driving the signal for U1520F a low impedance at high frequencies. If C 1400 is not present, this signal follows the threshold level of the schmitt input. U1520F inverts the pulses and improves their
risetime. The output drives a four-bit counter, U 1005 (upper-left of schematic 8). We'll come back to this counter shortly.

TS1 is asserted during its normal time-slot in both programmable and non wrogramable mainframes. VR810, R810 and Q904 shift the level of the TS-1 pulse and invert it. This pulse clears U1005 to insure that the counter always starts at zero during time-slot 1.

At the end of time-slot 1, pins 2 and 3 of $U 1005$ go low, enabling the counter. The output of U1520F (TSCLK Time-Slot Clock) al so goes low, and the falling edge clocks U1230. U1230 continues to count up in binary until time slot 1 is asserted again at the beginning of the next train of time slot pulses. When time slot 1 is asserted, the counter is cleared and the process repeats. As a result, the $Q_{A}-Q_{D}$ outputs of $U 1230$ count from zero to nine and start over. These outputs drive U1100 and U1000.

Selecting Row and Columan Currents. U1100 is a four-to-ten line decoder that asserts one output for each binary input word from zero through nine (outputs 3, 7 and 8 are used). Output pins 4, 9 and 10 drive current switches that sink the necessary row and column currents during time-slots 4, 8 and 9. For example, during time-slot 9, Q1308 sinks 400 microamps on the row current line.

U1000 takes the BCD coded time-slot number from U 1005 and the information on W1-W12 from the MPU. On the basis of these inputs U 1000 sinks a selected amount of current on pin 19. This current and the current in Q1304, Q1308 and Q1310 combine to form the row and column current for each time slot, and as a result, determine the character that is displayed during that time slot.

## MPO Control Example

We've discussed the components of the microprocessor system and the analog circuitry. Now let's see how the MPU controls the plug-in by going through one routine in the ROM program.

The sweep positioning, trigger level and hold-off circuits are very similar. We'll discuss the sweep positioning circuit as an example. This discussion of the circuit operation and microprocessor routines apply to all three circuits.

Recall from our discussion of the Positioning Cirouit that the DC positioning current came from a 10-bit Digital to Analog Converter (U330). The input word for the DAC is set by the microprocessor. When the plug-in is in remote mode, the POS command sets the value of this word.

If the 7B90P is in local mode, the POSITION control drives a tracking analog-tomdigital converter consisting of a comparator circuit, the microprocessor and the DAC. If, for example, the POSITION control (R130 upper-right of schematic 5) is turned clockwise (trace moves right), the voltage at FB 20 becomes less negative. The current through R814 increases and the voltage at pin 9 of $U 940 C$ and pin 10 of $U 940 \mathrm{D}$ becomes positive. As a result, the output of U940C switches high. We'll some back to this point shortly.

FPEN (Front Panel Enable) at the top-center of schematic 7 is asserted when the 7B90P is in local mode. The high on FPEN enables U010B so the phase-w clock pulses appear at its output high. These olock pulses go to U000A and $B$ and $U 420 \mathrm{~A}$ through $U 420 \mathrm{D}$ at the center of sohematic 7 . When the output of $U 940 C$ switches high as just discussed, the clock pulses appear at the output of $4420 B$ (SPDN Sweep Position Down). SPDN goes to one of the interrupt inputs of 4810 at the bottom-center of schematic 6 . The negative transition on this input sets the interrupt request flag in the control register of the PIA. At the same time, IRQA goes low, asserting the IRQ input of the MPU. The MPU saves its current status and sets the interrupt mask bit so that no further interrupts can occur until this one is serviced or the mask is cleared under program control. Then the processor jumps to a routine that reads the PIA control registers looking for a set interrupt flag. When the MPU reads control register $A$ of $U 810$, it finds the SPDN interrupt flag (bit 6 of the register) set. It then clears the flag and jumps to a service routine that takes control of the DAC input word.

The service routine checks if the interrupt was from the SPUP or SPDN comparators. For our example, the SPDN flag was set, so the service routine decrements the value of the DAC input word. As a result, the output current at pin 3 of U330 (top-left of schematic 4) decreases. The output of U340 swings less positive and begins to compensate for the shange in the voltage level at the input of the comparators (SPA). If the voltage at the inputs of the comparators is balanced within its stable "window", the outputs of both comparators go low and the SPUP and SPDN outputs both go high.

Since the POSITION control moves very slowly in relation to the MPU's speed, the POSITION service routine decrements the value of the DAC input
word by one and then checks all interrupts again. This allows other faster interrupts, such as transfers on the IEEE 488 bus, to be serviced while the operator adjusts the POSITION control. If the output of the DAC rises enough when the input word is decremented to bring the comparator inputs within their stable "window", no further interrupts occur. If the comparators' input voltage is not within this window, the SPDN interrupt occurs again on the next falling edge of the phase 2 clock. The MPU re-enters the interrupt routine and decrements the DAC input word again. This process is repeated until the SPUP interrupt stops occurring or the 10 -bit word reaches its maximum value. The positioning current that is added to the signal, as previously discussed, comes from the output of U340 (upper-left of schematic 4).

If no interrupts occur and the MPU becomes idle, it asserts the STOP line. This stops the clock to minimize noise in the analog circuits. When an interrupt occurs or an IFC message is received, the clock restarts and the MPU executes a power-up routine.

The MPU control the front-panel buttons, front-panel LED's and the GPIB interface in a similar interrupt-driven fashion.

## SECTION 6

## CALTBRATIOR

## Introduction

To assure instrument accuracy, check the calibration of the 7B90P every 1000 hours of operation or every six months if used infrequently. Before complete calibration, thoroughly clean and inspect this instrument as outlined in the Maintenance section.

Tektronix Field Service. Tektronix, Inc., provides complete instrument repair and recalibration service at local Field Service Centers and the Factory Service Center. Contact your local field office or representative for further information.

Performance Check. The performance of this instrument can be checked by performing only the steps listed in the Performance Check procedure. These steps check the instrument against the tolerances listed as Performance Requirements (see Specification section of this manual).

Limits and tolerances given in other check steps are calibration guides and should not be interpreted as instrument specifications. Operator front-panel adjustments are adjusted as part of the Performance Check procedure.

Calibration. To verify proper calibration of the $7890 P$ and to prevent unnecessary re-calibration of the entire instrument, perform the Adjustportion of a step only if the tolerance given in the Check part of the step is not met.

When performing a complete calibration procedure, make each adjustment to the exact setting even if the Check- is within allowable tolerance.

## Test Equipilient Required

General. The following test equipment and accessories, or its equivalent, is required for complete calibration of the 7B90P. Specifications given for the test equipment are the minimum necessary for accurate calibration. Therefore, the specifications of any test equipment used must meet or exceed the listed specifications. All test equipment is assumed to be correctly calibrated and operating within the listed specifications. Detailed operating instructions for the test equipment are not given in this procedure. Refer to the instruction manual for the test equipment if more information is needed. Table 6-1 shows a complete list of the test equipment.

Calibration Equipment Alternatives. All of the listed test equipment is required to completely check and adjust this instrument. The Calibration procedure is based on the first item of equipment given as an example of applicable equipment. When other equipment is substituted, control settings or the calibration set-up may need to be altered slightly to meet the requirements of the substitute equipment. If the exact item of test equipment given as an example in the Test Equipment list is not available, first check the Specifications column carefully to see if any other equipment is available which might suffice. Then check the Usage column to see what this item of test equipment is used for. If used for a check or adjustment which is of little or no importance to your measurement requirements, the item and corresponding step(s) can be deleted.

Checking Programmable Features. The basic 7B90P calibration procedure does not require a controller. If an IEEE 488 interfaced controller is available, the programmable features of the 7B90P can be tested by sending commands to the plug-in over the IEEE 488 bus. Refer to the programming section of this manual for specific information on the commands.

The commands can also be used to change the 7B90P settings during the calibration procedure. For example, a program could be written to set the 7B90P POSITION DAC (Digital to Analog Converter), simplifying the POSITION adjustments.

TABLE 6-1
Test Equipment

| Description | Minimu <br> Specifications | Usage | Examples of Applicable Equipment |
| :---: | :---: | :---: | :---: |
| Calibration Mainframe | 7900-Series programmable mainframe. | Used to provide display. | Tektronix 7912AD <br> Programmable <br> Digitizer and <br> necessary <br> peripherals <br> for display. |
| Wide-Band Amplifier Plug-in Unj.t | 7A-Series amplifier plug-in unit; Bandwidth - 500 MHz ; Deflection factor 50 mV to 5 V . | Used throughout procedure to provide vertical input to the mainframe. | Tektronix 7A 19 Amplifier. |
| Time-Mark Generator | Marker outputs - 2 ns to 5 s ; accuracy within $0.1 \%$. | Sweep timing check and adjustments. | Tektronix TG501 <br> Time-Mark <br> Generator. 1 |
| Low-Frequency Sine-Wave Generator | Frequency - 30 Hz to 50 KHz ; output amplitude - variable from 50 mV to 8 V into 50 ohms. | Low-frequency triggering checks and adjustments. | Tektronix FG503 Generator. 1 |
| Mid-Frequency Sine-Wave Generator | Frequency - 250 KHz to 250 MHz ; output amplitude range - 5 mV to 5.5 V peak-to-peak. | Mid - Frequency Trigger checks and adjustments | Tektronix SG503 Generator. 1 |
| High-Frequency <br> Sinewhave <br> Generator | Frequency - 245 MHz to 500 MHz output amplitude -0.5 V to 4.0 V . | High Frequency <br> Trigger checks and adjustments | Tektronix SG504 Generator. 1 |

TABLE 6-1 (cont.)
Test Equipment

| Description | Minimuim <br> Specifications | Usage | Examples of Applicable Equipment |
| :---: | :---: | :---: | :---: |
| $\begin{gathered} \text { Precision } \\ \text { Coaxial } \\ \text { Cable } \end{gathered}$ | 50 ohms. | $\begin{aligned} & \text { Used with } \\ & \text { SG503. } \end{aligned}$ | Tektronix part number: 012-0482-00. |
| Digital Volt Meter | Range - from 200 <br> millivolts to 10 volts. | Trigger Level adjustments. | Tektronix DM502. 1 |
| Test Oscilloscope | ```Vertical Sensitivity - 500 mV to 1 V Bandwidth - to }10\textrm{MHz``` | Holdoff check procedure. | Tektronix 465 Oscilloscope. |
| Plug-in <br> Extender | Tektronix 7000-series extender. | Provides access to internal adjustments and test points. | Tektronix <br> Calibration <br> fixture part <br> number: 067-0589-00.1 |
| Coaxial <br> Cables <br> (2 required) | Impedance - 50 ohms; <br> Type - RG58/U; length 18 and 42 inches; connectors - BNC. | Used throughout procedure | Tektronix part number: <br> 18 inch- <br> 012-0057-01 <br> 42 inch- <br> 012-0076-00. |
| BNC T Connector |  | External <br> trigger checks and adjustments. |  |
| X5 50 ohm attenuator |  | Trigger checks and adjustments. | Tektronix part number: 011-0060-01. |

1 Requires TM 500-Series Power Module.

## Performance Check Procedure

1. Insert the 7B90P and the vertical amplifer unit into the mainframe. Turn on the mainframe and monitor power and allow 20 minutes warmmup time before proceeding.
2. Set the 7912AD to LOCAL and TV modes.
3. Adjust the Graticule Intensity for a sharp graticule.
4. Adjust the Main Intensity and Focus for a visible trace with no blooming.
5. If a Tektronix IEEE 488 interfaced controller and TEK SPS BASIC software are available the $7 B 90 P$ programmable functions can be checked by running the 7912AD checkout software. Refer to the Cheokout Software Manual for information on loading and running the software. Checkout programs can be written for other IEEE 488 interfaced controllers by studying the information given in the programing section of this manual.


The writing beam intensity in the 7912AD is critical. Excessive intensity can permanently damage the scan converter target. Many portions of the Performance Check and Calibration procedures require adjustment of intensity or sweep speed. Remember that reducing sweep speed has the same effect as increasing intensity. Al ways reduce intensity before reducing sweep speed.

## Check Triggering Modes

Test Set-0p

Set the 7B90P controls as follows:

TIME/DIV: 20 microseconds/division
MAG: Off
MODE: PP AUTO
COUPLING: AC
SOURCE: INTernal
SLOPE: + (light on)
POSITION: Midrange
HOLD OFF: Fully counterclockwise

1. Connect the low-frequency sine wave generator to the amplifier unit input with the 42-inch BNC cable.
2. Set the generator frequency to 50 kilohertz.
3. Set the low-frequency sine-wave generator and the amplifier unit VOLTS/DIV for approximately two vertical divisions of deflection.
4. Check $-\cdots$ for a stable display at all TRIGGER LEVEL control settings.
5. Set the TRIGGERING MODE to NORMal.
6. Adjust the LEVEL control for a stable display ( (RIG'D light on).
7. Check- for no display (TRIG'D light off) when the LEVEL control is set fully clockwise and fully counterclockwise.
8. Set the LEVEL control for a stable display.
9. Set the TRIGGERING MODE to SinGLe SWeeP and the SOURCE to EXTernal.
10. Press the SinGle SWeeP RESET button (part of the SGL SWP MODE button).
11. Check- that the READY light is on.
12. Check- for one sweep when the INTernal SOURCE button is pressed. The Mainframe Intensity may need to be increased to view the single sweep display.
13. Check- that the READY light goes out at the end of the sweep.
14. Decrease the 7912AD Main Intensity and remove the signal connection.

## Check External Level Range

## Test Set-wp

Set the 7B90P controls as follows:

TIME/DIV: $\quad 500$ microseconds
MAG: Off
MODE: NORMal
COUPLING: AC
SOURCE: EXTernal (X1)
SLOPE: $\quad+$ (Light On)
POSITION: Midrange
HOLD OFF: Fully counterclockwise
TERM: IN- one Megohm

1. Connect the low frequency sine-wave generator to the EXT TRIG IN connector with a 42-inch, 50-ohm BNC cable, a X5, 50-ohm attenuator and a BNC T-connector. Connect the free output of the T-connector to the amplifier unit input with an 18-inch, 50-ohm cable.
2. Set the amplifier unit deflection factor to 0.1 volts per division. Set the sine-wave generator for a six division display at one kilohertz.
3. Check- that all levels of the positive slope may be selected for the sweep starting point as the TRIGGERING LEVEL control is rotated throughout its range.
4. Check- that the time base is not triggered at either end of the LEVEL control rotation.
5. Set the SLOPE to negative (light off).
6. Check- repeat steps three and four for the negetive slope of the waveform.

## Check Internal./External Low Frequency Triggering Sensitivity

## Test Set-Up

Set the 7B90P controls as follows:

TIME/DIV: One millisecond
MAG: Off
MODE: NORMal
COUPLING: AC
SOURCE: EXTernal (X1)
SLOPE: $\quad+$ (Light On)
POSITION: Midrange
HOLD OFF: Fully counterclockwise
TERM: $\quad$ IN- one Megohm

1. Set the amplifier unit VOLTS/DIV to ten millivolts.
2. Set the low-frequency sinewwave generator for 300 Hertz and an output amplitude of approximately 100 millivolts.
3. Reduce the generator output frequency to 30 Hertz.

## NOTE

The 7912AD turns the writing beam off when the time base is set for sweep speeds below 1 millisecond per division.
4. Adjust the LEVEL control for a stable display.
5. Check- for a stable display (TRIG'D light on) wish the coupling set to:
a. AC
b. $A C$ HF REJ
c. DC
(Adjust the LEVEL control as necessary).
6. Change the TRIGGERING SLOPE to negative (light off) and repeat step five.
7. Set the TRIGGERING MODE to PP AUTO and COUPLING to AC.
8. Set the amplifier unit to 0.1 volt/division and the sinemave generator for an output amplitude of 500 millivolts at 500 Hertz.
9. Reduce the generator output frequency to 50 Hertz.
10. Check- for a stable display (TRIG'D light on) at all settings of the LEVEL control with COUPLING set to:
a. AC
b. DC
11. Set the SLOPE to positive (light on), and repeat step ten.
12. Set the amplifier unit deflection factor for 50 millivolts/division and the sine-wave generator for approximately 125 millivolts at 200 Hertz.
13. Set the SLOPE to negative (light off) and Repeat steps ten and eleven.
14. Set the 7B90P TRIGGERING SOURCE to INTernal.
15. Set the amplifier unit to 0.1 Volt/Division. Adjust the generator for a display amplitude of 3 major divisions at 300 Hertz.
16. Set the amplifier unit to 1 Volt/Division and the generator output frequency to 30 Hertz.
17. Adjust the 7B90P LEVEL control for a stable display (TRIG'D lights on).
18. Check - for a stable display with coupling set to:
a. AC
b. $A C$ HF REJ
c. $D C$
(Adjust the LEVEL control as necessary).

## NOTE

The Tektronix FG502 signal generator has a DC offset control. This control must be off (pressed in) for the DC COUPLING checks.
19. Set the SLOPE to negative (light off) and repeat step 18.
20. Set the MODE to PP AUTO and COUPLING to $A C$.
21. Set the low-frequency sine-wave generator and the amplifier unit for two major divisions of vertical deflection at 500 Hertz.
22. Reduce the generator output frequency to 50 Hertz.
23. Check - for a stable display (TRIG'D light on) at all settings of the LEVEL control with COUPLING set to:
a. $A C$
b. $D C$
24. Set the SLOPE to positive (light on) and repeat step 23.
25. Set the sine-wave generator frequency to 200 Hertz and the amplifier unit to 0.1 volts per division.
26. Adjust the generator for five major divisions of vertical deflection.
27. Set the amplifier unit to 1 volt per division.
28. Set the SLOPE to negative and repeat steps 23 ard 24.

## Check Internal/External Mid-Frequency Triggering Sensitivity

## Test Set-Up

Set the 7B90P controls as follows:

TTME/DIV: $\quad 20$ nanoseconds
MAG: Off
MODE: NORMal
COUPLING: AC
SOURCE: EXTernal (X1)
SLOPE: + (Light On)
POSITION: Midrange
HOLD OFF: Fully counterclockwise

1. Disconnect the low-frequency sine-wave generator from the $T$ connector and connect the mid-frequency signal generator to the $T$-connector with a X 550 ohm attenuator.

## NOTE

The Tektronix SG503 requires a special BNC cable for calibrated operation. Refer to Table 6-1 for more information.
2. Set the amplifier unit deflection factor to 20 millivolts per division and the high-frequency generator for a display amplitude of five divisions at 50 megahertz.
3. Check- for a stable display (TRIG'D light on) with the COUPLING set to:
a. AC
b. AC LF REJ
c. DC
(Adjust the LEVEL control as necessary).
4. Set the SLOPE to negative and repeat step three.
5. Set the amplifier deflection factor to 50 millivolts/division and set the mid-frequency signal generator for a 2.5-division display.
6. Set the TRIGGERING MODE to PP AUTO.
7. Check- for a stable display (TRIG'D light on) at all settings of the LEVEL control with COUPLING set to:
a. $A C$
b. $D C$
8. Set the SLOPE to positive and repeat step seven.
9. Set the TRIGGERING MODE to NORM.
10. Set the amplifier unit to 0.1 Volts/Division and adjust the generator for three major divisions of vertical deflection at 50 Megahertz.
11. Set the amplifier unit to one volt per division.
12. Check - for a stable display (TRIG'D light on) with COUPLING set to:
a. AC
b. AC LF REJ
c. $D C$
(Adjust the LEVEL control as necessary).
13. Set the SLOPE to negative (light off) and repeat step 12.
14. Set the mid-frequency signal generator and the amplifier unit deflection factor for a 0.5 division display.
15. Set the TRIGGERING MODE to PP AUTO.
16. Check - for a stable display (TRIG'D light on) at all settings of the LEVEL control with COUPLING set to.
a. $A C$
b. $D C$
17. Set the SLOPE to positive (light on) and repeat step 16 .

## Check Internal/External High Frequency Triggering Sensitivity

Test Set-Up

Set the 7B90P controls as follows:

TIME/DIV: One nanosecond
MAG: Off
MODE: PP AUTO
COUPLING: AC
SOURCE: EXTernal (X1)
SLOPE: + (Light On)
POSITION: Midrange
HOLD OFF: Fully counterclockwise
TERM: IN- one Megohm

1. Disconnect the mid-frequency sine-wave generator and connect the high-frequency sine-wave generator to the T -connector with a X 5 , 50 -ohm attenuator.
2. Set the high-frequency signal generator for display amplitude of 7.5 divisions at 400 megahertz.
3. Check - for a stable display (TRIG'D light on) at all settings of the LEVEL control with COUPLING set to:
a. AC
b. $D C$
4. Set the SLOPE to negative and repeat step three.
5. Set the high-frequency signal generator for a five division display at 400 megahertz.
6. Set the TRIGGERING MODE to NORM and the SLOPE to positive.
7. Check for a stable display (TRIG'D light on) with coupling set to:
a. AC
b. AC LF REJ
c. DC
(Adjust the LEVEL control as necessary).
8. Set the SLOPE to negative (light off) and repeat step seven.
9. Set the high-frequency signal generator and the amplifier unit deflection factor for a 1.5 division display at 400 megahertz.
10. Set the TRIGGERING SOURCE to INTernal.
11. Repeat step seven for both positive and negative SLOPE.
12. Set the 7B90P MODE to PP AUTO.
13. Repeat step seven for both positive and negative SLOPE.

## Check Internal Trigger Jitter

## Test Setwop

Set the 7B90P controls as follows:

| TIME/DIV: | 500 picoseconds |
| :--- | :--- |
| MAG: | Off |
| MODE : | NORMal |
| COUPLING: | AC |
| SOURCE: | INTernal |
| SLOPE: | Positive (light on) |
| POSITION: | Midrange |
| HOLD OFF: | Fully counterclockwise |

1. Adjust the LEVEL control for a stable display (TRIG'D light on)
2. Check- for a stable display with no more than 0.2 major horizontal divisions ( 0.1 nanosecond) of jitter.

## Check Line Triggering

1. Remove all signal connections and reduce the 7912AD Main Intensity.
2. Set the TRTGGERING SOURCE to LINE and the TIME/DIV to one millisecond.
3. Set the LEVEL control to approximately midrange.
4. Check- that the TRIG'D light is on.
5. Check- that the TRIG'D light goes out when the LEVEL control is rotated to both limits.

## Check Sweep Length and Positioning

Test Set-Jp

Set the 7B90P controls as follows:

| TIME/DIV: | One millisecond |
| :--- | :--- |
| MAG: | Off |
| MODE: | PP AUTO |
| COUPLING: | AC |
| SOURCE: | INTernal |
| SLOPE: | Positive (light on) |
| POSITION: | Midrange |
| HOLDOFF: | Fully counterclockwise |
| TERM: | IN- one Megohm |

1. Connect the time-mark generator to the amplifier unit with a 50 ohm BNC cable. Set the time-mark generator for one millisecond markers.
2. Adjust the Main Intensity and Focus for a sharp display.
3. Set the 7B90P LEVEL control for a stable display (TRIG'D light on).
4. Set the amplifier for approximately two divisions of vertical deflection and center the display vertically.
5. Adjust the 7B90P POSITION control to place the second time marker on the first graticule line.
6. Check- that the end of the sweep is beyond 9.2 major graticule divisions, indicating that the sweep is at least 10.2 divisions long.
7. Set the POSITION control fully clockwise and note the position of the start of the sweep.
8. Check- that the start of the sweep is 1.4 divisions to the right of the graticule center $\pm 0.5$ div.
9. Set the POSITION control fully counterclockwise.
10. Check- that the end of the sweep is to the left of the graticule center.

## Check Magnifier Gain

1. Set the time-mark generator for one millisecond markers.
2. Set the 7B90P TIME/DIVISION to one millisecond.
3. Check - for one time mark per division over the center eight divisions.
4. Adjust - the front-panel screwdriver SWP CAL control for one marker per division over the center eight divisions.
5. Set the 7B90P MAG to on (button lit) and set the POSITION control to midrange.
6. Set the time-mark generator for 0.1 millisecond markers.
7. Check- for one marker per horizontal division over the center eight divisions. Adjust the 7B90P POSITION and mainframe Intensity control as necessary.

## Check Sweep Tixing

Test Set $-\mathbf{U p}$

Set the 7B90P controls as follows:

| MAG: | OFF |
| :--- | :--- |
| MODE : | NORMal |
| COUPLING: | AC |
| SOURCE: | INTernal |
| SLOPE: | Positive (light on) |
| POSITION: | Midrange |
| HOLD OFF: | Fully counterclockwise |
| TERM: | IN- one Megohm |

1. Check- the sweep accuracy for all unmagnified TIME/DIV ranges. Check for one marker per division over the center eight divisions. The time-marks should be within 0.12 division for sweep speeds below 100 nanoseconds and within 0.2 division for the faster sweep speeds. Adjust the TRIGGERING LEVEL, POSITION, and Mainframe Intensity controls as necessary.

## NOTE

> The sweep timing tolerances are for an ambient temperature range of +15 degrees centigrade to +35 degrees centigrade. If outside this range, see the Specifications section for allowable tolerances.

## Check Magnified Sweep Timing

1. Center the display horizontally on the graticule. Set the MAG to on (button lit) and the SOURCE to EXTernal.
2. Set the TERM to 50 mohms (button released).
3. Connect the time-mark generator trigger output to the EXT IRIG IN connector with a 50-ohm BNC cable.
4. Check- the sweep accuracy for all unmagnified IIME/DIV ranges. Check for one marker per division over the center eight divisions. (Use one nanosecond markers for 500 picoseconds/division and check for one marker in two divisions.) The time-marks should be within 0.2 division for sweep speeds below 5 nanoseconds and within 0.32 division for faster sweep speeds. Adjust the TRIGGERING LEVEL, POSITION and Mainfrane Intensity controls as necessary.

## Check Holdoff

1. Connect the GATE OUTput of the 7912AD to the vertical input of the test oscilloscope with a 42-inch BNC cable and a X5 50-ohm attenuator.
2. Adjust the test oscilloscope for a triggered display.
3. Check - that the displayed pulse width varys while rotating the 7B90P HOLDOFF control.
4. Repeat step three for all 7B90P TIME/DIV ranges below 200 microseconds and above 5 microseconds. Adjust the test oscilloscope as necessary.

## NOTE

The range of the HOLDOFF control is limited by the 7912AD in sweep speeds between 200 microseconds and 50 microseconds. The 7912AD disables the holdoff circuit in TIME/DIV ranges between 20 microseconds and 5 micro-seconds.

## Calibration Procedure

The following procedure is arranged so that the 7890 P can be calibrated with the least interaction of adjustments and re-connection of equipment. The control settings and test equipment set-up throughout the procedure continue from preceding steps unless otherwise noted. Refer to Fig. 6-1a and $b$ for the location of adjustments. The test set up is summarized at the beginning of each section of the calibration procedure. When performing the complete procedure, this summary provides a check of the instrument settings. Any part of the calibration procedure can be performed by setting up the equipment as shown in the tables.

## NOTE

Control titles that are printed on the 7B90P front panel are capitalized (e.g., POSITION). Internal adjustments and associated equipment controls are initially capitalized only (e.g., Pulse Amplitude).

## Preliminary Procedure for Calibration

1. Remove the 7B90P side panels.
2. Insert the plug-in extender into the mainframe horizontal compartment and plug the 7B90P into the extender.
3. Install the amplifier unit in the mainframe vertical compartment.
4. Turn the mainframe power on and adjust for a visible trace. Allow 20 minutes warm-up time before proceding.

## NOTE

This instrument should be calibrated at an ambient temperature of +20 degrees centigrade to +30 degrees centigrade for best overall accuracy. The performance of the instrument can be checked at any temperature from 0 degrees to 40 degrees Centigrade.


Fig. 6-1a. Test point and adjustment locations horizontal programming board.


Fig. 6-1b. Test point and adjustment locations sweep and trigger boards.

## Check/Adjust MPU Clock Speed

Test Set-Up

Set the 7B90P controls as follows:

TIME/DIV: 100 nanoseconds
MAG: Off
MODE: PP AUYO
COUPLING: AC
SOURCE: INTernal
SLOPE: Positive (light on)
POSITION: Midrange
HOLDOFF: Fully counterclockwise
TERM: IN- one Megohm

1. Remove the $\overline{S T O P}$ strap, P 1526 , on the Horizontal Programming Board (see Fig. 6-1a).
2. Install a X 10 , 50 ohm probe on the input of the amplifier unit. Set the VOLTS/DIV to one volt. If a X 10 , 50 -ohm probe is not available, use a high input impedance 7000-series amplifier unit and probe.
3. Connect the ground wire of the probe the the ground test point on the Horizontal Programming board (see Fig. 6-1a).
4. Connect the probe to TP132 on the Horizontal Programming Board.
5. Press the +SLOPE button once to start the clock, then press it again to reset the SLOPE to positive.
6. Check- that the pulse width (high-time) is between 480 nanoseconds and 520 nanoseconds.
7. Adjust- R022 for a 500 nanosecond pulse width.
8. Move the probe to TP130.
9. Check- that the pulse width is between 480 nanoseconds and 520 nanoseconds.
10. Adjust- R032 for a 500 nanosecond pulse width.
11. Remove the probe and install the $\overline{\text { STOP }}$ strap, P 1420.

## Check/Adjust Trigger Offset Null, Normal Balance and Slope Balance

 Test Set-OpSet the 7B90P controls as follows:

| TIME/DIV: | 20 microseonds |
| :--- | :--- |
| MAG: | Off |
| SLOPE: | Positive (light on) |
| MODE: | PP AUTO |
| COUPLING: | AC |
| SOURCE: | INTernal |
| POSITION: | Midrange |
| HOLD OFF: | Fully counterclockwise |
| TERM: | IN - one Megohm |

1. Connect the low frequency sine-wave generator to the amplifier unit with a $50-$ ohm BNC cable.
2. Set the amplifier unit to 0.1 volts per division.
3. Set the mainframe Intensity and Focus for a sharp display.
4. Adjust the sinewave generator for a display amplitude of three major divisions at 50 kilohertz.
5. Center the display vertically.
6. Set the amplifier-unit to one volt per division.
7. Set the 7Bgop TRIGGERING LEVEL to midrange.
8. Connect the negative lead of the Digital Voltmeter to the ground testpoint, TP030 on the sweep board (see Fig. 6-1b).
9. Measure the voltage at pins 5 through 12 of $\mathbf{U 1 0 6}$ on the trigger board.
10. Adjust the LEVEL control until pins 6 through 12 are all low $(0 \mathrm{~V}$ to $+0.8 \mathrm{~V})$ and pin 5 is high $(+2.4 \mathrm{~V}$ to $+5.5 \mathrm{~V})$.
11. Check- for a stable display with the TRIG'D light on.
12. Adjust- Offset Null, R243, for a stable display.
13. Set the SLOPE to negative (light off).
14. Check- for a stable display.
15. Adjust- Slope Balance, R240 for a stable display.
16. Interaction- Repeat adjustment of offset null (R243) and Slope Balance (R240) until a stable display is obtained while changing TRIGGERTNG SLOPE .
17. Set the TRIGGERING mode to NORMal.
18. Check- for a stable display.
19. Adjust- Normal Balance, R134 for a stable display.
20. Remove the test leads of the Digital Voltmeter.
21. To check the adjustments, perform the check procedure for the triggering functions. If any of the check procedures cannot be met, repeat the adjustments in this section.

## Check/Adjust External X1 and X10 Balance

## Test Set-Up

TTME/DIV: $\quad 20$ microseconds
MAG: Off
SLOPE: Positive (light on)
MODE: PP AUTO
COUPLING: AC
SOURCE: INTernal
POSITION: Midrange
HOLD OFF: Fully counterclockwise
TERM: $\quad$ IN- one megohm

1. Connect the low-frequency sine-wave generator to the EXT TRIG IN connector and the input of the amplifier unit through a BNC T-connector.
2. Set the amplifier unit to 0.1 volts per division.
3. Adjust the sine wave generator for a display amplitude of two major divisions at 50 kilohertz.
4. Center the display vertically.
5. Adjust the TRIGGERING LEVEL as described in step ten of the preceding section.
6. Switch the 7B90P SOURCE to EXT (X1) and the MODE to NORMal.
7. Check that the trigger point does not change when switching between INT and EXT.
8. Adjust- R524, Ext X1 Offset, for minimum shift in the trigger point while switching between INT and EXT.
9. Set the amplifier unit to 0.5 volts/division.
10. Adjust the sine-wave generator for a two division display.
11. Set the 7B90P SOURCE to EXT X10.
12. Check- that the trigger point does not change when switching between INT and EXT X10.
13. Adjust- R532, EXI X10 Offset, for minimum shift in the trigger point while switching between INT and EXT X10.
14. Disconnect the low-frequency sine-wave generator.

## Check/Adjust Sweep and Magnifier Calibration

## Test Set-Up

Set the 7B90P controls as follows:

TIME/DIV: One millisecond

MAG: Off
SLOPE: Positive (light on)
MODE: PP AUTO
COUPLING: AC
SOURCE: INTernal
POSITION: Midrange
HOLD OFF: Fully counterclockwise

1. Connect the time-mark generator to the amplifier unit input.
2. Set the time-mark generator for one millisecond markers.
3. Adjust the mainframe Intensity and Focus for a sharp display.
4. Check- for one marker per division over the center eight divisions. (Adjust the POSITION control as necessary.)
5. Adjust- the front panel SWP CAL for one marker per division over the center eight divisions.
6. Set the MAG to on (light on).
7. Set the time-mark generator for 0.1 millisecond markers.
8. Check- for one marker per division over the center eight divisions.
9. Adjust- R232, Magnifier Calibrate, for one marker per division over the center eight divisions.

## Check/Adjust Balance

Test Set-Up

Set the 7B90P controls as follows:

TIME/DIV: One millisecond
MAG: On
SLOPE: Positive (light on)
MODE: PP AUTO
COUPLING: AC

SOURCE: INTernal
POSITION: Midrange
HOLD OFF: Fully counterclockwise

1. Set the time-mark generator to one millisecond markers.
2. Adjust the POSITION control to place the first marker on the center graticule line.
3. Set the 7B90P MAG off.
4. Check- that the first time marker is positioned at the center graticule line.
5. Adjust- R238, Balance, for minimum shift in the marker position while switching the MAG on and off.

## Check/Adjust Sweep Timing

## Test Set-Tp

Set the 7B90P controls as follows:

TIME/DIV: $\quad 20$ nanoseconds
MAG: Off
SLOPE: Positive (light on)
MODE: PP AUTO
COUPLING: AC
SOURCE: INTernal
POSITION: Midrange
HOLD OFF: Fully counterclockwise
TERM: $\quad$ OUT $=50$ ohm

1. Set the time-mark generator to 20 nanosecond markers.
2. Adjust the mainframe Intensity and Focus for a sharp display.
3. Check- for one marker per division over the center eight divisions. (Adjust the POSI'IION control as necessary.)
4. Adjust C500, 20 ns . Adj., for one marker per division over the center eight divisions.
5. Set the 7B90P TIME/DIV to 10 nanoseconds and the time-mark generator for 10 nanosecond time markers.
6. Check- for one marker per division over the center eight divisions.
7. Set the 7B90P TIME/DIV to 50 nanoseconds and the time-mark generator for 50 nanosecond time markers.
8. Check- for one marker per division over the center eight divisions.
9. If the checks in steps five or seven are not met, adjust C500 for the best overall accuracy betweer 10, 20 and 50 nanoseconds. Then repeat steps one through eight.
10. Connect the trigger output of the time-mark generator to the EXT TRIG IN connector of the 7B90P.
11. Set the TRIGGERING SOURCE to EXTernal and adjust the LEVEL control for a stable display.
12. Set the time-mark generator for one nanosecond markers.
13. Set the 7B90P TIME/DIV to 500 picoseconds.
14. Adjust the mainframe Intensity and Focus for a sharp display.

## NOTE

Because of bandwidth limitations the tine-mark amplitude may be attenuated.
15. Check- for one marker per two divisions over the center eight divisions.
16. Adjust- R240, 500 ps. Adj., for one marker per two divisions over the center eight divisions.
17. Check the sweep timing on all TIME/DIV ranges as discussed in the Sweep Timing section of the Performance Check Procedure.
18. Decrease the mainframe intensity.

## Check/Adjust Sweep Position

## Test Set-Up

Set the 7B90P controls as follows:
TIME/DIV: One millisecond
MAG: Off
SLOPE: Positive (light on)
MODE: PP AUTO
COUPLTNG: AC
SOURCE: INT
POSITION: Midrange
HOLD OFF: Fully counterclockwise
TERM: OU'- 50 ohms

1. Set the time-mark generator for one millisecond markers.
2. Adjust the POSITION control so the first marker is on the center graticule mark.
3. Press the MAG button and set the first time mark as close as possible to the center graticule mark.
4. Connect the negative lead of the Digital Voltmeter to the ground test point, TP030 on the sweep board (see fig. 6-1b).
5. Measure the voltage on pins 4 through 13 of U330. Voltages in the range of 0 volts to 0.8 volts represent a logical zero. Voltages in the range of 2.0 volts to 5.0 volts represent a logical one. Record the binary word on pins 4 through 13, where pin 13 corresponds to the least significant bit and pin 4 corresponds to the most significant bit.
6. Add $800{ }_{10}\left(1100100000_{2}\right)$ to this binary word and record the sum.
7. Set the 7B90P MAG to off (light off).
8. Adjust the POSITTON control so the eleventh marker is on the center graticule mark.
9. Check the positioning of the mark by turning the MAG on.
10. Measure the binary word on U 330 as outlined in step five.
11. Check- that this binary word is equal to the sum computed in step six.
12. Adjust- R342, Position Gain, until the measured binary word is equal to the sum computed in step six.
13. Disconnect all test equipment.

This completes the 7B90P calibration procedure.

# REPLACEABLE <br> ELECTRICAL PARTS 

## PARTS ORDERING INFORMATION

Replacement parts are available from or through your local Tektronix, Inc. Field Office or representative.

Changes to Tektronix instruments are sometimes made to accommodate improved components as they become available, and to give you the benefit of the latest circuit improvements developed in our engineering department. It is therefore important, when ordering parts, to include the following information in your order: Part nurnber, instrument type or number, serial number, and modification number if applicable

If a part you have ordered has been replaced with a new or improved part, your local Tektronix, Inc. Field Office or representative will contact you concerning any change in part number.

Change information, if any, is located at the rear of this manual.

## SPECIAL NOTES AND SYMBOLS

X000 Part first added at this serial number
00X Part removed after this serial number

ITEM NAME
In the Parts List, an Item Name is separated from the description by a colon (:). Because of space limitations, an Item Name may sometimes appear as incomplete. For further Item Name identification, the U.S. Federal Cataloging Handbook H6-1 can be utilized where possible

ABBREVIATIONS

| ACTR | ACTUATOR | PLSTC | PLASTIC |
| :--- | :--- | :--- | :--- |
| ASSY | ASSEMBLY | QTZ | QUARTZ |
| CAP | CAPACITOR | RECP | RECEPYACLE |
| CER | CERAMIC | RES | RESISTOR |
| CKT | CIRCUIT | RF | RADIOFREQUENCY |
| COMP | COMPOSITION | SEL | SELECTED |
| CONN | CONNECTOR | SEMICOND | SEMICONDUCTOR |
| ELCTLT | ELECTROLYTIC | SENS | SENSITIVE |
| ELEC | ELECTRICAL | VAR | VARIABLE |
| INCANO | INCANDESCENT | WW | WIREWOUND |
| LED | LIGHT EMITTING DIODE | XFMR | TRANSFORMER |
| NONWIR | NON WIREWOUND | XTAL | CRYSTAL |

CROSS INDEX - MFR. CODE NUMBER TO MANUFACTURER

| Mfr. Code | Manufacturer | Address | City, State, Zip |
| :---: | :---: | :---: | :---: |
| 00853 | SANGAMO ELECTRIC CO., S. CAROLINA DIV. | P.O. BOX 128 | PICKENS, SC 29671 |
| 01121 | ALLEN-BRADLEY COMPANY | 1201 2ND STREET SOUTH | MIL.WAUKEE, WI 53204 |
| 01295 | TEXAS INSTRUMENTS, INC. |  |  |
|  | SEMICONDUCTOR GROUP | P.O. BOX 5012 | DA LLAS, TX 75222 |
| 02735 | RCA CORPORATION, SOLID STATE DIVISION | ROUTE 202 | SO VERVILLE, NY 08876 |
| 04222 | AVX CERAMICS, DIVISION OF AVX CORP. | POBOX 867 | MYRTLE BEACH, SC 29577 |
| 04713 | MOTOROLA, INC.. SEMICONDUCTOR PROD. DIV. | 5005 E MCDOWELL RD,PO BOX 20923 | PHOENIX, AZ 85036 |
| 07263 | FAIRCHILD SEMICONDUCTOR, A DIV. OF FAIRCHILD CAMERA AND INSTRUMENT CORP. |  |  |
| 09023 | CORNELL-DUBILIER ELECTRONIC DIVISION. | 464 ELLIS STREE | MOUNTAIN VIEW, CA 94042 |
|  | FEDERAL PACIFIC ELECTRIC CO. | 2652 DALRYMPLE ST. | SANFORD, NC 27330 |
| 12697 | CLAROSTAT MFG. CO., INC. | LOWER WASHINGTON STREET | DOVER. NH 03820 |
| 12954 | SIEMENS CORPORATION, COMPONENTS GROUP | 8700 E THOMAS RD, P O BOX 1390 | SCOTTSDALE, AZ 85252 |
| 13511 | AMPHENOL CARDRE DIV., BUNKER RAMO CORP. |  | LOS GATOS, CA 95030 |
| 14433 | ITT SEMICONDUCTORS | 3301 ELECTRONICS WAY |  |
|  |  | P O BOX 3049 | WEST PALM BEACH, FL 33402 |
| 14552 | MICRO SEMICONDUCTOR CORP. | 2830 E FAIRVIEW ST. | SANTA ANA, CA 92704 |
| 17856 | SILICONIX, INC. | 2201 LaURELWOOD DRIVE | SANTA CLARA, CA 95054 |
| 18324 | SIGNETICS CORP. | 811 E . ARQUES | SUINNYVALE, CA 94086 |
| 22229 | SOLITRON DEVICES, INC., |  | , |
|  | SEMICONDUCTOR GROUP | 8808 balboa avenue | SAN DIEGO OPERS, CA 92123 |
| 22526 | BERG ELECTRONICS, INC. | YOUK EXPRESSWAY | NEW CUMBERLAND, PA 17070 |
| 27014 32293 | NATIONAL SEMICONDUCTOR CORP. | 2900 SEMICONDUCTOR DR. | SANTA CLARA, CA 95051 |
| 32293 32997 | INTERSIL, INC. | 10900 N. TANTAU AVE. | CUPPERTINO, CA 95014 |
| 32997 34371 | BOURNS, INC., TAIMPOT PRODUCTS DIV. HARRIS SEMICONDUCTOR, DIV OF | 1200 COLUMBIA AVE. | RIVERSIDE, CA 92507 |
|  | HARRIS CORPORATION | P. O. BOX 883 | MEL.BOURNE, FL 32901 |
| 50434 | HEWLETT-PACKARD COMPANY | 640 PAGE MILL ROAD | PALO ALTO, CA 94304 |
| 51984 | NEC AMERICA INC. RADIO AND |  |  |
|  | TRANSMISSION DIV. | 2990 TELESTAR CT. SUITE 212 | FALLS CHURCH, VA 22042 |
| 53184 | XCITON CORPORATION | 5 HEMLOCK STREET | LAT-AM, NY 12110 |
| 54473 | MATSUSHITA ELECTRIC, CORP. OF AMERICA | 1 PANASONIC WAY | SECAUCUS. NJ 07094 |
| 55680 | NICHICON/AMERICA/CORP. | 6435 N PROESEL AVENUE |  |
| 56289 | SPRAGUE El.ectric co. | 87 MARSHALL ST. | NORTH ADAMS, MA 01247 |
| 57668 | R-OHM CORP. | 16931 MILLILKEN AVE. |  |
| 59660 | TUSONIXINC. | 2155 N FORBES BLVD | TUCSON, AZ 85705 |
| 71590 | CENTRALAB EIECTRONICS, DIV. OF |  |  |
|  | GLOBE-UNION, INC. | POBOX 858 | FORT DODGE, IA 50501 |
| 72982 | ERIE TECHNOLOGICAL PRODUCTS, INC. | 644 W .12 TH ST. | ERIE: PA 16512 |
| 73138 | BECKMAN INSTRUMENTS, INC., HELIPOT DIV. | 2500 HARBOR BLVD. | FUL-ERTON, CA 92634 |
| $74970$ | JOHNSON. E. F., CO. | 29910 TH AVE. S. W. | WASECA, MN 56093 |
| 75042 | TRW ELECTRONIC COMPONENTS, IRC FIXED RESISTORS, PHILADELPHIA DIVISION | 401 N. BROAD ST. | PHIL.ADELPHIA, PA 19108 |
| 78488 | Stackpole carbon co. |  | ST. VARYS, PA 15857 |
| 80009 | TEKTRONIX, INC. | P o box 500 | BEAVERTON OR 97077 |
| 91637 | DALE ELECTRONICS, INC. | P. O. BOX 609 | COLUMBUS NE 68601 |
| 96733 | SAN FERNANDO ELECTRIC MFG Co | 1501 FIRST ST | SAN FERNANDO, CA 91341 |


|  | Tektronix | Serial/Model No. |  | Name \& Description | Mfr Code | Mfr Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Component No. | Part No. | Eff | Dscont |  |  |  |
| ASSEMBLIES |  |  |  |  |  |  |
| A10 | 670-5038-00 | B010100 | B010156 | CKT BOARD ASSY:TRIGGER | 80009 |  |
| A10 | 670-5038-01 | B010157 | B019999 | CKT BOARD ASSY:TRIGGER | 80009 | 670-5038-00 |
| A10 | 670-5038-02 | B020000 | B030979 | CKT BOARD ASSY:TRIGGER | 80009 | 670-5038-02 |
| A10 | 670-5038-03 | B030980 |  | CKT BOARD ASSY:TRIGGER | 80009 | 670-5038-03 |
| A20 | 670-5037-00 | B010100 | B020454 | CKT BOARD ASSY:SWEEP | 80009 | 670-5037-00 |
| A20 | 670-5037-01 | B020455 | B031149 | CKT BOARD ASSY:SWEEP | 80009 | 670-5037-01 |
| A20 | 670-5037-02 | B031150 |  | CKT BOARD ASSY:SWEEP | 80009 | 670-5037-02 |
| A30 | 670-5041-00 |  |  | CKT BOARD ASSY:SWITCH | 80009 | 670-5041-00 |
| A40 | 670-5040-00 |  |  | CKT BOARD ASSY:LED | 80009 | 670-5040-00 |
| A50 | 670-5039-00 | B010100 | B030586 | CKT BOARD ASSY:HORIZONTAL PROGRAMMING | 80009 | 670-5039-00 |
| A50 | 670-5039-01 | B030587 | B030761 | CKT BOARD ASSY:HORIZONTAL PROGRAMMING | 80009 |  |
| A50 | 670-5039-02 | B030762 | B030837 | CKT BOARD ASSY:HORIZONTAL PROGRAMMING | 80009 | 670-5039-02 |
| A50 | 670-5039-03 | B030838 | B030844 | CKT BOARD ASSY:HORIZONTAL PROGRAMMING | 80009 | 670-5039-03 |
| A50 | 670-5039-04 | B030845 | B031149 | CKT BOARD ASSY:HORIZONTAL PROGRAMMING | 80009 | 670-5039-04 |
| A50 | 670-5039-05 | B031150 |  | CKT BOARD ASSY:HORIZONTAL PROGRAMMING | 80009 | 670-5039-05 |
| A51 | 670-7549-00 | B030762 | B030837 | CKT BOARD ASSY:MEMORY ADAPTEZ | 80009 | 670-7549-00 |

A10 TRIGGER

| A10 | 670-5038-00 | B010100 | B010156 | CKT BOARD ASSY:TRIGGER | 80009 |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A10 | 670-5038-01 | B010157 | B019999 | CKT BOARD ASSY:TRIGGER | $80009$ | 670-5038-01 |
| A10 | 670-5038-02 | B020000 | B030979 | CKT BOARD ASSY:TRIGGER | 80009 | 670-5038-02 |
| A10 | 670-5038-03 | B030980 |  | CKT BOARD ASSY:TRIGGER | 80009 | 670-5038-03 |
| A10C020 | 283-0010-00 |  |  | CAP.,FXD,CER DI:0.05UF, +100-20\%,50V | 56289 |  |
| A10C022 | 283-0111-00 |  |  | CAP.,FXD,CEA DI:0.1UF,20\%,50V | 56289 96733 | 1C10Z5U5032050B R2632 |
| A10C024 | 283-0220-00 |  |  | CAP.,FXD,CER DI:0.01UF, $20 \%, 50 \mathrm{~V}$ | 72982 | 8121 N075X7R0103M |
| A10C040 | 283-0221-00 |  |  | CAP.,FXD, CER DI:0.47UF,20\%,50V | 72982 |  |
| A10C050 | 290-0580-00 |  |  | CAP.,FXD,ELCTLT:0.27UF,20\%,50V | 56289 | $196 \mathrm{D} 274 \times 0050 \mathrm{HA1}$ |
| A10C106 | 283-0728-00 |  |  | CAP.,FXD,MICA D: $120 \mathrm{PF}, 1 \%, 500 \mathrm{~V}$ | 00853 | D155F121F0 |
| A10C110 | 283-0597-00 |  |  | CAP.,FXD,MICA D: $470 \mathrm{PF}, 10 \%, 300 \mathrm{~V}$ | 00853 | D155F471K0 |
| A10C112 | 283-0594-00 |  |  | CAP.,FXD,MICA D:0.001UF, $1 \%, 100 \mathrm{~V}$ | 00853 | D151F102F0 |
| A10C114 | 283-0698-00 |  |  | CAP.,FXD,MICA D:390PF, $1 \%, 500 \mathrm{~V}$ | 09023 | CD15FD391F03 |
| A10C116 | 283-0641-00 |  |  | CAP.,FXD,MICA D:180PF, $1 \%, 100 \mathrm{~V}$ | 00853 |  |
| A10C130 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A10C132 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A10C137 | 283-0605-00 |  |  | CAP.,FXD,MICA D: 678 PF, $1 \%, 300 \mathrm{~V}$ | 00853 | D153F6780F0 |
| A10C140 | 290-0580-00 |  |  | CAP.,FXD, ELCTLT:0.27UF, $20 \%, 50 \mathrm{~V}$ | 56289 | 196D274X0050HA1 |
| A10C146 | 283-0114-00 |  |  | CAP.,FXD,CER DI:0.0015UF,5\%,200V | 59660 | 805534 Y 5 DO 152 J |
| A10C148 | 283-0114-00 |  |  | CAP.,FXD,CER DI:0.0015UF,5\%,200V | 59660 | 805534Y500152J |
| A10C150 | 283-0220-00 |  |  | CAP.,FXD,CER DI:0.01UF,20\%,50V | 72982 | 8121N075X7R0103M |
| A10C152 | 283-0220-00 |  |  | CAP.,FXD,CER DI:0.01UF, $20 \%, 50 \mathrm{~V}$ | 72982 | 8121N075X7R0103M |
| A10C158 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A10C214 | 290-0745-00 |  |  | CAP.,FXD,ELCTLT:22UF, $+50-10 \%, 25 \mathrm{~V}$ | 54473 | ECE-A25V22i. |
| A10C220 | 283-0220-00 |  |  | CAP.,FXD, CER DI:0.01UF, $20 \%, 50 \mathrm{~V}$ | 72982 | 8121N075X7R0103M |
| A10C222 | 283-0220-00 |  |  | CAP.,FXD,CER DI:0.01UF,20\%,50V | 72982 |  |
| A10C224 | 283-0220-00 |  |  | CAP.,FXD,CER DI:0.01UF, $20 \%, 50 \mathrm{~V}$ | 72982 | 8121N075X7R0103M |
| A10C230 | 283-0111-00 |  |  | CAP.,FXD, CER DI:0.1UF, $20 \%, 50 \mathrm{~V}$ | 96733 | R2632 |
| A10C236 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A10C240 | 283-0111-00 |  |  | CAP.,FXD,CER DI: 0.1 UF, $20 \%, 50 \mathrm{~V}$ | 96733 | R2632 |
| A10C250 | 283-0600-00 |  |  | CAP.,FXD,MICA D: 43 PF, $5 \%, 500 \mathrm{~V}$ | 00853 | D105E430,0 |
| A10C308 | 290-0776-00 |  |  | CAP. FXD,ELCTLT:22UF, +50-10\%,10V | 55680 | ULA1A220TEA |


| Component No. |  | Serial/Model No. |  | Name \& Description | Mfr | Mfr Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Part No. | Eff | Dscont |  | Code |  |
| A10C312 | 283-0221-00 |  |  | CAP.,FXD,CER DI:0.47UF, $20 \%, 50 \mathrm{~V}$ | 72982 | 8131N087X7R0474M |
| A10C320 | 283-0642-00 |  |  | CAP.,FXD,MICA D:33PF, $+1-0.5 \mathrm{PF}, 300 \mathrm{~V}$ | 00853 | D10-5E330G |
| A10C322 | 290-0776-00 |  |  | CAP.,FXD, ELCTLT: $22 \mathrm{UF},+50-10 \%, 10 \mathrm{~V}$ | 55680 | ULA1A220TEA |
| A10C324 | 283-0597-00 |  |  | CAP.,FXD,MICA D:470PF, $10 \%, 300 \mathrm{~V}$ | 00853 | D155F471K0 |
| A10C326 | 283-0107-00 |  |  | CAP.,FXD,CER DI:51PF,5\%,200V | 96733 | R3017 |
| A10C336 | 283-0597-00 |  |  | CAP.,FXD,MICA D:470PF, $10 \%, 300 \mathrm{~V}$ | 00853 | D155F471K0 |
| A10C344 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A10C346 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A10C348 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A10C350 | 283-0220-00 |  |  | CAP.,FXD,CER DI:0.01UF,20\%,50V | 72982 | 8121N075X7R0103M |
| A10C360 | 283-0644-00 | B010100 | B010156 | CAP.,FXD,MICA D:150PF, $1 \%, 500 \mathrm{~V}$ | 00853 | D155F151F0 |
| A10C362 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A10C400 | 290-0745-00 |  |  | CAP.,FXD,ELCTLT:22UF, $+50-10 \%, 25 \mathrm{~V}$ | 54473 | ECE-A25V22L |
| A10C422 | 283-0221-00 |  |  | CAP.,FXD,CER DI:0.47UF, $20 \%, 50 \mathrm{~V}$ | 72982 | 8131N087X7R0474M |
| A10C424 | 283-0220-00 |  |  | CAP,,FXD,CER DI:0.01UF,20\%,50V | 72982 | 8121N075X7R0103M |
| A10C426 | 283-0220-00 |  |  | CAP,,FXD,CER DI:0.01UF, 20\%,50V | 72982 | 8121N075X7R0103M |
| A10C430 | 283-0107-00 |  |  | CAP.,FXD,CER DI:51PF, $5 \%, 200 \mathrm{~V}$ | 96733 | R3017 |
| A10C436 | 281-0795-00 |  |  | CAP.,FXD,CER DI:9.3PF, $0.1 \%, 500 \mathrm{~V}$ | 59660 | 374018 COH 0939 B |
| A10C438 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A10C440 | 283-0642-00 |  |  | CAP.,FXD,MICA D:33PF, $+1.0 .5 \mathrm{PF}, 300 \mathrm{~V}$ | 00853 | D10-5E330G |
| A10C442 | 283-0111-00 |  |  | CAP.,FXD, CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A10C448 | 283-0616-00 |  |  | CAP.,FXD,MICA D:75PF.5\%,500V | 00853 | D155E750J0 |
| A10C460 | 283-0644-00 |  |  | CAP,,FXD,MICA D:150PF, $1 \%, 500 \mathrm{~V}$ | 00853 | D155F151F0 |
| A10C462 | 283-0060-00 |  |  | CAP.,FXD,CER DI:100PF,5\%,200V | 59660 | 855-535U2J101J |
| A10C540 | 283-0693-00 |  |  | CAP.,FXD,MICA D:1730PF, $1 \%, 500 \mathrm{~V}$ | 00853 | D19-5F1731F0 |
| A10CR010 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A10CR022 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A10CR050 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A10CR140 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A10CR330 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A10CR332 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A10CR334 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A10CR336 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A10CR340 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A10CR342 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A10CR344 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A10CR346 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A10CR400 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A10CR402 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A10CR460 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON, $30 \mathrm{~V}, 150 \mathrm{MA}$ | 01295 | 1N4152R |
| A10CR462 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A10CR464 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A10CR466 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A10DS348 | 150-1036-00 |  |  | LAMP,LFD:RED, 3.0V,40MA | 01295 | TIL 209A |
| A10DS440 | 150-1036-00 |  |  | LAMP,LED: RED, $3.0 \mathrm{~V}, 40 \mathrm{MA}$ | 01295 | TIL 209A |
| A10J000 | 131-1003-00 |  |  | CONN,RCPT,ELEC:CKT BD MT, 3 PRONG | 80009 | 131-1003-00 |
| A10J212 | 131-1003-00 |  |  | CONN,RCPT,ELEC:CKT BD MT,3 PRONG | 80009 | 131-1003-00 |
| A10J264 | 131-1003-00 |  |  | CONN,RCPT,ELEC:CKT BD MT, 3 PRONG | 80009 | 131-1003-00 |
| A10J436 | 131-1003-00 |  |  | CONN,RCPT,ELEC:CKT BD MT. 3 PRONG | 80009 | 131-1003-00 |


| Component No. | Tektronix Part No. | Serial/ Eff | del No. Dscont | Name \& Descriptior | Mfr Code | Mfr Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A10LR230 | 108-0328-00 |  |  | COIL,RF:0.3UH | 80009 | 108-0328-00 |
| A10LR232 | 108-0328-00 |  |  | COIL,RF:0.3UH | 80009 | 108-0328-00 |
| A10LR310 | 108-0543-00 |  |  | COIL,RF:FIXED,1.1UH | 80009 | 108-0543-00 |
| A10LR314 | 108-0543-00 |  |  | COIL,RF:FIXED,1.1UH | 80009 | 108-0543-00 |
| A10LR320 | 108-0543-00 |  |  | COIL,RF:FIXED,1.1UH | 80009 | 108-0543-00 |
| A10LR410 | 108-0543-00 |  |  | COIL,RF:FIXED,1.1UH | 80009 | 108-0543-00 |
| A10P550 | 131-0608-00 |  |  | TERMINAL,PIN: $0.365 \mathrm{~L} \times 0.025 \mathrm{PH}$ BR 7 I GOLD | 22526 | 48283-036 |
| A10Q128 | 151.0220-00 |  |  | TRANSISTOR:SILICON, PNP | 07263 | S036228 |
| A10Q134 | 151-0220-00 |  |  | TRANSISTOR:SILICON, PNP | 07263 | S036228 |
| A10Q260 | 151-0325-00 | B010100 | B020414 | TRANSISTOR:SILICON,PNP,SEL FROM 2 N4258 | 80009 | 151-0325-00 |
| A10Q260 | 151-0221-00 | B020415 |  | TRANSISTOR:SILICON, PNP | 04713 | SPS246 |
| A10Q262 | 151-0325-00 | B010100 | B020414 | TRANSISTOR:SILICON,PNP,SEL FROM 2N4258 | 80009 | 151-0325-00 |
| A10Q262 | 151-0221-00 | B020415 |  | TRANSISTOR:SILICON,PNP | 04713 | SPS246 |
| A10Q264 | 151-0325-00 | 8010100 | B020414 | TRANSISTOR:SILICON,PNP,SEL FROM 2N4258 | 80009 | 151-0325-00 |
| A10Q264 | 151-0221-00 | B020415 | B030594 | TRANSISTOR:SILICON, PNP | 04713 | SPS246 |
| A10Q264 | 151-0342-00 | B030595 |  | TRANSISTOR:SILICON,PNP | 07263 | S035928 |
| A100266 | 151-0223-00 |  |  | TRANSISTOR:SILICON,NPN | 04713 | SPS8026 |
| A10Q326 | 151-1042-00 |  |  | SEMICOND DVC SE:MATCHED PAIR FET | 22229 | S2089 |
| A10Q326 | --- --- |  |  | (FURNISHED AS A UNIT WITH A10Q3ع.7) |  |  |
| A10Q327 | 151-1042-00 |  |  | SEMICOND DVC SE:MATCHED PAIR FET | 22229 | S2089 |
| A10Q327 | --...- ---- |  |  | (FURNISHED AS A UNIT WITH A10Q3E6) |  |  |
| A10Q439 | 151-1042-00 |  |  | SEMICOND DVC SE:MATCHED PAIR FET | 22229 | S2089 |
| A10Q439 | --... -...-- |  |  | (FURNISHED AS A UNIT WITH A10Q4C0) |  |  |
| A10Q440 | 151-1042-00 |  |  | SEMICOND DVC SE:MATCHED PAIR FET | 22229 | S2089 |
| A10Q440 | ----- .-...- |  |  | (FURNISHED AS A UNIT WITH A10Q439) |  |  |
| A10Q530 | 151-0220-00 |  |  | TRANSISTOR:SILICON,PNP | 07263 | S036228 |
| A100532 | 151-0220-00 |  |  | TRANSISTOR:SILICON,PNP | 07263 | S036228 |
| A10R002 | 315-0510-00 |  |  | RES.,FXD,CMPSN: 51 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB5105 |
| A10R010 | 315-0622-00 |  |  | RES.,FXD,CMPSN:6.2K OHM $, 5 \%, 0.25 \mathrm{~V}$ | 01121 | CB6225 |
| A10R040 | 315-0104-00 |  |  | RES.,FXD,CMPSN:100K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1045 |
| A10R042 | 315-0104-00 |  |  | RES.,FXD,CMPSN:100K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1045 |
| A10R044 | 321-0443-00 |  |  | RES.,FXD,FILM:402K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | CMF55116G40202F |
| A10R050 | 315-0205-00 |  |  | RES.,FXD,CMPSN:2M OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2055 |
| A10R052 | 321-0443-00 |  |  | RES.,FXD,FILM:402K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | CMF55116G40202F |
| A 10 R 100 | 321-0277-00 |  |  | RES.,FXD,FILM: 7.5 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G75000F |
| A10R102 | 315-0752-00 |  |  | RES.,FXD,CMPSN:7.5K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB7525 |
| A10R103 | 321-0289-00 |  |  | RES.,FXD,FILM:10K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G10001F |
| A10R106 | 321-0306-00 |  |  | RES.,FXD,FILM: 15 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G15001F |
| A10R110 | 315-0202-00 |  |  | RES.,FXD,CMPSN:2K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2025 |
| A10R112 | 315-0104-00 |  |  | RES.,FXD,CMPSN:100K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1045 |
| A10R114 | 315-0203-00 |  |  | RES.,FXD,CMPSN:20K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2035 |
| A10R116 | 315-0102-00 |  |  | RES.,FXD,CMPSN:1K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A10R120 | 315-0203-00 |  |  | RES.,FXD,CMPSN:20K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2035 |
| Al0R122 | 321-0431-00 |  |  | RES.,FXD,FILM:301K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G30102F |
| A10R124 | 315-0131-00 |  |  | RES.,FXD,CMPSN: 130 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1315 |
| A10R:125 | 315-0332-00 |  |  | RES.,FXD.CMPSN:3.3K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB3325 |
| A10R126 | 315-0512-00 |  |  | RES.,FXD,CMPSN:5.1K OHM, $5 \%, 0.25 \mathrm{~V}$ | 01121 | CB5125 |
| A10R127 | 315-0104-00 |  |  | RES.,FXD,CMPSN:100K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1045 |
| A10R128 | 315-0151-00 |  |  | RES.,FXD,CMPSN: 150 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1515 |
| A10R130 | 315-0124-00 |  |  | RES.,FXD,CMPSN:120K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1245 |
| A10R132 | 321-0274-00 |  |  | RES.,FXD,FILM: 6.98 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | CMF55116G69800F |


| Component No. | Tektronix <br> Part No. | Serial/Model No. <br> Eff Dscont | Name \& Description | Mfr Code | Mfr Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: |
| A10R134 | 311-1268-00 |  | RES.,VAR,NONWIR: 10 K OHM, $10 \%, 0.50 \mathrm{~W}$ | 32997 | 3329P-L58-103 |
| A10R135 | 315-0122-00 |  | RES.,FXD,CMFSN: 1.2 K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1225 |
| A10R136 | 315-0152-00 |  | RES.,FXD,CMPSN: 1.5 K OHM $.5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1525 |
| A10R137 | 315-0392-00 |  | RES.,FXD,CMPSN:3.9K OHM $.5 \%, 0.25 \mathrm{~W}$ | 01121 | CB3925 |
| A10R140 | 315-0205-00 |  | RES.,FXD,CMPSN: 2 M OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2055 |
| A10R142 | 315-0182-00 |  | RES.,FXD,CMPSN: 1.8 K OH: $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1825 |
| A10R144 | 315-0514-00 |  | RES.,FXD,CMPSN: 510 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB5145 |
| A10R146 | 315-0514-00 |  | RES.,FXD,CMPSN:510K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB5145 |
| A10R148 | 315-0622-00 |  | RES.,FXD,CMPSN: 6.2 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB6225 |
| A10R158 | 323-0155-00 |  | RES.,FXD,FILM 402 OHM, $1 \%, 0.50 \mathrm{~W}$ | 75042 | СеСТ0-4020F |
| A10R210 | 315-0202-00 |  | RES.,FXD,CMPSN:2K OHM $5 \%$ \% , 0.25 W | 01121 | CB2025 |
| A10R212 | 315-0203-00 |  | RES.,FXD,CMPSN:20K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2035 |
| A10R213 | 315-0560-00 |  | RES.,FXD,CMPSN: 56 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB5605 |
| A10R214 | 315-0102-00 |  | RES.,FXD,CMPSN: 1 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A10R215 | 315-0752-00 |  | RES.,FXD,CMPSN:7.5K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB7525 |
| A10R216 | 315-0183-00 |  | RES, FXD, CMPSN: 18 K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1835 |
| A10R218 | 315-0203-00 |  | RES.,FXD,CMPSN:20K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2035 |
| A10R220 | 315-0152-00 |  | RES.,FXD,CMPSN: 1.5 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1525 |
| A10R234 | 315-0470-00 |  | RES.,FXD,CMPSN: 47 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4705 |
| A10R236 | 315-0470-00 |  | RES.,FXD,CMPSN: 47 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4705 |
| A10R238 | 315-0680-00 |  | RES.,FXD,CMPSN: 68 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB6805 |
| A10R240 | 311-1268-00 |  | RES.,VAR,NONWIR: 10 K OHM, $10 \%, 0.50 \mathrm{~W}$ | 32997 | 3329P-L58-103 |
| A10R243 | 311-1594-00 |  | RES.,VAR,NONWIR: 10 OHM, $20 \%, 0.50 \mathrm{~W}$ | 73138 | 91-93-0 |
| A10R250 | 315-0911-00 |  | RES.,FXD,CMPSN:910 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB9115 |
| A10R252 | 315-0470-00 |  | RES.,FXD,CMPSN: 47 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4705 |
| A10R253 | 315-0181-00 | B020000 | RES.,FXD,CMPSN: 180 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1815 |
| A10R254 | 315-0470-00 |  | RES.,FXD,CMPSN: 47 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4705 |
| A 10 R 256 | 315-0102-00 |  | RES.,FXD,CMPSN: 1 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A10R260 | 321-0195-00 |  | RES.,FXD,FILM : 1.05 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G10500F |
| A10R264 | 315-0222-00 |  | RES.,FXD,CMPSN:2.2K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2225 |
| A10R266 | 315-0470-00 |  | RES.,FXD,CMPSN:47 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4705 |
| A10R322 | 315-0822-00 |  | RES.,FXD,CMPSN: 8.2 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB8225 |
| A10R324 | 315-0911-00 |  | RES.,FXD,CMPSN: 910 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB9115 |
| A10R330 | 315-0103-00 |  | RES.,FXD,CMPSN: 10 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A10R332 | 315-0681-00 |  | RES.,FXD,CMPSN: 680 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB6815 |
| A10R336 | 315-0911.00 |  | RES.,FXD,CMPSN: 910 OHM, $5 \%$, 0.25 W | 01121 | CB9115 |
| A10R340 | 315-0102-00 |  | RES., FXD,CMPSN:1K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A10R342 | 315-0102-00 |  | RES.,FXD,CMPSN:1K ОНM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A10R350 | 315-0102-00 |  | RES.,FXD,CMPSN:1K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A10R352 | 315-0103-00 |  | RES. FXD,CMPSN: 10 K OHM , $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A10R360 | 321-0198-00 |  | RES.,FXD,FILM 1.13 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G11300F |
| A 10R362 | 321-0260-00 |  | RES.,FXD,FILM:4.99K OHM, $1 \%$ \% 0.125 W | 91637 | MFF1816G49900F |
| A10R410 | 315-0222-00 |  | RES.,FXD,CMPSN:2.2K OHM, $5 \%$, 0.25 W | 01121 | CB2225 |
| A10R414 | 315-0332-00 |  | RES.,FXD,CMPSN:3.3K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB3325 |
| A10R420 | 315-0820-00 |  | RES.,FXD,CMPSN: 82 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | C88205 |
| A10R422 | 315-0560-00 |  | RES.,FXD,CMPSN:56 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB5605 |
| A10R424 | 321-0452-00 |  | RES.,FXD,FILM:499K OHM, $1 \%$, 0.125 W | 91637 | CMF55116G49902F |
| A10R430 | 315-0820-00 |  | RES.,FXD,CMPSN: 82 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB8205 |
| A10R432 | 315-0560-00 |  | RES.,FXD,CMPSN: 56 OHM, $5 \%$ \% 0.25 W | 01121 | CB5605 |
| A10R434 | 321-0437-00 |  | RES.,FXD,FILM 348 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | CMF55116G34802F |
| A10R436 | 315-0910-00 |  | RES.,FXD,CMPSN:91 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB9105 |
| A10R438 | 315-0472-00 |  | RES.,FXD,CMPSN: 4.7 K OHM $.5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4725 |
| A10R442 | 315-0100-00 |  | RES.,FXD,CMPSN: 10 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1005 |
| A10R444 | 321-0378-00 |  | RES.,FXD.FILM: 84.5 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G84501F |


| Component No. | Tektronix Part No. | Serial/M Eff | del No. Dscont | Name \& Description | Mfr Code | Mfr Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A10R448 | 321-0397-00 |  |  | RES.,FXD,FILM:133K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | CMF55116G13302F |
| A10R450 | 315-0103-00 |  |  | RES.,FXD,CMPSN: 10 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A10R452 | 315-0102-00 |  |  | RES.,FXD,CMPSN: 1 K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A10R454 | 321-0289-00 |  |  | RES.,FXD,FILM: 10 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G10001F |
| A10R460 | 315-0103-00 |  |  | RES., FXD,CMPSN: 10 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A10R500 | 315-0103-00 |  |  | RES.,FXD,CMPSN:10K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A10R522 | 315-0511-00 |  |  | RES.,FXD,CMPSN: 510 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB5115 |
| A10R524 | 311-1258-00 |  |  | RES.,VAR, NONWIR: 50 OHM, $10 \%, 0.50 \mathrm{~W}$ | 32997 | 3329P-L58-500 |
| A10R528 | 315-0153-00 |  |  | RES.,FXD,CMPSN: 15 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1535 |
| A10R530 | 315-0472-00 |  |  | RES.,FXD,CMPSN: 4.7 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4725 |
| A10R532 | 311-1258-00 |  |  | RES.,VAR,NONWIR: 50 OHM, 10\%,0.50W | 32997 | 3329P-L58-500 |
| A10R534 | 315-0511-00 |  |  | RES.,FXD,CMPSN: 510 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB5115 |
| A10R536 | 315-0153-00 |  |  | RES.,FXD,CMPSN: 15 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1535 |
| A10R537 | 315-0393-00 |  |  | RES.,FXD,CMPSN:39K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB3935 |
| A10R538 | 315-0393-00 |  |  | RES.,FXD,CMPSN:39K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB3935 |
| A10R550 | 321-0302-00 |  |  | RES.,FXD,FILM: 13.7 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G13701F |
| A10R552 | 315-0683-00 |  |  | RES.,FXD,CMPSN: 68 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB6835 |
| A10U030 | 156-0158-00 |  |  | MICROCIRCUIT,LI:DUAL OPERATIONAL. AMPLIFIER | 18324 | MC1458N |
| A10U106 | 156-0509-00 |  |  | MICROCIRCUIT,DI:8-BIT BINARY,MULT CUR | 04713 | MC1408L8 |
| A10U112 | 156-0158-00 |  |  | MICROCIRCUIT,LI:DUAL OPERATIONAL AMPLIFIER | 18324 | MC1458N |
| A10U150 | 156-0158-00 |  |  | MICROCIRCUIT,LI:DUAL OPERATIONAL AMPLIFIER | 18324 | MC1458N |
| A10U220 | 155-0126-00 |  |  | MICROCIRCUIT,LI:TRIGGER AMPLIFIEF | 80009 | 155-0126-00 |
| A10U240 | 155-0109-01 |  |  | MICROCIRCUIT, LI:MONOLITHIC TRIGGIER | 80009 | 155-0109-01 |
| A10U310 | 156-0723-00 | 8010100 | 8030654 | MICROCIRCUIT,LI:QUAD SPST ANALOG INVT SW | 80009 | 156-0723-00 |
| A10U310 | 156-0723-01 | B030655 |  | MICROCIRCUIT,LII:QUAD SPST ANALOG INVT SW | 80009 | 156-0723-01 |
| A10U350 | 156-0067-00 |  |  | MICROCIRCUIT,LI:OPERATIONAL AMPLIFIER | 01295 | MICROA 741 CP |
| A10U358 A10U358 | 156-0686-00 | $B 010100$ 8010157 | B010156 | MICROCIRCUIT,LI:OPNL AMPL,HIGH IMPEDANCE | 02735 | CA3130S |
| A10U358 | 156-1149-00 | B010157 |  | MICROCIRCUIT, LI: OPERATIONAL AMP,JFET INPUT | 27014 | GLEA134 |
| A10U410 | 156-0515-02 |  |  | MICROCIRCUIT,DI:TRIPLE 3-CHAN MUX ${ }^{\text {a }}$, SEL | 80009 | 156-0515-02 |
| A10U452 | 156-0132-00 | B010100 | B030979 | MICROCIRCUIT,LI:OPERATIONAL AMPLIFIER | 34371 | HA2-911-5 |
| A10U452 | 156-0770-00 | B030980 |  | MICROCIRCUIT,LI:OPERATIONAL AMPLIFIER | 27014 | LF356H |
| A10U510 | 156-0723-00 | B010100 | B030654 | MICROCIRCUIT,LI:QUAD SPST ANALOG INVT SW | 80009 | 156-0723-00 |
| A10U510 | 156-0723-01 | B030655 |  | MICROCIRCUIT,LI:QUAD SPST ANALOG INVT SW | 80009 | 156-0723-01 |


|  | Tektronix | Serial/Model No. |  | Mfr |
| :--- | :--- | :--- | :--- | :--- |
| Component No. | Part No. | Eff | Dscont | Name \& Description |


|  |  |  |  | A20 SWEEP |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A20 | 670-5037-00 | B010100 | B020454 | CKT BOARD ASSY:SWEEP | 80009 | 670-5037-00 |
| A20 | 670-5037-01 | B020455 | B031149 | CKT BOARD ASSY:SWEEP | 80009 | 670-5037-01 |
| A20 | 670-5037-02 | B031150 |  | CKT BOARD ASSY:SWEEP | 80009 | 670-5037-02 |
| A20C030 | 283-0178-00 |  |  | CAP.,FXD, CER DI:0.1UF. $+80-20 \%, 100 \mathrm{~V}$ | 72982 | 8131N145651 $104 Z$ |
| A20C032 | 283-0178-00 |  |  | CAP.,FXD,CER DI:0.1UF. $+80-20 \%, 100 \mathrm{~V}$ | 72982 | 8131 N145651 104Z |
| A20C033 | 283-0051-00 | B020455 |  | CAP.,FXD,CEF DI:0.0033UF, $5 \%, 100 \mathrm{~V}$ | 56289 | 1C20C0G332J100B |
| A20C034 | 290-0745-00 |  |  | CAP.,FXD,FILCTLT:22UF. $+50-10 \%, 25 \mathrm{~V}$ | 54473 | ECE-A25V22L |
| A20C036 | 290-0745-00 |  |  | CAP,,FXD, ELCTLT: $22 \mathrm{UF},+50-10 \%, 25 \mathrm{~V}$ | 54473 | ECE-A25V22L |
| A20C040 | 283-0111-00 |  |  | CAP,,FXD,CER D $: 0.1 \mathrm{UF}, 20 \%, 50 \mathrm{~V}$ | 96733 | R2632 |
| A20C044 | 290-0745-00 |  |  | CAP,,FXD,ELCTLT:22UF, $+50-10 \%, 25 \mathrm{~V}$ | 54473 | ECE-A25V22L |
| A20C045 | 290-0745-00 |  |  | CAP.,FXD,ELCTLT:22UF. $+50-10 \%, 25 \mathrm{~V}$ | 54473 | ECE-A25V22L |
| A20C046 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A20C050 | 283-0111-00 |  |  | CAP..FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A20C100 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A20C105 | 283-0067-00 |  |  | CAP.,FXD,CER DI: $0.001 \mathrm{UF}, 10 \% .200 \mathrm{~V}$ | 59660 | 835-515-Z5D0102K |
| A20C140 | 283-0111-00 |  |  | CAP.,FXD.CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A20C156 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A20C202 | 283-0642-00 |  |  | CAP.,FXD,MICA D:33PF, +1-0.5PF,300V | 00853 | D10-5E330G |
| A20C204 | 283-0616-00 |  |  | CAP.,FXD,MICA D:75PF,5\%,500V | 00853 | D155E750J0 |
| A20C220 | 281-0592-00 |  |  | CAP.,FXD, CER DI:4.7PF, +/-0.5PF,500V | 59660 | 0301080COH0479 D |
| A20C233 | 281-0541-00 |  |  | CAP.,FXD, CER DI:6.8PF, $10 \% .500 \mathrm{~V}$ | 59660 | 301-000C0H0689D |
| A20C248 | 283-0111-00 | B010100 | B010156 | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A 20 C 248 | 283-0220-00 | B010157 |  | CAP., FXD, CER DI:0.01UF,20\%,50V | 72982 | 8121N075X7R0103M |
| A20C252 | 281-0541-00 |  |  | CAP.,FXD, CER DI:6.8PF, $10 \%$,500V | 59660 | 301-000C0H0689D |
| A20C300 | 283-0594-00 |  |  | CAP.,FXD,MICA D:0.001UF, $1 \%, 100 \mathrm{~V}$ | 00853 | D151F102F0 |
| A20C302 | 283-0111-00 |  |  | CAP.,FXD, CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A20C304 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A20C310 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF, $20 \%, 50 \mathrm{~V}$ | 96733 | R2632 |
| A20C312 | 283-0111-00 |  |  | CAP.,FXD.CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A20C340 | 283-0060-00 |  |  | CAP.,FXD,CER DI:100PF, $5 \%, 200 \mathrm{~V}$ | 59660 | 855-535U2J101J |
| A20C352 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A20C358 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF, $20 \%, 50 \mathrm{~V}$ | 96733 | R2632 |
| A20C404 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A20C410 | 295-0187-00 |  |  | CAP SET,MATCHED:(1)EA 10UF,0.1UF,900PF | 80009 | 295-0187-00 |
| A20C412 | ---------- |  |  | (PART OF C410) |  |  |
| A20C420 | ---------- |  |  | (PART OF C410) |  |  |
| A20C446 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A20C448 | 283-0220-00 |  |  | CAP.,FXD, CER DI:0.01UF,20\%,50V | 72982 | 8121N075X7R0103M |
| A20C500 | 281-0166-00 |  |  | CAP,,VAR,AIR DI:1.9-15.7PF, 250V | 74970 | 187-0109-055 |
| A20C504 | 283-0111-00 |  |  | CAP.,FXD,CER DI:0.1UF, $20 \%$,50V | 96733 | R2632 |
| A20C510 | 283-0629-00 | 8010100 | B031149 | CAP.,FXD,MICA D:62PF, $1 \%, 500 \mathrm{~V}$ | 00853 | D105E620F0 |
| A20C510 | 283-0639-00 | B031150 |  | CAP.,FXD,MICA D:56PF, $1 \%, 100 \mathrm{~V}$ | 00853 | D151E560F0 |
| A20C512 | 283-0111-00 |  |  | CAP.,FXD.CER DI:0.1UF,20\%.50V | 96733 | R2632 |
| A20C530 | 283-0597-00 |  |  | CAP.,FXD.MICA D: $470 \mathrm{PF}, 10 \%, 300 \mathrm{~V}$ | 00853 | D155F471K0 |
| A20C538 | 283-0648-00 |  |  | CAP.,FXD,MICA D: $10 \mathrm{PF}, 5 \% .100 \mathrm{~V}$ | 00853 | D151C100D0 |
| A20C540 | 283-0111-00 |  |  | CAP.,FXD, CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A20C544 | 283-0644-00 |  |  | CAP.,FXD,MICA D: $150 \mathrm{PF}, 1 \%, 500 \mathrm{~V}$ | 00853 | D155F151F0 |
| A20C608 | 290-0267-00 |  |  | CAP.,FXD,ELCTLT:1UF,20\%,35V | 56289 | 173D105X0035V |
| A20C612 | 283-0111-00 |  |  | CAP..FXD,CER DI:0.1UF, $20 \%$,50V | 96733 | R2632 |
| A20C614 | 283-0599-00 |  |  | CAP..FXD,MICA D:98PF, $5 \%, 500 \mathrm{~V}$ | 00853 | D105F980J0 |
| A20C620 | 290-0282-00 |  |  | CAP.,FXD,ELCTLT:0.047UF, $10 \%$, 35 V | 56289 | 162D473×9035BC2 |


| Component No. | Tektronix Part No. | Serial/Model No. Eff Dscont | Name \& Description | Mfr Code | Mfr Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: |
| A20C622 | 283-0695-00 |  | CAP.,FXD,MICA D:4440PF, $1 \%, 300 \mathrm{~V}$ | 00853 | D195F4441F0 |
| A20C626 | 290-0244-00 |  | CAP.,FXD,ELCTLT:0.47UF,5\%,35V | 56289 | 173D474X5035U |
| A20C640 | 283-0111-00 |  | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A20C644 | 283-0698-00 |  | CAP.,FXD,MICA D:390PF, $1 \%, 500 \mathrm{~V}$ | 09023 | CD15FD391F03 |
| A20C650 | 290-0267-00 |  | CAP.,FXD,ELCTLT:1UF,20\%,35V | 56289 | 173D105X0035V |
| A20C654 | 283-0648-00 |  | CAP.,FXD,MICA D: 10 PF, $5 \%, 100 \mathrm{~V}$ | 00853 | D151C100D0 |
| A20C656 | 283-0111-00 |  | CAP.,FXD.CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A200730 | 290-0284-00 |  | CAP.,FXD,ELCTLT:4.7UF,10\%,35V | 56289 | 150D475X9035B2 |
| A20C750 | 290-0267-00 |  | CAP.,FXD,ELCTLT:1UF,20\%,35V | 56289 | 173 D 105 X 0035 V |
| A20C844 | 283-0111-00 |  | CAP.,FXD,CER DI:0.1UF,20\%,50V | 96733 | R2632 |
| A20CR138 | 152-0141-02 |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A20CR152 | 152-0322-00 |  | SEMICOND DEVICE:SILICON,15V,HOT CARRIER | 50434 | 5082-2672 |
| A20CR154 | 152-0141-02 |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A20CR204 | 152-0141-02 |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A20CR206 | 152-0141-02 |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A20CR222 | 152-0141-02 |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A20CR242 | 152-0141-02 |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A20CR310 | 152-0141-02 |  | SEMICOND DEVICE:SILICON, $30 \mathrm{~V}, 150 \mathrm{MA}$ | 01295 | 1N4152R |
| A20CR406 | 152-0322-00 |  | SEMICOND DEVICE:SILICON,15V,HOT CARRIER | 50434 | 5082-2672 |
| A20CR450 | 152-0141-02 |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A20CR452 | 152-0141-02 |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A20CR454 | 152-0141-02 |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A20CR544 | 152-0141-02 |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A20CR550 | 152-0141-02 |  | SEMICOND DEVICE: SILICON,30V,150MA | 01295 | 1N4152R |
| A20CR600 | 152-0141-02 |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A20CR602 | 152-0141-02 |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A20CR610 | 152-0141-02 |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A20CR732 | 152-0141-02 |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A20CR834 | 152-0075-00 |  | SEMICOND DEVICE:SW,GE,22V,40MA | 14433 | G866 |
| A20CR850 | 152-0075-00 |  | SEMICOND DEVICE:SW,GE,22V,40MA | 14433 | G866 |
| A20CR954 | 152-0075-00 |  | SEMICOND DEVICE:SW,GE,22V,40MA | 14433 | G866 |
| A200S001 | 150-1036-00 |  | LAMP,LED:RED,3.0V,40MA | 01295 | TIL 209A |
| A200S008 | 150-1036-00 |  | LAMP,LED:RED,3.0V,40MA | 01295 | TIL 209A |
| A20E148 | 276-0507-00 |  | SHIELDING BEAD,:FERRITE | 78488 | 57-3443 |
| A20E150 | 276-0507-00 |  | SHIELDING BEAD, FEPRRITE | 78488 | 57-3443 |
| A20J020 | 131-1003-00 |  | CONN,RCPT,ELEC:CKT BD MT, 3 PRONG | 80009 | 131-1003-00 |
| A20,022 | 131-1003-00 |  | CONN,RCPT,ELEC:CKT BD MT,3 PRONG | 80009 | 131-1003-00 |
| A20.J304 | 131-1003-00 |  | CONN,RCPT,ELEC:CKT BD MT,3 PRONG | 80009 | 131-1003-00 |
| A20L122 | 108-0509-00 |  | COIL,RF:2.45UH | 80009 | 108-0509-00 |
| A20LR032 | 108-0537-00 |  | COIL,RF:200UH | 80009 | 108-0537-00 |
| A20LR034 | 108-0537-00 |  | COIL,RF:200UH | 80009 | 108-0537-00 |
| A20LR040 | 108-0537-00 |  | COIL,RF:200UH | 80009 | 108-0537-00 |
| A20LR046 | 108-0537-00 |  | COIL,RF:200UH | 80009 | 108-0537-00 |
| A20LR050 | 108-0537-00 |  | COIL,RF:200UH | 80009 | 108-0537-00 |
| A20LR512 | 108-0271-00 |  | COIL,RF: 0.235 NH | 80009 | 108-0271-00 |
| A20P028 | 131-0608-00 |  | TERMINAL,PIN: $0.365 \mathrm{~L} \times 0.025$ PH BR 2 GOLD | 22526 | 48283-036 |


| Component No. | Tektronix | Serial/Model No. |  | Name \& Description | Mfr Code | Mfr Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Part No. | Eff | Dscont |  |  |  |
| A20Q120 | 151-0325-00 | B010100 | 8020414 | TRANSISTOR: SILICON,PNP, SEL FROM 2N4258 | 80009 | 151-0325-00 |
| A20Q120 | 151-0221-00 | B020415 |  | TRANSISTOR:SILICON,PNP | 04713 | SPS246 |
| A20Q130 | 151-0472-00 |  |  | TRANSISTOR:SILICON,NPN | 51984 | NE41632B |
| A20Q132 | 151-0472-00 |  |  | TRANSISTOR:SILICON NPN | 51984 | NE41632B |
| A20Q134 | 151-1113-00 |  |  | TRANSISTOR:FE,N CHANNEL, SILICON | 80009 | 151-1113-00 |
| A20Q142 | 151-0192-00 |  |  | TRANSISTOR:SILICON,NPN,SEL FROM MPS652 | 04713 | SPS8801 |
| A20Q144 | 151-0223-00 |  |  | TRANSISTOR:SILICON,NPN | 04713 | SPS8026 |
| A20Q146 | 151-0220-00 |  |  | TRANSISTOR:SILICON,PNP | 07263 | S036228 |
| A20Q148 | 151-0325-00 | B010100 | B020414 | TRANSISTOR:SILICON,PNP,SEL FROM 2N4258 | 80009 | 151-0325-00 |
| A20Q148 | 151-0221-00 | B020415 |  | TRANSISTOR:SILICON,PNP | 04713 | SPS246 |
| A20Q150 | 151-0223-00 |  |  | TRANSISTOR:SILICON,NPN | 04713 | SPS8026 |
| A20Q220 | 151-0220-00 |  |  | TRANSISTOR:SILICON.PNP | 07263 | S036228 |
| A20Q222 | 151-0220-00 |  |  | TRANSISTOR:SILICON,PNP | 07263 | S036228 |
| A200224 | 151-0220-00 |  |  | TRANSISTOR:SILICON,PNP | 07263 | S036228 |
| A20Q234 | 151-0220-00 |  |  | TRANSISTOR:SILICON,PNP | 07263 | S036228 |
| A20Q236 | 151-0220-00 |  |  | TRANSISTOR:SILICON, PNP | 07263 | S036228 |
| A20Q242 | 151-0325-00 | B010100 | B020414 | TRANSISTOR:SILICON,PNP,SEL FROM 2N4258 | 80009 | 151-0325-00 |
| A200242 | 151-0221-00 | B020415 |  | TRANSISTOR:SILICON,PNP | 04713 | SPS246 |
| A20Q250 | 151-0190-00 |  |  | TRANSISTOR:SILICON,NPN | 07263 | S032677 |
| A20Q252 | 151-0223-00 |  |  | TRANSISTOR:SILICON,NPN | 04713 | SPS8026 |
| A200300 | 151-0354-00 |  |  | TRANSISTOR:SILICON, PNP,DUAL | 32293 | ITS1200A |
| A20Q312 | 151-0223-00 |  |  | TRANSISTOR:SILICON,NPN | 04713 | SPS8026 |
| A20Q314 | 151-0223-00 |  |  | TRIANSISTOR:SILICON,NPN | 04713 | SPS8026 |
| A20Q320 | 151-0190-00 |  |  | TRANSISTOR:SILICON,NPN | 07263 | S0326\%7 |
| A20Q324 | 151-0220-00 |  |  | TRANSISTOR:SILICON,PNP | 07263 | S036228 |
| A20Q354 | 151-0301-00 |  |  | TRANSISTOR:SILICON,PNP | 27014 | 2N2907A |
| A20Q356 | 151-0220-00 |  |  | TRANSISTOR:SILICON,PNP | 07263 | S036228 |
| A20Q400 | 151-0367-00 |  |  | TRANSISTOR:SILICON,NPN, SEL FROM 3571TP | 04713 | SPS 8811 |
| A20Q406 | 151-0367-00 |  |  | TRANSISTOR:SILICON,NPN,SEL FROM 3571TP | 04713 | SPS 8811 |
| A200410 | 151-0223-00 |  |  | TRANSISTOR:SILICON.NPN | 04713 | SPS8026 |
| A20Q512 | 151-0410-00 |  |  | TRANSISTOR:SILICON,PNP | 80009 | 151-0410-00 |
| A20Q516 | 151-1036-00 |  |  | TRANSISTOR:SILICON,JFE,N-CHANNEL,DUAL | 17856 | DN1663 |
| A20Q542 | 151-0223-00 |  |  | TRANSISTOR:SILICON,NPN | 04713 | SPS8026 |
| A20Q544 | 151-0223-00 |  |  | TRANSISTOR:SILICON,NPN | 04713 | SPS8026 |
| A200546 | 151-0325-00 | B010100 | B020414 | TRANSISTOR:SILICON,PNP,SEL FROM 2N4258 | 80009 | 151-0325-00 |
| A200546 | 151-0221-00 | B020415 |  | TRANSISTOR:SILICON,PNP | 04713 | SPS246 |
| A20Q552 | 151-0325-00 | B010100 | B020414 | TRANSISTOR:SILICON,PNP,SEL FROM 2N4258 | 80009 | 151-0325-00 |
| A200552 | 151-0221-00 | B010415 |  | TRANSISTOR:SILICON,PNP | 34713 | SPS246 |
| A20Q558 | 151-0223-00 |  |  | TRANSISTOR:SILICON,NPN | 34713 | SPS8026 |
| A20Q604 | 151-022.0-00 |  |  | TRANSISTOR:SILICON,PNP | J7263 | S036228 |
| A20Q620 | 151-0437-00 |  |  | TRANSISTOR:SILICON,NPN,SEL FROM 2N5769 | 30009 | 151-0437-00 |
| A20Q646 | 151-0220-00 |  |  | TRANSISTOR:SILICON,PNP | 97263 | 5036228 |
| A200648 | 151-0223-00 |  |  | TRANSISTOR:SILICON,NPN | $1) 4713$ | SPS8026 |
| A200732 | 151-0220-00 |  |  | TRANSISTOR:SILICON,PNP | 97263 | S036228 |
| A20Q740 | 151-0223-00 |  |  | TRANSISTOR:SILICON,NPN | 1)4713 | SPS8026 |
| A20Q748 | 151-0220-00 |  |  | TRANSISTOR:SILICON,PNP | 97263 | S036228 |
| A20R002 | 315-0512-00 |  |  | RES.,FXD,CMPSN:5.1K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB5125 |
| A20R004 | 321-0385-04 |  |  | RES.,FXD,FILM:100K OHM, $0.1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816D10002B |
| A20R034 | 322-0224-00 |  |  | RES.,FXD.FILM:2.15K OHM, $1 \%, 0.25 \mathrm{~W}$ | 91637 | CMF6042G29400F |
| A20R054 | 315-0510-00 |  |  | RES.,FXD, CMPSN 51 OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB5105 |
| A20R100 | 315-0242-00 |  |  | RES.,FXD,CMPSN 2.4 K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2425 |
| A20R101 | 321-0385-04 |  |  | RES.,FXD,FILM: 100 K OHM $, 0.1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816D10002B |
| A20R102 | 321-0756-04 |  |  | RES.,FXD,FILM:50K OHM $, 0.1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816D50001B |


| Component No. | Tektronix Part No. |  | del No. Dscont | Name \& Description | Mfr Code | Mfr Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A20R103 | 321-0332-00 |  |  | RES.,FXD,FILM:28K OHM $1 \%, 0.125 \mathrm{~W}$ |  |  |
| A20R104 | 321-0258-00 |  |  | RES.,FXD.FILM 4.75 K OHM, $1 \% .0 .125 \mathrm{~W}$ | 91637 | MFF 1816 G 28001 F |
| A20R105 | 321-0238-00 |  |  | RES.,FXD,FILM 2.94 K OHM $, 1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G47500F |
| A20R106 | 315-0153-00 |  |  | RES.,FXD,CMPSN: 15 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1535 |
| A20R107 | 315-0153-00 |  |  | RES.,FXD,CMPSN: 15 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1535 |
| A20R108 | 321-0773-03 |  |  | RES.,FXD,FILM: 400 OHM $, 0.25 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816D400ROC |
| A20R116 | 315-0152-00 |  |  | RES.,FXD.CMPSN:1.5K OHM.5\%,0.25W | 01121 | CB1525 |
| A20R118 | 321-0210-00 |  |  | RES.,FXD,FILM: 1.5 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G15000F |
| A20R120 | $315-0473-00$ |  |  | RES.,FXD,CMPSN: 47 K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4735 |
| A20R121 | 323-0243-00 |  |  | RES., FXD, FILM:3.32K OHM $, 1 \%, 0.50 \mathrm{~W}$ | 91637 | MFF1226G33200F |
| A20R122 | 315-0682-00 |  |  | RES.,FXD,CMPSN:6.8K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB6825 |
| A20R123 | 321-0134-00 |  |  | RES.,FXD,FILM: 243 OHM $, 1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G243R0F |
| A20R126 | 321-0054-00 |  |  | RES.,FXD,FILM:35.7 OHM,5\%,0.125W | 91637 | MFF1816G35R70F |
| A20R128 | 315-0270-00 |  |  | RES.,FXD, CMPSN:27 OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2705 |
| A20R130 | 322-0224-00 |  |  | RES.,FXD,FILM $: 2.15 \mathrm{~K}$ OHM, $1 \%, 0.25 \mathrm{~W}$ | 91637 | CMF6042G29400F |
| A20R131 | 321-0134-00 |  |  | RES.,FXD, FILM: 243 OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G243R0F |
| A20R132 | 315-0393-00 |  |  | RES.,FXD,CMPSN:39K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB3935 |
| A20R133 | 321-0072-00 |  |  | RES.,FXD,FILM $54.9 \mathrm{OHM}, 1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G54R90F |
| A20R134 | 321-0072-00 |  |  | RES.,FXD,FILM:54.9 OHM, 1\%,0.125W | 91637 |  |
| A20R134 | 321-0054-00 |  |  | RES.,FXD,FILM:35.7 OHM $, 5 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G35R70F |
| A20R136 | 315-0105-00 |  |  | RES.,FXD, CMPSN: 1 M OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1055 |
| A20R138 | 315-0303-00 |  |  | RES.,FXD,CMPSN:30K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB3035 |
| A20R140 | 315-0202-00 |  |  | RES.,FXD,CMPSN:2K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2025 |
| A20R142 | 315-0270-00 |  |  | RES.,FXD,CMPSN: 27 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2705 |
| A20R143 | 315-0102-00 |  |  | RES.,FXD,CMPSN:1K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 |  |
| A20R144 | 315-0510-00 |  |  | RES.,FXD,CMPSN: 51 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB5105 |
| A20R145 | 315-0303-00 |  |  | RES.,FXD,CMPSN:30K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB3035 |
| A20R146 | 315-0101-00 |  |  | RES.,FXD,CMPSN: 100 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1015 |
| A20R148 A20R150 | 315-0332-00 |  |  | RES.,FXD,CMPSN:3.3K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB3325 |
| A20R150 | 321-0173-00 |  |  | RES.,FXD,FILM 619 OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G619R0F |
| A20R152 | 315-0101-00 |  |  | RES.,FXD,CMPSN: 100 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1015 |
| A20R156 A20R200 | 315-0132-00 |  |  | RES.,FXD,CMPSN: 1.3 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1325 |
| A20R200 | $321-0781-06$ $321-0924-07$ |  |  | RES.,FXD,FILM: 400 K OHM $, 0.25 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816C40002C |
| A20R203 | 321-0924-0926-07 |  |  | RES.,FXD,FILM:40K OHM.0.1\%,0.125W | 91637 | MFF1816C40001B |
| A20R204 | 321-0275-00 |  |  | RES.,FXD,FILM:4K OHM, $0.1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816C40000B |
|  | 32-0275-00 |  |  | RES.,FXD,FILM:7.15K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFFF1816G71500F |
| A20R206 | 315-0221-00 |  |  | RES.,FXD,CMPSN:220 OHM, $5 \%, 0.25 \mathrm{~W}$ |  |  |
| A20R207 | 315-0102-00 |  |  | RES.,FXD,CMPSN: 1 K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2215 CB1025 |
| A20R208 | 315-0470-00 |  |  | RES.,FXD, CMPSN: 47 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4705 |
| A20R210 | 315-0152-00 |  |  | RES.,FXD,CMPSN:1.5K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1525 |
| A20R212 | 315-0103-00 |  |  | RES.,FXD,CMPSN: 10 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A20R220 | 315-0101-00 |  |  | RES.,FXD,CMPSN: 100 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1015 |
| A20R221 | 315-0103-00 |  |  | RES.,FXD,CMPSN: 10 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 |  |
| A20R222 | 315-0153-00 |  |  | RES.,FXD,CMPSN: 15 K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1535 |
| A20R223 A20R224 | 315-0112-00 |  |  | RES.,FXD,CMPSN:1.1K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1125 |
| A20R226 | 315-0100-00 |  |  | RES.,FXD,CMPSN: 10 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1005 |
| A20R227 | $315-0511-00$ $315-0153-00$ |  |  | RES , FXD,CMPSN: $510 \mathrm{OHM}, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB5115 |
| A20R227 | 315-0153-00 |  |  | RES.,FXD,CMPSN:15K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1535 |
| A20R228 | 321-0193-00 |  |  | RES.,FXD,FILM:1K OHM,1\%,0.125W |  |  |
| A20R230 | 323-0187-00 |  |  | RES., FXD,FILM: 866 OHM, $1 \%, 0.50 \mathrm{~W}$ | 75042 | ORD BY DESCR CECTO-8660F |
| A20R232 | 311-1423-00 |  |  | RES.,VAR,NONWIR: 20 OHM, $20 \%, 0.50 \mathrm{~W}$ | 73138 | 72PM R20-1A |
| A20R233 | 315-0622-00 |  |  | RES.,FXD,CMPSN:6.2K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB6225 |
| A20R235 | 321-0193-00 |  |  | RES..FXD,FILM: 1 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 01121 | ORD BY DESCR |
|  | 321-0122-00 |  |  | RES.,FXD,FILM: 182 OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G182R0F |


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| :---: | :---: | :---: | :---: | :---: | :---: |
| A20R236 | 321-0356-00 |  | RES.,FXD,FILM: 49.9 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G49901F |
| A20R237 | 321-0193-00 |  | RES.,FXD,FILM: 1 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 01121 | ORD BY DESCR |
| A20R238 | 311-1226-00 |  | RES.,VAR,NONWIR: 2.5 K OHM, $20 \%, 0.50 \mathrm{~W}$ | 32997 | 3386F-T04-252 |
| A20R240 | 311-1228-00 |  | RES.,VAR,NONWIR: 10 K OHM, $20 \%, 0.50 \mathrm{~W}$ | 32997 | 3386F-T04-103 |
| A20R241 | 315-0682-00 |  | RES.,FXD,CMPSN:6.8K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB |
| A20R242 | 315-0103-00 |  | RES.,FXD,CMPSN:10K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A20R243 | 315-0270-00 |  | RES.,FXD,CMPSN:27 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2705 |
| A20R244 | 315-0473-00 |  | RES.,FXD,CMPSN:47K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4735 |
| A20R246 | 315-0510-00 |  | RES.,FXD,CMPSN: 51 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB5105 |
| A20R248 | 315-0510-00 |  | RES.,FXD,CMPSN: 51 OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB5105 |
| A20R250 | 321-0210-00 |  | RES.,FXD,FILM: 1.5 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G1500 |
| A20R252 | 315-0152-00 |  | RES.,FXD,CMPSN:1.5K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | 1525 |
| A20R253 | 315-0510-00 |  | RES.,FXD,CMPSN: 51 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB5105 |
| A20R254 | 315-0302-00 |  | RES.,FXD,CMPSN:3K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB3025 |
| A20R255 | 315-0472-00 |  | RES.,FXD,CMPSN:4.7K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4725 |
| A20R256 | 315-0202-00 |  | RES.,FXD,CMPSN: 2 K OHM, $5 \%, 0,25 \mathrm{~W}$ | 01121 | CB2025 |
| A20R257 | 315-0222-00 |  | RES, FXD,CMPSN:2.2K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2225 |
| A20R258 | 315-0222-00 |  | RES.,FXD,CMPSN:2.2K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2225 |
| A20R300 | 323-0172-00 |  | RES.,FXD,FILM:604 OHM, 1\%,0.50W | 91637 | MFF1226G604R0F |
| A20R302 | 315-0470-00 |  | RES.,FXD,CMPSN: 47 OHM , $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4705 |
| A20R310 | 315-0102-00 |  | RES.,FXD,CMPSN:1K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A20R312 | 315-0102-00 |  | RES.,FXD,CMPSN:1K OHM, 5\%,0.25W | 01121 | CB1025 |
| A20R314 | 315-0102-00 |  | RES.,FXD,CMPSN: 1 K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A20R320 | 315-0152-00 |  | RES.,FXD,CMPSN:1.5K OHM $.5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1525 |
| A20R330 | 321-0193-00 |  | RES.,FXD,FILM:1K OHM, $1 \%, 0.125 \mathrm{~W}$ | 01121 | ORD BY DESCR |
| A20R340 | 315-0202-00 |  | RES.,FXD,CMPSN:2K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2025 |
| A20R341 | 315-0122-00 |  | RES.,FXD,CMPSN:1.2K OHM, $5 \%, 0.25 \mathrm{~W}$ | 21 | CB1225 |
| A20R342 | 311-1266-00 |  | RES.,VAR,NONWIR: 2.5 K OHM, $10 \%, 0.50 \mathrm{~W}$ | 32997 | 3329P-L.58-252 |
| A20R343 | 321-0178-00 |  | RES.,FXD,FILM: 698 OHM, $1 \% .0 .125 \mathrm{~W}$ | 91637 | MFF1816G698R0F |
| A20R344 | 321-0188-00 |  | RES.,FXD,FILM: 887 OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G887ROF |
| A20R346 | 321-0271-00 |  | RES.,FXD,FILM:6.49K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G64900F |
| A20R350 | 315-0430-00 |  | RES.,FXD,CMPSN:43 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4305 |
| A20R352 | 315-0103-00 |  | RES.,FXD,CMPSN: 10 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A20R354 | 315-0102-00 |  | RES.,FXD,CMPSN: 1 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A20R356 | 315-0472-00 |  | RES.,FXD,CMPSN:4.7K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4725 |
| A20R35\% | 315-0103-00 |  | RES.,FXD,CMPSN:10K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A20R358 | 315-0202-00 |  | RES.,FXD,CMPSN:2K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2025 |
| A20R400 | 315-0101-00 |  | RES.,FXD,CMPSN: 100 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1015 |
| A20R402 | 315-0221-00 |  | RES.,FXD,CMPSN:220 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2215 |
| A20R404 | 315-0100-00 |  | RES.,FXD,CMPSN: 10 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1005 |
| A20R405 | 315-0102-00 |  | RES.,FXD,CMPSN:1K OHM $.5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A20R406 | 315-0101-00 |  | RES.,FXD.CMPSN: $100 \mathrm{OHM}, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1015 |
| A20R410 | 307-0114-00 |  | RES.,FXD,CMPSN:6.2 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB62G5 |
| A20R412 | 315-0102-00 |  | RES.,FXD,CMPSN: 1 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A20R430 | 321-0193-00 |  | RES.,FXD,FILM: 1 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 01121 | ORD BY DESCR |
| A20R432 | 321-0193-00 |  | RES.,FXD,FILM: 1 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 01121 | ORD BY DESC |
| A20R440 | 315-0393-00 |  | RES.,FXD,CMPSN:39K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB3935 |
| A20R442 | 321-0231-00 |  | RES.,FXD,FILM:2.49K OHM $, 1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G24900F |
| A20R444 | 321-0175-00 |  | RES.,FXD,FILM: 649 OHM, 1\%,0.125W | 91637 | CMF55116G649R0F |
| A20R446 | 321-0280-00 |  | RES.,FXD,FILM:8.06K OHM $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G80600F |
| A20R454 | 315-0622-00 |  | RES.,FXD,CMPSN:6.2K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB6225 |
| A20R502 | 315-0101-00 |  | RES.,FXD,CMPSN: 100 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1015 |
| A20R505 | 315-0153-00 |  | RES.,FXD.CMPSN: 15 K OHM. $5 \% .0 .25 \mathrm{~W}$ | 01121 | CB1535 |
| A20R506 | 315-0180-00 |  | RES.,FXD, CMPSN: 18 OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1805 |


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| :---: | :---: | :---: | :---: | :---: | :---: |
| A20R512 | 315-0470-00 |  | RES.,FXD,CMPSN: 47 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4705 |
| A20R514 | 315-0510-00 |  | RES.,FXD,CMPSN: 51 OHM , $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB5105 |
| A20R516 | 315-0101-00 |  | RES.,FXD,CMPSN: 100 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1015 |
| A20R520 | 315-0470-00 |  | RES.,FXD,CMPSN: 47 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4705 |
| A20R522 | 315-0470-00 |  | RES.,FXD,CMPSN: 47 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4705 |
| A20R524 | 315-0272-00 |  | RES.,FXD,CMPSN:2.7K OHM, $5 \%, 0.25 \mathrm{~V}$ | 01121 | CB2725 |
| A20R526 | 315-0470-00 |  | RES.,FXD,CMPSN: 47 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4705 |
| A20R530 | 315-0102-00 |  | RES.,FXD,CMPSN:1K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A20R534 | 321-0168-00 |  | RES.,FXD,FILM: 549 OHM, 1\%,0.125W | 91637 | MFF1816G549R0F |
| A20R536 | 321-0235-00 |  | RES.,FXD,FILM: 2.74 K OHM, $1 \%, 0.125 \mathrm{~W}^{\prime}$ | 91637 | MFF1816G27400F |
| A20R540 | 315-0472-00 |  | RES,,FXD,CMPSN:4.7K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4725 |
| A20R542 | 315-0391-00 |  | RES.,FXD,CMPSN: 390 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB3915 |
| A20R543 | 315-0202-00 |  | RES.,FXD,CMPSN:2K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2025 |
| A20R544 | 321-0239-00 |  | RES.,FXD,FILM 3.01 K OHM $, 1 \%, 0.125 \mathrm{~W}^{\prime}$ | 91637 | MFF1816G30100F |
| A20R546 | 315-0162-00 |  | RES.,FXD,CMPSN:1.6K OHM , $5 \%, 0.25 \mathrm{~V}$ | 01121 | CB1625 |
| A20R547 | 315-0432-00 |  | RES.,FXD,CMPSN: 4.3 K OHM, $5 \%, 0.25 \mathrm{~V}$ | 01121 | CB4325 |
| A20R548 | 315-0242-00 |  | RES.,FXD,CMPSN:2.4K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2425 |
| A20R549 | 315-0821-00 |  | RES.,FXD,CMPSN:820 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB8215 |
| A20R550 | 315-0223-00 |  | RES.,FXD,CMPSN: 22 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2235 |
| A20R554 | 315-0102-00 |  | RES.,FXD,CMPSN:1K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A20R556 | 315-0202-00 |  | RES.,FXD,CMPSN:2K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2025 |
| A20R558 | 315-0102-00 |  | RES.,FXD,CMPSN:1K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A20R600 | 315-0101-00 |  | RES.,FXD,CMPSN: $100 \mathrm{OHM}, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1015 |
| A20R602 | 321-0301-00 |  | RES.,FXD,FILM: 13.3 K OHM, $1 \%, 0.125 \mathrm{~W}^{\prime}$ | 91637 | MFF1816G13301F |
| A20R610 | 315-0101-00 |  | RES.,FXD,CMPSN: 100 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1015 |
| A20R614 | 315-0272-00 |  | RES.,FXD,CMPSN: 2.7 K OHM, $5 \%, 0.25 \mathrm{~V}$ | 01121 | CB2725 |
| A20R616 | 321-0202-00 |  | RES.,FXD,FILM:1.24K OHM, $1 \%, 0.125 \mathrm{~W}^{\prime}$ | 91637 | MFF1816G12400F |
| A20R622 | 315-0470-00 |  | RES.,FXD,CMPSN: 47 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4705 |
| A20R640 | 315.0391-00 |  | RES.,FXD,CMPSN:390 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB3915 |
| A20R642 | 315-0102-00 |  | RES.,FXD,CMPSN:1K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A20R650 | 315-0184-00 |  | RES.,FXD,CMPSN:180K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1845 |
| A20R652 | 321-0256-00 |  | RES.,FXD,FILM: 4.53 K OHM, $1 \% .0 .125 \mathrm{~W}$ | 91637 | MFF1816G45300F |
| A20R656 | 321-0256-00 |  | RES.,FXD,FILM:4.53K OHM, $1 \%, 0.125 \mathrm{~W}^{\prime}$ | 91637 | MFF1816G45300F |
| A20R710 | 321-0202-00 |  | RES.,FXD,FILM 1.24 K OHM $, 1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G12400F |
| A20R730 | 321-0257-00 |  | RES.,FXD,FILM: 4.64 K OHM, $1 \%, 0.125 \mathrm{~W}^{\prime}$ | 91637 | MFF1816G46400F |
| A20R732 | 315.0431-00 |  | RES.,FXD,CMPSN: 430 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4315 |
| A20R740 | 315-0222-00 |  | RES.,FXD,CMPSN:2.2K OHM, $5 \%, 0.25 \mathrm{~V}$ | 01121 | CB2225 |
| A20R742 | 321-0807-00 |  | RES.,FXD,FILM:900K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | CMF55116G90000F |
| A20R744 | 315-0472-00 |  | RES ,FXD, CMPSN: 4.7 K OHM, $5 \%, 0.25 \mathrm{~V}$ | 01121 | CB4725 |
| A20R746 | 321-0263-00 |  | RES.,FXD,FILM 5.36 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G53600F |
| A20R747 | 315-0474-00 |  | RES.,FXD,CMPSN:470K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4745 |
| A20R748 | 315-0242-00 |  | RES.,FXD,CMPSN:2.4K OHM $, 5 \%, 0.25 \mathrm{~V}$ | 01121 | CB2425 |
| A20R750 | 321-0377-00 |  | RES.,FXD,FILM: 82.5 K OHM, $1 \%, 0.125 \mathrm{~W}^{\prime}$ | 91637 | MFF1816G82501F |
| A20R752 | 315-0102-00 |  | RES.,FXD,CMPSN:1K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A20R754 | 315-0752-00 |  | RES.,FXD,CMPSN:7.5K OHM $, 5 \%, 0.25 \mathrm{~V}$ J | 01121 | CB7525 |
| A20R758 | 315-0510-00 |  | RES.,FXD,CMPSN:51 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB5105 |
| A20R814 | 321-0340-00 |  | RES.,FXD,FILM:34K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | CMF55116G34001F |
| A20R830 | 315-0103-00 |  | RES.,FXD,CMPSN:10K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A20R832 | 315-0103-00 |  | RES.,FXD,CMPSN:10K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A20R834 | 315-0153-00 |  | RES.,FXD,CMPSN:15K OHM , $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1535 |
| A20R836 | 321-0333-00 |  | RES.,FXD,FILM:28.7K OHM, $1 \%, 0.125 \mathrm{~W}^{\prime}$ | 91637 | CMF55116G28701F |
| A20R838 | 321-0289-00 |  | RES.,FXD,FILM:10K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G10001F |
| A20R840 | 315-0103-00 |  | RES.,FXD,CMPSN:10K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A20R842 | 315-0103-00 |  | RES,,FXD,CMPSN:10K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |


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| A20R850 | 321-0335-00 |  | RES, FXD,FILM: 30.1 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | CMF55116G30101F |
| A20R852 | 321-0349-00 |  | RES,.FXD,FILM:42.2K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G42201F |
| A20R854 | 315-0910-00 |  | RES , FXD,CMPSN: 91 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB9105 |
| A20R930 | 315-0103-00 |  | RES.,FXD,CMPSN: 10 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A20R932 | 315-0103-00 |  | RES.,FXD,CMPSN: 10 K OHM, $5 \%$, 0.25 W | 01121 | CB1035 |
| A20R936 | 315-0150-00 |  | RES.,FXD,CMPSN: 15 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1505 |
| A20R944 | 315-0470-00 |  | RES.,FXD,CMPSN:47 OHM $.5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4705 |
| A20R950 | 315-0155-00 |  | RES.,FXD,CMPSN:1.5M OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1555 |
| A20R952 | 321-0308-00 |  | RES.,FXD,FILM 15.8 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G15801F |
| A20R956 | 321-0340-00 |  | RES.,FXD,FILM:34K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | CMF55116G34001F |
| A20T028 | 120-0444-00 |  | XFMR,TOROID: 5 TURNS, BIFILAR | 80009 | 120-0444-00 |
| A20u020 | 156-0846-00 |  | MICROCIRCUIT, LI:VOLTAGE REGULATOR | 04713 | MC7905CT |
| A20U110 | 156-0197-00 |  | MICROCIRCUIT,LI: 5 TRANSISTOR ARRAY | 02735 | CA3086 |
| A20U116 | 156-0512-00 |  | MICROCIRCUIT,LI:OPERATIONAL AMPLIFIER | 27014 | LM308N |
| A20U212 | 156-0514-02 |  | MICROCIRCUIT, Di:DIFF 4 CHANNEL MUX, SEL | 80009 | 156-0514-02 |
| A20U320 | 156-0140-02 |  | MICROCIRCUIT,DI:HEX BUFFERS W/OC HV OUT | 01295 | SN7417 (NP3) |
| A20U330 | 156-0927-00 |  | MICROCIRCUIT,LI:DIGITAL TO ANALOG CONVERTER | 80009 | 156-0927-00 |
| A20U340 | 156-0067-00 |  | MICROCIRCUIT,LI:OPERATIONAL AMPLIFIER | 01295 | MICROA741CP |
| A20U430 | 156-0515-02 |  | MICROCIRCUIT, DI:TRIPLE 3-CHAN MUX,SEL | 80009 | 156-0515-02 |
| A20U448 | 155-0049-02 |  | MICROCIRCUIT,DI:SWEEP CONTROL,W/LOCKOUT | 80009 | 155-0049-02 |
| A20U530 | 156-0251-00 |  | MICROCIRCUIT, DI:VOLTAGE COMPENSATOR | 27014 | LM361H |
| A20U630 | 156-0513-02 |  | MICROCIRCUIT,DI:8-CHANNEL MUX,SEL | 80009 | 156-0513-02 |
| A20U644 | 156-0096-00 |  | MICROCIRCUIT,LI:VOLTAGE COMPARATOR | 27014 | LM311H |
| A20U654 | 156-0067-00 |  | MICROCIRCUIT,LI:OPERATIONAL AMPLIFIER | 01295 | MICROA741CP |
| A20U710 | 156-0509-00 |  | MICROCIRCUIT, DI:8-BIT BINARY,MULT CUR | 04713 | MC1408L8 |
| A20U800 | 156-0140-02 |  | MICROCIRCUIT, DI:HEX BUFFERS W/OC HV OUT | 01295 | SN7417 (NP3) |
| A20U820 | 156-0140-02 |  | MICROCIRCUIT, DI:HEX BUFFERS W/OC HV OUT | 01295 | SN7417 (NP3) |
| A20U840 | 156-0411-00 |  | MICROCIRCUIT,LI:QUAD-COMP,SGL SUPPLY | 27014 | LM339N |
| A20U900 | 156-0093.02 |  | MICROCIRCUIT,DI:HEX INV BUFFER,BURN-IN | 01295 | SN74LS00 (NP3) |
| A20U910 | 156-0140-02 |  | MICROCIRCUIT,DI:HEX BUFFERS W/OC HV OUT | 01295 | SN7417 (NP3) |
| A20U920 | 156-0140-02 |  | MICROCIRCUIT,DI:HEX BUFFERS W/OC HV OUT | 01295 | SN7417 (NP3) |
| A20U940 | 156-0411-00 |  | MICROCIRCUIT,LI:QUAD-COMP,SGL SUPPLY | 27014 | LM339N |
| A20VR210 | 152-0281-00 |  | SEMICOND DEVICE:ZENER,0.4W,22V.5\% | 12954 | 1 N969B |
| A20VR430 | 152-0195-00 |  | SEMICOND DEVICE:ZENER,0.4W, 5.1V.5\% | 04713 | SZ11755 |
| A20VR434 | 152-0195-00 |  | SEMICOND DEVICE:ZENERT,0.4W, 5.1V.5\% | 04713 | SZ11755 |
| A20VR532 | 152-0127-00 |  | SEMICOND DEVICE:ZENER, $0.4 \mathrm{~W}, 7.5 \mathrm{~V}, 5 \%$ | 04713 | SZG35009к2 |




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| Component No. | Part No. Eff | Dscont | Name \& Description | Code | Mfr Part Number |


| A50 | 670-5039-00 | B010100 | B030586 | CKT BOARD ASSY:HORIZONTAL PROGRAMMING | 80009 | 670-5039-00 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A50 | 670-5039-01 | B030587 | B030761 | CKT BOARD ASSY:HORIZONTAL PROGRAMMING | 80009 | 670-5039-01 |
| A50 | 670-5039-02 | B030762 | B030837 | CKT BOARD ASSY:HORIZONTAL PROGRAMMING | 80009 | 670-5039-02 |
| A50 | 670-5039-03 | B030838 | B030844 | CKT BOARD ASSY:HORIZONTAL PRCIGRAMMING | 80009 | 670-5039-03 |
| A50 | 670-5039-04 | B030845 | B031149 | CKT BOARD ASSY:HORIZONTAL PRCIGRAMMING | 80009 | 670-5039-04 |
| A50 | 670-5039-05 | B031150 |  | CKT BOARD ASSY:HORIZONTAL PROGRAMMING | 80009 | 670-5039-05 |
| A50C000 | 283-0010-00 |  |  | CAP.,FXD, CER DI:0.05UF, $+100-20 \%, 50 \mathrm{~V}$ | 56289 | 1C1025U5032050B |
| A50C020 | 283-0644-00 |  |  | CAP, FXX, MICA D: $150 \mathrm{PF}, 1 \%, 500 \mathrm{~V}$ | 00853 | D155F151F0 |
| A50C024 | 283-0060-00 |  |  | CAP.,FXD,CER DI: $100 \mathrm{PF}, 5 \%, 200 \mathrm{~V}$ | 59660 | 855-535U2J101J |
| A50C030 | 283-0644-00 |  |  | CAP.,FXD,MICA D: 150 PF, $1 \%$, 500 V | 00853 | D155F151F0 |
| A50C032 | 283-0060-00 |  |  | CAP.,FXD,CER DI: $100 \mathrm{PF}, 5 \%, 200 \mathrm{~V}$ | 59660 | 855-535U2J101J |
| A50C100 | 283-0010-00 |  |  | CAP.,FXD, CER DI:0.05UF, + 100-20\%,50V | 56289 | 1C10Z5U5032050B |
| A50C220 | 283-0010-00 |  |  | CAP.,FXD, CER DI: $0.05 \mathrm{UF},+100-20 \%, 50 \mathrm{~V}$ | 56289 | 1C10z5U503z050B |
| A50C410 | 283-0032-00 |  |  | CAP.,FXD,CER DI: 470 PF, $5 \%, 500 \mathrm{~V}$ | 59660 | 0831085Z5E00471J |
| A50C420 | 283-0010-00 |  |  | CAP.,FXD, CER DI:0.05UF. $+100-20 \%, 5 \mathrm{~V}$ | 56289 | 1C10Z5U503z050B |
| A50C528 | 283-0032-00 |  |  | CAP.,FXD,CER DI:470PF, $5 \%$,500V | 59660 | 0831085Z5E00471J |
| A50C530 | 283-0010-00 | B010100 | B030837 | CAP.,FXD, CER DI:0.05UF, $+100-20 \%, 5 \mathrm{JV}$ | 56289 | 1C10Z5U503Z050B |
| A50C628 | 283-0010-00 |  |  | CAP.,FXD.CER DI:0.05UF, $+100-20 \%, 55 \mathrm{~V}$ | 56289 | 1C10z5U503z050B |
| A50C728 | 283-0010-00 | B010100 | B030837 | CAP.,FXD, CER DI:0.05UF, $+100-20 \%, 5 \mathrm{JV}$ | 56289 | 1C1025u503z050b |
| A50C800 | 283-0000-00 |  |  | CAP.,FXD, CER DI:0.001UF, $+100-0 \%, 530 \mathrm{~V}$ | 59660 | 831610Y5U0102P |
| A50C1020 | 283-0010-00 |  |  | CAP.,FXD,CER DI:0.05UF, +100-20\%,5JV | 56289 | 1C1025U503z050b |
| A50C1100 | 283.0010-00 |  |  | CAP, FXD,CER DI:0.05UF, $+100-20 \%$,5JV | 56289 | 1C1025U503z050B |
| A50C1106 | 283-0010-00 |  |  | CAP, FXD, CER DI:0.05UF, $+100-20 \%, 5 \mathrm{~V}$ | 56289 | 1C10z5U503z050B |
| A50C1120 | 283-0010-00 |  |  | CAP.,FXD,CER DI:0.05UF, $+100-20 \%$,5JV | 56289 | 1C1025U503z050в |
| A50C1300 | 283-0010-00 |  |  | CAP.,FXD, CER DI:0.05UF, $+100-20 \%, 5 \mathrm{~V}$ | 56289 | 1C10Z5U503z050B |
| A50C1400 | 283-0060-00 |  |  | CAP.,FXD,CER DI:100PF,5\%,200V | 59660 | 855-535U2J101J |
| A50C1516 | 290-0746-00 |  |  | CAP.,FXD,ELCTLT:47UF, $+50-10 \%, 16 \mathrm{~V}$ | 55680 | ULA1C470TEA |
| A50C1518 | 283-0000-00 |  |  | CAP.,FXD,CER DI:0.001UF, $+100-0 \%, 500 \mathrm{~V}$ | 59660 | 831610Y5U0102P |
| A50C1520 | 283-0010-00 |  |  | CAP, FXD, CER DI:0.05UF, +100-20\%,50V | 56289 | 1C10Z5U503Z050B |
| A50C1530 | 283-0119-00 |  |  | CAP.,FXD, CER DI:2200PF,5\%,200V | 59660 | 855-536Y5E0222J |
| A50C1620 | 290-0745-00 |  |  | CAP.,FXD,ELCTLT: $22 \mathrm{UF},+50-10 \%, 25 \mathrm{~V}$ | 54473 | ECE-A25V22L. |
| A50C1622 | 290-0745-00 |  |  | CAP.,FXD,ELCTLT:22UF, $+50-10 \%, 25 \mathrm{~V}$ | 54473 | ECE-A25V22L |
| A50C1630 | 283-0119-00 | B010100 | B030586 | CAP.,FXD,CER DI:2200PF, $5 \%$, 200 V | 59660 | 855-536Y5E0222J |
| A50C1630 | 283-0032-00 | B030587 |  | CAP.,FXD,CER DI:470PF,5\%,500V | 59660 | 0831085Z5E00471J |
| A50C1720 | 290-0746-00 | B010100 | B030837 | CAP.,FXD,ELCTLT:47UF, $+50-10 \%, 16 \mathrm{~V}$ | 55680 | ULA1C470TEA |
| A50C1730 | 283-0010-00 |  |  | CAP.,FXD, CER DI: $0.05 \mathrm{UF},+100-20 \%, 50 \mathrm{~V}$ | 56289 | 1C10Z5U503Z050B |
| A50CR234 | 152-0075-00 | B010100 | B030837 | SEMICOND DEVICE:SW,GE,22V,40MA | 14433 | G866 |
| A50CR800 | 152-0322-00 |  |  | SEMICOND DEVICE:SILICON, 15 V ,HOT CARRIER | 50434 | 5082-2672 |
| A50CR1310 | 152-0141-02 |  |  | SEMICOND DEVICE:SILICON,30V,150MA | 01295 | 1N4152R |
| A50E730 | 276-0507-00 | B010100 | B030837 | Shiel.ding bead : FERRITE | 78488 | 57-3443 |
| A50LP1310 | 108-0543-00 |  |  | COIL,RF:FIXED,1.1UH | 80009 | 108-0543-00 |
| A50LR1612 | 108-0537-00 |  |  | COIL,RF:200UH | 80009 | 108-0537-00 |
| A50LR1614 | 108-0543-00 |  |  | COIL,RF:FiXED, 1.1UH | 80009 | 108-0543-00 |
| A50LR1616 | 108-0543-00 |  |  | COIL,RF:FIXED, 1.1UH | 80009 | 108-0543-00 |
| A50P130 | 131-0608-00 |  |  | TERMINAL,PIN:0.365 L X 0.025 PH BRZZ GOLD | 22526 | 48283-036 |
| A50P232 | 131-0608-00 |  |  | TERMINAL,PIN: $0.365 \mathrm{~L} \times 0.025 \mathrm{PH}$ BR:Z GOLD | 22526 | $48283-036$ |
| A50P1526 | 131-0608-00 |  |  | TERMINAL,PIN: $0.365 \mathrm{~L} \times 0.025 \mathrm{PH}$ BR:Z GOLD | 22526 | 48283-036 |


| Component No. | Tektronix Part No. | Serial/M <br> Eff | del No. Dscont | Name \& Description | Vfr Sode | Mfr Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A50Q126 | 151-0216-00 |  |  | TRANSISTOR:SILICON,PNP | 04713 | SPS8803 |
| A50Q135 | 151-0216-00 |  |  | TRANSISTOR:SILICON,PNP | 194713 | SPS8803 |
| A50Q730 | 151-0190-00 | B010100 | B030837 | TRANSISTOR:SILICON,NPN | 07263 | S032677 |
| A50Q904 | 151-0190-00 |  |  | TRANSISTOR:SILICON,NPN | 17263 | 5032677 |
| A50Q1304 | 151-0190-00 |  |  | TRANSISTOR:SILICON,NPN | 1)7263 | S032677 |
| A50Q1308 | 151-0190-00 |  |  | TPANSISTOR:SILICON,NPN | 07263 | S032677 |
| A50Q1310 | 151-0190-00 |  |  | TRANSISTOR:SILICON,NPN | 07263 | S032677 |
| A50Q1404 | 151-0190-00 |  |  | TRANSISTOR:SILICON,NPN | 07263 | S032677 |
| A50Q1412 | 151-0216-00 |  |  | TRANSISTOR:SILICON.PNP | 04713 | SPS8803 |
| A50Q1514 | 151-0221-00 |  |  | TRANSISTOR:SILICON.PNP | 04713 | SPS246 |
| A50R020 | 315-0222-00 |  |  | RES.,FXD,CMPSN:2.2K OHM, 5\%,0.25W | $0 \dagger 121$ | CB2225 |
| A50R022 | 311-1267-00 |  |  | RES.,VAR,NONWIR:5K OHM, 10\%,0.50W | 32997 | 3329P-L58-502 |
| A50R024 | 315-0221-00 |  |  | RES.,FXD,CMPSN:220 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2215 |
| A50R030 | 315-0222-00 |  |  | RES.,FXD,CMPSN: 2.2 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2225 |
| A50R032 | 311-1267-00 |  |  | RES.,VAR,NONWIR:5K OHM, $10 \%, 0.50 \mathrm{~W}$ | 32997 | 3329P-L58-502 |
| A50R034 | 315-0221-00 |  |  | RES.,FXD,CMPSN:220 OHM , $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2215 |
| A50R120 | 315-0102-00 |  |  | RES.,FXD,CMPSN: 1 K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A50R122 | 315-0102-00 |  |  | RES.,FXD,CMPSN: 1 K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A50R124 | 315-0220-00 |  |  | RES.,FXD,CMPSN:22 OHM, 5\%,0.25W | 01121 | CB2205 |
| A50R126 | 315-0102-00 |  |  | RES.,FXD,CMPSN: 1 K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A50R130 | 315-0102-00 |  |  | RES.,FXD,CMPSN:1K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A50R132 | 315-0102-00 |  |  | RES.,FXD,CMPSN:1K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A50R134 | 315-0220-00 |  |  | RES.,FXD,CMPSN: 22 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2205 |
| A50R136 | 315-0103-00 |  |  | RES.,FXD,CMPSN: 10 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A50R220 | 315-0102-00 |  |  | RES.,FXD,CMPSN:1K OHM, $5 \%, 0.25 \mathrm{~W}$ | $(1121$ | CB1025 |
| A50R230 | 321-0324-00 | B010100 | B030837 | RES.,FXD,FILM:23.2K OHM $, 1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G23201F |
| A50R232 | 321-0705-00 | B010100 | B030837 | RES.,FXD,FILM: 41.7 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G41701F |
| A50R234 | 321-0341-00 |  |  | RES.,FXD,FILM 34.8 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | CMF55116G34801F |
| A50R235 | 321-0354-00 |  |  | RES.,FXD,FILM: 47.5 K OHM, $1 \%, 0.125 \mathrm{~W}$ | ${ }_{6} 1637$ | CMF55116G47501F |
| A50R332 | 315-0681-00 |  |  | RES.,FXD,CMPSN: 680 OHM, $5 \%, 0.25 \mathrm{~W}$ | 011121 | CB6815 |
| A50R334 | 321-0378-00 | B010100 | B030837 | RES.,FXD,FILM:84.5K OHM, $1 \%, 0.125 \mathrm{~W}$ | $¢ 1637$ | MFF1816G84501F |
| A50R336 | 321-0334-00 | B010100 | B030837 | RES.,FXD,FILM:29.4K OHM, $1 \%, 0.125 \mathrm{~W}$ | ¢1637 | MFF1816G29401F |
| A50R4 10 | 315-0103-00 |  |  | RES.,FXD,CMPSN:10K OHM, $5 \%, 0.25 \mathrm{~W}$ | C1121 | CB1035 |
| A50R430 | 315-0104-00 |  |  | RES.,FXD,CMPSN: 100 K OHM, $5 \%, 0.25 \mathrm{~W}$ | C1121 | CB1045 |
| A50R530 | 315-0331-00 |  |  | RES.,FXD,CMPSN:330 OHM $, 5 \%, 0.25 \mathrm{~W}$ | C1121 | CB3315 |
| A50R532 | 315-0103-00 |  |  | RES.,FXD,CMPSN: 10 K OHM $, 5 \%, 0.25 \mathrm{~W}$ | ¢1121 | CB1035 |
| A50R534 | 315-0103-00 |  |  | RES.,FXD, CMPSN: 10 K OHM, $5 \%, 0.25 \mathrm{~W}$ | C1121 | CB1035 |
| A50R536 | 315-0433-00 |  |  | RES.,FXD,CMPSN:43K OHM, $5 \%, 0.25 \mathrm{~W}$ | C1121 | CB4335 |
| A50R538 | 315-0103-00 |  |  | RES.,FXD,CMPSN:10K OHM, $5 \%, 0.25 \mathrm{~W}$ | C1121 | CB1035 |
| A50R539 | 315-0102-00 |  |  | RES.,FXD,CMPSN: 1 K OHM, $5 \%, 0.25 \mathrm{~W}$ | C1121 | CB1025 |
| A50R730 | 315-0102-00 | B010100 | B010160 | RES.,FXD,CMPSN:1K OHM, $5 \%, 0.25 \mathrm{~W}$ | C1121 | CB1025 |
| A50R730 | 315-0331-00 | B010161 | B030837 | RES.,FXD,CMPSN:330 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB3315 |
| A50R800 | 321-0344-00 |  |  | RES.,FXD,FILM: 37.4 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G37401F |
| A50R810 | 315-0473-00 |  |  | RES.,FXD,CMPSN:47K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4735 |
| A50R812 | 315-0683-00 |  |  | RES.,FXD,CMPSN:68K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB6835 |
| A50R900 | 315-0102-00 |  |  | RES.,FXD, CMPSN:1K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A50R1102 | 315-0102-00 |  |  | RES.,FXD.CMPSN:1K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A50R1104 | 315-0102-00 |  |  | RES.,FXD,CMPSN: 1 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A50R1200 | 315-0102-00 |  |  | RES.,FXD,CMPSN:1K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A50R1204 | 315-0102-00 |  |  | RES.,FXD, CMPSN:1K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A50R1206 | 315-0102-00 |  |  | RES.,FXD,CMPSN:1K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A50R1220 | 315-0103-00 |  |  | RES.,FXD,CMPSN: 10 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A50R1300 | 315-0753-00 |  |  | RES.,FXD,CMPSN:75K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB7535 |


|  | Tektronix | Serial/Model No. |  | Name \& Description | Mfr Code | Mir Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Component No. | Part No. | Eff | Dscont |  |  |  |
| A50R1302 | 315-0154-00 |  |  | RES.,FXD,CMPSN:150K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1545 |
| A50R1304 | 315-0103-00 |  |  | RES.,FXD,CMPSN:10K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A50R1306 | 315-0103-00 |  |  | RES.,FXD,CMPSN:10K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A50R1310 | 315-0103-00 |  |  | RES.,FXD,CMPSN:10K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A50R1400 | 321-0344-00 |  |  | RES.,FXD, FILM: 37.4 K OHM, $1 \%, 0.125 \mathrm{~W}$ | 91637 | MFF1816G37401F |
| A50R1402 | 315-0753-00 |  |  | RES.,FXD, CMPSN:75K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB7535 |
| A50R1404 | 315-0513.00 |  |  | RES.,FXD,CMPSN: 51 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB5135 |
| A50R1406 | 315-0222-00 |  |  | RES.,FXD,CMPSN:2.2K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2225 |
| A50R1408 | 315-0154-00 |  |  | RES.,FXD,CMPSN: 150 K OHM $, 5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1545 |
| A50R1410 | 315-0102-00 |  |  | RES.,FXD,CMPSN:1K OHM, $5 \%, 0,25 \mathrm{~W}$ | 01121 | CB1025 |
| A50R1412 | 315-0103-00 |  |  | RES.,FXD,CMPSN: 10 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A50R1514 | 315-0472-00 |  |  | RES.,FXD,CMPSN:4.7K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB4725 |
| A50R1526 | 315-0103-00 |  |  | RES.,FXD,CMPSN: 10 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1035 |
| A50R1532 | 315-0203-00 |  |  | RES.,FXD,CMPSN:20K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB2035 |
| A50R1620 | 315-0101-00 |  |  | RES.,FXD,CMPSN: 100 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1015 |
| A50R1630 | 315-0133-00 | B010100 | B030586 | RES.,FXD,CMPSN: 13 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1335 |
| A50R1630 | 315-0103-00 | B030587 |  | RES.,FXD,CMPSN:10K OHM,5\%,0.25W | 01121 | CB1035 |
| A50R1730 | 315-0102-00 |  |  | RES.,FXD,CMPSN: 1 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A50R1732 | 315-0102-00 |  |  | RES.,FXD,CMPSN: 1 K OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1025 |
| A50U000 | 156-0382-02 |  |  | MICROCIRCUIT DI:QUAD 2-INP NAND GATE | 01295 | SN74LS00 |
| A50U010 | 156-0480-02 |  |  | MICROCIRCUIT,DI:QUAD 2 INP \& GATE | 01295 | SN74LS08NP3 |
| A50U030 | 156-0733-02 |  |  | MICROCIRCUIT, DI:DUAL MONOSTABLE MV,SCRN | 01295 | SN74LS221N3 |
| A50U100 | 156-0479-02 |  |  | MICROCIRCUIT, DI:QUAD 2-INP OR GATE | 01295 | SN74LS32NP3 |
| A50U110 | 156-0736-02 |  |  | MICROCIRCUIT,DI:BCD TO DECIMAL DCDR | 80009 | 156-0736-02 |
| A50U200 | 156-0541-02 |  |  | MICROCIRCUIT, DI:DUAL 2 TO 4 LINE DCDR | 01295 | SN74LSt39NP3 |
| A50U230 | 156-0093-02 |  |  | MICROCIRCUIT,DI:HEX INV BUFFER,BUFIN-IN | 01295 | SN74LS00 (NP3) |
| A50U310 | 156-0464-02 |  |  | MICROCIRCUIT,DI:DUAL 4 INP NAND GATE | 01295 | SN74LS20 |
| A50U420 | 156-0382-02 |  |  | MICROCIRCUIT, DI:QUAD 2-INP NAND GATTE | 01295 | SN74LS00 |
| A50U430 | 156-0411-00 |  |  | MICROCIRCUIT,LI:QUAD-COMP,SGL SUPPLY | 27014 | LM339N |
| A50U500 | 156-0480-02 |  |  | MICROCIRCUIT, DI:QUAD 2 INP \& GATE | 01295 | SN74LS08NP3 |
| A50U510 | 156-0427-00 | B010100 | B030629 | MICROCIRCUIT,DI:PERIPHERAL INTERFACE ADPTR | 04713 | MC6820(L OR P) |
| A50U510 | 156-0427-04 | B030630 |  | MICROCIRCUIT, DI:PERIPHERAL INTERFACE ADPTR | 80009 | 156-0427-04 |
| A50U530 | 156-0467-02 |  |  | MICROCIRCUIT, DI:QUAD 2-INP NAND BFR,SCRN | 01295 | SN74LS38 |
| A50U630 | 156-0541-02 |  |  | MICROCIRCUIT, DI:DUAL 2 TO 4 LINE DCDR | 01295 | SN74LS139NP3 |
| A50U710 | 156-0427-00 | B010100 | B030629 | MICROCIRCUIT, DI:PERIPHERAL INTERFACE ADPTR | 04713 | MC6820(L OR P) |
| A50U710 | 156-0427-04 | B030630 |  | MICROCIRCUIT,DI:PERIPHERAL INTERFF.CE ADPTR | 80009 | 156-0427-04 |
| A50U810 | 156-0427-00 | B010100 | B030629 | MICROCIRCUIT,DIPERIPHERAL INTERFF.CE ADPTR | 04713 | MC6820(L OR P) |
| A50U810 | 156-0427-04 | B030630 |  | MICROCIRCUIT, OI:PERIPHERAL INTERFF:CE ADPTR | 80009 | 156-0427-04 |
| A50U830 | 156-0978-00 | B010100 | 8030761 | MICROCIRCUIT,DI: $2048 \times 8$ ROM,CUSTAIM MASK | 80009 | 156-0978-00 |
| A50U830 | 160-1514-00 | B030838 |  | MICROCIRCUIT,DI:2048 $\times 8$ EPROM, PRO 3 RAMMED | 80009 | 160-1514-00 |
| A50U1000 | 155-0135-00 |  |  | MICROCIRCUIT,DI:DUAL. IN LINE,20 LEAC) | 80009 | 155-0135-00 |
| A50U1005 | 156-0656-02 |  |  | MICROCIRCUIT,DI:DECADE COUNTER,BURN-IN | 01295 | SN74LS90 |
| A50U1030 | 156-0979-00 | B010100 | B030761 | MICROCIRCUIT,DI: $2048 \times 8$ ROM,CUSTOM MASK | 80009 | 156-0979-00 |
| A50U1030 | 160-1513-00 | B030838 |  | MICROCIRCUIT,DI:2048 $\times 8$ EPROM, PROGRAMMED | 80009 | 160-1513-00 |
| A50U1100 | 156-0736-02 |  |  | MICROCIRCUIT,DI:BCD TO DECIMAL DCIDR | 80009 | 156-0736-02 |
| A50U1105 | 156-0093-02 |  |  | MICROCIRCUIT,DI:HEX INV BUFFER,BURN-IN | 01295 | SN74LS00 (NP3) |
| A50U1110 | 156-0427-00 | B010100 | B030629 | MICROCIRCUIT,DI:PERIPHERAL INTERFACE ADPTR | 04713 | MC6820(L OR P) |
| A50U1110 | 156-0427-04 | B030630 |  | MICROCIRCUIT, DI:PERIPHERAL INTERFACE ADPTR | 80009 | 156-0427-04 |
| A50U1130 | 156-0716-00 |  |  | MICROCIRCUIT,DI:RAM, $128 \times 8$ STATIC | 04713 | MCM6810S |
| A50U1220 | 156-0426-00 | B010100 | B020511 | MICROCIRCUIT,DI:MICROPROCESSOR | 04713 | MC6800S |
| A50U1220 | 156-0426-05 | B020512 |  | MICROCIRCUIT, Di:MICROPROCESSOR,SCREENED | 80009 | 156-0426-05 |
| A50U1310 | 156-0541-02 |  |  | MICROCIRCUIT,DI:DUAL 2104 LINE DCDR | 01295 | SN74LS139NP3 |


|  | Tektronix | Serial/Model No. |  | Name \& Description | Mfr |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Component No. | Part No. | Eff | Dscont |  | Code | Mfr Part Number |
| A50U1420 | 156-0427-00 | 8010100 | B030629 | MICROCIRCUIT,DI:PERIPHERAL INTERFACE ADPTR | 04713 | MC6820(L. OR P) |
| A50U1420 | 156-0427-04 | 8030630 |  | MICROCIRCUIT,DI:PERIPHERAL INTERFACE ADPTR | 80009 | 156-0427-04 |
| A50U1500 | 155-0164-00 |  |  | MICROCIRCUIT,DI:TIME SLOT SWITCH | 80009 | 155-0164-00 |
| A50U1520 | 156-0645-00 |  |  | MICROCIRCUIT,DI:EX INV ST NAND GATES,SCRN | 01295 | SN741LS14(NP3) |
| A50U1530 | 156-0405-03 |  |  | MICROCIRCUIT,DI:DUAL RETRIG MONOSTABLE MV | 07263 | 9602 |
| A50U1620 | 156-0479-02 |  |  | MICROCIRCUIT, DI:QUAD 2-INP OR GATE | 01295 | SN74LS32NP3 |
| A50U1630 | 156-0382-02 |  |  | MICROCIRCUIT, DI:QUAD 2-INP NAND GATE | 01295 | SN74LS00 |
| A50U1720 | 156-0467-02 |  |  | MICROCIRCUIT,DI:QUAD 2-INP NAND BFR,SCRN | 01295 | SN74LS38 |
| A50U1730 | 156-0388-03 |  |  | MICROCIRCUIT, DI:DUAL D FLIP-FLOP | 07263 | 74LS74A |
| A50VR1204 | 152-0590-00 |  |  | SEMICOND DEVICE:ZENER,18V,5\% AI 7MA | 04713 | SZG35014K2 |
| A50VR1206 | 152-0590-00 |  |  | SEMICOND DEVICE:ZENER, $18 \mathrm{~V}, 5 \%$ AT 7MA | 04713 | SZG35014K2 |
| A50VR1300 | 152-0590-00 |  |  | SEMICOND DEVICE:ZENER, $18 \mathrm{~V}, 5 \%$ AT 7MA | 04713 | SZG35014K2 |
| A50VR330 | 152-0437-00 | B010100 | $B 031149$ | SEMICOND DEVICE:ZENER, $51,8.2 \mathrm{~V}, 2 \%, 0.4 \mathrm{~W}$ | 14552 | TD332679 |
| A50VR330 | 152-0127-00 | B031150 |  | SEMICOND DEVICE:ZENER, $0.4 \mathrm{~W}, 7.5 \mathrm{~V}, 5 \%$ | 04713 | SZG35009K2 |
| A50VR730 | 152-0508-00 | B010100 | B030837 | SEMICOND DEVICE:ZENER,0.4W, 12.6V,5\% | 80009 | 152-0508-00 |
| A50VR810 | 152-0217-00 |  |  | SEMICOND DEVICE:ZENER,0.4W,8.2V,5\% | 04713 | SZG20 |
| A50W320 | 131-0566-00 |  |  | BUS CONDUCTOR:DUMMY RES,2.375,22 AWG | 57668 | JWW-0200E0 |
| A50W635 | 131-0566-00 | B030837 |  | BUS CONDUCTOR:DUMMY RES, 2.375,22 AWG | 57668 | JWW-0200E0 |
| A50W730 | 131-0566-00 | B030837 |  | BUS CONDUCTOR:DUMMY RES, 2.375,22 AWG | 57668 | JWW-0200E0 |
| A50W735 | 131-0566-00 | B030837 |  | BUS CONDUCTOR:DUMMY RES, 2.375,22 AWG | 57668 | JWW-0200E0 |
| A50W1720 | 131-0566-00 | B010100 | 8030837 | BUS CONDUCTOR:DUMMY RES,2.375,22 AWG | 57668 | JWW-0200E0 |


| Component No. | Tektronix | Serial/Model No. |  | Name \& Description | Mfr |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Part No. | Eff | Dscont |  | Code | Mfr Part Number |
|  |  |  |  | A51 MEMORY ADAPTER |  |  |
| A51 | 670-7549-00 | B030762 | B030837 | CKT BOARD ASSY:MEMORY ADAFTER | 80009 | 670-7549-00 |
| A51U830 | 160-1514-00 | B030762 | B030837 | MICROCIRCUIT,DI:2048 $\times 8$ EPROM,PROGRAMMED | 80009 | 160-1514-00 |
| A51U1030 | 160-1513-00 | B030762 | B030837 | MICROCIRCUIT,DI:2048 $\times 8$ EPROM,PROGRAMMED | 80009 | 160-1513-00 |
|  |  |  |  | CHASSIS PARTS |  |  |
| C5402 | 281-0621-00 |  |  | CAP.,FXD, CER DI: $12 \mathrm{PF}, 1 \%, 500 \mathrm{~V}$ | 59660 | 301-080 COGO-12F |
| C5412 | 281-0593-00 |  |  | CAP.,FXD,CER DI:3.9PF, $10 \%, 500 \mathrm{~V}$ | 04222 | 7001-C0J-3R9C |
| E5412 | 276-0507-00 |  |  | SHIELDING BEAD,:FERRITE | 78488 | 57-3443 |
| J001 | 131-0955-00 |  |  | CONN,RCPT,ELEC:BNC,FEMALE | 13511 | 31-279 |
| L5412 | 108-0170-01 |  |  | COIL,RF:FIXED,36ONH | 80009 | 108-0170-01 |
| R5400 | 321-0481-00 |  |  | RES.,FXD,FILM:1M OHM, 1\%,0.125W | 91637 | CMF55116G10003F |
| R5402 | 315-0910-00 |  |  | RES.,FXD,CMPSN: 91 OHM, $5 \%, 0.25 \mathrm{~N}$ | 01121 | CB9105 |
| R5410 | 315-0181-00 |  |  | RES.,FXD,CMPSN: 180 OHM, $5 \%, 0.25 \mathrm{~W}$ | 01121 | CB1815 |
| R5412 | 315-0510-00 |  |  | RES , FXD CMPSN: 51 OHM, $5 \%, 0.25 \mathrm{~N}$ | 01121 | CB5105 |
| R5414 | 315-0680-00 |  |  | RES.,FXD,CMPSN:68 OHM, $5 \%, 0.25 \mathrm{~N}$ | 01121 | CB6805 |
| S001 | 260-1666-00 |  |  | SWITCH,PUSH:1 STA,2 POLE,PUSH-PUSH | 71590 | 2KAB010000-674 |

## DIAGRAMS AND CIRCUIT BOARD ILLUSTRATIONS

## Symbols and Reference Designators

Electrical components shown on the diagrams are in the following units unless noted otherwise:

| Capacitors $=$ | Values one or greater are in picofarads $(\mathrm{pF})$. |
| :--- | :--- |
|  | Values less than one are in microfarads $(\mu \mathrm{F})$. |
| Resistors $=$ | Ohms $(\Omega)$. |

Graphic symbols and class designation letters are based on ANSI Standard Y32.2-1975.
Logic symbology is based on ANSI Y32.14-1973 in terms of positive logic. Logic symbols depict the logic function performed and may differ from the manufacturer's data.
The overline on a signal name indicates that the signal performs its intended function when it goes to the low state. Abbreviations are based on ANSI Y1.1-1972.
Other ANSI standards that are used in the preparation of diagrams by Tektronix, Inc. are:
Y14.15, 1966 Drafting Practices.
Y14.2, 1973 Line Conventions and Lettering.
Y10.5, 1968 Letter Symbols for Quantities Used in Electrical Science and Electrical Engineering.

The following prefix letters are used as reference designators to identify components or assemblies on the diagrams.

| A | Assembly, separable or repairable <br> (circuit board, etc) |
| :--- | :--- |
| AT | Attenuator, fixed or variable |
| B | Motor |
| BT | Battery |
| C | Capacitor, fixed or variable |
| CB | Circuit breaker |
| CR | Diode, signal or rectifier |
| DL | Delay line |
| DS | Indicating oevice (lamp) |
| E | Spark Gap, Ferrite bead |
| F | Fuse |
| FL | Filter |


| H | Heat dissipating device (heat sink, <br> heat radiator, etc) |
| :--- | :--- |
| HR | Heater |
| HY | Hybrid circuit |
| J | Connector, stationary portion |
| K | Relay |
| L | Inductor, fixed or variable |
| M | Meter |
| P | Connector, movable portion |
| Q | Transistor or silicon-controlled |
|  | rectifier |
| R | Resistor, fixed or variable |
| RT | Thermistor |


| S | Switch or contactor |
| :--- | :--- |
| T | Transformer |
| TC | Thermocouple |
| TP | Test point |
| U | Assembly, inseparable or non-repairable <br>  <br> (integrated circuit, etc.) |
| V | Electron tube |
| VR | Voltage regulator (zener diode, etc.) |
| W | Wirestrap or cable |
| Y | Crystal |
| Z | Phase shifter |

Resistor, fixed or variabl
Thermistor
The following special symbols may appear on the diagrams:



Fig. 8-1. P/O Trigger Board A10.
*See Parts List for serial number ranges.

P/O TRIGGER BOARD A10

| COMPONENT" NUMBER | SCHEMATIC LOCATION |  | BOARD LOCATION |  | COMPONENT NUMBEA | SCHEMATIC LOCATION |  | BOARD LOCATION |  | COMPONENT NUMBER | SCHEMATIC LOCATION |  | BOARD LOCATION |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | COL | ROW | COL | ROW |  | COL | ROW | COL | ROW |  | COL | ROW | COL | ROW |
| C020 | E | 5 | 0 | 2 | CR330 | D | 6 | 3 | 3 | R122. | F | 4 | 1 | 2 |
| C022 | E | 4 | 0 | 2 | CR332 | D | 6 | 3 | 3 | R124 | F | 4 | 1 | 2 |
| C024 | F | 4 | 0 | 2 | CR334 | D | 5 |  | 3 | R125 | F | 4 | 1 | 2 |
| C040 | E | 6 | 0 | 4 | CR336 | D | 5 |  | 3 | R126 | F | 4 | 1 | 2 |
| C050 | E | 6 | 0 | 5 | CR340 | D | 6 |  | 4 | R127 | G | 3 | 1 | 2 |
| C106 | D | 1 | 1 | 0 | CR342 | D | 5 | 3 | 4 | R128 | G | 3 | 1 | 2 |
| C110 | B | 1 | 1 | 1 | CR344 | D | 5 | 3 | 4 | R130 | E | 4 | 1 | 3 |
| C112 | E | 1 | 1 | 1 | CR346 | D | 5 | 3 | 4 | R132 | F | 5 | 1 | 3 |
| C114 | B | 1 | 1 | 1 | CR400 | C | 3 | 4 | 0 | R134 | G | 3 | 1 | 3 |
| C116 | G | 3 | 1 | 2 | CR402 | C | 3 | 4 | 0 | R135 | G | 4 | 1 | 3 |
| C130 | E | 5 | 1 | 3 | CR460 | E | 3 | 4 | 6 | R136 | G | 5 | 1 | 3 |
| C132 | F | 5 | 1 | 3 | CR462 | E | 3 | 4 | 6 | R137 | G | 5 | 1 | 3 |
| C137 | G | 5 | 1 | 3 | CR464 | C | 4 | 5 | 6 | R140 | F | 5 | 1 | 4 |
| C140 | E | 5 | 1 | 4 | CR466 | C | 4 | 5 | 5 | R142 | E | 4 | 1 | 4 |
| C146 | G | 5 | 1 | 4 | DS348 | D | 5 | 3 | 4 | R144 | G | 5 | 1 | 4 |
| C148 | G | 6 | 1 | 4 | DS440 | D | 6 | 4 | 4 | R146 | G | 6 | 1 | 4 |
| C150 | F | 6 | 1 | 5 | E5412 | A | 4 | 4 | 1 | R148 | $E$ | 5 | 1 | 4 |
| C152 | F | 5 | 1 | 5 | J000 | A | 1 | 0 | 0 | R158 | G | 4 | 1 | 5 |
| C158 | F | 3 | 1 | 5 | J212 | A | 1 | 2 | 1 | R210 | B | 1 | 2 | 1 |
| C214 | H | 2 | 2 | 1 | J300 | B | 4 | 3 | $0 \dagger$ | R212 | B | 1 | 2 | 1 |
| C220 | B | 1 | 2 | 2 | J436 | B | 4 | 4 | 3 | R213 | A | 1 | 2 | 1 |
| C222 | F | 4 | 2 | 2 | J5001 | A | 4 | CHASSI |  | R214 | C | 1 | 2 | 1 |
| C224 | B | 3 | 2 | 2 | L5412 | A | 4 | 4 | 1 | R215 | B | 2 | 2 | 1 |
| C230 | F | 3 | 2 | 3 | LR230 | F | 5 | 2 | 3 | R216 | C | 3 | 2 | 1 |
| C236 | G | 4 | 2 | 3 | LR232 | G | 5 | 2 | 3 | R218 | B | 1 | 2 | 1 |
| C240 | G | 3 | 2 | 4 | LR310 | H | 2 | 3 | 1 | R220 | C | 2 | 2 | 2 |
| C308 | G | 3 | 3 | 0 | LR314 | H | 2 | 3 | 1 | R234 | G | 4 | 2 | 3 |
| C312 | B | 2 | 3 | 1 | LR320 | H | 3 | 3 | 2 | R236 | G | 4 | 2 | 3 |
| C322 | G | 3 | 3 | 2 | LR410 | H | 2 | 4 | 1 | R238 | G | 4 | 2 | 3 |
| C324 | C | 5 | 3 | 2 | P550 | B | 4 | 5 | 5 | R240 | G | 3 | 2 | 4 |
| C326 | C | 5 | 3 | 2 | Q128 | G | 5 | 1 | 2 | R243 | G | 4 | 2 | 4 |
| C336 | D | 5 | 3 | 3 | Q134 | E | 4 | 1 | 3 | R252 | H | 3 | 2 | 5 |
| C344 | D | 5 | 3 | 4 | Q326 | C | 5 | 3 | 2 | R253 | H | 4 | 2 | 5 |
| C346 | D | 5 | 3 | 4 | Q327 | C | 5 | 3 | 3 | R254 | H | 4 | 2 | 5 |
| C348 | C | 5 | 3 | 4 | Q439 | D | 5 | 4 | 3 | R32? | A | 3 | 3 | 2 |
| C360* | G | 3 | 3 | 6 | Q440 | D | 5 | 4 | 4 | R324 | C | 5 | 3 | 2 |
| C400 | H | 2 | 4 | 0 | Q530 | B | 5 | 5 | 3 | R330 | E | 3 | 2 | 2 |
| C422 | D | 3 | 4 | 2 | Q532 | B | 6 | 5 | 3 | R332 | E | 3 | 3 | 3 |
| C424 | B | 5 | 4 | 2 | R002 | A | 1 | 0 | 0 | R336 | 0 | 5 | 3 | 3 |
| C426 | B | 5 | 4 | 2 | R010 | C | 2 | 0 | 1 | R340 | D | 5 | 3 | 4 |
| C430 | D | 5 | 4 | 3 | Reso | E | 6 | 0 | 4 | R342 | D | 5 | 3 | 4 |
| C436 | B | 5 | 4 | 3 | R042 | E | 5 | 0 | 4 | R350 | F | 2 | 3 | 5 |
| C438 | B | 6 | 4 | 3 | R044 | $E$ | 5 | 0 | 4 | R352 | F | 2 | 3 | 5 |
| C440 | 8 | 6 | 4 | 4 | R050 | F | 6 | 0 | 5 | R.410 | B | 3 | 4 | 1 |
| C442 | C | 5 | 4 | 4 | R052 | $E$ | 6 | 0 | 5 | R414 | A | 3 | 4 | 1 |
| C448 | C | 4 | 4 | 4 | R100 | D | 1 | 1 | 0 | R420 | C | 5 | 4 | 2 |
| C462 | E | 2 | 4 | 6 | R102 | D | 1 | 1 | 0 | R422 | C | 5 | 4 | 2 |
| C540 | C | 4 | 5 | 4 | R103 | E | 1 | 1 | 0 | R424 | B | 5 | 4 | 2 |
| C5402 | B | 4 | 3 | 0 | R106 | D | 1 | 1 | 0 | R430 | D | 5 | 4 | 3 |
| C5412 | A | 4 | 4 | 0 | R110 | A | 1 | 1 | 1 | R432 | B | 6 | 4 | 3 |
| CR010 | G | 5 | 0 | 1 | R112 | E | 2 | 1 | 1 | R434 | B | 5 | 4 | 3 |
| CR022 | D | 2 | 0 | 2 | R114 | B | 1 | 1 | 1 | R436 | B | 5 | 4 | 3 |
| CR050 | F | 6 | 0 | 5 | R116 | G | 3 | 1 | 2 | R438 | B | 6 | 4 | 3 |
| CR140 | F | 5 | 1 | 4 | R120 | A | 1 | 1 | 2 | R442 | B | 6 | 4 | 4 |

†P/O LEO BOARD.

P/O TRIGGER BOARD A10

| COMPONENT NUMBER | SCHEMATIC LOCATION |  | BOARD LOCATION |  | COMPONENT NUMBER | SCHEMATIC LOCATION |  | $\begin{aligned} & \text { BOARD } \\ & \text { LOCATION } \end{aligned}$ |  | CONIPONENT NUMBER | SCHEMATIC LOCATION |  | BOARD LOCATION |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | COL | ROW | COL | ROW |  | COL | ROW | COL | ROW |  | COL | ROW | COL | ROW |
| CR330 | D | 6 | 3 | 3 | R122 | $F$ | 4 | 1 | 2 | R444 | B | 6 | 4 | 4 |
| CR332 | D | 6 | 3 | 3 | R124 | F | 4 |  | 2 | R448 | C | 4 | 4 | 4 |
| CR334 | D | 5 | 3 | 3 | R125 | F | 4 |  | 2 | R450 | F | 2 | 4 | 5 |
| CR336 | D | 5 | 3 | 3 | R126 | F | 4 |  | 2 | R452 | E | 3 | 4 | 5 |
| CR340 | D | 6 | 3 | 4 | R127 | G | 3 |  | 2 | R454 | D | 3 | 4 | 5 |
| CR342 | D | 5 | 3 | 4 | R128 | G | 3 |  | 2 | R522 | B | 5 | 5 | 2 |
| CR344 | D | 5 | 3 | 4 | R130 | E | 4 | 1 | 3 | R524 | B | 5 | 5 | 2 |
| CR346 | D | 5 | 3 | 4 | R132 | F | 5 | 1 | 3 | R528 | A | 6 | 5 | 2 |
| CR400 | C | 3 | 4 | 0 | R134 | G | 3 | 1 | 3 | R530 | B | 5 | 5 | 3 |
| CR402 | C | 3 | 4 | 0 | R135 | G | 4 | 1 | 3 | R532 | B | 6 | 5 | 3 |
| CR460 | E | 3 | 4 | 6 | R136 | G | 5 | 1 | 3 | R534 | B | 5 | 5 | 3 |
| CR462 | E | 3 | 4 | 6 | R137 | G | 5 | 1 | 3 | R536 | A | 5 | 5 | 3 |
| CR464 | C | 4 | 5 | 6 | R140 | F | 5 | 1 | 4 | 18537 | A | 5 | 5 | 4 |
| CR466 | C | 4 | 5 | 5 | R142 | E | 4 | 1 | 4 | R538 | A | 6 | 5 | 3 |
| DS348 | D | 5 | 3 | 4 | R144 | G | 5 | 1 | 4 | 12550 | C | 4 | 5 | 5 |
| DS440 | D | 6 | 4 | 4 | R146 | G | 6 | 1 | 4 | R552 | C | 4 | 5 | 5 |
| E5412 | A | 4 |  | 1 | R148 | E | 5 | 1 | 4 | 125400 | A | 4 | 3 | 0 |
| J000 | A | 1 | 0 | 0 | R158 | G | 4 | 1 | 5 | 1R5402 | A | 4 | 4 | 0 |
| J212 | A | 1 | 2 | 1 | R210 | B | 1 | 2 | 1 | 185410 | A | 4 | 4 | 1 |
| J300 | B | 4 | 3 | $0 \dagger$ | R212 | B | 1 | 2 | 1 | 185412 | A | 4 | 4 | 1 |
| J436 | B | 4 | 4 | 3 | R213 | A | 1 | 2 | 1 | R5414 | A | 4 | 4 | 1 |
| J5001 | A | 4 | CHASSI |  | R214 | C | 1 | 2 | 1 | TP060 | H | 5 | 0 | 6 |
| L5412 | A | 4 | 4 | 1 | R215 | B | 2 | 2 | 1 | TP220 | H | 3 | 2 | 2 |
| LR230 | F | 5 | 2 | 3 | R216 | C | 3 | 2 | 1 | TP460 | G | 2 | 4 | 6 |
| LR232 | G | 5 | 2 | 3 | R218 | B | 1 | 2 | 1 | TP462 | G | 2 | 4 | 6 |
| LR310 | H | 2 | 3 | 1 | R220 | C | 2 | 2 | 2 | TP464 | G | 3 | 4 | 6 |
| LR314 | H | 2 | 3 | 1 | R234 | G | 4 | 2 | 3 | TP466 | G | 2 | 4 | 6 |
| LR320 | H | 3 | 3 | 2 | R236 | G | 4 | 2 | 3 | U030A | E | 4 | 0 | 3 |
| LR410 | H | 2 | 4 | 1 | R238 | G | 4 | 2 | 3 | U030B | E | 6 | 0 | 3 |
| P550 | B | 4 | 5 | 5 | R240 | G | 3 | 2 | 4 | U106 | D | 1 | 1 | 0 |
| Q128 | G | 5 | 1 | 2 | R243 | G | 4 | 2 | 4 | U112A | B | 1 | 1 | 1 |
| Q134 | E | 4 | 1 | 3 | R252 | H | 3 | 2 | 5 | U112B | E | 1 | 1 | 1 |
| Q326 | C | 5 | 3 | 2 | R253 | H | 4 |  | 5 | U150A | F | 6 | 1 | 5 |
| Q327 | C | 5 | 3 | 3 | R254 | H | 4 | 2 | 5 | U150B | F | 5 | 1 | 5 |
| Q439 | D | 5 | 4 | 3 | R322 | A | 3 | 3 | 2 | U220 | F | 3 | 2 | 2 |
| Q440 | D | 5 | 4 | 4 | R324 | C | 5 | 3 | 2 | 1240 | G | 4 | 2 | 4 |
| Q530 | B | 5 | 5 | 3 | R330 | E | 3 | 2 | 2 | U310 | F | 1 | 3 | 1 |
| Q532 | B | 6 | 5 | 3 | R332 | E | 3 | 3 | 3 | U310A | B | 2 | 3 | 1 |
| R002 | A | 1 | 0 | 0 | R336 | D | 5 | 3 | 3 | L310B | B | 1 | 3 | 1 |
| R010 | C | 2 | 0 | 1 | R340 | D | 5 |  | 4 | U310C | B | 2 | 3 | 1 |
| R040 | E | 6 | 0 | 4 | R342 | D | 5 |  | 4 | U3100 | B | 2 | 3 | 1 |
| R042 | E | 5 | 0 | 4 | R350 | F | 2 |  | 5 | U350 | F | 2 | 3 | 5 |
| R044 | $E$ | 5 | 0 | 4 | R352 | F | 2 |  | 5 | L1358 | F | 2 | 3 | 5 |
| R050 | F | 6 | 0 | 5 | R410 | B | 3 | 4 | 1 | U410 | F | 1 | 4 | 1 |
| R052 | E | 6 | 0 | 5 | R414 | A | 3 | 4 | 1 | U410A | A | 3 | 4 | 1 |
| R100 | 0 | 1 | 1 | 0 | R420 | C | 5 | 4 | 2 | U410B | C | 2 | 4 | 1 |
| R102 | D | 1 | 1 | 0 | R422 | C | 5 | 4 | 2 | U410C | B | 3 | 4 | 1 |
| R103 | E | 1 | 1 | 0 | R424 | $B$ | 5 | 4 | 2 | U452 | C | 4 | 4 | 5 |
| R106 | D | 1 | 1 | 0 | R430 | D | 5 | 4 | 3 | U510 | F | 1 | 5 | 1 |
| R110 | A | 1 | 1 | 1 | R432 | B | 6 | 4 | 3 | U510A | D | 3 | 5 | 1 |
| R112 | E | 2 | 1 | 1 | R434 | B | 5 | 4 | 3 | U510B | D | 3 | 5 | 1 |
| R114 | B | 1 | 1 | 1 | R436 | B | 5 | 4 | 3 | U510C | D | 4 | 5 | 1 |
| R116 | G | 3 | 1 | 2 | P438 | B | 6 |  | 3 | U5100 | C | 3 | 5 | 1 |
| R120 | A | 1 | 1 | 2 | R442 | B | 6 |  | 4 |  |  |  |  |  |





Fig. 8-2. P/O Sweep Board A20.


P/O SWEEP BOARD A20

| COMPONENT NUMBER | SCHEMATIC <br> LOCATION |  | BOARD LOCATION |  | COMPONENT NUMBER | SCHEMATIC <br> LOCATION |  | BOARD LOCATION |  | COMPONENT NUMBER | SCHEMATIC <br> LOCATION |  | BOARD LOCATION |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | COL | ROW | COL | ROW |  | COL | ROW | COL | ROW |  | COL | ROW | COL | ROW |
| C030 | H | 1 | 0 | 3 |  |  |  |  |  |  |  |  |  |  |
| C032 | G | 1 | 0 | 3 | Q300 | B | 4 | 3 | 0 | R430 | G | 4 | 4 | 3 |
| C033 | G | 1 | 0 | 3 | Q312 | E | 3 | 3 | 1 | R432 | G | 4 | 4 | 3 |
| C034 | G | 3 | 0 | 3 | Q314 | E | 3 | 3 | 1 | R502 | C | 4 | 5 | 0 |
| C036 | G | 2 | 0 | 4 | Q400 | C | 5 | 4 | 0 | R505 | C | 4 | 5 | 0 |
| C040 | G | 2 | 0 | 4 | Q406 | D | 5 | 4 | 0 | R506 | D | 4 | 5 | 1 |
| C044 | G | 1 | 0 | 4 | Q410 | E | 3 | 4 | 1 | R512 | E | 4 | 5 | 1 |
| C045 | G | 2 | 0 | 5 | Q512 | C | 3 | 5 | 1 | R514 | E | 4 | 5 | 1 |
| C046 | G | 1 | 0 | 4 | Q516A | F | 4 | 5 | 1 | R516 | F | 4 | 5 | 2 |
| C050 | G | 2 | 0 | 5 | Q516B | F | 4 | 5 | 1 | R520 | F | 4 | 5 | 2 |
| C100 | A | 3 | 1 | 0 | Q604 | C | 4 | 6 | 0 | R522 | F | 4 | 5 | 2 |
| C105 | B | 3 | 1 | 0 | Q620 | F | 4 | 6 | 2 | R524 | F | 4 | 5 | 2 |
| C140 | G | 2 | 1 | 4 | Q732 | G | 5 | 7 | 3 | R526 | F | 4 | 5 | 2 |
| C202 | A | 4 | 2 | 0 | Q748 | G | 5 | 7 | 4 | R534 | G | 4 | 5 | 3 |
| C204 | B | 4 | 2 | 0 | R002 | D | 2 | 0 | 0 | R536 | G | 3 | 5 | 3 |
| C300 | B | 4 | 3 | 0 | R004 | C | 2 | 0 | 0 | R600 | C | 4 | 6 | 0 |
| C302 | C | 4 | 3 | 0 | R100 | D | 2 | 1 | 0 | R602 | B | 4 | 6 | 0 |
| C304 | C | 6 | 3 | 0 | R101 | C | 2 | 1 | 0 | R610 | F | 4 | 6 | 1 |
| C310 | C | 3 | 3 | 1 | R102 | D | 2 | 1 | 0 | R622 | F | 5 | 6 | 2 |
| C312 | E | 2 | 3 | 1 | R103 | C | 1 | 1 | 0 | R730 | F | 4 | 7 | 3 |
| C404 | D | 6 | 4 | 0 | R104 | C | 1 | 1 | 0 | R732 | G | 5 | 7 | 3 |
| C410 | E | 3 | 4 | 1 | R105 | C | 1 | 1 | 0 | R744 | G | 5 | 7 | 4 |
| C412 | E | 3 | 4 | 1 | R106 | A | 2 | 1 | 0 | R746 | G | 5 | 7 | 4 |
| C420 | E | 3 | 4 | 2 | R107 | A | 2 | 1 | 0 | R747 | F | 5 | 7 | 4 |
| C500 | D | 4 | 5 | 0 | R108 | A | 3 | 1 | 0 | R748 | G | 5 | 7 | 4 |
| C504 | B | 4 | 5 | 0 | R116 | B | 3 | 1 | 1 | R758 | G | 5 | 7 | 5 |
| C510 | D | 4 | 5 | 1 | R144 | G | 1 | 1 | 4 | TP010 | G | 2 | 0 | 1 |
| C512 | F | 4 | 6 | 1 | R200 | A | 3 | 2 | 0 | TP030 | H | 2 | 0 | 3 |
| C530 | G | 4 | 5 | 3 | R202 | A | 3 | 2 | 0 | TP040 | G | 5 | 0 | 4 |
| C612 | F | 5 | 6 | 1 | R203 | A | 3 | 2 | 0 | TP420 | F | 4 | 4 | 2 |
| CR204 | B | 4 | 2 | 0 | R204 | B | 4 | 2 | 0 | TP422 | F | 4 | 4 | 2 |
| CR206 | B | 4 | 2 | 0 | R207 | B | 4 | 2 | 0 | U020 | G | 2 | 0 | 2 |
| CR310 | C | 3 | 3 | 1 | R208 | A | 4 | 2 | 0 | U110 | D | 2 | 1 | 1 |
| CR406 | C | 4 | 4 | 0 | R210 | B | 3 | 2 | 1 | U116 | B | 3 | 1 | 1 |
| CR600 | C | 4 | 6 | 0 | R212 | C | 3 | 2 | 1 | U212 | B | 3 | 2 | 1 |
| CR602 | C | 4 | 6 | 0 | R300 | C | 6 | 3 | 0 | U320A | D | 3 | 3 | 2 |
| CR610 | F | 4 | 6 | 1 | R302 | B | 4 | 3 | 0 | U320B | D | 3 | 3 | 2 |
| CR732 | G | 5 | 7 | 3 | R310 | E | 3 | 3 | 1 | U320C | D | 3 | 3 | 2 |
| DS001 | D | 2 | 0 | 0 | R312 | E | 3 | 3 | 1 | U320C | A | 2 | 3 | 2 |
| DS008 | D | 2 | 0 | 0 | R314 | E | 2 | 3 | 1 | U320D | A | 2 | 3 | 2 |
| LR032 | G | 1 | 0 | 3 | R400 | C | 6 | 4 | 0 | U320E | A | 3 | 3 | 2 |
| LR034 | G | 1 | 0 | 4 | R402 | C | 6 | 4 | 0 | U320F | A | 1 | 3 | 2 |
| LR040 | G | 2 | 0 | 4 | R404 | D | 6 | 4 | 0 | U530 | G | 4 | 5 | 3 |
| LR046 | G | 1 | 0 | 4 | R405 | C | 4 | 5 | 0 | U820F | A | 2 | 8 | 2 |
| LR050 | G | 2 | 0 | 5 | R406 | D | 6 | 4 | 0 | VR210 | B | 3 | 2 | 1 |
| LR512 | D | 3 | 5 | 1 | R410 | E | 3 | 4 | 1 | VR434 | G | 4 | 4 | 3 |
| P028 | G | 3 | 0 | 2 | R412 | A | 1 | 4 | 1 | VR532 | G | 5 | 5 | 3 |

P/O TRIGGER BOARD A10. See Fig. 8-1

| C250 | B | 5 | 2 | 5 | Q260 | B | 5 | 2 | 6 | R264 | B | 6 | 2 | 6 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| C350 | B | 6 | 3 | 5 | Q262 | A | 5 | 2 | 6 | R266 | B | 5 | 2 | 6 |
| C362 | A | 6 | 3 | 6 | Q264 | B | 5 | 3 | 6 | R360 | A | 6 | 3 | 6 |
| C460 | B | 6 | 4 | 6 | Q266 | B | 6 | 3 | 6 | R362 | A | 6 | 3 | 6 |
| J264 | B | 5 | 3 | 6 | R250 | B | 5 | 2 | 5 | R460 | B | 6 | 4 | 6 |
| J304 | C | 5 | 3 | 0 | R260 | B | 6 | 2 | 6 |  |  |  |  |  |









Fig. 8-3. P/O Sweep Board A20.


Fig. 8-3. P/O Sweep Board A20.

SWEEP LOGIC


P/O SWEEP BOARD AZO





Fig. 8-4. P/O Sweep Board A20.


Fig. 8-4. P/O Sweep Board A20.

HORIZ PREAMP

P/O SWEEP BOARD A20





SEE PARTS LIST FOR EARLIER VALUES AND SERIAL NUMBER Ranges of parts outlined OR DEPICTED IN GREY.


Fig. 8-5. P/O LED Board.


Fig. 8-6. P/O Switch Board.

FRONT PANEL


P/O LED BOARD A40





Fig. 8-7A. P/O Horizontal Programming Board A50 (SN B030837 \& below).

P/O PROGRAMMING LOGIC BOARD A50 (SN B030837 \& below).

| COMPONENT <br> NUMBER | SCHEMATIC <br> LOCATION | BOARD <br> LOCATION |  | COMPONENT <br> COMBER | SCHEMATIC <br> LOCATION |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | COL ROW | COL | ROW |  |  |
| COL |  |  |  |  |  |



MICROPROCESSOR
P/O PROGRAMMING LOGIC BOARD A50 (SN B030838 \& up).

| SCHEMATIC <br> LOCATION |  | BOARDLOCATION |  | COMPONENT NUMBER | SCHEMATIC LOCATION |  | BOARD LOCATION |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| COL | ROW | COL | ROW |  | COL | Row | COL | ROW |
| G | 1 | 4 | 1 | U310A | G | 1 | 3 | 1 |
| B | 3 | 2 | 3 | U500 | B | 4 | 5 | 0 |
| C | 3 | 15 | 2 | U510 | B | 6 | 5 | 1 |
| G | 2 | 2 | 2 | U630 | B | 2 | 6 | 3 |
| B | 3 | 2 | 3 | U710 | D | 5 | 7 | 1 |
| G | 1 | 4 | 1 | U810 | G | 6 | 8 | 1 |
| A | 2 | 12 | 2 | U830 | E | 2 | 8 | 1 |
| C | 3 | 15 | 2 | U1030 | F | 2 | 10 | 3 |
| A | 2 | 12 | 2 | U1110 | E | 6 | 11 | 1 |
| A | 3 | 1 | 3 | U1130 | D | 1 | 11 | 3 |
| G | 2 | 1 | 0 | U1220 | B | 1 | 12 | 2 |
| G | 2 | 1 | 0 | U1310 | C | 2 | 13 | 1 |
| G | 3 | 2 | 0 | U1620A | c | 3 | 16 | 2 |
| G | 2 | 2 | 0 | W635 | D | 3 | 7 | 3 |
| B | 3 | 2 | 3 | W730 | D | 3 | 7 | 2 |
| G | 1 | 3 | 1 | W735 | D | 3 | 7 | 3 |







Fig. 8-8. P/O Horizontal Programming Board A50.


Fig. 8-8. P/O Horizontal Programming Board A50.
*See Parts List for serial number ranges.

P/O PROGRAMMING LOGIC BOARD A50.

| COMPONENT NUMBER | SCHEMATIC <br> LOCATION |  | BOARDLOCATION |  | COMPONENT NUMBER | SCHEMATIC LOCATION |  | BOARD LOCATION |  | COMPONENT NUMBER | SCHEMATIC LOCATION |  | BOARD LOCATION |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | COL | ROW | COL | ROW |  | COL | ROW | COL | ROW |  | COL | ROW | COL | ROW |
| C000 | F | 1 | 0 | 0 | R120 | G | 3 | 1 | 2 | U230A | G | 3 | 2 | 3 |
| C020 | F | 3 | 0 | 2 | R122 | H | 3 |  | 2 | U230B | G | 3 | 2 | 3 |
| C024 | G | 3 | 0 | 2 | R124 | H | 3 |  | - 2 | U310B | G | 5 | 3 | 1 |
| C030 | G | 3 | 0 | 3 | R130 | G | 3 |  | 3 | U420A | D | 4 | 4 | 2 |
| C032 | G | 4 | 0 | 3 | R132 | G | 3 |  | 3 | U420B | D | 4 | 4 | 2 |
| C100 | F | 1 | 1 | 1 | R134 | G | 3 | 1 | 3 | U420C | D | 3 | 4 | 2 |
| C220 | $F$ | 1 | 2 | 2 | *R230 | D | 5 | 2 | 3 | U420D | D | 3 | 4 | 2 |
| C420 | F | 1 | 4 | 2 | *R232 | D | 5 | 3 | 3 | U430A | E | 4 | 4 | 3 |
| C528 | E | 6 | 5 | 2 | R234 | D | 6 | 2 | 3 | U430B | E | 5 | 4 | 3 |
| C530 | G | 2 | 6 | 3 | R235 | D | 6 | 2 | 3 | U430C | D | 6 | 4 | 3 |
| C628 | F | 2 | 6 | 2 | R332 | D | 5 | 3 | 3 | U430D | F | 6 | 4 | 3 |
| C1020 | F | 2 | 10 | 2 | *R334 | D | 5 | 3 | 3 | U530A | B | 3 | 5 | 3 |
| C1100 | F | 2 | 11 | 0 | *R336 | D | 4 | 3 | 3 | U530B | F | 6 | 5 | 3 |
| C1106 | F | 2 | 11 | 0 | R430 | E | 6 | 4 | 3 | U530C | C | 5 | 5 | 3 |
| C1120 | F | 3 | 11 | 2 | R530 | E | 6 | 5 | 3 | U530D | B | 3 | 5 | 3 |
| C1400 | C | 1 | 14 | 0 | R532 | F | 5 | 5 | 3 | U1310B | C | 5 | 13 | 1 |
| C1516 | H | 1 | 15 | 1 | R534 | E | 5 | 5 | 3 | U1420 | E | 2 | 14 | 2 |
| C1518 | C | 3 | 15 | 2 | R536 | E | 6 | 5 | 3 | U1500 | B | 2 | 15 | 0 |
| C1520 | F | 3 | 15 | 2 | R538 | F | 5 | 5 | 3 | U1520B | B | 6 | 15 | 2 |
| C1530 | H | 5 | 15 | 3 | R539 | B | 3 | 5 | 3 | U1520C | B | 6 | 15 | 2 |
| C1620 | H | 1 | 16 | 2 | R1406 | C | 1 | 14 | 0 | U1520D | A | 4 | 15 | 2 |
| C1622 | G | 2 | 16 | 2 | R1410 | B | 1 | 14 | 0 | U1520E | C | 3 | 15 | 2 |
| C1630 | B | 5 | 15 | 3 | R1412 | B | 1 | 14 | 1 | U1520F | C | 1 | 15 | 2 |
| ${ }^{*} \mathrm{C} 1720$ | G | 2 | 17 | 2 | R1514 | B | 1 |  | 1 | U1530A | H | 5 | 15 | 3 |
| C1730 | F | 3 | 17 | 3 | R1532 | H | 5 | 15 | 3 | U1530B | B | 5 | 15 | 3 |
| * CR234 | E | 5 | 3 | 3 | R1620 | C | 4 | 16 | 2 | U1620B | C | 1 | 16 | 2 |
| CR1310 | B | 1 | 13 | 1 | R1630 | B | 5 | 15 | 3 | U1620C | C | 5 | 16 | 2 |
| CR1614 | H | 2 | 16 | 1 | R1730 | F | 1 | 17 | 3 | U1620D | A | 6 | 16 | 2 |
| CR1616 | H | 1 | 16 | 1 | R1732 | B | 3 | 17 | 3 | U1630A | B | 4 | 16 | 3 |
| LR1612 | H | 1 | 16 | 1 | TP130 | H | 3 | 1 | 3 | U1630B | G | 5 | 16 | 3 |
| P130 | G | 3 | 1 | 3 | TP132 | H | 4 | 1 | 3 | U1630C | B | 6 | 16 | 3 |
| Q126 | G3 | 1 | 2 |  | TP138 | F | 6 | 2 | 3 | U1630D | B | 4 | 16 | 3 |
| Q135 | G | 3 | 1 | 3 | TP330 | H | 2 | 3 | 2 | U1720A | B | 4 | 17 | 2 |
| Q1404 | C | 1 | 14 | 0 | TP636 | E | 2 | 6 | 3 | U1720B | C | 6 | 17 | 2 |
| Q1412 | B | 1 | 14 | 1 | TP1622 | E | 2 | 16 | 2 | U1720C | B | 6 | 17 | 2 |
| 01514 | B | 1 | 15 | 1 | U000A | D | 3 | 0 | 0 | U1720D | B | 4 | 17 | 2 |
| R020 | F | 3 | 0 | 2 | U000B | D | 3 | 0 | 0 | U1730A | A | 5 | 17 | 3 |
| R022 | F | 3 | 0 | 2 | U010A | G | 5 | 0 | 1 | U1730B | B | 4 | 17 | 3 |
| R024 | G | 3 | 0 | 2 | U010B | E | 3 | 0 | 1 | VR330 | D | 5 | 3 | 3 |
| R030 | G | 3 | 0 | 3 | U010C | H | 5 | 0 | 1 | W320 | F | 2 | 3 | 2 |
| R032 | G | 2 | 0 | 3 | U030A | F | 3 | 0 | 3 | *W1720 | H | 2 | 17 | 2 |
| R034 | G | 4 | 0 | 3 | U030B | G | 3 | 0 | 3 |  |  |  |  |  |





Fig. 8-9. P/O PROG. Logic Board A50.

READOUT


P/O PROGRAMMING LOGIC BOARD A50.


9. P/O PROG. Logic Board A50.

READOUT


OGRAMMING LOGIC BOARD A50.



M


## REPLACEABLE MECHANICAL PARTS

## PARTS ORDERING INFORMATION

Replacement parts are available from or through your local Tektronix, Inc. Fiefd Office or representative

Changes to Tektronix instruments are sometimes made to accommodate improved components as they become available. and to give you the benefit of the latest circuit improvements deveioped in our engineering department. It is therefore important, when ordering parts, to include the following information in your order: Part number, instrument type or number, serial number, and modification number if applicable

If a part you have ordered has been replaced with a new or improved part. your local Tektronix, Inc. Field Office or representative will contact you concerning any change in part number.

Change information, if any, is located at the rear of this manual.

## SPECIAL NOTES AND SYMBOLS

X000 Part first added at this serial number
00X Part removed after this serial number

FIGURE AND INDEX NUMBERS
Items in this section are referenced by figure and index numbers to the illustrations

## INDENTATION SYSTEM

This mechanical parts list is indented to indicate item relationships. Following is an example of the indentation system used in the description column

12345
Name \& Description
Assembly and/or Component
Attaching parts for Assembly andior Component

-     -         - . -

Detail Part of Assembly and/or Component
Attaching parts for Detail Part

- . .

Parts of Detail Part
Attaching parts for Parts of Detail Part

Attaching Parts always appear in the same indentation as the item it mounts, while the detail parts are indented to the right. Indented items are part of, and included with, the next higher indentation. The separation symbol--**--indicates the end of attaching parts.

Attaching parts must be purchased separately, unless otherwise specified.

## ITEM NAME

In the Parts List, an Item Name is separated from the description by a colon (i). Because of space limitations, an Item Name may sometimes appear as incomplete. For further Item Name identification, the U.S. Federal Cataloging Handbook H6-1 can be utilized where possible.

| ABBREVIATIONS |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| " | INCH | ELCTRN | ELECTRON | IN | 1 NCH | SE | SINGLE END |
| \# | NUMBER SIZE | ELEC | ELECTRICAL | INCAND | INCANDESCENT | SECT | SECTION |
| ACTA | ACTUATOR | ELCTLT | ELECTROLYTIC | INSUL | INSULATOR | SEMICOND | SEMICONDUCTOR |
| ADPTA | ADAPTER | ELEM | ELEMENT | INTL | INTERNAL | SHLD | SHIELD |
| ALIGN | ALIGNMENT | EPL | ELECTRICAL PARTS LIST | LPHLDR | LAMPHOLDER | SHLDR | SHOULDERED |
| AL | ALUMINUM | EQPT | EQUIPMENT | MACH | MACHINE | SKT | SOCKET |
| ASSEM | ASSEMBLED | EXT | EXTERNAL | MECH | MECHANICAL | SL | SLIDE |
| ASSY | ASSEMBLY | FIL | FILLISTER HEAD | MT'G | MOUNTING | SLFLKG | SELF-LOCKING |
| ATTEN | ATTENUATOR | FLEX | flexible | NIP | NIPPLE | SLVG | SLEEVING |
| AWG | AMERICAN WIRE GAGE | FL.H | FLAT HEAD | NON WIRE | NOT WIRE WOUND | SPR | SPRING |
| BD | BOARD | FLTR | FILTER | OBD | ORDEA BY DESCRIPTION | SQ | SQUARE |
| BRKT | BRACKET | FR | FRAME or FRONT | OD | OUTSIDE DIAMET $=$ R | SST | STAINLESS STEEL |
| BRS | BRASS | FSTNR | FASTENER | OVH | OVAL HEAD | STL | STEEL |
| BRZ | BRONZE | FT | FOOT | PH BRZ | PHOSPHOR BRONZE | SW | SWITCH |
| BSHG | BUSHING | FXD | FIXED | PL | PLAIN OR PLATE | T | TUBE |
| CAB | CABINET | GSKT | GASKET | PLSTC | PLASTIC | TERM | TERMINAL |
| CAP | CAPACITOR | HDL | HANDLE | PN | PART NUMBER | THD | THREAD |
| CER | CERAMIC | HEX | HEXAGON | PNH | PAN HEAD | THK | THICK |
| CHAS | CHASSIS | HEX HD | HEXAGONAL HEAD | PWR | POWER | TNSN | TENSION |
| CKT | CIRCUIT | HEX SOC | HEXAGONAL SOCKET | RCPT | AECEPTACLE | TPG | TAPPING |
| COMP | COMPOSITION | HLCPS | HELICAL COMPRESSION | RES | RESISTOA | TRH | TRUSS HEAD |
| CONN | CONAECTOA | HLEXT | helical extension | RGD | RIGID | $V$ | VOLTAGE |
| COV | COVER | HV | HIGH VOLTAGE | RLF | RELIEF | VAR | VARIABLE |
| CPLG | COUPLING | 1 C | INTEGRATEO CIRCUIT | RTNR | FETAINEA | W | WITH |
| CAT | CATHODE RAY TUBE | ID | INSIDE DIAMETER | SCH | SOCKET HEAD | WSHR | WASHER |
| DEG | DEGREE | IDENT | IDENTIFICATION | SCOPE | OSCILLOSCOPE | XFMR | TRANSFORMER |
| OWR | DRAWER | IMPLR | IMPELLER | SCR | SCREW | XSTA | TRANSISTOR |

CROSS INDEX -...MFR. CODE NUMBER TO MANUFACTURER

| Mir. Code | Manufacturer | Address | City, State, Zip |
| :---: | :---: | :---: | :---: |
| 000CY | NORTHWEST FASTENEA SALES, INC. | 7923 SW CIRRUS DRIVE, | BEAVERTON, OR 97005 |
| 000FW | WESTERN SINTERING CO INC. | 2620 STEVENS DRIVE | RICHLAND, WA 99352 |
| 00779 | AMP, INC. | P.O. BOX 3608 | HAFRISBURG, PA 17105 |
| 09922 | BURNDY CORPORATION | RICHARDS AVENUE | NOFIWALK, CT 05852 |
| 13511 | AMPHENOL CARDRE DIV., BUNKER RAMO CORP. |  | LOS GATOS. CA 95030 |
| 22526 | BERG ELECTRONICS, INC. | YOUK EXPRESSWAY | NEW/ CUMBERLAND, PA 17070 |
| 22599 | ESNA, DIV. OF AMERACE CORPORATION | 16150 STAGG STREET | VAN NUYS, CA 91409 |
| 73803 | TEXAS INSTRUMENTS, INC., METALLURGICAL |  |  |
|  | MATERIALS DIV. | 34 FOREST STREET | ATTLEBORO, MA 02703 |
| 80009 | TEKTRONIX, INC. | P O BOX 500 | BEAVERTON, OR 97077 |
| 83385 | CENTRAL SCREW CO. | 2530 CRESCENT DR. | BROADVIEW, IL 60153 |
| 87308 | N. L. INDUSTRIES, INC., SOUTHERN SCREW DIV. | P. O. BOX 1360 | STATESVILLE, NC 28677 |
| 92101 | SCHULZE MFG, 50 INGOLD RD BURLINGAME, CA 94010 |  |  |



## Replaceable Mechanical Parts-7B90P

| Fig. 8 Index No. | Tektronix Part No. | Serial/Mo Eff | Dscont | Qty 12345 Name \& Description |  | Mfr Code | Mfr Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1. | ----- .-.- |  |  | - | CKT BOARD ASSY INCLUDES: |  |  |
| -42 | 131-0608-00 |  |  | 8 | .TERMINAL,PIN:0.365 L X 0.025 PH BRZ GOLD | 22526 | 48283-036 |
| -43 | 214-0579-00 |  |  | 8 | TERM,TEST POINT:BRS CD PL | 80009 | 214-0579-00 |
| -44 | 136-0578-00 | B010100 | B029999 | 4 | .SKT,PL-IN ELEK:MICROCKT, 24 PIN,LOW PRFL | 09922 | DILB24P-108 |
|  | 136-0578-00 | B030000 | B030761 | 5 | .SKT,PL-IN ELEK:MICROCKT,24 PiN,LOW PRFL | 09922 | DILB24P-108 |
|  | 136-0578-00 | B030762 | B030844 | 4 | .SKT,PL-IN ELEK:MICROCKT, 24 PIN,LOW PRFL | 09922 | DILB24P-108 |
|  | 136-0757-00 | B030845 |  | 4 | .SKT,PL-IN ELEK:MICROCKT,40 PIN | 09922 | DILB40P-108 |
| -44.1 | ( | B030762 | B030837 | 2 | .CKT BOARD ASSY:MEMORY ADAPTERISEE A51 REP |  |  |
| -44.2 | 131-0608-00 | B030762 | B030837 | 48 | ..TERMINAL, PIN: 0.365 L X 0.025 PH BRZ GOLD | 22526 09922 | 48283-036 |
| -44.3 | 136-0578-00 | B030762 | B030837 | 2 | ..SKT,PL-IN ELEK:MICROCKT, 24 PIN,LOW PRFL | 22526 | 75377-001 |
| -45 | 136-0263-04 |  |  | 109 | SOCKET,PIN TERM:FOR 0.025 INCH SQ PIN SOCKET, PLUG-IN:40 DIP,LOW PROFILE | 09922 | DILB40P-108 |
| -46 | 136-0623-00 | B010100 | B030629 | 6 | SOCKET,PLUG-IN:40 DIP,LOW PROFILE SOCKET,PLUG-IN:40 DIP,LOW PROFILE | 09922 | DILB40P-108 |
|  | 136-0623-00 | B030630 | B030844 | 1 | SOCKET,PLUG-IN:40 DIP,LOW PROFILE SKT PL-IN ELEK:MICROCKT, 40 PIN | 09922 | DILB40P-108 |
|  | 136-0757-00 | B030845 | B031109 | 1 | .SKT,PL-IN ELEK:MICROCKT, 40 PIN | 73803 | Cs9002-20 |
| -47 | 136-0634-00 | B010100 | B030844 | 1 | SOCKET,PLUG-IN:20 LEAD DIP,CKT BD MTG | 09922 | DILB20P-108 |
|  | 136-0752-00 | B030845 |  | 1 | SKT,PL-IN ELEK:MICROCIRCUIT,20 DIP BUSCONDUCTOR: 2 WIRE BLACK | 00779 | 850100-01 |
| -48 | 131-0993-00 |  |  | 4 | .BUS,CONDUCTOR: 2 WIRE BLACK SPACER, SLEEVE: $2500 \times 0.34 \mathrm{INCH}$ LONG | 80009 | $361-0238-00$ |
| -49 | 361-0238-00 |  |  | 6 | .SPACER,SLEEVE:0.25 OD $\times 0.34$ INCH LONG | 880009 | 386-1402-00 |
| -50 | 386-1402-00 |  |  | 1 | PANEL,REAR: <br> ************(ATTACHING PARTS) ${ }^{* * * * * * * * * ~}$ |  | 386-1402-00 |
| -51 | 213-0192-00 |  |  | 4 | SCR,TPG,THD FOR: $6-32 \times 0.50 \mathrm{INCH}, \mathrm{PNH}$ STL <br> (END ATTACHING PARTS) ${ }^{*+\ldots \ldots * *}$ | 87308 | ORD BY DESCR |
| -52 | 388-5632-01 |  |  | 1 | CIRCUIT BOARD:SHIEID | 80009 | 388-5632-01 |
| -53 | 214-1140-00 |  |  | 6 | SPRING,HLCPS: 0.251 OD $\times 0.375^{\prime \prime}$ L,SST WIRE | 80009 | 214-1140-00 |
| -54 | 386-1657-00 |  |  | 3 | SUPPORT,CKT BD: | 80009 | 386-1657-00 |
| -55 | ----------- |  |  | 1 | CKT BOARD ASSY:SWEEP(SEE A20 REPL) <br> '(ATTACHING PARTS)** |  |  |
| -56 | 211-0116-00 |  |  | 3 | SCR,ASSEM WSHR:4-40 0.312 INCH,PNH BRS *(END ATTACHING PARTS)****** | 83385 | ORD BY DESCR |
|  |  |  |  | - | CKT BOARD ASSY INCLUDES: |  |  |
| -57 | 351-0188-00 |  |  | 2 | .GUIDE-POST, LOCK:0.65 INCH LONG | 80009 | 351-0188-00 |
| -58 | 131-0592-00 |  |  | 14 | CONTACT,ELEC: 0.885 INCH LONG | 22526 | 47353 |
| -59 | 351-0186-00 |  |  | 4 | .GUIDE-POST,LOCK:0.84 INCH LONG | 80009 | 351-0186-00 |
| -60 | 131-0595-00 |  |  | 17 | CONTACT, ELEC 1.37 INCH LONG | 2526 | 47355 |
|  | 131-0590-00 |  |  | 92 | CONTACT,ELIEC:0.71 INCH LONG | 22526 | 47351 |
|  | 131-0608-00 |  |  | 2 | TERMINAL,PIN: $0.365 \mathrm{~L} \times 0.025$ PH BRZ GOLD | 22526 | 48283-036 |
| -61 | 136-0634-00 | B010100 | 8010999 | 1 | SOCKET,PLUG-IN:20 LEAD DIP,CKT BD MTG | 73803 | CS9002-20 |
|  | 136-0752-00 | B011000 |  | 1 | SKT,PL-IN ELEK:MICROCIRCUIT, 20 DIP | 09922 | DILB20P-108 |
| -62 | 136-0260-02 | B010100 | B010999 | 5 | .SKT,PL-IN ELEK:MICROCIRCUIT, 16 DIP,LOW CL | 09922 | DILB16P-108T |
|  | 136-0729-00 | B011000 | B031109 | 5 | SKT,PL-IN ELEK:MICROCKT, 16 CONTACT | 09922 | DILB16P-108 |
| -63 | 214-0579-00 |  |  | 7 | TERM,TEST POINT:BRS CD PL | 80009 | 214-0579-00 |
| -64 | 136-0252-07 | B010100 | B031109 | 16 | SOCKET,PIN CONN:W/O DIMPLE | 22526 | 75060-012 |
|  | 136-0252-07 | 8031110 |  | 6 | .SOCKET,PIN CONN:W/O DIMPLE | 22526 | ${ }^{75060-012}$ DILB149P-108 |
| -65 | 136-0269-02 | B010100 | B010999 | 1 | SKT,PL-IN ELEK:MICROCIRCUIT, 14 DIP | 09922 | DILB149P-108 |
|  | 136-0728-00 | B011000 | B031109 | 1 | SKT,PL-IN ELEK:MICROCKT, 14 CONTACT | 09922 | DILB14P-108 |
| -66 | 131-0993-00 |  |  | 1 | BUS,CONDUCTOR:2 WIRE BLACK | 00779 | 850100-01 |
| -67 | 131-1003-00 |  |  | 3 | CONN,RCPTTELEC:CKT BD MT,3 PRONG | 80009 | 131-1003-00 |
| -68 | 351-0185-00 |  |  | 4 | .GUIDE-POST,LOCK:0.65 INCH LONG | 80009 | 351-0185-00 |
| -69 | 220-0547-01 |  |  | 3 | NUT,BLOCK:0.38 $\times 0.26 \times 0.282$ (2)4-40 | 000FW | ORD BY DESCR |
| -70 | 211-0105-00 |  |  | 3 | ********(END ATTACHING PARTS)****** <br> SPRING,GROUND:FLAT | 83385 | ORD bY DESCR |
| -71 | 214-1061-00 |  |  | 1 |  | 80009 | 214-1061-00 |
| -72 | 344-0210-00 |  |  | 1 | CLIP,SPR TNSN: | 80009 | 344-0210-00 |
| -73 | 426-0505-01 |  |  | 1 | FR SECT, PLUG-IN:TOP | 80009 | 426-0505-01 |
| -74 | 214-1054-00 |  |  | 1 | SPRING,FLAT:0.825 $\times 0.322$, SST | 80009 | 214-1054-00 |
| -75 | 105-0075-00 |  |  | 1 | BOLT, LATCH:7A \& 7B SER PL-IN | 80009 | 105-0075-00 |
| -76 | 426-0499-01 |  |  | 1 | FR SECT,PLUG-IN:BOTTOM | 80009 | 426-0499-01 |
|  | 198-3730-00 |  |  | 1 | WIRE SET, ELEC: | 80009 | 198-3730-00 |



REV MAR 1982

(74)

Fig. \&

| index <br> No. | Tektronix <br> Part No. | Serial/Model No. Eff Dscont | Qty | 12345 | Name \& Description | Mfr <br> Code | Mfr Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 070-2309- |  | 1 | MANUAL, TEC | CTION | 80009 | 070-2309-00 |


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| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | Single sweep arm (write oniy) |  |




## aiv פninwvevord

ESVG 3NL d06el
For additional copies of this programing aid, contact your ocal Tektronix Sales Engineer

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|  | 8 | $\frac{\pi}{8}$ | $\frac{\mathfrak{m}}{8}$ | 0 |

The <address> and <data> bytes are taken from Table A. The < checksum > is the 2's complement of the modulo- 256 sum of the
preceding message byes.
0า0: -

## Remove programming aid at perforation. Fold along <br> 


${ }^{1}$ To set read-out and front panel, sweep rate low blts must be changed also. If not
changed, sweep is still magnified, but $T / D$ ? will return $T / D$ error.
070」 momsux
Description
Select pp Auto trigger mode. Select Normal trigger mode. Select Singie-sweep triggering. Set trigger coupling to $A C$.
Set trigger coupling to $D C$. Set coupling to AC \&FREJ. Set coupling to AC HFREJ. Select internal trigger source. Select Line-freq trigger source. Select External trigger source. Select EXT $\div 10$ trigger source.
Set trigger slope positive. Set trigger slope negative. Set trigger level ( -6.4 to +6.35 div). Set TIME/DIV ( $5 \mathrm{E}-10$ to $5 \mathrm{E}-1 \mathrm{sec}) .{ }^{2}$ Turn sweep magnifier on. rurn sweep magnifier oft. Set trace position ( -6.4 to +6.39 div). ${ }^{1}$ Set trigger hold-off (Oto 63, uncai). ${ }^{3}$ Enable end-of-sweep SRQ. Disable end-of-sweep SRQ.
Trigger light is on. \} query only
Trigger light is off.\} query only Arm single sweep. Single sweep is disarmed. Returns plug-in type. Queries all functions. (except the query-on
Argumen
 Header

${ }^{1}$ Accepts < $\mathrm{NR} 1>$, <NR2>, or < NR3>; query returns <NR2>.
${ }^{2}$ Acccpts <NR1>, <NR2>, or <NR3>; query returns <NR3>.
${ }^{3}$ Accepts <NR1>; query returns <NR1>.

## MANUAL CHANGE INFORMATION

At Tektronix, we continually strive to keep up with latest electronic developments by adding circuit and component improvements to our instruments as soon as they are developed and tested.

Sometimes, due to printing and shipping requirements, we can't get these changes immediately into printed manuals. Hence, your manual may contain new change information on following pages.

A single change may affect several sections. Since the change information sheets are carried in the manual until all changes are permanently entered, some duplication may occur. If no such change pages appear following this page, your manual is correct as printed.

