

General Description
The MAX1151 is a parallel flash analog-to-digital converter (ADC) capable of digitizing full-scale ( 0 V to -2 V ) inputs into 8-bit digital words at an update rate of 750 Msps . The ECL-compatible outputs are demuxed into two separate output banks, each with differential data-ready outputs to ease the task of data capture. The MAX1151's wide input bandwidth and low capacitance eliminate the need for external track/hold amplifiers for most applications. A proprietary decoding scheme reduces metastable errors to 1LSB. This device operates from a single -5.2 V supply, with a nominal power dissipation of 5.5 W .

Ordering Information

| PART | TEMP. RANGE | PIN-PACKAGE |
| :---: | :--- | :--- |
| MAX1151AIZS | $-20^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 80 MQUAD |
| MAX1151BIZS | $-20^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | 80 MQUAD |

Features

- 1:2 Demuxed ECL-Compatible Outputs
- Wide Input Bandwidth: 900MHz
- Low Input Capacitance: 15pF
- Metastable Errors Reduced to 1LSB
- Single -5.2V Supply

Applications
Digital Oscilloscopes
Data Acquisition
Transient-Capture Applications
Radar, EW, ECM
Direct RF/IF Downconversion
Pin Configuration appears on last page.


## 8-Bit, 750Msps Flash ADC

## ABSOLUTE MAXIMUM RATINGS

| Supply Voltages |  |
| :---: | :---: |
| Negative Supply Voltage (VEE to GND) | ....-7.0V to +0.5V |
| Ground Voltage Differential | ...-0.5V to +0.5 V |
| Input Voltages |  |
| Analog Input Voltage | . +0.5 V to $\mathrm{V}_{\mathrm{EE}}$ |
| Reference Input Voltage | +0.5V to $\mathrm{V}_{\mathrm{EE}}$ |
| Digital Input Voltage | +0.5 V to $\mathrm{V}_{\mathrm{EE}}$ |
| Reference Current (V) | . 35 mA |

Digital Output Current ...........................................0mA to -28mA Operating Temperature Range ........................... $20^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$
Case Temperature .......................................................... $125^{\circ} \mathrm{C}$ Junction Temperature ..................................................... $+150^{\circ} \mathrm{C}$
Lead Temperature (soldering, 10sec). ............................ $+300^{\circ} \mathrm{C}$
Storage Temperature Range ............................. $65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS

$\left(\mathrm{V}_{\mathrm{EE}}=-5.2 \mathrm{~V}, \mathrm{~V}_{\mathrm{RB}}=-2.00 \mathrm{~V}, \mathrm{~V}_{\mathrm{RM}}=-1.00 \mathrm{~V}, \mathrm{~V}_{\mathrm{RT}}=0.00 \mathrm{~V}, \mathrm{f} \mathrm{CLK}=750 \mathrm{MHz}\right.$, duty cycle $=50 \%$, typical thermal impedance $\left(\theta_{\mathrm{JC}}\right)=4^{\circ} \mathrm{C} / \mathrm{W}$, $\left.\mathrm{T}_{\mathrm{j}}=\mathrm{T}_{\mathrm{C}}=\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}.\right)($ Note 1$)$

| PARAMETER | CONDITIONS | $\begin{aligned} & \text { TEST } \\ & \text { LEVEL } \end{aligned}$ | MAX1151A |  |  | MAX1151B |  |  | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN | TYP | MAX | MIN | TYP | MAX |  |
| Resolution |  |  |  | 8 |  |  | 8 |  | Bits |
| DC ACCURACY |  |  |  |  |  |  |  |  |  |
| Integral Nonlinearity | $\mathrm{fCLK}=100 \mathrm{kHz}$ | I | -1.0 |  | 1.0 | -1.5 |  | 1.5 | LSB |
| Differential Nonlinearity | $\mathrm{fCLK}=100 \mathrm{kHz}$ | 1 | -0.85 |  | 0.95 | -0.95 |  | 1.5 | LSB |
| No Missing Codes |  |  |  | arante |  |  | arante |  |  |
| ANALOG INPUT |  |  |  |  |  |  |  |  |  |
| Input Voltage Range |  | I | VRB |  | $\mathrm{V}_{\text {RT }}$ | VRB |  | VRT | V |
| Input Bias Current | $\mathrm{V}_{\mathrm{IN}}=0 \mathrm{~V}$ | I |  | 0.75 | 2.0 |  | 0.75 | 2.0 | mA |
| Input Resistance |  | V |  | 15 |  |  | 15 |  | $\mathrm{k} \Omega$ |
| Input Capacitance | Over full input range | V |  | 15 |  |  | 15 |  | pF |
| Input Bandwidth | Small signal | V |  | 900 |  |  | 900 |  | MHz |
|  | Large signal | V |  | 500 |  |  | 500 |  |  |
| Offset Error $\mathrm{V}_{\mathrm{RT}}$ |  | IV | -30 |  | 30 | -30 |  | 30 | mV |
| Offset Error VRB |  | IV | -30 |  | 30 | -30 |  | 30 | mV |
| Input Slew Rate |  | V |  | 5 |  |  | 5 |  | $\mathrm{V} / \mathrm{ns}$ |
| REFERENCE INPUT |  |  |  |  |  |  |  |  |  |
| Ladder Resistance |  | I | 60 | 80 |  | 60 | 80 |  | $\Omega$ |
| Reference Bandwidth |  | V |  | 30 |  |  | 30 |  | MHz |
| TIMING CHARACTERISTICS |  |  |  |  |  |  |  |  |  |
| Maximum Sample Rate |  | I | 750 |  |  | 750 |  |  | MHz |
| Aperture Jitter |  | V |  | 2 |  |  | 2 |  | ps |
| Acquisition Time |  | V |  | 250 |  |  | 250 |  | ps |
| CLK to DATA READY Delay |  | IV | 0.9 | 1.4 | 1.9 | 0.9 | 1.4 | 1.9 | ns |
| Clock to Data Delay |  | IV | 1.25 | 1.75 | 2.25 | 1.25 | 1.75 | 2.25 | ns |

## 8-Bit, 750Msps Flash ADC

## ELECTRICAL CHARACTERISTICS (continued)

$\left(\mathrm{V}_{E E}=-5.2 \mathrm{~V}, \mathrm{~V}_{\mathrm{RB}}=-2.00 \mathrm{~V}, \mathrm{~V}_{\mathrm{RM}}=-1.00 \mathrm{~V}, \mathrm{~V}_{\mathrm{RT}}=0.00 \mathrm{~V}\right.$, fcLK $=750 \mathrm{MHz}$, duty cycle $=50 \%$, typical thermal impedance $(\theta \mathrm{JC})=4^{\circ} \mathrm{C} / \mathrm{W}$, $\mathrm{T}_{\mathrm{j}}=\mathrm{T}_{\mathrm{C}}=\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$.) (Note 1)

| PARAMETER | CONDITIONS | $\begin{aligned} & \text { TEST } \\ & \text { LEVEL } \end{aligned}$ | MAX1151A |  |  | MAX1151B |  |  | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | MIN | TYP | MAX | MIN | TYP | MAX |  |
| DYNAMIC PERFORMANCE |  |  |  |  |  |  |  |  |  |
| Signal-to-Noise Ratio (without harmonics) | $\mathrm{fin}=50 \mathrm{MHz}$ | I | 46 |  |  | 44 |  |  | dB |
|  | $\mathrm{fiN}_{\mathrm{I}}=250 \mathrm{MHz}$ | 1 | 44 |  |  | 42 |  |  |  |
| Total Harmonic Distortion | $\mathrm{fin}=50 \mathrm{MHz}$ | I | -45 |  |  | -43 |  |  | dBc |
|  | $\mathrm{fiN}_{\mathrm{I}}=250 \mathrm{MHz}$ | I | -37 |  |  | -35 |  |  |  |
| Signal-to-Noise and Distortion | $\mathrm{f}_{\mathrm{IN}}=50 \mathrm{MHz}$ | I | 43 |  |  | 41 |  |  | dB |
|  | $\mathrm{fIN}=250 \mathrm{MHz}$ | 1 | 36 |  |  | 34 |  |  |  |
| Spurious-Free Dynamic Range | $\mathrm{f} \mathrm{IN}=50 \mathrm{MHz}$ | 1 | 48 |  |  | 44 |  |  | dB |
|  | $\mathrm{fiN}_{\mathrm{N}}=250 \mathrm{MHz}$ | I | 40 |  |  | 36 |  |  |  |
| DIGITAL INPUTS |  |  |  |  |  |  |  |  |  |
| Input High Voltage (CLK, NCLK) |  | 1 | -1.1 | $-0.7$ |  | -1.1 | -0.7 |  | V |
| Input Low Voltage (CLK, NCLK) |  | 1 |  | -1.8 | -1.5 |  | -1.8 | -1.5 | V |
| Clock Pulse Width High (tpwh) |  | I | 0.67 | 0.5 |  | 0.67 | 0.5 |  | ns |
| Clock Pulse Width Low (tpwL) |  | I | 0.67 | 0.5 |  | 0.67 | 0.5 |  | ns |
| Clock Synchronous Input Currents |  | V |  | 2 |  |  | 2 |  | $\mu \mathrm{A}$ |
| DIGITAL OUTPUTS |  |  |  |  |  |  |  |  |  |
| Logic "1" Voltage |  | I | -1.1 | -0.9 |  | -1.1 | -0.9 |  | V |
| Logic "0" Voltage |  | I |  | -1.8 | -1.5 |  | -1.8 | -1.5 | V |
| POWER-SUPPLY REQUIREMENTS |  |  |  |  |  |  |  |  |  |
| Supply Voltage (VEE) |  | IV | -4.95 | -5.2 | -5.45 | -4.95 | -5.2 | -5.45 | V |
| Supply Current (IEE) |  | I |  | 1.05 | 1.2 |  | 1.05 | 1.2 | A |
| Power Dissipation |  | 1 |  | 5.5 | 6.25 |  | 5.5 | 6.25 | W |

Note 1: All parameters having min/max specifications are guaranteed. The Test Level column indicates the specific device testing actually performed during production and Quality Assurance inspection. Unless otherwise noted, all tests are pulsed tests; therefore, $\mathrm{T}_{\mathrm{j}}=\mathrm{T}_{\mathrm{C}}=\mathrm{T}_{\mathrm{A}}$.

## TEST LEVEL TEST PROCEDURE

I $100 \%$ production tested at the specified temperature.
II $100 \%$ production tested at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, and sample tested at the specified temperatures.
III QA sample tested only at the specified temperatures.
IV Parameter is guaranteed (but not tested) by design and characterization data.
$V \quad$ Parameter is a typical value for information purposes only.
$\mathrm{VI} \quad 100 \%$ production tested at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$. Parameter is guaranteed over specified temperature range.

## 8-Bit, 750Msps Flash ADC

Pin Description

| PIN | NAME | FUNCTION |
| :---: | :---: | :---: |
| 1, 2, 3 | D2B, D3B, D4B | Data Output Bank, Bits 2, 3, and 4 |
| $\begin{gathered} 4,5,19,20,22,23,27,28,38,39 \\ 40,46,47,49,60,67,79 \end{gathered}$ | VEe | Negative Supply, nominally -5.2V |
| 6 | D5B | Data Output Bank B, Bit 5 |
| $\begin{gathered} 7,9,11,54,56,58, \\ 69,71,73,75,77 \end{gathered}$ | DGND | Digital Ground |
| 8 | D6B | Data Output Bank B, Bit 6 |
| 10 | D7B | Data Output Bank B, Bit 7 (MSB) |
| 12 | D8B | Data Output Bank B, Bit 8 (OVR) |
| 13, 14, 31, 34, 41, 63, 64 | N.C. | No Connection. Not internally connected. |
| $\begin{gathered} 15-18,25,26,29,30,36 \\ 37,44,45,51,52 \end{gathered}$ | AGND | Analog Ground |
| 21 | VRBF | Reference-Voltage Force Bottom |
| 24 | VRBS | Reference-Voltage Sense Bottom |
| 32, 33 | VIN | Analog Input Voltage. Can be either voltage or sense. |
| 35 | VRM | Reference-Voltage Middle, nominally -1V |
| 42 | VRTF | Reference-Voltage Force Top |
| 43 | VRTS | Reference-Voltage Sense Top |
| 48 | NCLK | Inverse Clock Input |
| 50 | CLK | Clock Input |
| 53 | DRA | Data Ready Bank A |
| 55 | NDRA | Not Data Ready Bank A |
| 57 | DOA | Data Output Bank A, Bit 0 (LSB) |
| 59, 61, 62, 65, 66, 68 | D1A-D6A | Data Output Bank A, Bits 1-6 |
| 70 | D7A | Data Output Bank A, Bit 7 (MSB) |
| 72 | D8A | Data Output Bank A, Bit 8 (OVR) |
| 74 | NDRB | Not Data Ready Bank B |
| 76 | DRB | Data Ready Bank B |
| 78 | D0B | Data Output Bank B, Bit 0 (LSB) |
| 80 | D1B | Data Output Bank B, Bit 1 |

## Detailed Description

The MAX1151 is one of the fastest monolithic, 8 -bit, parallel, flash analog-to-digital converters (ADCs) available today. The nominal conversion rate is 750 Msps , and the analog bandwidth is in excess of 900 MHz . A major advance over previous flash converters is the inclusion of 255 input preamplifiers between the reference ladder and input comparators (see Functional Diagram). This not only reduces clock transient kickback to the input and reference ladder but also reduces the effect of the input
signal's dynamic state on the input comparators' latching characteristics. The preamplifiers act as buffers to stabilize the input capacitance so that it remains constant over different input voltage and frequency ranges, making the part easier to drive than previous flash converters. The preamplifiers also add a gain of +2 to the input signal, so that each comparator has a wider overdrive or threshold range to trip into or out of the active state. This gain reduces metastable states that can cause errors at the output.

# 8-Bit, 750Msps Flash ADC 

The MAX1151 has true differential analog and digital data paths from the preamplifiers to the output buffers (current-mode logic) for reducing potential missing codes while rejecting common-mode noise.
Signature errors are also reduced by careful layout of the analog circuitry. The device's output drive capability can provide full ECL swings into $50 \Omega$ loads.

Typical Interface Circ uit
The circuit of Figure 1 shows a method of achieving the least error by correcting for integral linearity, inputinduced distortion, and power-supply/ground noise. This is achieved with the use of external reference-ladder tap connections, an input buffer, and supply decoupling. Contact the factory for the MAX1150/MAX1151 evaluation kit manual, which contains more details on interfacing the MAX1151. The function of each pin and external connections to other components are described in the following sections.

VEE, AGND, DGND
VEE is the supply pin with AGND as ground for the device. The power-supply pins should be bypassed as close to the device as possible with at least a $0.01 \mu \mathrm{~F}$ ceramic capacitor. A $1 \mu \mathrm{~F}$ tantalum can also be used for low-frequency suppression. DGND is the ground for the ECL outputs, and should be referenced to the output pulldown voltage and appropriately bypassed, as shown in Figure 1.

## VIN (Analog Input)

There are two analog input pins that are tied to the same point internally. Either one may be used as an analog input sense, while the other is used for input force. This is convenient for testing the source signal to see if there is sufficient drive capability. The pins can also be tied together and driven by the same source. The MAX1151 is superior to similar devices due to a preamplifier stage before the comparators. This makes the device easier to drive because it has constant capacitance and induces less slew-rate distortion.

CLK, NCLK (Clock Inputs) The clock inputs are designed to be driven differentially with ECL levels. The duty cycle of the clock should be kept at $50 \%$, to avoid causing larger second harmonics. If this is not important to the intended application, duty cycles other than $50 \%$ may be used.

D0 to D8, DR, NDR (A and B)
The digital outputs can drive $50 \Omega$ to ECL levels when pulled down to -2 V . When pulled down to -5.2 V , the outputs can drive $130 \Omega$ to $1 \mathrm{k} \Omega$ loads. All digital outputs are gray code, with the coding as shown in Table 1.

Table 1. Output Coding

\begin{tabular}{|c|c|c|}
\hline VIN (V) \& D8 \& D7... D0 <br>
\hline 0 \& \multirow[t]{6}{*}{1

0} \& 10000000 <br>
\hline \& \& 10000001 <br>
\hline \& \& 10000011 <br>
\hline \& \& - <br>
\hline \& \& - <br>
\hline \multirow{5}{*}{-0.5} \& \& 10100001 <br>
\hline \& \multirow[t]{4}{*}{0} \& 10100000 <br>
\hline \& \& 11100000 <br>
\hline \& \& - <br>
\hline \& \& - <br>
\hline \multirow{6}{*}{-1.0} \& \& 11000001 <br>
\hline \& \multirow[t]{5}{*}{0} \& 11000000 <br>
\hline \& \& 01000000 <br>
\hline \& \& . <br>
\hline \& \& 01100001 <br>
\hline \& \& 01100001 <br>
\hline \multirow[t]{6}{*}{-1.5} \& \multirow[t]{6}{*}{0} \& 01100000 <br>
\hline \& \& 00100000 <br>
\hline \& \& - <br>
\hline \& \& - <br>
\hline \& \& 00000011 <br>
\hline \& \& 00000001 <br>
\hline -2.0 \& 0 \& 00000000 <br>
\hline
\end{tabular}

VRBF, VRBS, VRTF, VRTS, VRM
(Reference Inputs)
There are two reference inputs and one external reference voltage tap. These are -2V (VRB force and sense), mid-tap (VRM), and AGND (VRT force and sense). The reference pins and tap can be driven by op amps (as shown in Figure 1), or VRM can be bypassed for limited temperature operation. These voltage inputs can be bypassed to AGND for further noise suppression, if desired.

Thermal Management
The typical thermal impedance ( $\theta \mathrm{CA}$ ) for the MQUAD package has been measured at $\theta \mathrm{CA}=17^{\circ} \mathrm{C} / \mathrm{W}$, in still air with no heatsink.
To ensure rated performance, we highly recommend using this device with a heatsink that can provide adequate air flow. We have found that a Thermalloy 17846 heatsink with a minimum air flow of 1 meter/second (200 linear feet per minute) provides adequate thermal performance under laboratory tests. Application-specific conditions should be taken into account to ensure that the device is properly heat sinked.

## 8-Bit, 750Msps Flash ADC



Figure 1. Typical Interface Circuit

## 8-Bit, 500Msps Flash ADC


#### Abstract

Operation The MAX1151 has 255 preamplifier/comparator pairs; each is supplied with the voltage from VRT to VRB, divided equally by the resistive ladder as shown in the Functional Diagram. This voltage is applied to the positive input of each preamplifier/comparator pair. An analog input voltage applied at VIN is connected to the negative inputs of each preamplifier/comparator pair. The comparators are then clocked through each one's individual clock buffer. When the CLK pin is in the low state, the master or input stage of the comparators compares the analog input voltage to the respective reference voltage. When CLK changes from low to high, the comparators are latched to the state prior to the clock transition and output logic codes in sequence from the top comparators, closest to VRT (OV), down to the point where the magnitude of the input signal changes sign (thermometer code). The output of each comparator is then registered into four 64-to-6 bit decoders when CLK is changed from high to low. At the output of the decoders is a set of four 7-bit latches that are enabled (track) when the clock changes from high to low. From here, the output of the latches is coded into six LSBs from four columns, and four columns are coded into two MSBs. Finally, eight ECL output latches and buffers are used to drive the external loads. The conversion takes one clock cycle from the input to the data outputs.




Figure 2. Timing Diagram


Figure 3. Subcircuit Schematics

## 8-Bit, 750Msps Flash ADC

TOP VIEW


Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

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