## **General Description**

The MAX5762–MAX5765 32-channel, 16-bit, voltageoutput, digital-to-analog converters (DACs) are ideal for applications requiring a high number of programmable voltages. The MAX5762–MAX5765 can be programmed to calibrate the 16-bit input data word for gain and offset errors before updating the DAC output.

An SPI™-/QSPI™-/MICROWIRE™- or DSP-compatible serial interface controls the MAX5762–MAX5765. Each DAC channel has its own input, gain, and offset register. These three registers pass data through one of the two multiplier accumulator units (MACs), resulting in a voltage output corrected for gain and offset error. Each DAC channel has a double-buffered input structure to minimize the digital-noise feedthrough from the digital inputs to the outputs, and allows for synchronous or asynchronous updating of the outputs. The DAC outputs update independently or simultaneously with a single software or hardware command. The MAX5762–MAX5765 also provide a digital output (DOUT) that allows for readback or daisy chaining of multiple devices.

All DAC outputs are buffered and drive  $10k\Omega$  in parallel with 100pF. The MAX5762 has a 0 to +5V output range; the MAX5763 has a 0 to +10V output range; the MAX5764 has a -2.5V to +7.5V output range; and the MAX5765 has a -5V to +5V output range.

The MAX5762–MAX5765 are available in a 68-pin, 10mm x 10mm, TQFN package and a 64-pin, 12mm x 12mm, TQFP package. The MAX5762–MAX5765 are specified over the 0°C to +85°C temperature range. Refer to the MAX5732–MAX5735 data sheet for similar 16-bit, 32-channel DACs without offset and gain calibration.

### **Applications**

- Automatic Test Systems
- **Optical Router Controls**
- Industrial Process Controls
- Arbitrary Function Generators
- Avionics Equipment
- Minimum Component Count Analog Systems
- Digital Offset/Gain Adjustment

SPI/QSPI are trademarks of Motorola, Inc. MICROWIRE is a trademark of National Semiconductor Corp.

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**\_ Features** 

- Digital Offset and Gain Correction for Each DAC Channel
- Guaranteed Monotonic to 16 Bits
- 32 Individual DACs in a 10mm x 10mm, 68-Pin TQFN Package or 12mm x 12mm, 64-Pin TQFP Package
- Output Voltage Ranges

   0 to +5V (MAX5762)
   0 to +10V (MAX5763)
   -2.5V to +7.5V (MAX5764)
   -5V to +5V (MAX5765)
- ♦ Buffered Voltage Outputs Drive 10kΩ || 100pF
- Glitch-Free Power-Up
- SPI/QSPI/MICROWIRE- and DSP-Compatible 33MHz Serial Interface

# \_Ordering Information

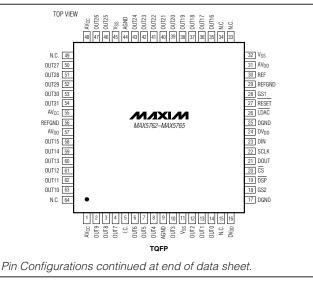
Pin Configurations

PART	PIN- PACKAGE	INL (LSB)	OUTPUT VOLTAGE RANGE (V)	PKG CODE
MAX5762UTK**	68 TQFN-EP*	±16	0 to +5	T6800-3
MAX5762UCB	64 TQFP	±16	0 to +5	C64-12

Ordering Information continued at end of data sheet.

**Note:** All devices are specified over the 0°C to +85°C temperature range.

\*EP = Exposed paddle. Connect to V<sub>SS</sub> or leave unconnected. \*\*Future product—contact factory for availability.



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For pricing, delivery, and ordering information, please contact Maxim/Dallas Direct! at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

### **ABSOLUTE MAXIMUM RATINGS**

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### ELECTRICAL CHARACTERISTICS—MAX5762 (0 to +5V Output Voltage Range)

 $(AV_{CC} = +5.25V \text{ to } +5.5V \text{ (Note 1)}, AV_{DD} = +5V \pm 5\%, DV_{DD} = +2.7V \text{ to } AV_{DD}, V_{SS} = AGND = DGND = REFGND = GS1 = GS2 = 0, V_{REF} = 3.0V, R_L = \infty, C_L = 50pF$  referenced to V\_{SS}, T\_A = T\_{MIN} \text{ to } T\_{MAX}, SCLK = 0, OFF-REG<sub>0-31</sub>= 0000h, GAIN-REG<sub>0-31</sub> = 0000h, unless otherwise noted. Typical values are at T\_A = +25°C.)

PARAMETER	SYMBOL	CONDITIONS		ТҮР	MAX	UNITS
DC CHARACTERISTICS						
Resolution	Ν		16			Bits
Integral Nonlinearity	INL	(Note 2)		±8	±16	LSB
Differential Nonlinearity	DNL	Guaranteed monotonic			±1	LSB
Offset Error (Without Digital Calibration)	V <sub>OS</sub>	$V_{SS} = -0.5V, AV_{CC} = +5.25V$ (Note 3)		±8	±40	mV
Offset Error (with Digital Calibration)	V <sub>OS-CAL</sub>	$V_{SS} = -0.5V, AV_{CC} = +5.25V$		±300		μV
Gain Error (Without Digital Correction)		(Note 3)		±0.1	±0.5	%FSR
Gain Error (with Digital Correction)				±0.05		%FSR
Gain Temperature Coefficient				20		ppm FSR/°C
DC Crosstalk		V <sub>SS</sub> = -0.5V, AV <sub>CC</sub> = +5.25V (Note 4)		50	250	μV
DYNAMIC CHARACTERISTICS						
Output-Voltage Settling Time	ts	Full-scale change to $\pm 0.5$ LSB		20		μs
Voltage-Output Slew Rate				1		V/µs
Digital Feedthrough		(Note 5)		5		nV-s
Digital-to-Analog Glitch Impulse		Major carry transition		120		nV-s
DAC-to-DAC Crosstalk		(Note 6)		15		nV-s
Output-Noise Spectral Density at 1kHz		Full-scale code		250		nV/√Hz

### ELECTRICAL CHARACTERISTICS—MAX5762 (0 to +5V Output Voltage Range) (continued)

 $(AV_{CC} = +5.25V \text{ to } +5.5V \text{ (Note 1)}, AV_{DD} = +5V \pm 5\%, DV_{DD} = +2.7V \text{ to } AV_{DD}, V_{SS} = AGND = DGND = REFGND = GS1 = GS2 = 0, V_{REF} = 3.0V, R_L = \infty, C_L = 50pF \text{ referenced to } V_{SS}, T_A = T_{MIN} \text{ to } T_{MAX}, SCLK = 0, OFF-REG_{0-31} = 0000h, GAIN-REG_{0-31} = 0000h, unless otherwise noted. Typical values are at T_A = +25°C.)$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
ANALOG OUTPUTS (OUT0 TO O	UT31)	1				
Output Voltage Range		V <sub>SS</sub> = -0.5V, AV <sub>CC</sub> = +5.25V (Note 1)	0		5	V
Resistive Load to Ground			10	50		kΩ
Capacitive Load to Ground					100	рF
DC-Output Impedance				0.1		Ω
Ohard Oireadt Ormand		Sourcing, full scale, output connected to AGND		+5		
Short-Circuit Current		Sinking, zero scale, output connected to AV <sub>CC</sub>		-5		mA
GROUND-SENSE ANALOG INPU	TS (GS1 AN	D GS2)				
Input Voltage Range	V <sub>GS</sub>	Relative to AGND	-0.5		+0.5	V
Ground-Sense Gain	Ags		0.993	1.000	1.005	V/V
Input Resistance		$-0.5V \le V_{GS} \le +0.5V, V_{SS} = -0.5V$	35			kΩ
REFERENCE INPUT (REF)						
Input Resistance			1			MΩ
Reference Input Voltage Range	VREF	Referred to REFGND	2.900	3.000	3.100	V
DIGITAL INPUTS (CS, SCLK, DIN	, LDAC, RES	SET, DSP)				
Input-Voltage High	Maria	$+2.7V \le DV_{DD} \le + 3.6V$	0.7 x DV	DD		V
	VIH	$+3.6V < DV_{DD} \le +5.25V$	2.4			V
Input-Voltage Low	VIL				0.8	V
Input Capacitance	CIN			10		pF
Input Current	lin	Digital inputs = 0 or $DV_{DD}$			±1	μA
DIGITAL OUTPUT (DOUT)						
Output-Voltage Low	VOL	$I_{SINK} = 1$ mA, $DV_{DD} = +2.7V$ to $+3.6V$			0.4	V
Output-Voltage High	VOH	$I_{\text{SOURCE}} = 0.2 \text{mA}, \text{DV}_{\text{DD}} = +2.7 \text{V to } +3.6 \text{V}$	0.8 x DV	DD		V
Tri-State Leakage Current	١L				±10	μA
Tri-State Output Capacitance	COUT			10		pF
POWER REQUIREMENTS (AV <sub>CC</sub>	V <sub>SS</sub> , AGND	, AV <sub>DD</sub> , DV <sub>DD</sub> , DGND)				
Output-Amplifier Positive Supply Voltage	AV <sub>CC</sub>		4.75		5.50	V
Output-Amplifier Negative Supply Voltage	V <sub>SS</sub>		-0.5		0	V
Output-Amplifier Supply Voltage Difference		AV <sub>CC</sub> - V <sub>SS</sub>			5.75	V
Analog Supply Voltage	AV <sub>DD</sub>		4.75		5.25	V
Digital Supply Voltage	DVDD		2.7		AV <sub>DD</sub>	V
Analog Supply Correct	ΔI	$V_{OUT0}$ through $V_{OUT31} = 0$		10	15	mA
Analog Supply Current	AI <sub>DD</sub>	Software shutdown		10	-	μA



### ELECTRICAL CHARACTERISTICS—MAX5762 (0 to +5V Output Voltage Range) (continued)

(AV<sub>CC</sub> = +5.25V to +5.5V (Note 1), AV<sub>DD</sub> = +5V ±5%, DV<sub>DD</sub> = +2.7V to AV<sub>DD</sub>, V<sub>SS</sub> = AGND = DGND = REFGND = GS1 = GS2 = 0, V<sub>REF</sub> = 3.0V, R<sub>L</sub> = ∞, C<sub>L</sub> = 50pF referenced to V<sub>SS</sub>, T<sub>A</sub> = T<sub>MIN</sub> to T<sub>MAX</sub>, SCLK = 0, OFF-REG<sub>0-31</sub> = 0000h, GAIN-REG<sub>0-31</sub> = 0000h, unless otherwise noted. Typical values are at  $T_A = +25^{\circ}C$ .)

PARAMETER	SYMBOL		CONDITIONS	MIN	ТҮР	MAX	UNITS
Digital Supply Current		$DV_{DD} = +5V, V_{IH} = DV_{DD}, V_{IL} = 0,$ f <sub>SCLK</sub> = 20MHz, continuous writethrough			2.0	3.0	
	DI <sub>DD</sub>	$\label{eq:DVDD} \begin{split} DV_{DD} &= +5V, \ V_{IH} = +2.4V, \ V_{IL} = +0.8V, \\ f_{SCLK} &= 20MHz, \ continuous \ writethrough \end{split}$			5.0	6.0	mA
		Software shutdown			75		nA
Output-Amplifier Positive Supply	Alaa	Vouto through	V <sub>OUT31</sub> = 0		4	8	mA
Current	AICC	Software shutdo	own		20		μA
Output-Amplifier Negative Supply	laa		$V_{OUT0}$ through $V_{OUT31} = 0$		-4	-10	mA
Current	ISS	V <sub>SS</sub> = -0.5V Software shutdown			-20		μΑ
Power-Supply Rejection Ratio	PSRR	For AV <sub>CC</sub> , AV <sub>DD</sub> , and V <sub>SS</sub>			-95		dB

### ELECTRICAL CHARACTERISTICS—MAX5763 (0 to +10V Output Voltage Range)

 $(AV_{CC} = +10.5V \text{ to } +11V \text{ (Note 1)}, \text{ } AV_{DD} = +5V \pm 5\%, \text{ } DV_{DD} = +2.7V \text{ to } AV_{DD}, \text{ } V_{SS} = \text{AGND} = \text{DGND} = \text{REFGND} = \text{GS1} = \text{GS2} = 0, \text{ } P_{CC} = +10.5V \text{ to } +11V \text{ (Note 1)}, \text{ } AV_{DD} = +5V \pm 5\%, \text{ } DV_{DD} = +2.7V \text{ to } AV_{DD}, \text{ } V_{SS} = \text{AGND} = \text{DGND} = \text{REFGND} = \text{GS1} = \text{GS2} = 0, \text{ } P_{CC} = +10.5V \text{ to } +11V \text{ (Note 1)}, \text{ } AV_{DD} = +5V \pm 5\%, \text{ } DV_{DD} = +2.7V \text{ to } AV_{DD}, \text{ } V_{SS} = \text{AGND} = \text{DGND} = \text{REFGND} = \text{GS1} = \text{GS2} = 0, \text{ } P_{CC} = +10.5V \text{ to } +11V \text{ (Note 1)}, \text{ } P_{CC} = +10.5V \text{ to } +11V \text{ (Note 1)}, \text{ } P_{CC} = +10.5V \text{ to } +11V \text{ (Note 1)}, \text{ } P_{CC} = +10.5V \text{ to } +11V \text{ (Note 1)}, \text{ } P_{CC} = +10.5V \text{ to } +11V \text{ (Note 1)}, \text{ } P_{CC} = +10.5V \text{ to } +1000 \text{ cm}^{-1}$ VREF = +3.0V, RL = ∞, CL = 50pF referenced to VSS, TA = TMIN to TMAX, SCLK = 0, all offset and gain registers = 0000h, unless otherwise noted. Typical values are at  $T_A = +25^{\circ}C$ .)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
DC CHARACTERISTICS	•	•	•			•
Resolution	Ν		16			Bits
Integral Nonlinearity	INL	(Note 2)		±8	±16	LSB
Differential Nonlinearity	DNL	Guaranteed monotonic			±1	LSB
Offset Error (Without Digital Calibration)	Vos	$V_{SS} = -0.5V, AV_{CC} = +10.5V (Note 3)$		±8	±40	mV
Offset Error (with Digital Calibration)	Vos-cal	V <sub>SS</sub> = -0.5V, AV <sub>CC</sub> = +10.5V		±300		μV
Gain Error (Without Digital Calibration)		(Note 3)		±0.1	±0.5	%FSR
Gain Error (with Digital Calibration)				±0.05		%FSR
Gain Temperature Coefficient				20		ppm FSR/°C
DC Crosstalk		V <sub>SS</sub> = -0.5V, AV <sub>CC</sub> = +10.5V (Note 4)		50	250	μV
DYNAMIC CHARACTERISTICS (	fsc∟k = 20M⊦	łz)				
Output-Voltage Settling Time	ts	Full-scale change to $\pm 0.5$ LSB		20		μs
Voltage-Output Slew Rate				1		V/µs
Digital Feedthrough		(Note 5)		5		nV-s
Digital-to-Analog Glitch Impulse		Major carry transition		120		nV-s
DAC-to-DAC Crosstalk		(Note 6)		25		nV-s





### ELECTRICAL CHARACTERISTICS—MAX5763 (0 to +10V Output Voltage Range)(continued)

 $(AV_{CC} = +10.5V \text{ to } +11V \text{ (Note 1)}, AV_{DD} = +5V \pm 5\%, DV_{DD} = +2.7V \text{ to } AV_{DD}, V_{SS} = AGND = DGND = REFGND = GS1 = GS2 = 0, V_{REF} = +3.0V, R_L = \infty, C_L = 50pF \text{ referenced to } V_{SS}, T_A = T_{MIN} \text{ to } T_{MAX}, SCLK = 0, all offset and gain registers = 0000h, unless otherwise noted. Typical values are at T_A = +25°C.)$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS	
Output-Noise Spectral Density at 1kHz	Full-scale code			250		nV/√Hz	
ANALOG OUTPUTS (OUT0 to OL	JT31)					•	
Output Voltage Range		V <sub>SS</sub> = -0.5V, AV <sub>CC</sub> = +10.5V (Note 1)	0		10	V	
Resistive Load to Ground			10	50		kΩ	
Capacitive Load to Ground					100	pF	
DC-Output Impedance				0.1		Ω	
Short-Circuit Current		Sourcing, full scale, output connected to AGND		+5		mA	
Short-Circuit Current		Sinking, zero scale, output connected to $AV_{CC}$		-5		MA	
GROUND-SENSE ANALOG INPU	TS (GS1 and	I GS2)					
Input Voltage Range	V <sub>GS</sub>	Relative to AGND	-0.5		+0.5	V	
Ground-Sense Gain	Ags		0.993	1.000	1.005	V/V	
Input Resistance		$-0.5V \leq V_{GS} \leq +0.5V, V_{SS} = -0.5V$	70			kΩ	
<b>REFERENCE INPUT (REF)</b>							
Input Resistance			1			MΩ	
Reference Input Voltage Range	VREF	Referred to REFGND	2.900	3.000	3.100	V	
DIGITAL INPUTS (CS, SCLK, DIN	, IDAC, RES	SET, DSP)					
Input Voltage High	Maria	$+2.7V \le DV_{DD} \le + 3.6V$	0.7 x DV	DD		V	
Input-Voltage High	VIH	$+3.6V < DV_{DD} \le +5.25V$	2.4			v	
Input-Voltage Low	VIL				0.8	V	
Input Capacitance	CIN			10		pF	
Input Current	lin	Digital inputs = 0 or $DV_{DD}$			±1	μA	
DIGITAL OUTPUT (DOUT)							
Output-Voltage Low	VOL	I <sub>SINK</sub> = 1mA			0.4	V	
Output-Voltage High	V <sub>OH</sub>	I <sub>SOURCE</sub> = 0.2mA	0.8 x DV	DD		V	
Tri-State Leakage Current	١L				±10	μA	
Tri-State Output Capacitance	COUT			10		pF	
POWER REQUIREMENTS (AVCC	, V <sub>SS</sub> , AGND	, AV <sub>DD</sub> , DV <sub>DD</sub> , DGND)					
Output-Amplifier Positive Supply Voltage	AV <sub>CC</sub>		10		11	V	
Output-Amplifier Negative Supply Voltage	V <sub>SS</sub>		-0.5		0	V	
Output-Amplifier Supply Voltage Difference		AV <sub>CC</sub> - V <sub>SS</sub>			11	V	
Analog Supply Voltage	AV <sub>DD</sub>		4.75		5.25	V	
Digital Supply Voltage	DVDD		2.7		AVDD	V	



### ELECTRICAL CHARACTERISTICS—MAX5763 (0 to +10V Output Voltage Range)(continued)

 $(AV_{CC} = +10.5V \text{ to } +11V \text{ (Note 1)}, AV_{DD} = +5V \pm 5\%, DV_{DD} = +2.7V \text{ to } AV_{DD}, V_{SS} = AGND = DGND = REFGND = GS1 = GS2 = 0, V_{REF} = +3.0V, R_L = \infty, C_L = 50pF \text{ referenced to } V_{SS}, T_A = T_{MIN} \text{ to } T_{MAX}, SCLK = 0, all offset and gain registers = 0000h, unless otherwise noted. Typical values are at T_A = +25°C.)$ 

PARAMETER	SYMBOL		CONDITIONS	MIN	ТҮР	МАХ	UNITS
Appleg Supply Surrent	Alaa	VOUT0 through	$V_{OUT31} = 0$		10	15	mA
Analog Supply Current	AI <sub>DD</sub>	Software shutd	own		10		μA
Digital Supply Current		55	i <sub>H</sub> = DV <sub>DD</sub> , V <sub>IL</sub> = 0, ε, continuous writethrough		2.0	3.0	~
	DI <sub>DD</sub>	55	IH = +2.4V, VIL = +0.8V, c, continuous writethrough		5.0 6.0		mA
		Software shutd	own		75		nA
Output-Amplifier Positive Supply	Alcc	V <sub>OUT0</sub> through	$V_{OUT31} = 0$		4	8	mA
Current	AICC	Software shutd	own		20		μA
Output-Amplifier Negative Supply			$V_{OUT0}$ through $V_{OUT31} = 0$		-4	-10	mA
Current	I <sub>SS</sub>	$V_{SS} = -0.5V$	Software shutdown		-20		μA
Power-Supply Rejection Ratio	PSRR				-95		dB

### ELECTRICAL CHARACTERISTICS—MAX5764 (-2.5V to +7.5V Output Voltage Range)

 $(AV_{CC} = +7.75V \text{ to } +8.25V \text{ (Note 1)}, AV_{DD} = +5V \pm 5\%, DV_{DD} = +2.7V \text{ to } AV_{DD}, V_{SS} = -2.75V \text{ to } -3.25V, AGND = DGND = REFGND = GS1 = GS2 = 0, program the offset DAC to 4000h. V_{REF} = +3.0V, R_L = ∞, C_L = 50pF referenced to V_{SS}, T_A = T_{MIN} \text{ to } T_{MAX}, SCLK = 0, all offset and gain registers = 0000h, unless otherwise noted. Typical values are at T_A = +25°C.)$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
DC CHARACTERISTICS		·	•			•
Resolution	N		16			Bits
Integral Nonlinearity	INL	(Note 2)		±8	±16	LSB
Differential Nonlinearity	DNL	Guaranteed monotonic			±1	LSB
Offset Error (Without Digital Calibration)	V <sub>OS</sub>	V <sub>SS</sub> = -3.25V, AV <sub>CC</sub> = +7.75V (Note 3)		±8	±40	mV
Offset Error (with Digital Calibration)	Vos-cal	$V_{SS} = -3.25V, AV_{CC} = +7.75V$		±300		μV
Gain Error (Without Digital Calibration)		(Note 3)		±0.1	±0.5	%FSR
Gain Error (with Digital Calibration)				±0.05		%FSR
Gain Temperature Coefficient				20		ppm FSR/°C
DC Crosstalk		V <sub>SS</sub> = -3.25V, AV <sub>CC</sub> = +7.75V (Note 4)		50	250	μV

### ELECTRICAL CHARACTERISTICS—MAX5764 (-2.5V to +7.5V Output Voltage Range) (continued)

 $(AV_{CC} = +7.75V \text{ to } +8.25V \text{ (Note 1)}, AV_{DD} = +5V \pm 5\%, DV_{DD} = +2.7V \text{ to } AV_{DD}, V_{SS} = -2.75V \text{ to } -3.25V, AGND = DGND = REFGND = GS1 = GS2 = 0, program the offset DAC to 4000h. V_{REF} = +3.0V, R_L = \infty, C_L = 50pF referenced to V_{SS}, T_A = T_{MIN} \text{ to } T_{MAX}, SCLK = 0, all offset and gain registers = 0000h, unless otherwise noted. Typical values are at T_A = +25°C.)$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
DYNAMIC CHARACTERISTICS (f	SCLK = 20MI	Hz)	1			
Output-Voltage Settling Time	ts	Full-scale change to $\pm 0.5$ LSB		20		μs
Voltage-Output Slew Rate		-		1		V/µs
Digital Feedthrough		(Note 5)		5		nV-s
Digital-to-Analog Glitch Impulse		Major carry transition		120		nV-s
DAC-to-DAC Crosstalk		(Note 6)		25		nV-s
Output-Noise Spectral Density at 1kHz		Full-scale code		250		nV/√Hz
ANALOG OUTPUTS (OUT0 to OL	JT31)		1			
Output Voltage Range		V <sub>SS</sub> = -2.75V, AV <sub>CC</sub> = +7.75V (Note 1)	-2.5		+7.5	V
Resistive Load to Ground			10	50		kΩ
Capacitive Load to Ground					100	рF
DC Output Impedance				0.1		Ω
		Sourcing, full scale, output connected to AGND		+5		mA
Short-Circuit Current		Sinking, zero scale, output connected to AV <sub>CC</sub>		-5		
GROUND-SENSE ANALOG INPU	TS (GS1 and	GS2)				
Input Voltage Range	VGS	Relative to AGND	-0.5		+0.5	V
Ground-Sense Gain	AGS		0.993	1.000	1.005	V/V
Input Resistance		$-0.5V \le V_{GS} \le +0.5V, V_{SS} = -2.75V$	70			kΩ
REFERENCE INPUT (REF)						
Input Resistance			1			MΩ
Reference Input Voltage Range	VREF	Referred to REFGND	2.900	3.000	3.100	V
DIGITAL INPUTS (CS, SCLK, DIN	I, LDAC, RES	SET, DSP)				
Input-Voltage High	VIH	$+2.7V \le DV_{DD} \le +3.6V$	0.7 x DV <sub>DD</sub>			V
		$+3.6V < DV_{DD} \le +5.25V$	2.4			
Input-Voltage Low	VIL				0.8	V
Input Capacitance	CIN			10		pF
Input Current	l <sub>IN</sub>	Digital inputs = 0 or $DV_{DD}$			±1	μA
DIGITAL OUTPUT (DOUT)						
Output-Voltage Low	Vol	I <sub>SINK</sub> = 1mA			0.4	V
Output-Voltage High	Vон	I <sub>SOURCE</sub> = 0.2mA	0.8 x DV <sub>DD</sub>			V
Tri-State Leakage Current	١L				±10	μA
Tri-State Output Capacitance	Cout			10		pF



### ELECTRICAL CHARACTERISTICS—MAX5764 (-2.5V to +7.5V Output Voltage Range) (continued)

 $(AV_{CC} = +7.75V \text{ to } +8.25V \text{ (Note 1)}, AV_{DD} = +5V \pm 5\%, DV_{DD} = +2.7V \text{ to } AV_{DD}, V_{SS} = -2.75V \text{ to } -3.25V, AGND = DGND = REFGND = GS1 = GS2 = 0, program the offset DAC to 4000h. V_{REF} = +3.0V, R_L = ∞, C_L = 50pF referenced to V_{SS}, T_A = T_{MIN} \text{ to } T_{MAX}, SCLK = 0, all offset and gain registers = 0000h, unless otherwise noted. Typical values are at T_A = +25°C.)$ 

PARAMETER	SYMBOL		CONDITIONS	MIN	TYP	MAX	UNITS
POWER REQUIREMENTS (AV <sub>CC</sub> ,	V <sub>SS</sub> , AGND,	, AV <sub>DD</sub> , DV <sub>DD</sub> , D	GND)				
Output-Amplifier Positive Supply Voltage	AV <sub>CC</sub>			7.50		8.25	V
Output-Amplifier Negative Supply Voltage	V <sub>SS</sub>					-2.50	V
Output-Amplifier Supply Voltage Difference		AV <sub>CC</sub> - V <sub>SS</sub>				11	V
Analog Supply Voltage	AV <sub>DD</sub>			4.75		5.25	V
Digital Supply Voltage	DVDD			2.7		AV <sub>DD</sub>	V
Analog Supply Current	Aldd	V <sub>OUT0</sub> through	$V_{OUT31} = 0$		10	15	mA
Analog Supply Current	AIDD	Software shutd	own		10		μA
	DI <sub>DD</sub>	55 / 1	H = DV <sub>DD</sub> , VIL = 0, z, continuous writethrough		2.0	3.0	
Digital Supply Current		55	H = +2.4V, V <sub>IL</sub> = +0.8V, c, continuous writethrough		5.0	6.0	mA
		Software shutd	own		75		nA
Output-Amplifier Positive Supply	Alee	VOUT0 through	$V_{OUT31} = 0$		4	8	mA
Current	Alcc	Software shutd	own		20		μA
Output-Amplifier Negative Supply	100	Vss = -2.75V	$V_{OUT0}$ through $V_{OUT31} = 0$		-4	-10	mA
Current	I <sub>SS</sub>	v <sub>55</sub> = -2.75v	Software shutdown		-20		μA
Power-Supply Rejection Ratio	PSRR				-95		dB

# ELECTRICAL CHARACTERISTICS—MAX5765 (-5V to +5V Output Voltage Range)

 $(AV_{CC} = +5.25V \text{ to } +5.5V (\text{Note 1}), AV_{DD} = +5V \pm 5\%, DV_{DD} = +2.7V \text{ to } AV_{DD}, V_{SS} = -5.25V \text{ to } -5.5V, AGND = DGND = REFGND = GS1 = GS2 = 0, program the offset DAC to 8000h. V_{REF} = +3.0V, R_L = ∞, C_L = 50pF referenced to V_{SS}, T_A = T_{MIN} \text{ to } T_{MAX}, SCLK = 0, all offset and gain registers = 0000h, unless otherwise noted. Typical values are at T_A = +25°C.)$ 

PARAMETER	SYMBOL	CONDITIONS		TYP	MAX	UNITS
DC CHARACTERISTICS						
Resolution	N		16			Bits
Integral Nonlinearity	INL	(Note 2)		±8	±16	LSB
Differential Nonlinearity	DNL	Guaranteed monotonic			±1	LSB
Offset Error (Without Digital Calibration)	V <sub>OS</sub>	V <sub>SS</sub> = -5.25V, AV <sub>CC</sub> = +5.25V (Note 3)		±8	±40	mV
Offset Error (with Digital Calibration)	V <sub>OS-CAL</sub>	$V_{SS} = -5.25V, AV_{CC} = +5.25V$		±300		μV

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### ELECTRICAL CHARACTERISTICS—MAX5765 (-5V to +5V Output Voltage Range) (continued)

(AV<sub>CC</sub> = +5.25V to +5.5V (Note 1), AV<sub>DD</sub> = +5V  $\pm$ 5%, DV<sub>DD</sub> = +2.7V to AV<sub>DD</sub>, V<sub>SS</sub> = -5.25V to -5.5V, AGND = DGND = REFGND = GS1 = GS2 = 0, program the offset DAC to 8000h. V<sub>REF</sub> = +3.0V, R<sub>L</sub> =  $\infty$ , C<sub>L</sub> = 50pF referenced to V<sub>SS</sub>, T<sub>A</sub> = T<sub>MIN</sub> to T<sub>MAX</sub>, SCLK = 0, all offset and gain registers = 0000h, unless otherwise noted. Typical values are at T<sub>A</sub> = +25°C.)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
Gain Error (Without Digital Calibration)		(Note 3)		±0.1	±0.5	%FSR
Gain Error (with Digital Calibration)				±0.05		%FSR
Gain Temperature Coefficient				20		ppm FSR/°C
DC Crosstalk		V <sub>SS</sub> = -5.50V, AV <sub>CC</sub> = +5.25V (Note 4)		50	250	μV
<b>DYNAMIC CHARACTERISTICS (f</b>	SCLK = 20MH	łz)				
Output-Voltage Settling Time	ts	Full-scale change to $\pm 0.5$ LSB		20		μs
Voltage-Output Slew Rate				1		V/µs
Digital Feedthrough		(Note 5)		5		nV-s
Digital-to-Analog Glitch Impulse		Major carry transition		120		nV-s
DAC-to-DAC Crosstalk		(Note 6)		25		nV-s
Output-Noise Spectral Density at 1kHz		Full-scale code		250		nV/√Hz
ANALOG OUTPUTS (OUT0 to OU	JT31)					
Output Voltage Range		V <sub>SS</sub> = -5.25V, AV <sub>CC</sub> = +5.25V (Note 1)	-5		+5	V
Resistive Load to Ground			10	50		kΩ
Capacitive Load to Ground					100	рF
DC Output Impedance				0.1		Ω
		Sourcing, full scale, output connected to AGND		+5		
Short-Circuit Current		Sinking, zero scale, output connected to AV <sub>CC</sub>	-5			- mA
GROUND-SENSE ANALOG INPU	TS (GS1 and	GS2)				
Input Voltage Range	V <sub>GS</sub>	Relative to AGND	-0.5		+0.5	V
Ground-Sense Gain	AGS		0.993	1.000	1.005	V/V
Input Resistance		$-0.5V \le V_{GS} \le +0.5V, V_{SS} = -5.25V$	70			kΩ
REFERENCE INPUT (REF)						
Input Resistance			1			MΩ
Reference Input Voltage Range	V <sub>REF</sub>	Referred to REFGND	2.900	3.000	3.100	V
DIGITAL INPUTS (CS, SCLK, DIN	, LDAC, RES	ET, DSP)				
Input-Voltage High	VIH	$+2.7V \le DV_{DD} \le +3.6V$	0.7 x DV <sub>DD</sub>			V
		$+3.6V < DV_{DD} \le 5.25V$	2.4			
Input-Voltage Low	VIL				0.8	V
Input Capacitance	CIN			10		pF



### ELECTRICAL CHARACTERISTICS—MAX5765 (-5V to +5V Output Voltage Range) (continued)

 $(AV_{CC} = +5.25V \text{ to } +5.5V (\text{Note 1}), AV_{DD} = +5V \pm 5\%, DV_{DD} = +2.7V \text{ to } AV_{DD}, V_{SS} = -5.25V \text{ to } -5.5V, AGND = DGND = REFGND = GS1 = GS2 = 0, program the offset DAC to 8000h. V_{REF} = +3.0V, R_L = ∞, C_L = 50pF referenced to V_{SS}, T_A = T_{MIN} \text{ to } T_{MAX}, SCLK = 0, all offset and gain registers = 0000h, unless otherwise noted. Typical values are at T_A = +25°C.)$ 

PARAMETER	SYMBOL		CONDITIONS	MIN	ТҮР	MAX	UNITS
DIGITAL OUTPUT (DOUT)							
Output-Voltage Low	VOL	I <sub>SINK</sub> = 1mA				0.4	V
Output-Voltage High	V <sub>OH</sub>	ISOURCE = 0.2n	nA	0.8 x DV <sub>DD</sub>			V
Tri-State Leakage Current	١L					±10	μA
Tri-state Output Capacitance	COUT				10		pF
POWER REQUIREMENTS (AV <sub>CC</sub>	, V <sub>SS</sub> , AGND	, AV <sub>DD</sub> , DV <sub>DD</sub> , D	GND)				
Output-Amplifier Positive Supply Voltage	AV <sub>CC</sub>			4.75		5.50	V
Output-Amplifier Negative Supply Voltage	V <sub>SS</sub>			-5.50		-4.75	V
Output-Amplifier Supply Voltage Difference		AV <sub>CC</sub> - V <sub>SS</sub>				11	V
Analog Supply Voltage	AVDD			4.75		5.25	V
Digital Supply Voltage	DVDD			2.7		AV <sub>DD</sub>	V
Analog Supply Current	Aloo	VOUT0 through	V <sub>OUT31</sub> = 0		10	15	mA
Analog Supply Current	AIDD	Software shutdo	own		10		μA
		$DV_{DD} = +5V, V$ f <sub>SCLK</sub> = 20MHz		2.0	3.0		
Digital Supply Current	DI <sub>DD</sub>		IH = +2.4V, VIL = +0.8V, , continuous writethrough		5.0	6.0	mA
		Software shutdo	own		75		nA
Output-Amplifier Positive Supply	Alee	V <sub>OUT0</sub> through	Vout31 = 0		4	8	mA
Current	Alcc	Software shutdo	Software shutdown		20		μA
Output-Amplifier Negative Supply	ISS	V <sub>SS</sub> = -5.25V	$V_{OUT0}$ through $V_{OUT31} = 0$		-4	-10	mA
Current	188	v 350.20V	Software shutdown		-20		μΑ
Power-Supply Rejection Ratio	PSRR				-95		dB

### **TIMING CHARACTERISTICS**

(Figures 1 and 2,  $AV_{DD} = +4.75V$  to +5.25V,  $DV_{DD} = +4.75V$  to +5.25V, AGND = DGND = REFGND = GS1 = GS2 = 0,  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at  $T_A = +25^{\circ}C$ .)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
Serial Clock Frequency	<b>f</b> SCLK		0		33	MHz
SCLK Pulse-Width High	tсн		10			ns
SCLK Pulse-Width Low	tCL		10			ns
SCLK Fall to $\overline{CS}$ Fall Setup Time	tscs		6			ns

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### TIMING CHARACTERISTICS (continued)

(Figures 1 and 2,  $AV_{DD} = +4.75V$  to +5.25V,  $DV_{DD} = +4.75V$  to +5.25V, AGND = DGND = REFGND = GS1 = GS2 = 0,  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted. Typical values are at  $T_A = +25^{\circ}C$ .)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
CS Fall to SCLK Fall Setup Time	tcss		5			ns
CS Rise to SCLK Fall	tCS1	At end of cycle in SPI mode only	15			ns
SCLK Fall to $\overline{CS}$ Rise Setup Time	tCS2		0			ns
DIN to SCLK Fall Setup Time	tDS		10			ns
DIN to SCLK Fall Hold Time	tDH		2			ns
SCLK Fall to DOUT Fall	tSCL	Load capacitance = 20pF			20	ns
SCLK Fall to DOUT Rise	tsdh	Load capacitance = 20pF			20	ns
CS Pulse-Width High	<b>t</b> CSPWH		50			ns
CS Pulse-Width Low	tCSPWL	(Note 7)	20			ns
LDAC Pulse-Width Low	<b>t</b> LDAC		20			ns
RESET Pulse-Width Low	<b>t</b> CLR		50			ns

### TIMING CHARACTERISTICS

(Figures 1 and 2,  $AV_{DD} = +4.75V$  to +5.25V, **DV<sub>DD</sub> = +2.7V** to +5.25V, AGND = DGND = REFGND = GS1 = GS2 = 0, T<sub>A</sub> = T<sub>MIN</sub> to T<sub>MAX</sub>, unless otherwise noted. Typical values are at T<sub>A</sub> = +25°C.)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	МАХ	UNITS
Serial Clock Frequency	<b>f</b> SCLK		0		25	MHz
SCLK Pulse-Width High	tсн		10			ns
SCLK Pulse-Width Low	tCL		10			ns
SCLK Fall to CS Fall Setup Time	tscs		10			ns
CS Fall to SCLK Fall Setup Time	tcss		10			ns
CS Rise to SCLK Fall	tCS1	At end of cycle in SPI mode only	18			ns
SCLK Fall to CS Rise Setup Time	tCS2		0			ns
DIN to SCLK Fall Setup Time	tDS		10			ns
DIN to SCLK Fall Hold Time	t <sub>DH</sub>		2			ns
SCLK Fall to DOUT Fall	tSCL	Load capacitance = 20pF			25	ns
SCLK Fall to DOUT Rise	tsdh	Load capacitance = 20pF			25	ns
CS Pulse-Width High	<b>t</b> CSPWH		50			ns
CS Pulse-Width Low	tCSPWL	(Note 7)	20			ns
LDAC Pulse-Width Low	<b>T</b> LDAC		20			ns
RESET Pulse-Width Low	t <u>CLR</u>		50			ns

Note 1: AV<sub>CC</sub> should be at least 0.25V higher than the maximum output voltage required from the DAC.

Note 2: Linearity guaranteed for full code range.

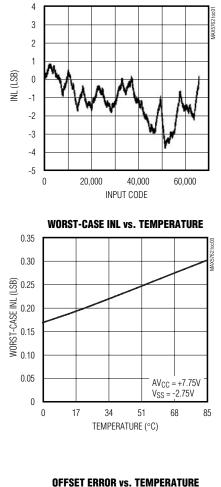
Note 3: Offset error is measured at code 0. Gain error is measured at code FFFFh.

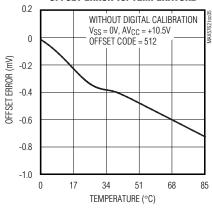
- Note 4: DC crosstalk is the change in the output level of one DAC at midscale in response to the full-scale output change of all other DACs.
- **Note 5:** Digital feedthrough is a measure of the impulse injected into the analog outputs from the digital control inputs when the device is not being written to. It is measured with a worst-case change on the digital inputs.
- **Note 6:** DAC-to-DAC crosstalk is the glitch impulse that appears at the output of one converter due to both the digital change and subsequent analog output change at another converter.
- **Note 7:** In DSP mode, maintain the maximum  $\overline{CS}$  pulse width low to  $\leq$  16 SCLK cycles.



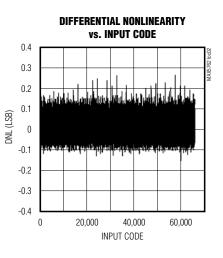
**INTERNAL NONLINEARITY vs. INPUT CODE** 

 $(T_A = +25^{\circ}C, unless otherwise noted.)$ 

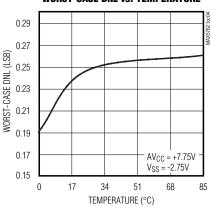


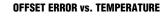


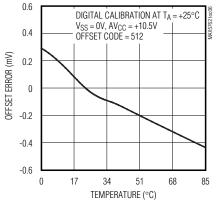
### **Typical Operating Characteristics**



**WORST-CASE DNL vs. TEMPERATURE** 

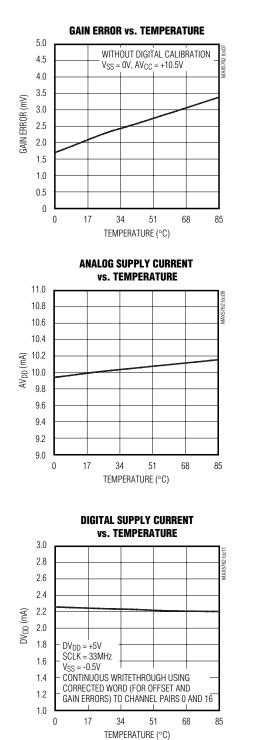


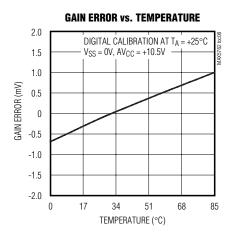




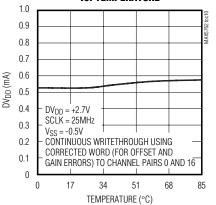
# **Typical Operating Characteristics (continued)**

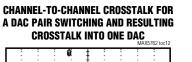
 $(T_A = +25^{\circ}C, unless otherwise noted.)$ 

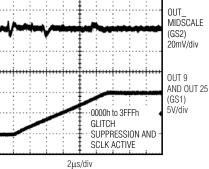




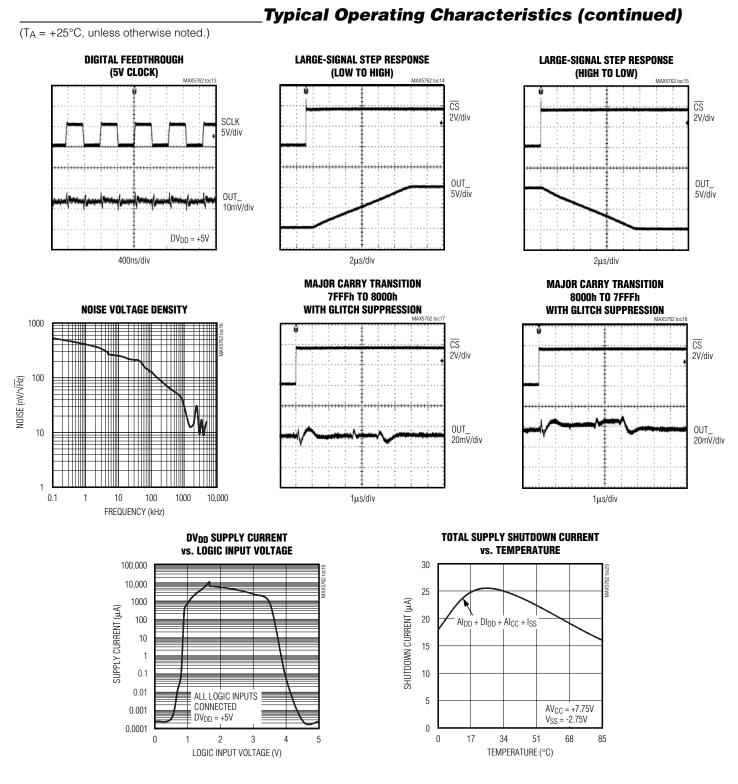
#### DIGITAL SUPPLY CURRENT vs. TEMPERATURE







#### 



# \_Pin Description

PIN			
68 TQFN	64 TQFP	NAME	FUNCTION
1, 16, 34, 35, 36, 51, 52, 53, 68	15, 33, 34, 49, 64	N.C.	No Connection. Not internally connected.
2, 50, 59	1, 48, 55	AV <sub>CC</sub>	Output-Amplifier Positive Supply Input. Bypass to AGND with a 0.1µF capacitor.
3	2	OUT9	DAC9 Buffered Analog Output Voltage
4	3	OUT8	DAC8 Buffered Analog Output Voltage
5	4	OUT7	DAC7 Buffered Analog Output Voltage
6	5	I.C.	Internally Connected. Do not make any connections to I.C.
7	6	OUT6	DAC6 Buffered Analog Output Voltage
8	7	OUT5	DAC5 Buffered Analog Output Voltage
9	8	OUT4	DAC4 Buffered Analog Output Voltage
10, 46	9, 44	AGND	Analog Ground
11	10	OUT3	DAC3 Buffered Analog Output Voltage
12, 33, 47	11, 32, 45	Vss	Bypass to AGND with a 0.1µF Capacitor
13	12	OUT2	DAC2 Buffered Analog Output Voltage
14	13	OUT1	DAC1 Buffered Analog Output Voltage
15	14	OUT0	DAC0 Buffered Analog Output Voltage
17, 25	16, 24	DVDD	Digital Power-Supply Input. Bypass to DGND with a 0.1µF capacitor.
18, 26	17, 25	DGND	Digital Ground
19	18	GS2	Ground-Sense Analog Input 2. Offsets the DAC amplifier outputs OUT0, OUT1, OUT2, OUT16, OUT17, and OUT18 by $\pm 0.5V$ to compensate for a remote system ground potential difference.
20	19	DSP	Digital Serial-Interface Mode-Select Input. Drive low for DSP interface mode. Drive high for SPI interface mode.
21	20	CS	Active-Low Digital Chip-Select Input
22	21	DOUT	Digital Serial Data Output. Use DOUT to daisy chain or read the contents of the internal registers. DOUT data clocks out on the falling edge of SCLK, MSB first.
23	22	SCLK	Digital Serial Clock Input
24	23	DIN	Digital Serial Data Input. Data clocks in on the falling edge of SCLK.
27	26	LDAC	Active-Low Digital Load DAC Input. Drive this asynchronous input low to transfer the contents of the input register to their respective DAC registers and update all DAC outputs accordingly.
28	27	RESET	Active-Low Reset Input. Drive this asynchronous input low to initiate a power-on reset. See the <i>Power-On Reset</i> section for further information.

# **Pin Description (continued)**

PI	N		
68 TQFN	64 TQFP	NAME	FUNCTION
29	28	GS1	Ground-Sense Analog Input 1. Offsets the DAC amplifier outputs OUT3–OUT15 and OUT19–OUT31 by $\pm 0.5V$ to compensate for a remote system ground potential difference.
30, 60	29, 56	REFGND	Reference Ground
31	30	REF	Analog Reference Voltage Input. Connect a +3V reference to REF and bypass to REFGND with a 0.1 $\mu\text{F}$ capacitor.
32, 61	31, 57	AV <sub>DD</sub>	Analog Power-Supply Input. Bypass to AGND with a 0.1µF capacitor.
37	35	OUT16	DAC16 Buffered Analog Output Voltage
38	36	OUT17	DAC17 Buffered Analog Output Voltage
39	37	OUT18	DAC18 Buffered Analog Output Voltage
40	38	OUT19	DAC19 Buffered Analog Output Voltage
41	39	OUT20	DAC20 Buffered Analog Output Voltage
42	40	OUT21	DAC21 Buffered Analog Output Voltage
43	41	OUT22	DAC22 Buffered Analog Output Voltage
44	42	OUT23	DAC23 Buffered Analog Output Voltage
45	43	OUT24	DAC24 Buffered Analog Output Voltage
48	46	OUT25	DAC25 Buffered Analog Output Voltage
49	47	OUT26	DAC26 Buffered Analog Output Voltage
54	50	OUT27	DAC27 Buffered Analog Output Voltage
55	51	OUT28	DAC28 Buffered Analog Output Voltage
56	52	OUT29	DAC29 Buffered Analog Output Voltage
57	53	OUT30	DAC30 Buffered Analog Output Voltage
58	54	OUT31	DAC31 Buffered Analog Output Voltage
62	58	OUT15	DAC15 Buffered Analog Output Voltage
63	59	OUT14	DAC14 Buffered Analog Output Voltage
64	60	OUT13	DAC13 Buffered Analog Output Voltage
65	61	OUT12	DAC12 Buffered Analog Output Voltage
66	62	OUT11	DAC11 Buffered Analog Output Voltage
67	63	OUT10	DAC10 Buffered Analog Output Voltage
EP	—	EP	Exposed Paddle. Connect to V <sub>SS</sub> .

### **Detailed Description**

The MAX5762–MAX5765 are 32-channel, 16-bit, voltage-output DACs. All devices accept a +3.0V external reference input at REF. An internal offset DAC allows all outputs to be offset up to -5V (see Table 1).

A 33MHz SPI-/QSPI-/MICROWIRE- or DSP-compatible serial interface (see Figure 1) controls the MAX5762–MAX5765. Use DSP to select the DSP mode or the SPI/QSPI/MICROWIRE mode.

Each DAC has a double-buffered input structure to minimize the digital-noise feedthrough from the digital inputs to the outputs. The two buffers are organized as an input register followed by a DAC register. Input registers update the DAC registers independently or simultaneously with a single software or hardware command. Each DAC channel has its own offset and gain registers that calibrate offset and gain errors for a given channel. The MAX5762–MAX5765 are divided into two banks of 16 channels to provide real-time calibration of channel pairs. Channel bank pairs are ordered as (bank 0:bank 1) OUT0:OUT16, OUT1:OUT17 ... OUT14:OUT30, OUT15:OUT31. An offset DAC allows all the outputs to be offset up to -5V and dual ground-sensing inputs (GS1 and GS2) allow the output voltages to be referenced to remote grounds (see the *Functional Diagram*). Three output channels of each channel bank (bank 0 and bank 1) are referenced to GS2 (OUT0, OUT1, OUT2 in bank 0, and OUT16, OUT17, OUT18 in bank 1). All other output channels are referenced to GS1.

The MAX5762–MAX5765 analog and digital sections have separate power inputs (AV<sub>DD</sub> and DV<sub>DD</sub>). Separate power inputs are also provided for the output buffers (AV<sub>CC</sub> and V<sub>SS</sub>). Proprietary deglitch circuitry prevents output glitches at power-up and eliminates the need for power sequencing. A software-shutdown mode facilitates efficient power management. When shut down, the MAX5762–MAX5765 consume 50 $\mu$ A (typ) of supply current.

The MAX5762–MAX5765 have a 5µs (typ) wake-up time from shutdown mode to normal DAC operation.

All DACs provide buffered outputs that can drive 10k $\Omega$  in parallel with 100pF. The MAX5762 has a 0 to +5V output; the MAX5763 has a 0 to +10V output range; the MAX5764 has a -2.5V to +7.5V output range; and the MAX5765 has a -5V to +5V output range.

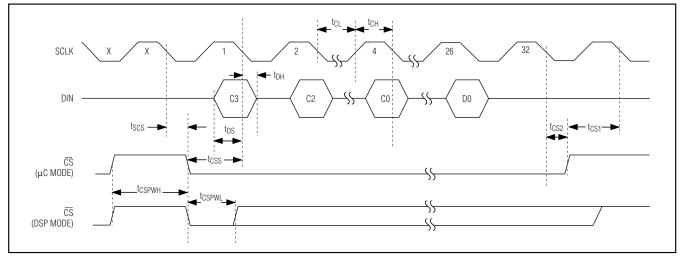


Figure 1. Serial-Interface Timing

#### Table 1. Offset DAC Codes

PART	D15	D14	D13	D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
MAX5762	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
MAX5763	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
MAX5764	0	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0
MAX5765	1	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0



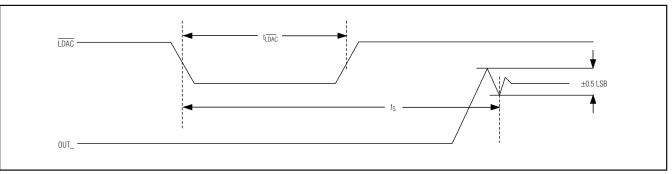


Figure 2. LDAC Timing

#### **External Reference Input (REF)**

The REF voltage sets the full-scale output voltage for all 32 DACs. REF accepts a  $+3.0V \pm 3\%$  input. Reference voltages outside these limits result in a degradation of device performance.

REF is a buffered input. The input impedance is  $1M\Omega$  minimum, and it does not vary with code. Use a highaccuracy, low-noise voltage reference such as the MAX6126AASA30 (3ppm/°C temperature drift and 0.02% initial accuracy) to improve static accuracy. REF does not accept AC signals.

#### Ground Sense (GS1 and GS2)

The MAX5762–MAX5765 include ground-sense inputs (GS1 and GS2) that allow the output voltages to be referenced to a remote ground. The ground-sense input GS1 is connected to channels OUT3–OUT15 (bank 0) and to channels OUT19–OUT31 (bank 1). The ground-sense input GS2 is connected to channels OUT0, OUT1, OUT2 (bank 0), and OUT16, OUT17, OUT18 (bank 1). Channels connected to the same ground sense can be accessed in pairs. For example, OUT0 and OUT16 can be accessed as a pair (see Figure 3 for pairing of output channels).

The ground-sense input voltage range (V<sub>GS1</sub> or V<sub>GS2</sub>) is -0.5V to +0.5V with respect to AGND. V<sub>GS</sub> is added to the output voltage with unity gain. Ensure that the resulting output voltage is within the valid output voltage range set by the power supplies. See the *Output Amplifiers (OUTO–OUT31)* section for the effect of the ground-sense inputs on the DAC outputs.

**Offset DAC** The MAX5762–MAX5765 feature an offset DAC that determines the output voltage range. While each device provides an allowable output voltage range, the offset DAC determines the end-point voltages of the range. Table 1 shows the offset DAC code necessary for each device's output voltage range. The MAX5762–MAX5765 offset DAC can be programmed with any of the three output voltage ranges. The specifications in the *Electrical Characteristics* table are only guaranteed (production tested) for the offset code associated with each particular part number.

The offset DAC is summed with the GS\_ input voltage (see the *Functional Diagram*). Any change in the offset DAC affects all 32 DACs.

The offset DAC is also double buffered with an input and DAC register.

Software commands for the MAC-bypass for all channels and load-DAC for all channels do not affect the offset DAC.

The data format for writing to the offset DAC is: control bits C3-C0 = 0011, address bits A5-A0 = 110000, 16 data bits, and 6 don't-care bits as shown in Table 2.

### Table 2. Offset DAC Input Data Format

CONTROL BITS	DATA BITS		DON'T-CARE BITS		
C3–C0	A5-A0	D15–D0	6 Don't-Care Bits		
0011	110000	See Table 1	XXXXXX		

# Table 3. Software Load-DAC Input DataFormat

CONTROL BITS	ADDRESS BITS	DATA BITS	DON'T-CARE BITS
C3–C0	A5–A0	D15–D0	6 Don't-Care Bits
0010	111111	XXXXXXXXXXX XXXXXX	XXXXXX



ADD	RESS E	BITS		CHANNEL BANK(S) SELECTED								
A5		A4	4									
0			HANNEL BANK									
0			HANNEL BANK									
1				IEL AND ALL CHANN	IEL OPERATION							
NPUT F BAN			REGIST		EGISTER, AND	OFFSET	REGISTE ADDRES		SS*			
A5	A4	A5	A4	A5	A4							
						A3	A2	A1	AO			
0	0	0	1	1	0	<u> </u>						
CHANI		CHAN	NEL 16	CHANNEL 0 AND	UCHANNEL 16	0	0	0	0			
CHAN	NEL 1	CHAN	NEL 17	CHANNEL 1 ANI	O CHANNEL 17	0	0	0	1			
CHAN	NEL 2	CHAN	NEL 18	CHANNEL 2 ANI	CHANNEL 18	0	0	1	0			
CHAN	VEL 3	CHAN	NEL 19	CHANNEL 3 ANI	CHANNEL 19	0	0	1	1			
CHAN	NEL 4	CHAN	NEL 20	CHANNEL 4 ANI	CHANNEL 20	0	1	0	0			
CHAN	NEL 5	CHAN	NEL 21	CHANNEL 5 ANI	CHANNEL 21	0	1	0	1			
CHAN	NEL 6	CHAN	NEL 22	CHANNEL 6 ANI	CHANNEL 22	0	1	1	0			
CHAN	CHANNEL 7		NEL 23	CHANNEL 7 ANI	CHANNEL 23	0	1	1	1			
CHAN	NEL 8	CHAN	NEL 24	CHANNEL 8 ANI	CHANNEL 24	1	0	0	0			
CHAN	NEL 9	CHAN	NEL 25	CHANNEL 9 ANI	) CHANNEL 25	1	0	0	1			
CHANN	EL 10	CHAN	NEL 26	CHANNEL 10 AM	ID CHANNEL 26	1	0	1	0			
CHANN	EL 11	CHAN	NEL 27	CHANNEL 11 AM	ID CHANNEL 27	1	0	1	1			
CHANN	EL 12	CHAN	NEL 28	CHANNEL 12 AM	ID CHANNEL 28	1	1	0	0			
CHANN	EL 13	CHAN	NEL 29	CHANNEL 13 AM	ID CHANNEL 29	1	1	0	1			
CHANN	EL 14	CHAN	NEL 30	CHANNEL 14 AM	ID CHANNEL 30	1	1	1	0			
CHANN	EL 15	CHAN	NEL 31	CHANNEL 15 AM	ID CHANNEL 31	1	1	1	1			
		FSET C	HANNEL		HANNELS (AD							
10	-					REGIS	TER SEL	ECTED				
<b>A3</b> 0		A2 0	<b>A1</b> 0	<b>AU</b> 0			OFFSET CHAN	NEL				
		0001 THROL			ALL CL	HANNELS (SOFT)	UNUSED		RADVCC			
1		1	1	1	ALL UI		COMMAND:		JILAGO			
ACCESS		FSET C ADDRE A2	HANNEL SS BITS A1	AND ALL C	HANNELS (AC	DDRESS B	ITS A5 AN TER SELI	ID A4 = 1 ECTED				
U	-			U			UNUSED					
1		1	1	1	ALL CH	TAININELS (SUFT)			STPASS			
						6 C2 C0 -		·				
ITS A5	AND A	4 = 00)	SS BITS						UKES			
A3 A2		A1	A0		REGISTER SELECTED							
<u>A3</u>		0	0	0		CONFIGURATION REGISTER (SEE TABLE 8) UNUSED						

Figure 3. Address Space When Accessing DAC Channel Register(s), Offset Register(s), Gain Register(s), and Special Registers



#### **Output Amplifiers (OUT0-OUT31)**

All DAC outputs are internally buffered. The internal buffers provide gain, improved load regulation, and glitch suppression for the DAC outputs. The output buffers slew at 1V/µs and can drive 10k $\Omega$  in parallel with 100pF. The output buffers are powered by AV<sub>CC</sub> and V<sub>SS</sub>. AV<sub>CC</sub> and V<sub>SS</sub> determine the maximum output voltage range of the device.

The input code, the voltage reference, the offset DAC output, the voltage on GS1 (or GS2), and the gain of the output amplifier determine the output voltage. Calculate V<sub>OUT</sub> as follows:

$$V_{OUT} = \frac{GAIN \times V_{REF} \times (DAC \ CODE - OFFSET \ DAC \ CODE)}{2^{16}} + V_{GS}$$

where GAIN = 5/3 for the MAX5762 and GAIN = 10/3 for the MAX5763/MAX5764/MAX5765.

#### Load-DAC (LDAC) Input

The MAX5762–MAX5765 feature an active-low  $\overline{LDAC}$  input that allows the outputs (OUT\_) to update asynchronously. Keep  $\overline{LDAC}$  high during normal operation (when the device is controlled only through the serial interface). Drive  $\overline{LDAC}$  low to simultaneously update all DAC outputs with data from their respective input registers. Figure 2 shows the  $\overline{LDAC}$  timing with respect to OUT\_.

A software command can also perform the  $\overline{\text{LDAC}}$  operation. To initiate  $\overline{\text{LDAC}}$  by software, set control bits C3–C0 = 0010, address bits A5–A0 = 111111, and all data bits to don't care. See Table 3 for the data format. This operation updates all DAC outputs simultaneously. The software load-DAC command for all channels does not affect the offset DAC.

#### **Software MAC-Bypass**

The MAX5762–MAX5765 feature a software MAC-bypass command that loads data into the DAC directly from DIN. Software MAC-bypass loads one DAC, a pair of DACs, or all 32 DACs with a data word (D15–D0) entered at DIN and the selected DAC output(s) are simultaneously updated. Software MAC-bypass bypasses gain and offset calibration, sending the input data directly to the DAC register immediately updating the selected DAC outputs. After executing MAC-bypass on a channel(s), previously calibrated data can be reloaded into the DAC by executing software load-DAC or hardware LDAC (see Figure 4). Using software MAC-bypass, the DAC output(s) can be set to the ground-sense value or any arbitrary value within the DAC output voltage range.

To activate software MAC-bypass, set control bits C3-C0 = 0111. The address bits (A5-A0) select the

DAC(s) to be updated, and the data bits (D15–D0) control the DAC output-voltage value. Table 4 shows the input data format for the software-controlled MACbypass command.

#### Reset (RESET)

The MAX5762–MAX5765 feature an active-low RESET logic input that asynchronously sets all the registers to code 0000h (power-up state). The serial interface can also issue a software-reset command. Setting the control bits C3-C0 = 1111 performs the same function as driving the logic input RESET low. Table 5 shows the reset data format for the software-controlled reset command. The software reset does not work in daisy-chain mode. Reprogram the offset DAC after asserting a software or hardware reset.

#### **Serial Interface**

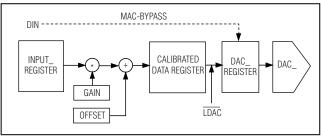
The MAX5762–MAX5765 allow channel updates either individually or in pairs. This is achieved by dividing the 32 channels into two channel banks, with 16 channels in each bank. Channel bank 0 contains output channels OUT0–OUT15, and channel bank 1 contains channels OUT16–OUT31. A channel from bank 0 is paired with a channel from bank 1 and is ordered as OUT0:OUT16, OUT1:OUT17...OUT14:OUT30, OUT15:OUT31.

### Table 4. MAC-Bypass Data Format

CONTROL BITS	ADDRESS BITS	DATA BITS	DON'T-CARE BITS
C3–C0	A5–A0	D15–D0	6 Don't-Care Bits
0111	See Figure 3	D15–D0	XXXXXX

### Table 5. Reset Data Format

CONTROL BITS	ADDRESS BITS	DATA BITS	DON'T-CARE BITS
C3–C0	A5-A0	D15–D0	6 Don't-Care Bits
1111	XXXXXX	XXXXXXXXXXX XXXXXX	XXXXXX







### **Table 6. Serial Data Format**

CONTROL BITS			6 DON'T- CARE BITS	
MSB			LSB	
INISD			LOD	

# Table 7. Gain and Offset Register InputData Format

REGISTER	CONTROL BITS (C3–C0)	ADDRESS BITS (A5–A0)	DATA BITS (D15–D0)	6 DON'T- CARE BITS	
Offset Register	1001	See Figure 3	See Table 10	XXXXXX	
Gain Register	1000	See Figure 3	See Table 9	XXXXXX	

### Table 8. Control-Bit Functions

A 3-wire SPI-/QSPI-/MICROWIRE- and DSP-compatible serial interface controls the MAX5762–MAX5765. The interface requires a 32-bit command word to control the device. The command word consists of 4 control bits, 6 address bits, 16 data bits, and 6 don't-care bits. Table 6 shows the general serial data format. The control bits control various write and read commands, as well as the load DAC and MAC-bypass commands. Table 8 shows the control-bit functions. The address bits select the register(s) to update. Figure 3 shows the address functions. The data bits control the voltage value of the DAC outputs. DIN data is clocked in at the falling edge of SCLK (Figure 1).

4	4 CONTROL BITS		s					
C3	C2	C1	C0	CONTROL-BIT DESCRIPTION				
0	0	0	0	No operation (NOP). No internal registers change state. The NOP command can be passed to DOUT depending on the state of the configuration register. Address bits A5–A0 and data bits D15–D0 are ignored.				
0	0	0	1	This instruction writes and calibrates the 16-bit input data word for gain and offset errors. Drive $\overline{\text{LDAC}}$ low or use a software load-DAC command to update the selected DAC outputs.				
0	0	1	0	Software load-DAC command. Updates the output of the selected DAC channel(s). Depending on the address bits, this command updates one DAC output, a pair of DAC outputs, or all the DAC outputs simultaneously. Data bits D15–D0 are ignored.				
0	0	1	1	This instruction writes and calibrates the 16-bit input data word for gain and offset errors and immediately updates the DAC outputs for the selected address.				
0	1	0	0	Read command. Depending on the address bits, one of the input register values is read back through DOUT. Data bits D15–D0 are ignored. See the <i>Daisy-Chain Operation</i> section.				
0	1	0	1	Reserved; do not use.				
0	1	1	0	Reserved; do not use.				
0	1	1	1	MAC-bypass command. Depending on the address bits, one, two, or all DAC registers are loaded with a 16-bit data word at DIN. The input data is not calibrated for gain and offset errors (see the <i>Software MAC-Bypass</i> section). Selected DAC output(s) are immediately updated.				
1	0	0	0	Loads D15–D0 into one or two of the gain register(s) for the selected address. The data for the selected address is calibrated for gain error. Drive LDAC low or use a software load-DAC command to update the selected DAC outputs.				
1	0	0	1	Loads D15–D0 into one or two of the offset register(s) for the selected address. The data for the selected address is calibrated for offset error. Drive LDAC low or use a software load-DAC command to update the selected DAC outputs.				
1	0	1	0	Read command. Reads one of the gain registers and presents the data at DOUT.				
1	0	1	1	Read command. Reads one of the offset registers and presents the data at DOUT.				
1	1	0	0	Write command. Loads D15–D0 into the configuration register.				
1	1	0	1	Read command. Reads the contents of the configuration register.				
1	1	1	0	Read command. Reads the DAC register for the selected address.				
1	1	1	1	Reset instruction.				

### 

#### **Gain and Offset Registers**

The MAX5762–MAX5765 contain a gain and offset register associated with each channel to correct real-time gain and offset errors associated with each channel. The gain and offset registers can be accessed individually or in pairs.

The gain word range is limited between 0 and 1. The gain word is an unsigned 16-bit data word. A gain word of 0000h is a special mapping to provide a gain of absolute one, i.e., no gain correction. For all the other gain word codes, the amount of gain correction varies proportionally to the gain word's decimal value. For

example, a gain word of 0001h is equivalent to a gain of  $0.5^{16}$  and a gain of FFFFh is equivalent to 1 -  $0.5^{16}$  (see Table 9). To access the gain register, set control bits C3–C0 = 1000 (see Tables 7 and 8).

The offset has a range from -FS/2 to + (FS/2 - 1 LSB). The offset word is a 16-bit data word represented in two's complement. For example, an offset word equivalent to 7FFFh would provide an offset of FS/2 - 1 LSB and offset word of 8000h would provide an offset of -FS/2 (see Table 10). To access the offset register, set control bits C3–C0 = 1001 (see Tables 7 and 8).

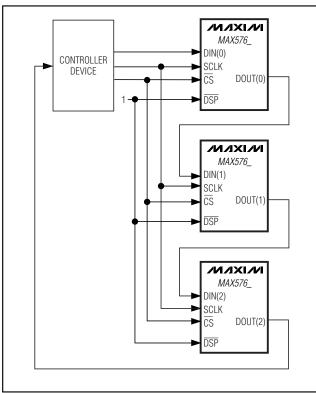


Figure 5. Daisy-Chain Configuration

GAIN VALUE
Unity gain
0.00001526
0.00003052
0.49998474
0.5
0.99996948
0.99998474

#### **Table 9. Gain Register Code Values**

#### Table 10. Offset Register Code Values

CODE (DATA BITS D15–D0)	OFFSET VALUE*		
7FFFh	FS/2 -1 LSB		
7FFEh	FS/2 - 2 LSB		
0002h	2 LSB		
0001h	1 LSB		
0000h	0		
FFFFh	-1 LSB		
FFFEh	-2 LSB		
8001h	-FS / 2 + 1 LSB		
8000h	-FS / 2		
1.0			

 $*1 LSB = FS / 2^{16}$ 



#### **Configuration Register**

The configuration register controls the advanced features of the MAX5762–MAX5765. Write to the configuration register by setting control bits C3-C0 = 1100 and address bits 000000. Table 11 shows the configuration register data format for the D15–D0 data bits. Table 12 shows the commands controlled by the configuration register.

#### DSP Mode ( $\overline{DSP}$ )

The MAX5762–MAX5765 provide a hardware-selectable DSP-interface mode. The active-low DSP logic input selects the microcontroller ( $\mu$ C)-interface or DSP-interface mode. Drive DSP high for  $\mu$ C-interface mode. Drive DSP low for DSP-interface mode. DSP mode, when active, allows chip select (CS) to go high before the entire 32-bit command word is clocked in. Figure 1 illustrates serial timing for both  $\mu$ C- and DSP-interface modes.

### Table 11. Configuration Register Data Format

	16 DATA BITS														
D15	D15 D14 D13 D12 D11 D10 D9 D8 D7 D6 D5 D4 D3 D2 D1 D0								D0						
ERRF	SING	GLT	DT	SHDN	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х	Х

X = Don't care.

### **Table 12. Configuration Register Commands**

DATA BIT	NAME	DESCRIPTION
D15	ERRF	<ul> <li>Error flag. ERRF goes logic-high when an invalid command is attempted. ERRF is cleared each time the configuration register is read back to DOUT. Reset-register commands C3–C0 = 1111 reset ERRF. Conditions that trigger ERRF include:</li> <li>Attempted read of a wrong or invalid address bits A5–A0</li> <li>Access to reserved addresses</li> <li>The default is logic-low (no error flags); ERRF is read only.</li> </ul>
D14	SING	Single device. SING determines daisy-chain or stand-alone mode. Logic-high sets the device to operate in stand-alone mode or in parallel with other devices. Only 16 data bits are output to DOUT when SING is logic-high. When SING is logic-low, the entire 32-bit command word is output to DOUT. The default is logic low (daisy-chain mode). For daisy-chain operation, set SING to logic-low and DSP must be set high (µC mode). SING is read/write.
D13	GLT	Glitch-suppression enable. The MAX5762–MAX5765 feature glitch-suppression circuitry on the analog outputs that minimizes the output glitch during a major carry transition. A logic-low disables the internal glitch-suppression circuitry. Logic-high enables glitch suppression, suppressing up to a 120nV-s glitch impulse on the DAC outputs. Default is logic-low (glitch suppression is disabled). GLT is read/write.
D12	DT	Digital output enable. A logic-low enables DOUT. Logic-high disables DOUT. Disabling DOUT reduces power consumption and digital noise feedthrough to the DAC outputs from the DOUT output buffer. Default is logic-low (DOUT enabled); DT is read/write.
D11	SHDN	Shutdown. A logic-high shuts down all 32 DACs. The logic interface remains active, and the data is retained in the DAC and input registers. Read/write operations can be performed while the device is shut down; however, no changes can occur at the device outputs. Logic-low powers up all 32 DACs. Upon waking up (5µs (typ)), the DAC outputs return to the last stored value in the DAC registers. Default is logic-low (normal operation). SHDN is read/write.
D10-D0	Х	Don't care.

#### SING

When SING = 0, the MAX5762–MAX5765 are in daisychain mode. For daisy-chain operation, set  $\overline{\text{DSP}}$  high (µC mode) and SING = 0.

In daisy-chain operation, DOUT follows DIN after 32 clock cycles for a write command. For a read command, DOUT provides only the 16 data bits in the next cycle following the  $\overline{CS}$  falling edge. Data is provided MSB first at DOUT on the falling edge of SCLK.

When SING = 1, the device is in stand-alone mode. To reduce the time it takes to read data out, the read data is provided MSB first at DOUT on the last 16 cycles of the current command word. The device acts on an

incoming command word independent of the rising edge of CS. SING functionality is ignored in DSP mode.

#### **Daisy-Chain Operation**

Daisy chain any number of the MAX5762–MAX5765 devices by connecting the DOUT of one device to the DIN of another. Set  $\overrightarrow{\text{DSP}}$  high and  $\overrightarrow{\text{SING}} = 0$  for all devices in the daisy chain (see Figure 5).

The MAX5762–MAX5765 support daisy-chain connections of multiple devices. The default (power-up) configuration for the MAX5762–MAX5765 assumes that the device may be part of a daisy chain of devices (SING = 0 and  $\overline{\text{DSP}}$  = 1). For a write command, DOUT follows DIN after 32 clock cycles in the default configuration. Figures 6 and 7 show examples of daisy-chain data sequences.

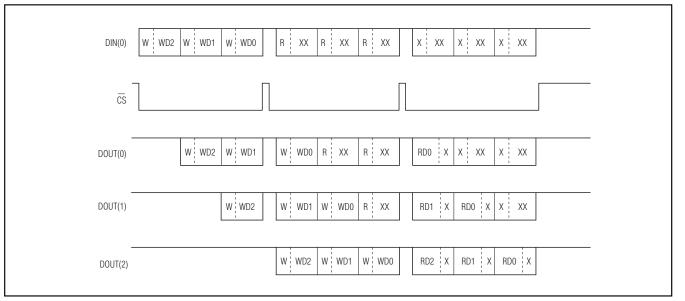


Figure 6. Example 1 of a Daisy-Chain Data Sequence

*W/WD0 = 32-bit word with a write command; WD0 writes data for device 0. The 0 refers to the position in the daisy chain (0 is closest to the bus master). Devices 1 and 2 are devices further down the chain.* 

RDn/X = 32-bit word with a 16-bit read output from device n followed by don't-care bits.

X = Don't care (for X in the data or command position).

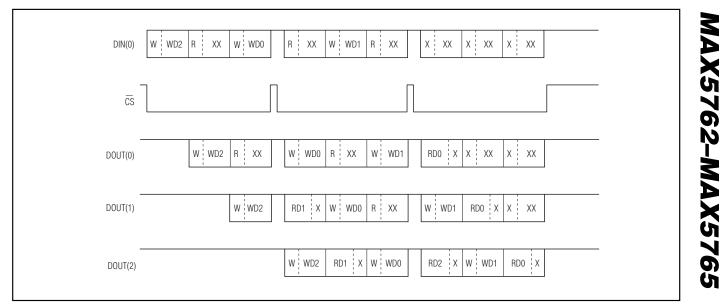


Figure 7. Example 2 of a Daisy-Chain Data Sequence

*W/WD0* = 32-bit word with a write command; WD0 writes data for device 0. The 0 refers to the position in the daisy chain (0 is closest to the bus master). Devices 1 and 2 are devices further down the chain.

RDn/X = 32-bit word with a 16-bit read output from device n followed by don't-care bits.

X = Don't care (for X in the data or command position).

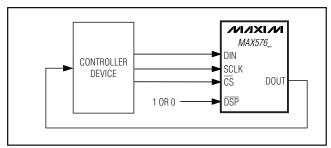


Figure 8. Stand-Alone Configuration

#### **Data Readback**

Read the contents of the MAX5762–MAX5765 DAC and configuration registers at DOUT by issuing a read-data command. Control bits C3–C0 configure the device for the read-data modes (see Table 8). The address bits

select the register(s) to be read. The contents of the register(s) are clocked out MSB first at DOUT on the falling edge of SCLK. The output data format depends on the status of DSP and SING. Table 13 shows the manner in which data is written to DOUT.

#### **Shutdown Mode**

The MAX5762–MAX5765 feature a software-controlled, low-power shutdown mode.

Setting bit 9 of the configuration register to a logic-high disables the analog section of the device, forcing the outputs to go high impedance. In shutdown, supply current is reduced to 50µA typical. Data stored in the DAC and input registers is retained, and the device outputs return to their previous values upon exiting shutdown. Wake-up time is 5µs (typ). The serial interface remains active while the device is in shutdown.

# Table 13. Read-Data Modes with SING and DSP Controls

DSP	SING	CONFIGURATION	READ DATA AT DOUT
0	Х	Stand-alone DSP mode	DOUT provides the 16 data bits only. The 16 data bits are clocked out MSB first at DOUT, on the last 16 clock edges of the current read command word. See Figures 8, 9, and 10.
1	0	Standard daisy-chain configuration µC mode	For write commands, DOUT follows DIN after 32 clocks. The entire 32-bit write command word (both command word and data) is clocked out MSB first at DOUT. For read commands, the 16 data bits of the selected register are valid at DOUT starting with the first clock falling edge after the falling edge of $\overline{\text{CS}}$ . See Figures 6 and 7.
1	1	Stand-alone µC mode	DOUT provides the 16 data bits of the selected register from the current read command word. The 16 data bits are clocked out MSB first at DOUT on the last 16 clock edges of the current read command word. See Figures 8, 9, and 10.

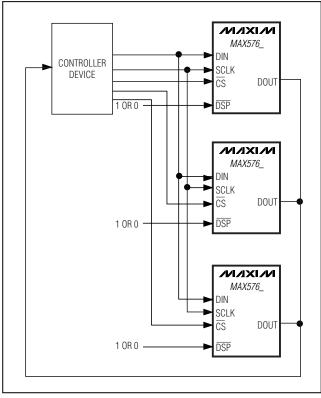


Figure 9. Example of a Parallel Configuration with Readback

#### **Power-Up State**

The MAX5762–MAX5765 monitor the four power supplies and maintain the output buffers in a known state until sufficient voltage is available to ensure that no output glitches occur. Once the minimum voltage threshold has been exceeded, the device outputs come up in the clear state (all outputs = 0).

# Applications Information

#### Automatic Test Equipment (ATE) Applications

The MAX5762–MAX5765 include many features suited for ATE applications. These devices are the most highly integrated level-setting solution available for high-density pin electronics boards, and provide the output voltage ranges required by most ATE applications. The offset DAC simultaneously adjusts the voltage range of all 32 DACs, allowing optimization to the application. The remote-sense feature (GS1 and GS2) allows the pin electronic voltages to be referenced to the ground potential at the DUT site. An integrated offset and gain feature eliminates the need for costly external circuitry.

The pipelined register architecture allows all 32 DACs to be updated simultaneously. This is valuable during test setups, as all values in the tester can be set and then updated in unison with a single command. Accessing the serial interface or the LDAC input updates all 32 DACs simultaneously.

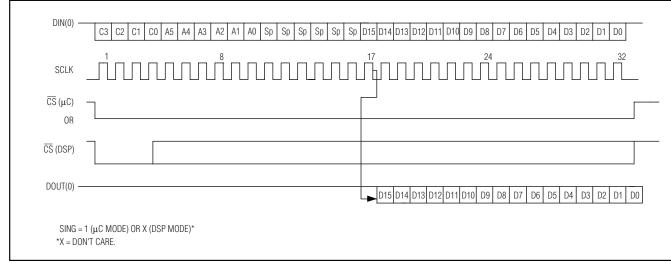


Figure 10. Read Data Timing When Not Daisy Chained

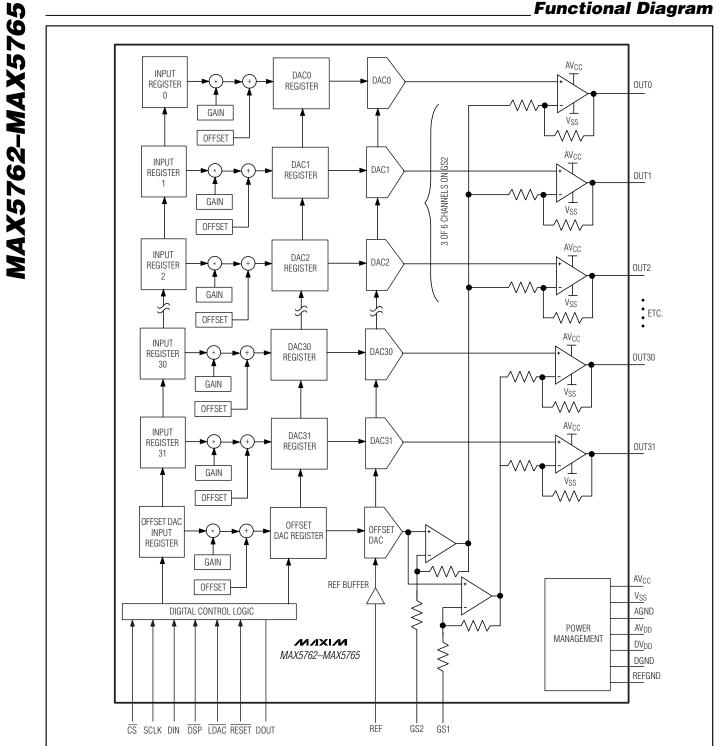
The low output noise of the MAX5762–MAX5765 allows direct connection to the pin electronics, eliminating the cost and PC board area of external filtering.

Modern pin electronics integrated circuits (PEICs) are typically fabricated on high-speed processes with low breakdown voltages. Some devices require external protection on their reference inputs to satisfy absolute maximum ratings. The MAX5762–MAX5765 feature outputs that are almost rail-to-rail allowing the AV<sub>CC</sub> and V<sub>SS</sub> supplies to be set to voltages within the absolute maximum ratings of the PEIC to guarantee that the PEIC is protected in all situations.

### Power Supplies, Bypassing, \_\_\_\_Decoupling, and Layout

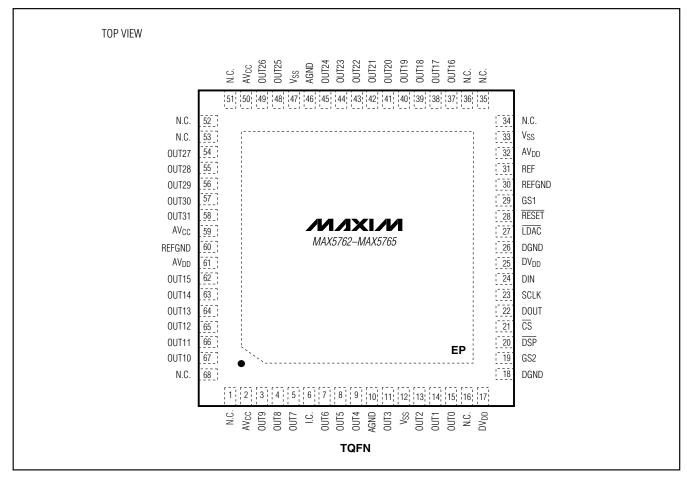
Grounding and power-supply decoupling strongly influence device performance. Digital signals can couple through the reference input, power supplies, and ground connection. Proper grounding and layout can reduce digital feedthrough and crosstalk. For noisy environments, bypass all power supplies with  $0.1\mu$ F and  $1\mu$ F on each pin, as close to the device as possible.

The MAX5762–MAX5765 have four separate power supplies. AV\_DD powers the internal analog circuitry (except for the output buffers), and DV\_DD powers the digital section of the device. AV\_CC and V\_SS power the output buffers.



**Functional Diagram** 

# **Pin Configurations (continued)**



# \_Ordering Information (continued)

PIN- PACKAGE	INL (LSB)	OUTPUT VOLTAGE RANGE (V)	PKG CODE
68 TQFN-EP*	±16	0 to +10	T6800-3
64 TQFP	±16	0 to +10	C64-12
68 TQFN-EP*	±16	-2.5 to +7.5	T6800-3
64 TQFP	±16	-2.5 to +7.5	C64-12
68 TQFN-EP*	±16	-5 to +5	T6800-3
64 TQFP	±16	-5 to +5	C64-12
	PACKAGE           68 TQFN-EP*           64 TQFP           68 TQFN-EP*           64 TQFP           68 TQFN-EP*           64 TQFP	PACKAGE         (LSB)           68 TQFN-EP*         ±16           64 TQFP         ±16           68 TQFN-EP*         ±16           64 TQFP         ±16           64 TQFP         ±16           64 TQFN-EP*         ±16           64 TQFN-EP*         ±16	PIN- PACKAGE         INL (LSB)         VOLTAGE RANGE (D)           68 TQFN-EP*         ±16         0 to ±10           64 TQFP         ±16         0 to ±10           68 TQFN-EP*         ±16         -2.5 to ±7.5           64 TQFP         ±16         -2.5 to ±7.5           64 TQFN-EP*         ±16         -5.5 to ±5.5

**Note:** All devices are specified over the 0°C to +85°C temperature range.

\*EP = Exposed paddle. Connect to V<sub>SS</sub> or leave unconnected. \*\*Future product—contact factory for availability. **Chip Information** 

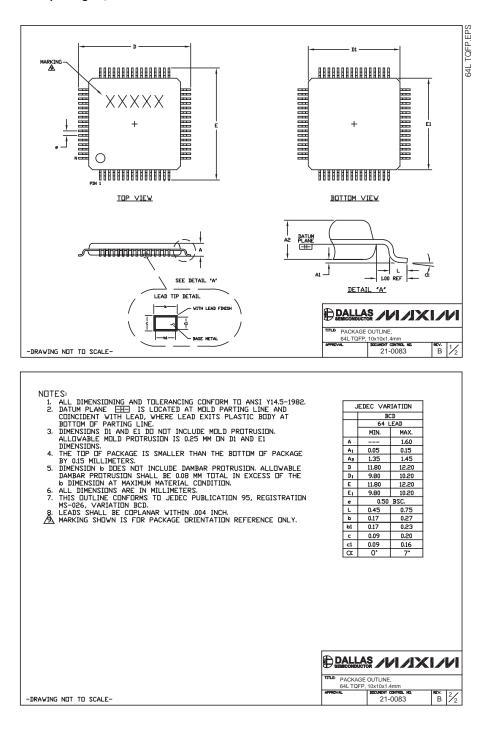
PROCESS: BiCMOS

### **Revision History**

Pages changed at Rev 1: 1, 2, 4, 6, 8, 30, 31.

## \_Package Information

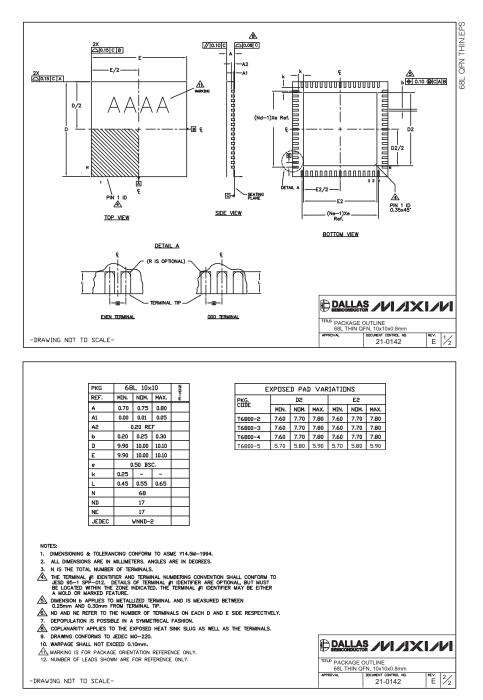
(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to **www.maxim-ic.com/packages**.)



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## Package Information (continued)

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to **www.maxim-ic.com/packages**.)



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